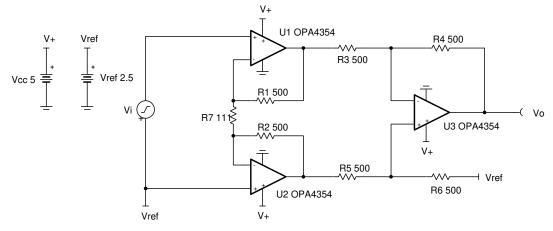
TEXAS INSTRUMENTS

Design Goals

Input		Output		Bandwidth	Supply		
V _{iMin}	V _{iMax}	V _{oMin}	V _{oMax}	BW	V _{cc}	V _{ee}	V _{ref}
-0.24V	+0.24V	+0.1V	+4.9V	10MHz	+2.5V	0V	2.5V

Design Description

This design uses 3 op-amps to build a discrete wide bandwidth instrumentation amplifier. The circuit converts a differential, high frequency signal to a single-ended output.



Design Notes

- 1. Reduce the capacitance on the output of each op amp to avoid stability issues.
- 2. Use low gain configurations to maximize the bandwidth of the circuit.
- 3. Use precision resistors to achieve high DC CMRR performance.
- 4. Use small resistors in op-amp feedback to maintain stability.
- 5. Set the reference voltage, V_{ref}, at mid-supply to allow the output to swing to both supply rails.
- 6. Phase margin of 45° or greater is required for stable operation.
- 7. R₇ sets the gain of the instrumentation amplifier.
- Linear operation depends upon the input common–mode and the output swing ranges of the discrete op amps used. The linear output swing ranges are specified under the A_{ol} test conditions in the op amps datasheets.
- 9. V_{ref} also sets the common-mode voltage of the input, V_i, to ensure linear operation.

1



Design Steps

1. The transfer function of the circuit is given below.

$$V_o = V_i \times \left(1 + \frac{2 \times R_1}{R_7}\right) \times \left(\frac{R_6}{R_5}\right)$$

where V_i is the differential input voltage

Vref is the reference voltage provided to the amplifier

$$Gain = \left(1 + \frac{2 \times R_1}{R_7}\right) \times \left(\frac{R_6}{R_5}\right)$$

2. To maximize the usable bandwidth of design, set the gain of the diff amp stage to 1V/V. Use smaller value resistors to minimize noise.

Choose $R_3 = R_4 = R_5 = R_6 = 500 \Omega$ (Standard value)

3. Choose values for resistors R_1 and R_2 . Keep these values low to minimize noise.

 $R_1 = R_2 = 500 \Omega$ (Standard value)

4. Calculate resistor R_7 to set the gain of the circuit to 10V/V

$$G = \left(1 + \frac{2 \times R_1}{R_7}\right) = 10 \frac{V}{V} \rightarrow \frac{2 \times 500\Omega}{R_7} = 9 \frac{V}{V}$$
$$R_7 = \frac{1000\Omega}{9 \frac{V}{V}} = 111.11\Omega \rightarrow R_7 = 111\Omega \text{ (Standard Value)}$$

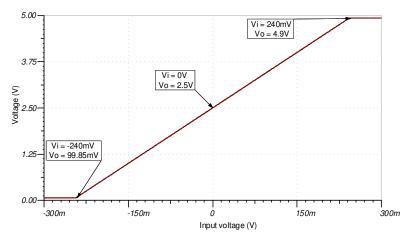
5. Calculate the reference voltage to bias the input to mid-supply. This will maximize the linear output swing of the instrumentation amplifier. See References for more information on the linear operating region of instrumentation amplifiers.

$$V_{ref} = \frac{V_s}{2} = \frac{5}{2} = 2.5$$
 V

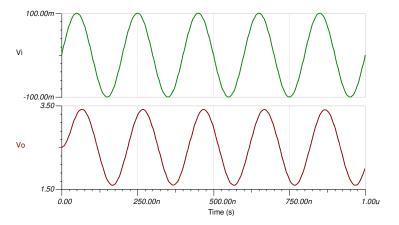


Design Simulations

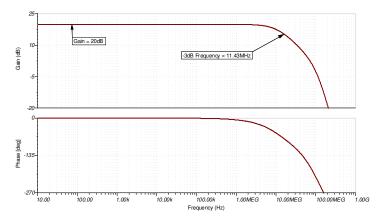
DC Simulation Results



Transient Simulation Results



AC Simulation Results





References

- 1. Analog Engineer's Circuit Cookbooks
- 2. SPICE Simulation File SBOMAU6
- 3. TI Precision Labs
- 4. Instrumentation Amplifier V_{CM} vs. V_{OUT} Plots
- 5. Common-mode Range Calculator for Instrumentation Amplifiers

Design Featured Op Amp

OPA354				
V _{ss}	2.5V to 5.5V			
V _{inCM}	Rail-to-rail			
V _{out}	Rail-to-rail			
V _{os}	2mV			
lq	4.9mA/Ch			
I _b	ЗрА			
UGBW	250MHz			
SR	150V/µs			
#Channels	1,2,4			
www.ti.com/product/opa354				

Design Alternate Op Amp

OPA322				
V _{ss}	1.8V to 5.5V			
V _{inCM}	Rail–to–rail			
V _{out}	Rail–to–rail			
V _{os}	500µV			
Ιq	1.6mA/Ch			
l _b	0.2pA			
UGBW	20MHz			
SR	10V/µs			
#Channels	1,2,4			
www.ti.com/product/opa322				

Revision History

Revision	Date	Change
A	December 2020	Updated R11 to R7 for resistor number consistency

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