



TPS7B4254-Q1 跟踪容差为 4mV 的 150mA、40V 电压跟踪 LDO

1 特性

- 适用于汽车电子 应用
- 符合 AEC-Q100 标准的下列结果
 - 器件温度等级 1: 环境运行温度范围为 -40°C 至 125°C
 - 器件人体放电模式 (HBM) 静电放电 (ESD) 分类等级 3A
 - 器件组件充电模式 (CDM) ESD 分类等级 C6
- -40V 至 45V 宽输入电压范围 (最大值)
- 可调输出电压范围: 2V 到 40V
- 150mA 输出电流能力
- $\pm 4\text{mV}$ 超低输出跟踪容差
- 160mV 低压降 ($I_{\text{OUT}} = 100\text{mA}$ 时)
- 低静态电流 (I_{Q}):
 - $< 4\mu\text{A}$ ($\text{ADJ} = \text{低电平}$ 时)
 - 轻负载时的典型值为 $60\mu\text{A}$
- 超宽等效串联电阻 (ESR) 范围
 - $10\mu\text{F}$ 至 $500\mu\text{F}$ 的陶瓷输出电容
 - 等效串联电阻 (ESR) 范围为 $1\text{m}\Omega$ 至 20Ω
- 反极性保护
- 限流和热关断保护
- 针对接地和电源的输出短路保护
- 输出引脚感性钳位
- 8 引脚小外形尺寸 (SO) PowerPAD™ 封装, 带外露散热焊盘

2 应用

- 非板载传感器电源
- 高精度电压跟踪
- 非板载负载的电源开关

3 说明

对于汽车非板载传感器和低电流非板载模块, 电源通过一条长电缆连接主板。在这类情况下, 电源器件需要为非板载负载提供保护, 防止板载组件在接地短路或者电缆破损造成的电池短路期间受损。非板载传感器需要与板载组件采用相同的电源, 以确保数据采集的高精度。

TPS7B4254-Q1 器件设计用于 具有 45V 负载突降问题的汽车类应用。该器件可用作一个跟踪低压降 (LDO) 稳压器或者电压跟踪器, 通过板载主电源为非板载传感器构建一个封闭电源环路。该器件的输出由 ADJ 引脚的基准电压精准调节。

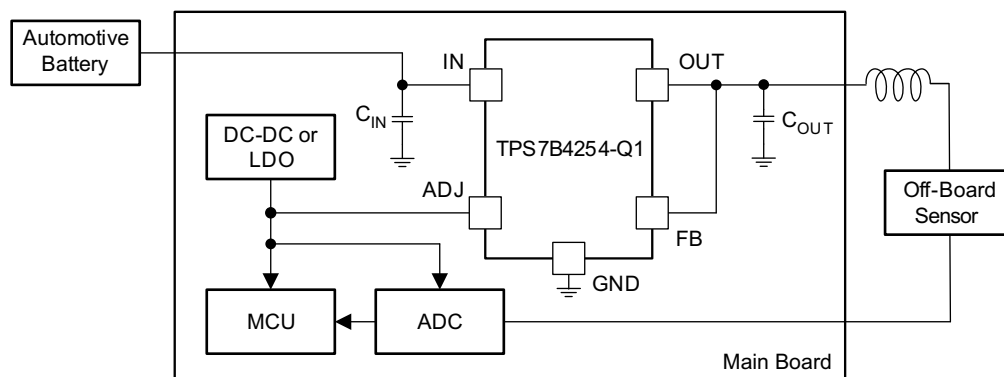
为了给非板载模块提供精确的电源, 该器件的 ADJ 与 FB 引脚间在运行温度范围内具有 4mV 超低跟踪容差。PMOS 背靠背拓扑消除了反极性条件下对外部二极管的需求。TPS7B4254-Q1 器件还包括热关断、感性钳位、过载和电池短路保护, 防止板载组件在极限条件下受损。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
TPS7B4254-Q1	SO PowerPAD (8)	4.89mm × 3.90mm

(1) 要了解所有可用封装, 请参见数据表末尾的可订购产品附录。

典型应用电路原理图



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目录

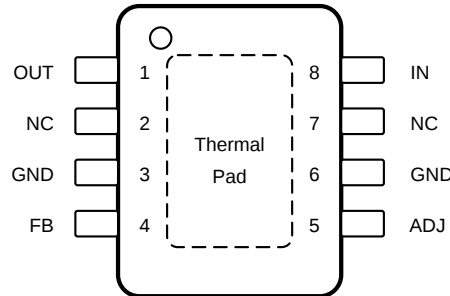
1	特性	1	7.3	Feature Description	10
2	应用	1	7.4	Device Functional Modes	13
3	说明	1	8	Application and Implementation	14
4	修订历史记录	2	8.1	Application Information	14
5	Pin Configuration and Functions	3	8.2	Typical Applications	14
6	Specifications	4	9	Power Supply Recommendations	17
6.1	Absolute Maximum Ratings	4	10	Layout	18
6.2	ESD Ratings	4	10.1	Layout Guidelines	18
6.3	Recommended Operating Conditions	4	10.2	Layout Example	18
6.4	Thermal Information	5	11	器件和文档支持	19
6.5	Electrical Characteristics	5	11.1	社区资源	19
6.6	Typical Characteristics	6	11.2	商标	19
7	Detailed Description	10	11.3	静电放电警告	19
7.1	Overview	10	11.4	Glossary	19
7.2	Functional Block Diagram	10	12	机械、封装和可订购信息	19

4 修订历史记录

Changes from Original (April 2016) to Revision A	Page
• 已将数据表状态由“产品预览”改为“量产数据”	1

5 Pin Configuration and Functions

**DDA Package
8-Pin HSOP With Exposed Thermal Pad
Top View**



NC – No internal connection

Pin Functions

PIN		TYPE ⁽¹⁾	DESCRIPTION
NAME	NO.		
ADJ	5	I	Connect the reference to this pin. A low signal disables the device and a high signal enables the device. The reference voltage can be connected directly or by a voltage divider for lower output voltages. To compensate for line influences, connect a capacitor close to the device pin.
FB	4	I	This pin is the feedback pin, which can connect to the external resistor divider to select the output voltage.
GND	3, 6	G	Ground reference
IN	8	I	This pin is the device supply. To compensate for line influences, connect a capacitor close to the device pin.
NC	2, 7	NC	Not internally connected.
OUT	1	O	Block to GND with a capacitor close to the device pins with respect to the capacitance and ESR requirements listed in the Output Capacitor section.
Exposed thermal pad		—	Connect the thermal pad to the GND pin or leave it floating.

(1) I = input, O = output, G = ground, NC = no internal connection

6 Specifications

6.1 Absolute Maximum Ratings

over operating ambient temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Unregulated input voltage	IN ⁽²⁾	–40	45	V
Regulated output voltage	OUT ⁽²⁾⁽³⁾	–1	45	V
Voltage difference between the input and output	IN – OUT	–40	45	V
Reference voltage	ADJ ⁽²⁾	–0.3	45	V
Feedback input voltage for the tracker	FB ⁽²⁾	–1	45	V
Reference voltage minus the input voltage	ADJ – IN ⁽⁴⁾		18	V
Operating junction temperature, T _J		–40	150	°C
Storage temperature, T _{stg}		–65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to the GND pin.
- (3) An internal diode is connected between the OUT and GND pins with 600-mA dc current capability for inductive clamp protection.
- (4) When the (ADJ – IN) voltage is higher than 18 V, the (ADJ – OUT) voltage should be maintained lower than 18 V, otherwise the device can be damaged.

6.2 ESD Ratings

			VALUE	UNIT
V _(ESD) Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾	All pins except NC	±4000	V
		NC pins	±2000	
	Charged-device model (CDM), per AEC Q100-011		±1000	

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

over operating ambient temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
V _{IN}	Unregulated input voltage ⁽²⁾	4	40	V
V _{ADJ}	Reference input voltage	1.5	18	V
V _{FB}	Feedback input voltage for the tracker	1.5	18	V
V _{OUT}	Regulated output voltage	1.5	40	V
C _{OUT}	Output capacitor requirements ⁽³⁾	10	500	μF
	Output ESR requirements ⁽⁴⁾	0.001	20	Ω
T _J	Operating junction temperature	–40	150	°C

- (1) Within the functional range the device operates as described in the circuit description. The electrical characteristics are specified within the conditions given in the related [Electrical Characteristics](#) table.
- (2) V_{IN} > V_{ADJ} + V_{DROPOUT}
- (3) The minimum output capacitance requirement is applicable for a worst-case capacitance tolerance of 30%. When a resistor divider is connected between the OUT and FB pins (the output voltage is higher than reference voltage), a 47-nF feedforward capacitor is required to be connected between the OUT and FB pins for loop stability, and the ESR range of the output capacitor is required to be from 0.001 to 10 Ω.
- (4) Relevant ESR value at f = 10 kHz

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾		TPS7B4254-Q1	UNIT
		DDA (HSOP)	
		8 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	45.4	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	51.1	°C/W
R _{θJB}	Junction-to-board thermal resistance	27	°C/W
ψ _{JT}	Junction-to-top characterization parameter	8.2	°C/W
ψ _{JB}	Junction-to-board characterization parameter	26.9	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	6.4	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

