3.3V 至 18V Thunderbolt 电源多路复用器

查询样品: TPS22981

特性

- 3.3V 电源供电
- 4.5V 至 19.8V 高电压开关
- 3V 至 3.6V 开关
- 可调节限流
- 热关断
- 先合后断开关
- 低电压闭合之前的高电压放电
- 反向电流阻断

应用范围

- 笔记本电脑
- 台式机
- 电源管理系统