6.5 Electrical Characteristics

V_{IN} = 13.5 V, V_{ADJ} ≥ 2 V, T_J = –40°C to 150°C, over operating ambient temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V _{IN(UVLO)} IN undervoltage detection	V _{IN} rising			3.65	V
	V _{IN} falling			2.8	V
ΔV _{OUT} Output voltage tracking accuracy ⁽¹⁾	I _{OUT} = 100 μA to 150 mA, V _{IN} = 4 to 40 V V _{ADJ} < V _{IN} – 1 V 2 V < V _{ADJ} < 18 V	–4		4	mV
ΔV _{OUT(ΔIO)} Load regulation, steady-state	I _{OUT} = 0.1 to 150 mA, V _{ADJ} = 5 V			4	mV
ΔV _{OUT(ΔVI)} Line regulation, steady-state	I _{OUT} = 10 mA, V _{IN} = 6 to 40 V, V _{ADJ} = 5 V			4	mV
PSRR Power-supply ripple rejection	f _{rip} = 100 Hz, V _{rip} = 0.5 VPP, C _{OUT} = 10 μF, I _{OUT} = 100 mA		70		dB
V _{DROPOUT} Dropout voltage (V _{DROPOUT} = V _{IN} – V _{OUT})	I _{OUT} = 100 mA, V _{IN} = V _{ADJ} ≥ 4 V ⁽²⁾		160	260	mV
I _{OUT(LIM)} Output current limitation	V _{ADJ} = 5 V, OUT short to GND	151	450	520	mA
I _{R(IN)} Reverse current at IN	V _{IN} = 0 V, V _{OUT} = 40 V, V _{ADJ} = 5 V	–2		0	μA
I _{R(–IN)} Reverse current at negative IN	V _{IN} = –40 V, V _{OUT} = 0 V, V _{ADJ} = 5 V	–10			μA
T _{SD} Thermal shutdown temperature			175		°C
T _{SD_hys} Thermal shutdown hysteresis			15		°C
I _Q Current consumption	4 V ≤ V _{IN} ≤ 40 V, V _{ADJ} = 0 V		2	4	μA
	4 V ≤ V _{IN} ≤ 40 V, V _{ADJ} = 5 V, I _{OUT} < 100 μA		60	100	μA
	4 V ≤ V _{IN} ≤ 40 V, V _{ADJ} = 5 V, I _{OUT} < 150 mA		210	260	μA
I _{Q(DROPOUT)} Current consumption in dropout region	V _{IN} = V _{ADJ} = 5 V, I _{OUT} = 100 μA		70	140	μA
I _{ADJ} Reference input current	V _{ADJ} = V _{FB} = 5 V			5.5	μA
V _{ADJ(LOW)} Reference low signal valid	V _{OUT} = 0 V	0		0.7	V
V _{ADJ(HIGH)} Reference high signal valid	V _{OUT} – V _{ADJ} < 4 mV	2		18	V
I _{FB} FB bias current	V _{ADJ} = V _{FB} = 5 V			0.5	μA

(1) The tracking accuracy is specified when the FB pin is directly connected to the OUT pin which means V_{ADJ} = V_{OUT}, external resistor divider variance is not included.

(2) Measured when the output voltage, V_{OUT}, has dropped 10 mV from the nominal value.

6.6 Typical Characteristics

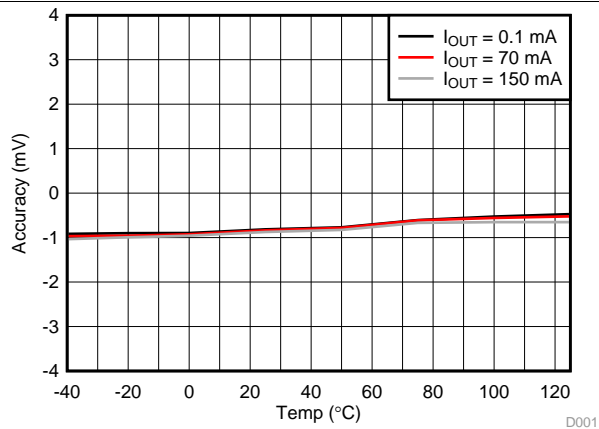


Figure 1. Tracking Accuracy vs Ambient Temperature

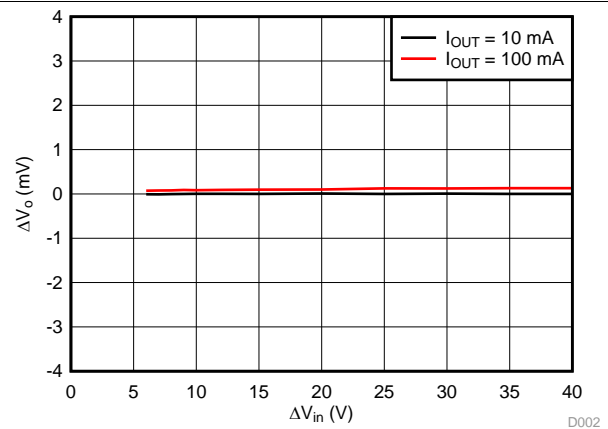


Figure 2. Line Regulation

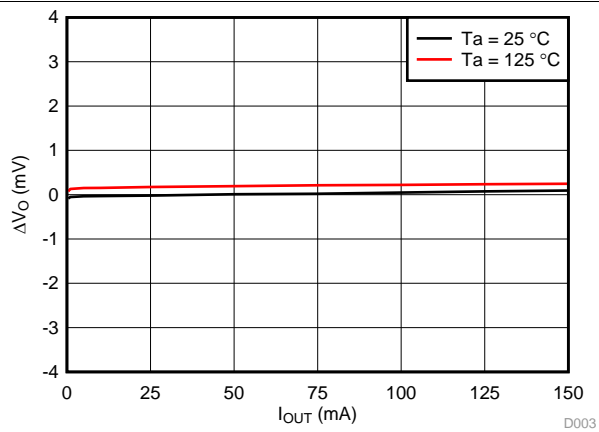


Figure 3. Load Regulation

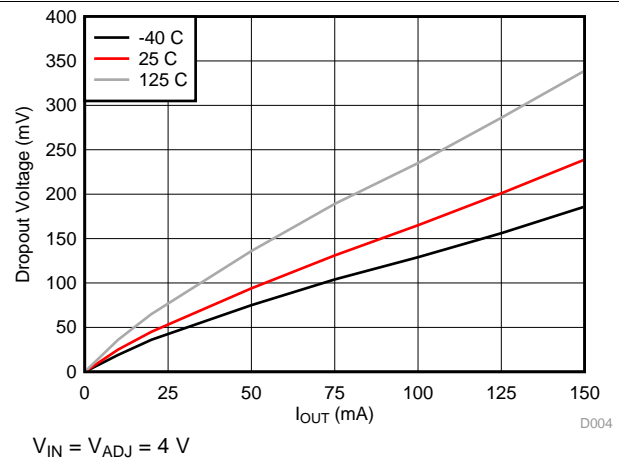


Figure 4. Dropout Voltage vs Output Current

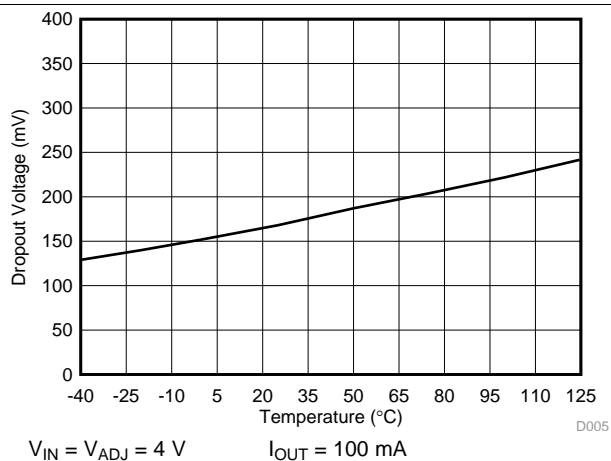


Figure 5. Dropout Voltage vs Ambient Temperature

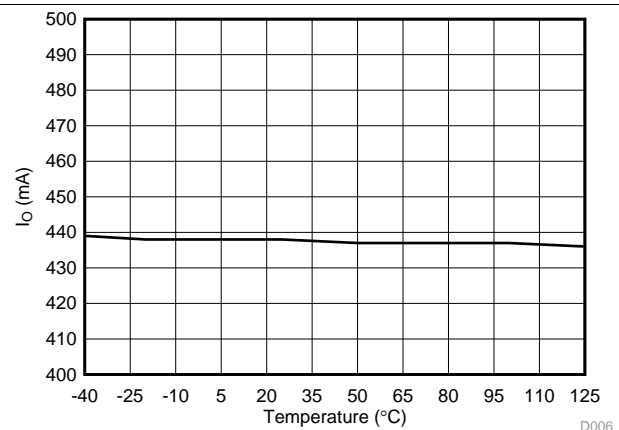


Figure 6. Current Limit ($I_{OUT(LIM)}$) vs Ambient Temperature

Typical Characteristics (continued)

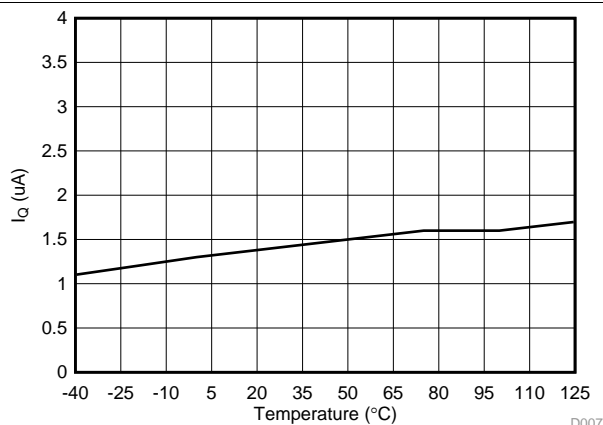


Figure 7. Shutdown Current vs Ambient Temperature

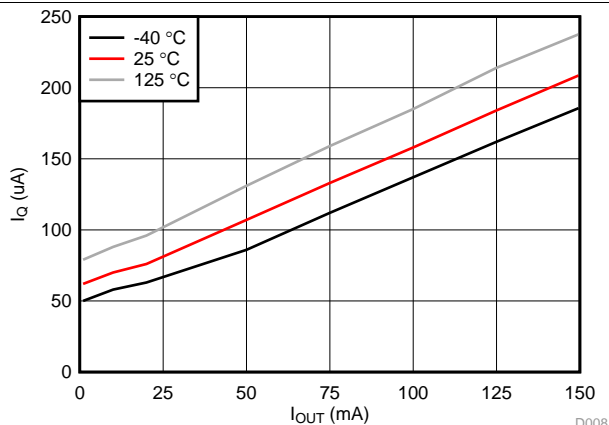


Figure 8. Quiescent Current vs Output Current

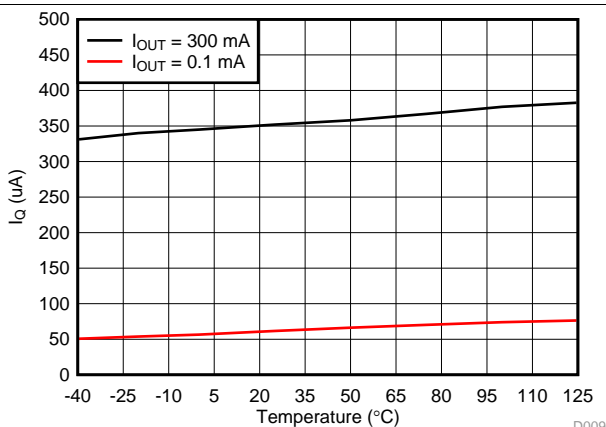


Figure 9. Quiescent Current vs Ambient Temperature

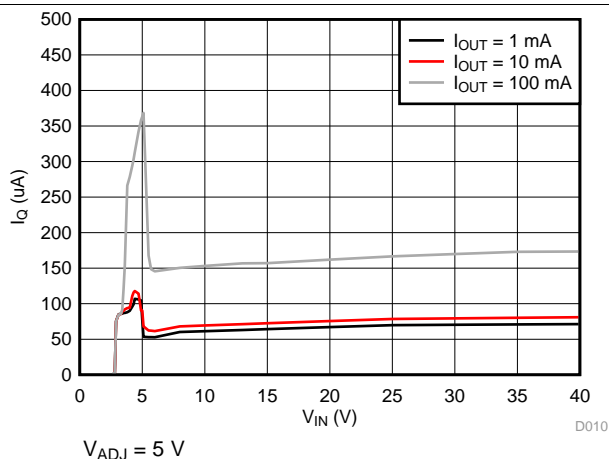


Figure 10. Quiescent Current vs Input Voltage

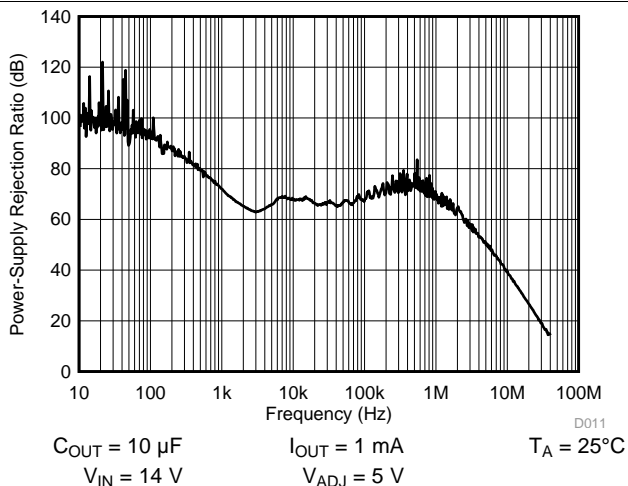


Figure 11. PSRR

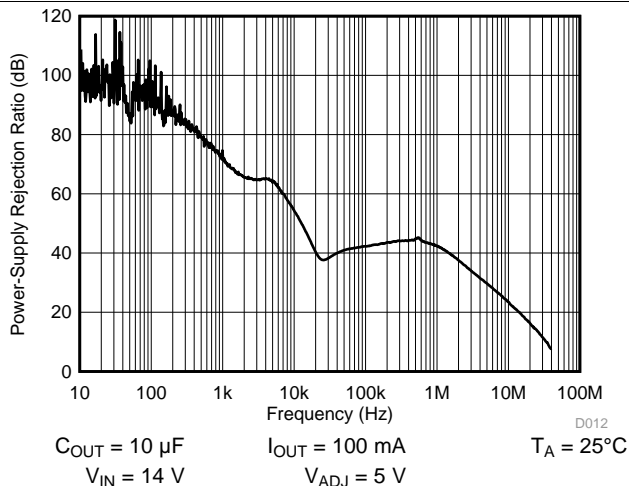


Figure 12. PSSR

Typical Characteristics (continued)

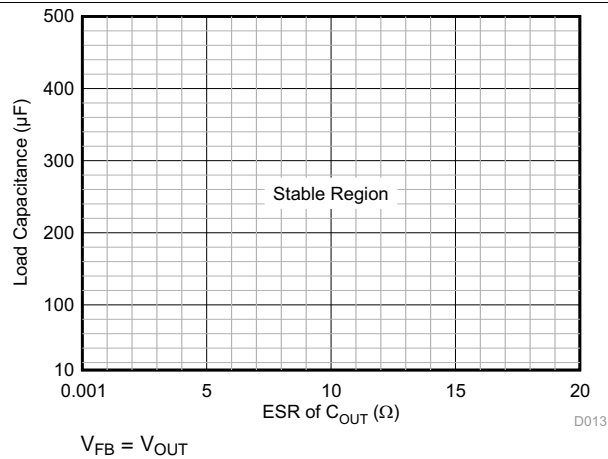


Figure 13. ESR Stability vs Load Capacitance

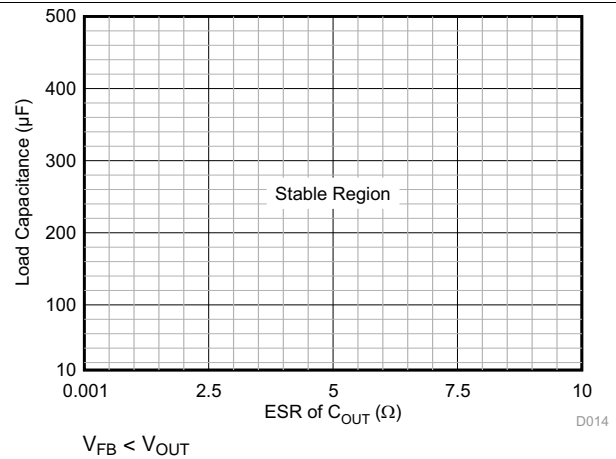


Figure 14. ESR Stability vs Load Capacitance

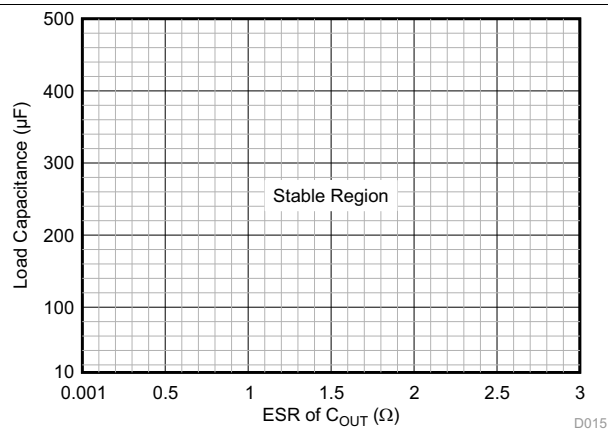


Figure 15. ESR Stability vs Load Capacitance (Multiple Output Capacitors in parallel)

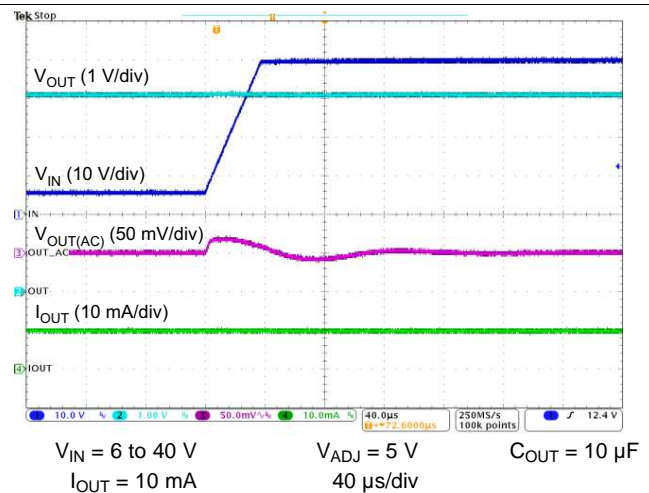


Figure 16. 6-V to 40-V Line Transient

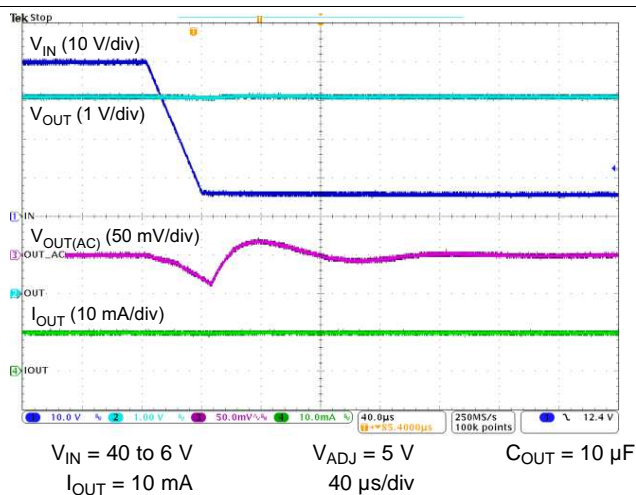


Figure 17. 40-V to 6-V Line Transient

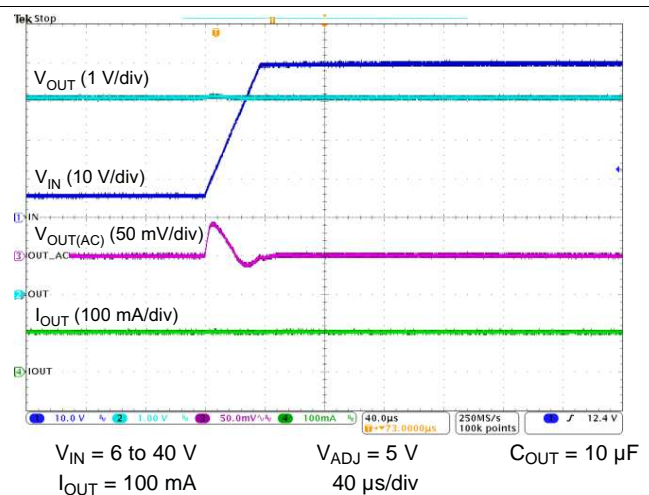


Figure 18. 6-V to 40-V Line Transient

Typical Characteristics (continued)

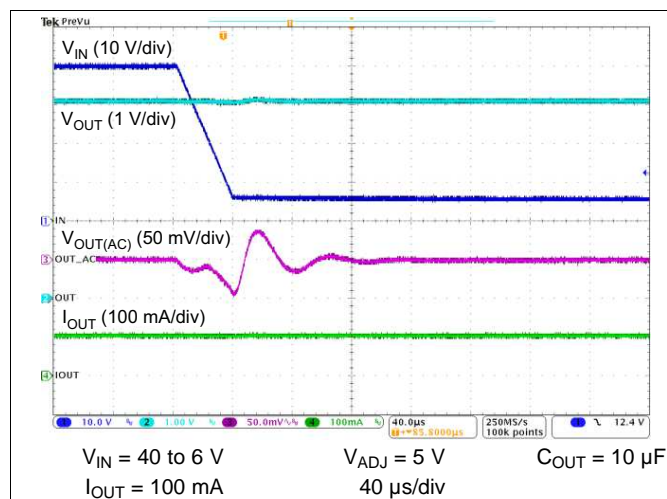


Figure 19. 40-V to 6-V Line Transient

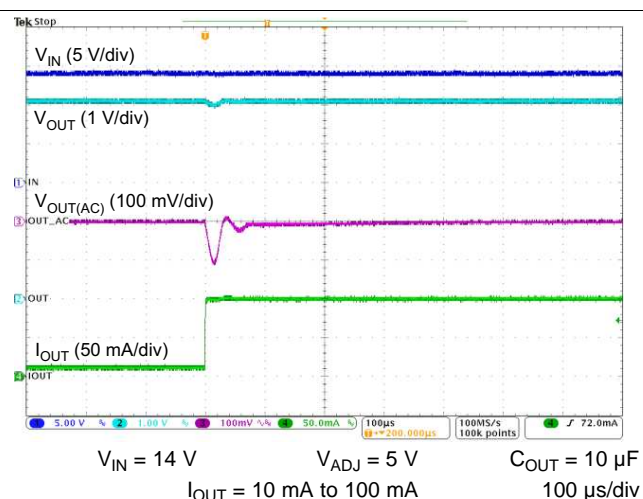


Figure 20. 10-V to 100-mA Load Transient

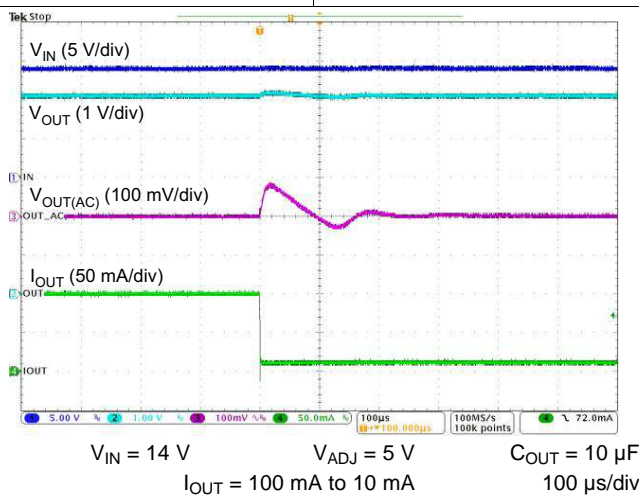


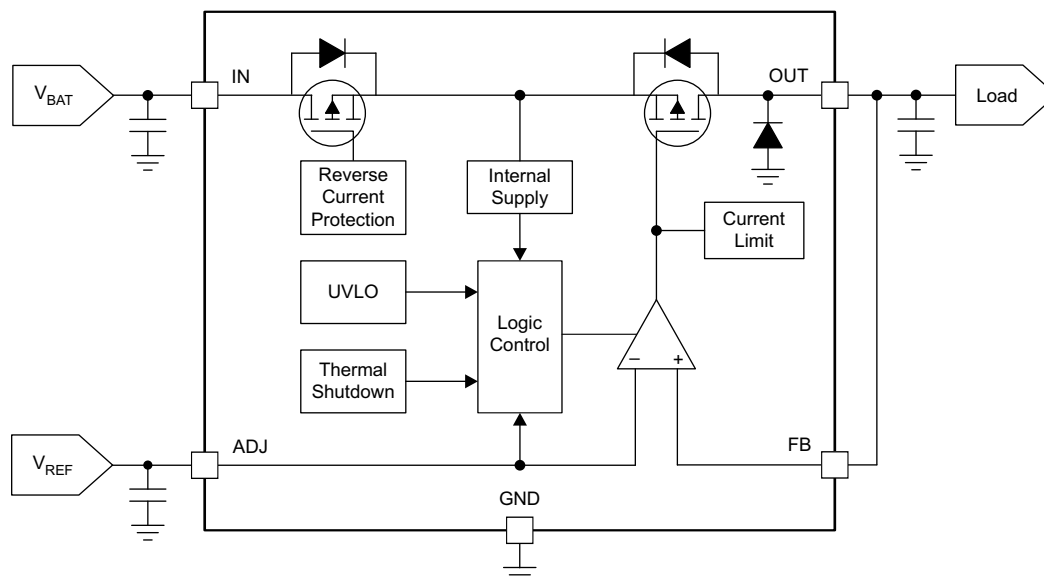
Figure 21. 100-mA to 10-mA Load Transient

7 Detailed Description

7.1 Overview

The TPS7B4254-Q1 device is a monolithic integrated low-dropout voltage tracker with an ultralow tracking tolerance. Key protection circuits are integrated in the device, including output current limitation, reverse polarity protection, inductive load clamp, output short-to-battery protection, and thermal shutdown in case of an overtemperature event.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 Short-Circuit and Overcurrent Protection

The TPS7B4254-Q1 device features integrated fault protection, which makes the device ideal for automotive applications. To keep the device in a safe area of operation during certain fault conditions, internal current-limit protection is used to limit the maximum output current. This protection protects the device from excessive power dissipation. For example, during a short-circuit condition on the output, the current through the pass element is limited to $I_{OUT(LIM)}$ to protect the device from excessive power dissipation.

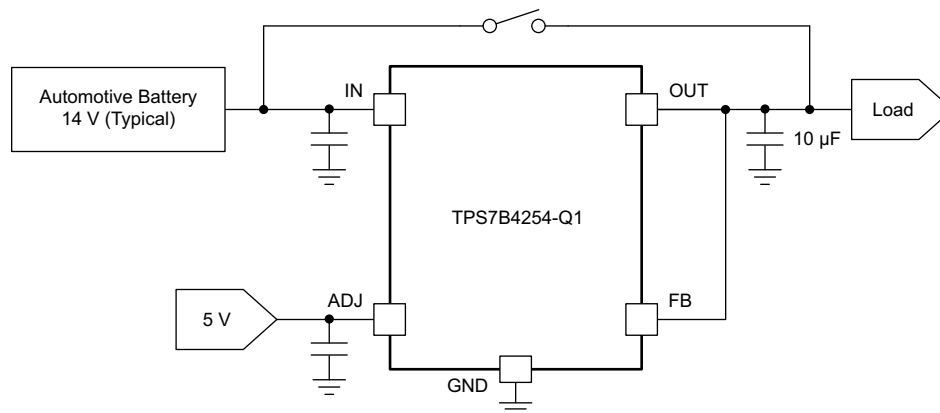
7.3.2 Integrated Inductive Clamp Protection

During output turnoff, the cable inductance continues to source the current from the output of the device. The device integrates an inductive clamp at the OUT pin to help to dissipate the inductive energy stored in the cable. An internal diode is connected between the OUT and GND pins with a dc-current capability of 600 mA for inductive clamp protection.

7.3.3 OUT Short-to-Battery and Reverse-Polarity Protection

The TPS7B4254-Q1 device can withstand a short to battery on the output, as shown in [Figure 22](#). Therefore, no damage to the device occurs.

Feature Description (continued)

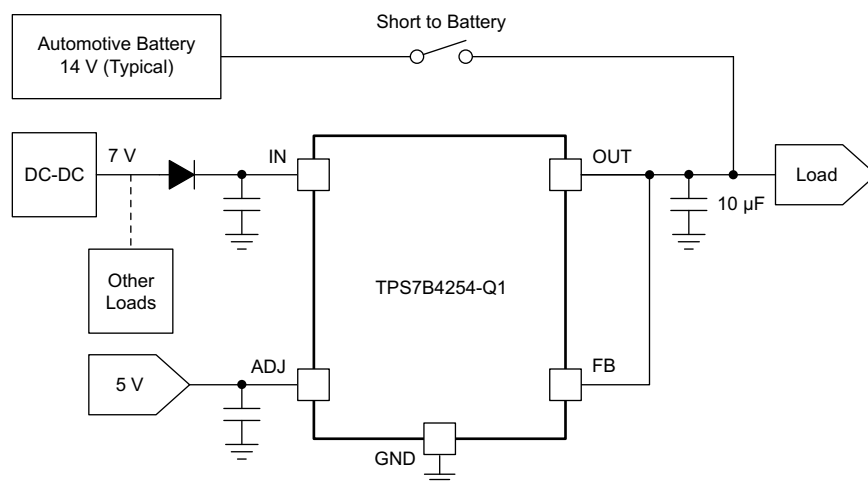


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Figure 22. OUT Short to Battery, $V_{IN} = V_{BAT}$

A short to the battery can also occur when the device is powered by an isolated supply at lower voltage, as shown in Figure 23. In this case, the TPS7B4254-Q1 supply-input voltage is set to 7 V when a short to battery (14 V typical) occurs on the OUT pin, which operates at 5 V. The internal back-to-back PMOS remains on for 1 ms, during which the input voltage of the TPS7B4254-Q1 device charges up to the battery voltage. A diode connected between the output of the dc-dc converter and the input of the TPS7B4254-Q1 device is required in case the other loads connected behind the dc-dc converter cannot withstand the voltage of an automotive battery. To achieve a lower dropout voltage, TI recommends using a Schottky diode. This diode can be eliminated if the output of the dc-dc converter and the loads connected behind it withstand automotive battery voltage.

The internal back-to-back PMOS is switched to OFF when reverse polarity or a short to battery occurs for 1 ms. After that, the reverse current that flows out through the IN pin is less than 10 µA. Meanwhile, a special ESD structure implemented at the input ensures the device can withstand –40 V.



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Figure 23. OUT Short to Battery, $V_{IN} < V_{BAT}$

In most cases, the output of the TPS7B4254-Q1 device is shorted to the battery through an automotive cable. The parasitic inductance on the cable results in LC oscillation at the output of the TPS7B4254-Q1 device when the short to battery occurs. The peak voltage at the output of the TPS7B4254-Q1 device must be lower than the absolute-maximum voltage rating (45 V) during LC oscillation.

Feature Description (continued)

7.3.4 Undervoltage Shutdown

The device has an internally fixed undervoltage-shutdown threshold. Undervoltage shutdown activates when the input voltage on IN drops below UVLO. This activation ensures the regulator is not latched into an unknown state during a low input-supply voltage. If the input voltage has a negative transient that drops below the UVLO threshold and then recovers, the regulator shuts down and then powers up with a standard power-up sequence when the input voltage is above the required level.

7.3.5 Thermal Protection

The device incorporates a thermal shutdown (TSD) circuit as a protection from overheating. During continuous normal operation, the junction temperature should not exceed the TSD trip point. If the junction temperature exceeds the TSD trip point, the output turns off. When the junction temperature decreases to 15°C (typical) lower than the TSD trip point, the output turns on.

NOTE

The purpose of the internal protection circuitry of the TPS7B4254-Q1 device is to protect against overload conditions and is not intended as a replacement for proper heat-sinking. Continuously running the device into thermal shutdown degrades device reliability.

7.3.6 Regulated Output (OUT)

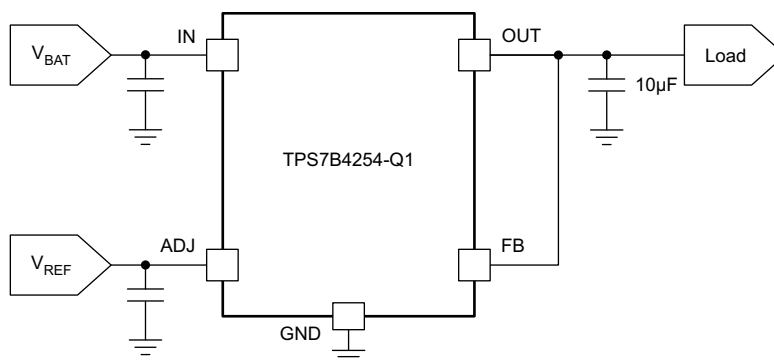
The OUT pin is the regulated output based on the required voltage. The output has current limitation. During initial power up, the regulator has an incorporated soft-start feature to control the initial current through the pass element.

7.3.7 Adjustable Output Voltage (FB and ADJ)

7.3.7.1 OUT Voltage Equal to the Reference Voltage

With the reference voltage applied directly at the ADJ pin and the FB pin connected to the OUT pin, the voltage at the OUT pin equals to the reference voltage at the ADJ pin, as shown in [Figure 24](#).

$$V_{OUT} = V_{ADJ} \quad (1)$$



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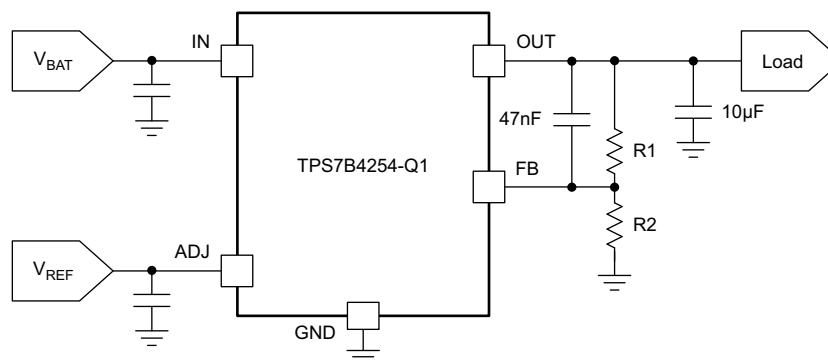
Figure 24. OUT Voltage Equal to the Reference Voltage

7.3.7.2 OUT Voltage Higher Than Reference Voltage

By using an external resistor divider connected between the OUT and FB pins, an output voltage higher than reference voltage can be generated as shown in [Figure 25](#). Use [Equation 2](#) to calculate the value of the output voltage. The recommended range for R1 and R2 is from 10 kΩ to 100 kΩ.

$$V_{OUT} = V_{ADJ} \times \left(1 + \frac{R1}{R2} \right) \quad (2)$$

Feature Description (continued)



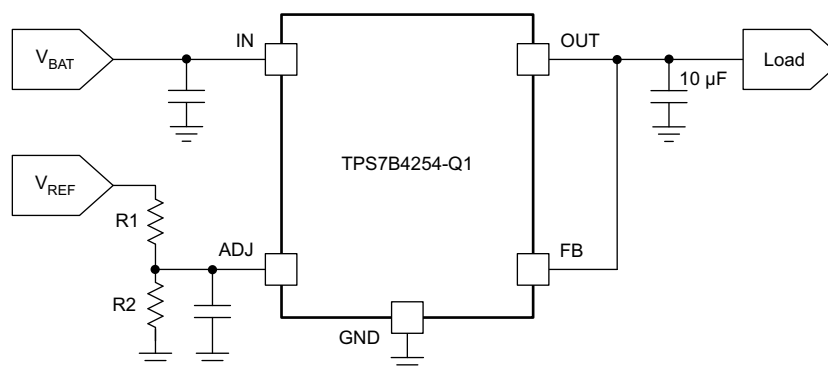
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Figure 25. OUT Voltage Higher Than the Reference Voltage

7.3.7.3 Output Voltage Lower Than Reference Voltage

By using an external resistor divider connected at the ADJ pin, an output voltage lower than reference voltage can be generated as shown in Figure 26. Use Equation 3 to calculate the output voltage. The recommended value for both R1 and R2 is less than 100 kΩ.

$$V_{OUT} = V_{REF} \times \frac{R2}{R1 + R2} \quad (3)$$



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Figure 26. OUT Voltage Lower Than the Reference Voltage

7.4 Device Functional Modes

7.4.1 Operation With $V_{IN} < 4$ V

The maximum UVLO voltage is 3.65 V, and the device generally operates at an input voltage above 4 V. The device can also operate at a lower input voltage; no minimum UVLO voltage is specified. At an input voltage below the actual UVLO voltage, the device does not operate.

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

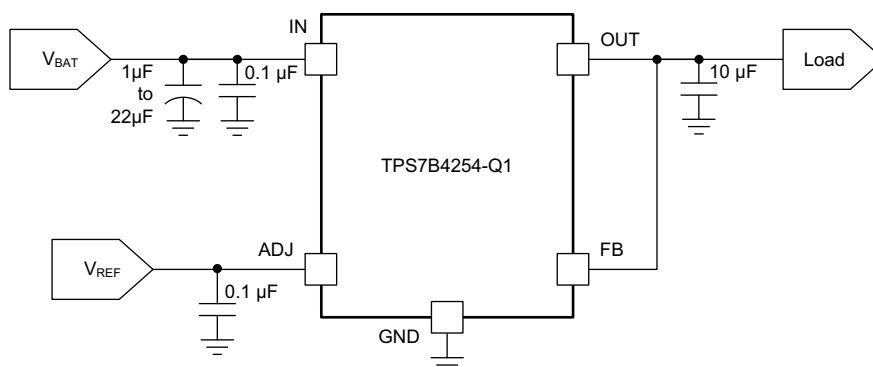
8.1 Application Information

The TPS7B4254-Q1 device is a 150-mA low-dropout tracking regulator with ultralow tracking tolerance. The PSpice transient model is available for download on the product folder and can be used to evaluate the basic function of the device.

8.2 Typical Applications

8.2.1 Application With Output Voltage Equal to the Reference Voltage

Figure 27 shows a typical application circuit for the TPS7B4254-Q1 device. Different values of external components can be used, depending on the end application. An application may require a larger output capacitor during fast load steps to prevent a large drop on the output voltage. TI recommends using a low-ESR ceramic capacitor with a dielectric of type X5R or X7R.



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Figure 27. Output Voltage Equals the Reference Voltage

8.2.1.1 Design Requirements

For this design example, use the parameters listed in Table 1 as the design parameters.

Table 1. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Input voltage	4 V to 40 V
Output voltage	2 V to 40 V
ADJ voltage	2 V to 18 V
Output capacitor	10 µF to 500 µF
Output capacitor ESR range	0.001 Ω to 20 Ω

8.2.1.2 Detailed Design Procedure

To begin the design process, determine the following:

- Input voltage range
- Output voltage
- Reference voltage
- Output current

- Current limit

8.2.1.2.1 Input Capacitor

The device requires an input decoupling capacitor, the value of which depends on the application. The typical recommended value for the decoupling capacitor is 10 μ F with a 0.1 μ F ceramic bypass capacitor in parallel. The voltage rating must be greater than the maximum input voltage.

8.2.1.2.2 Output Capacitor

To ensure the stability of the TPS7B4254-Q1 device, the device requires an output capacitor with a value in the range from 10 μ F to 500 μ F and with an ESR range from 0.001 Ω to 20 Ω when the FB pin is directly connected to the OUT pin. TI recommends selecting a ceramic capacitor with low ESR to improve the load transient response.

To achieve an output voltage higher than the reference voltage, a resistor divider is connected between the OUT pin and the FB pin. In this case, a 47-nF feedforward capacitor must be connected between the OUT and FB pins for loop stability. The ESR of the output capacitor must be from 0.001 Ω to 10 Ω .

When multiple capacitors (two or more) are connected in parallel at the OUT pin, the ESR range of each output capacitor must be from 0.001 Ω to 3 Ω for loop stability.

In case the FB pin is shorted to ground, the TPS7B4254-Q1 device functions as a power switch with no need for the output capacitor.

8.2.1.3 Application Curve

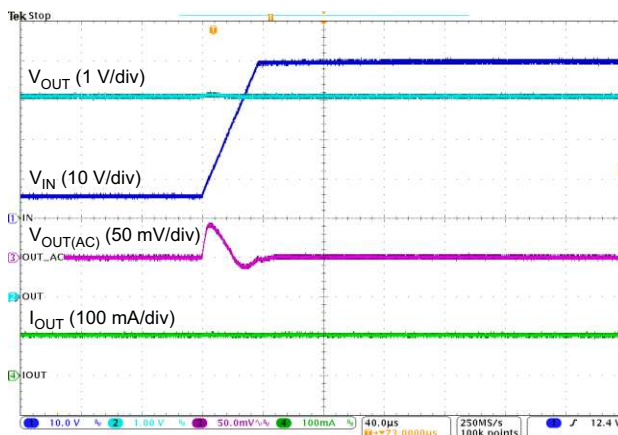
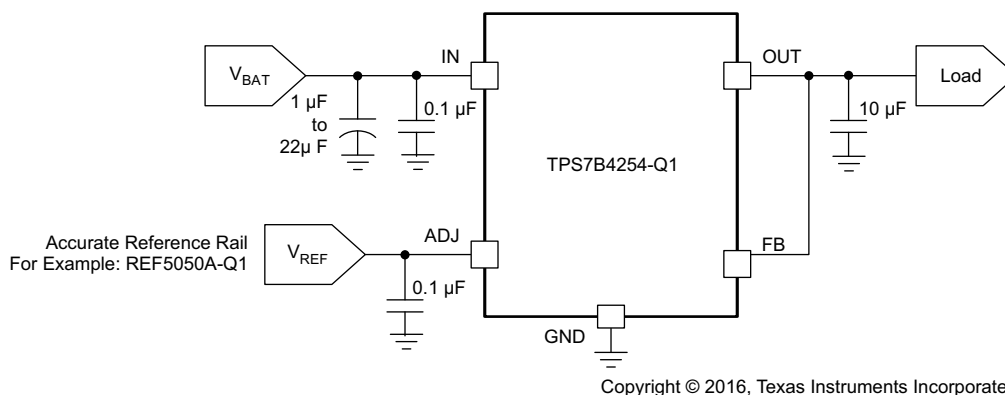


Figure 28. 6-V to 40-V Line Transient

8.2.2 High-Accuracy LDO

With an accurate voltage rail, the TPS7B4254-Q1 device can be used as an LDO with ultrahigh-accuracy output voltage by configuring the device as shown in [Figure 29](#).


Figure 29. High-Accuracy LDO Application

For example, assume the reference voltage is a 5-V rail with 0.1% accuracy. Because the tracking accuracy between the ADJ and OUT pins is specified below 4 mV across temperature, the output accuracy of the TPS7B4254-Q1 device can be calculated with [Equation 4](#).

$$\text{Accuracy of } V_{\text{OUT}} = \frac{V_{\text{ADJ}} \times 0.1\% + 4 \text{ mV}}{V_{\text{OUT}}} \times 100\% = \frac{5 \times 0.1\% + 0.004}{5} \times 100\% = 0.18\% \quad (4)$$

9 Power Supply Recommendations

The device is designed to operate with an input voltage supply from 4 V to 40 V. This input supply must be well regulated. If the input supply is more than a few inches away from the TPS7B4254-Q1 device, TI recommends adding an electrolytic capacitor with a value of 10 μ F and a ceramic bypass capacitor at the input.

10 Layout

10.1 Layout Guidelines

For the layout of the TPS7B4254-Q1 device, place the input and output capacitors close to the devices as shown in the [Functional Block Diagram](#). To enhance the thermal performance, TI recommends surrounding the device with some vias. Minimize equivalent series inductance (ESL) and ESR to maximize performance and ensure stability. Place every capacitor as close as possible to the device and on the same side of the PCB as the regulator.

Do not place any of the capacitors on the opposite side of the PCB from where the regulator is installed. TI strongly discourages the use of vias and long traces for the path between the output capacitor and the OUT pins because vias can negatively impact system performance and even cause instability.

10.2 Layout Example

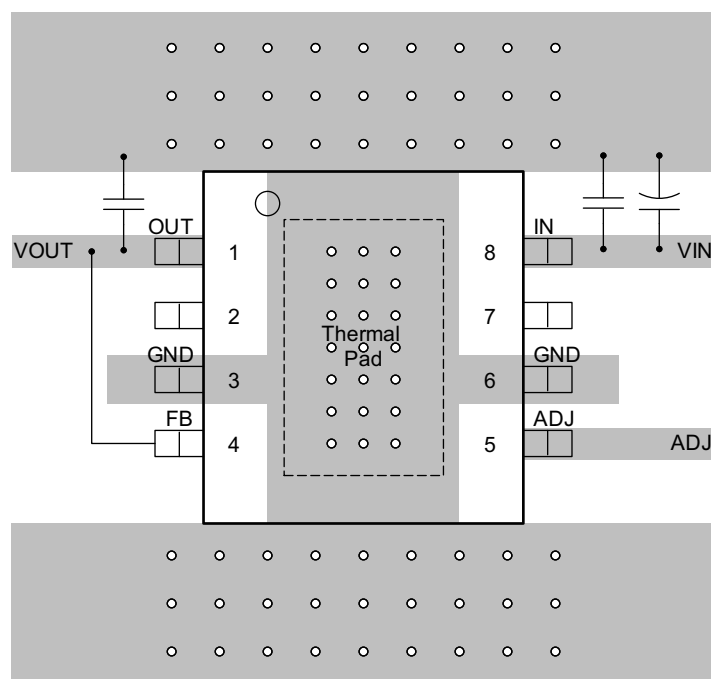


Figure 30. TPS7B4254-Q1 Layout Example

11 器件和文档支持

11.1 社区资源

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11.4 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

12 机械、封装和可订购信息

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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS7B4254QDDARQ1	ACTIVE	SO PowerPAD	DDA	8	2500	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	4254	Samples

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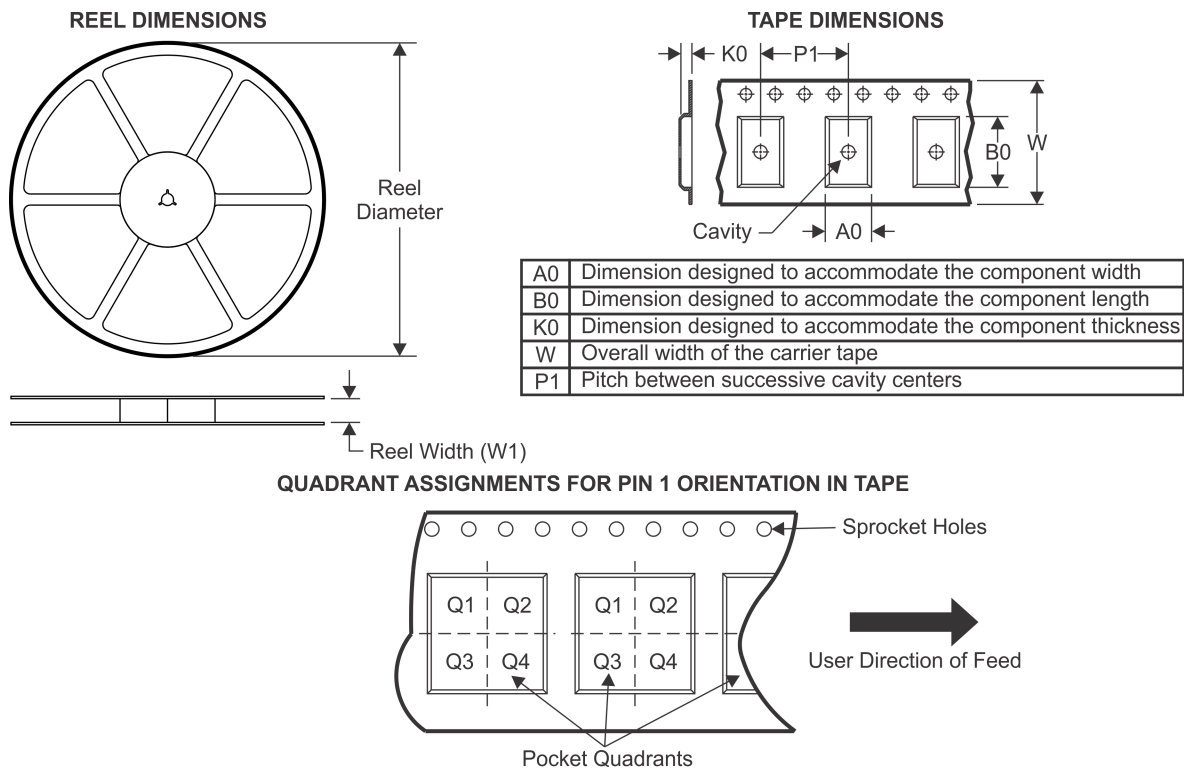
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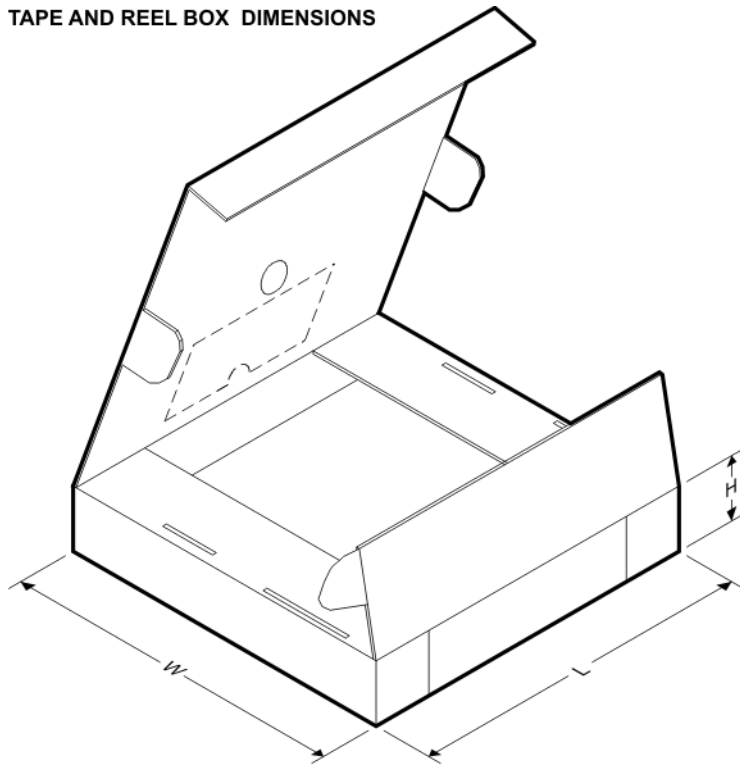
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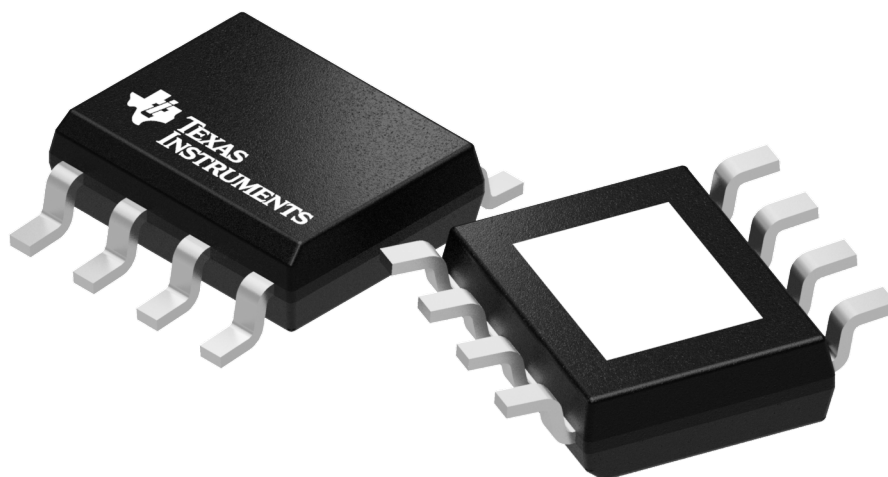
Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS7B4254QDDARQ1	SO Power PAD	DDA	8	2500	330.0	12.8	6.4	5.2	2.1	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS



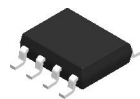
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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS7B4254QDDARQ1	SO PowerPAD	DDA	8	2500	366.0	364.0	50.0



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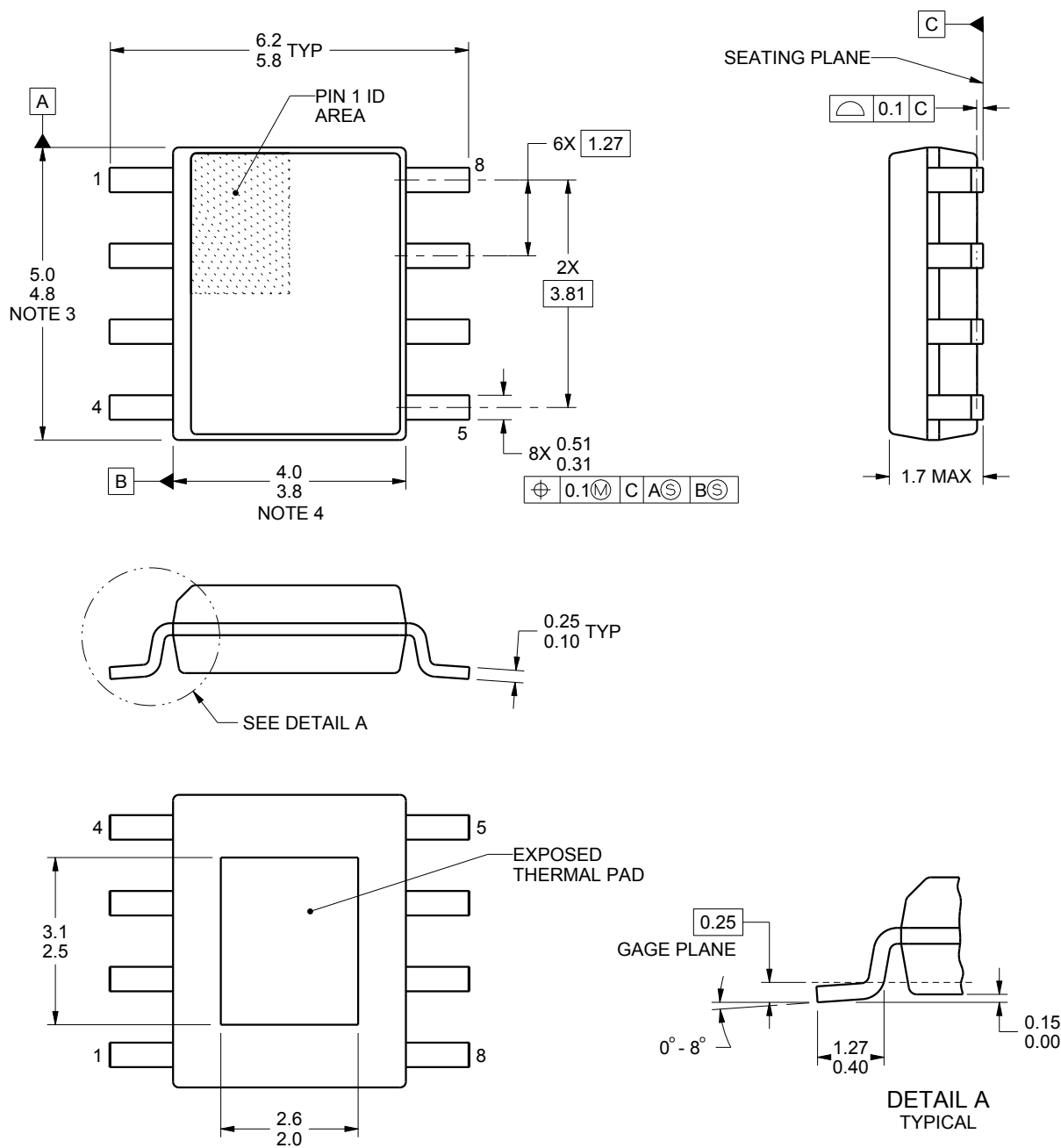
DDA0008J



PACKAGE OUTLINE

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PLASTIC SMALL OUTLINE



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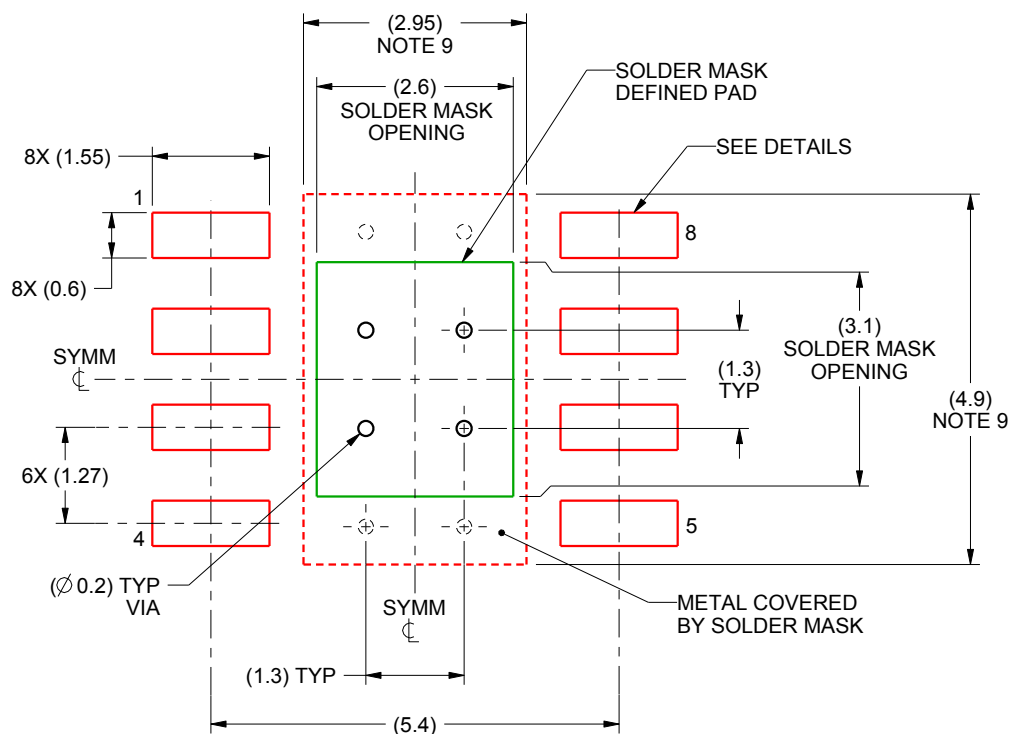
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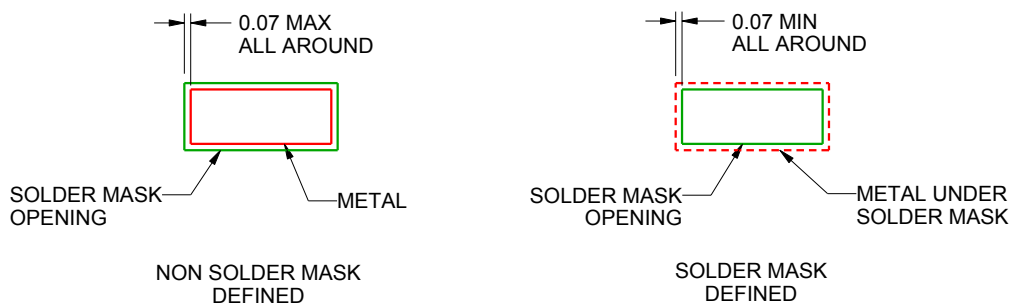
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PLASTIC SMALL OUTLINE



LAND PATTERN EXAMPLE
SCALE:10X



SOLDER MASK DETAILS

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NOTES: (continued)

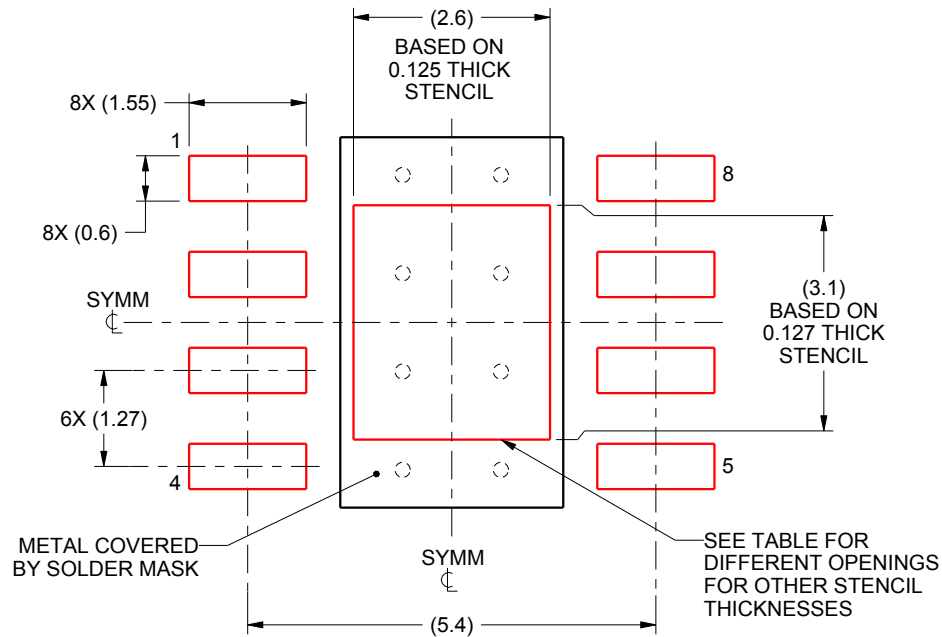
6. Publication IPC-7351 may have alternate designs.
7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
8. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature numbers SLMA002 (www.ti.com/lit/slma002) and SLMA004 (www.ti.com/lit/slma004).
9. Size of metal pad may vary due to creepage requirement.

EXAMPLE STENCIL DESIGN

DDA0008J

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PLASTIC SMALL OUTLINE



SOLDER PASTE EXAMPLE
EXPOSED PAD
100% PRINTED SOLDER COVERAGE BY AREA
SCALE:10X

STENCIL THICKNESS	SOLDER STENCIL OPENING
0.1	2.91 X 3.47
0.125	2.6 X 3.1 (SHOWN)
0.150	2.37 X 2.83
0.175	2.20 X 2.62

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NOTES: (continued)

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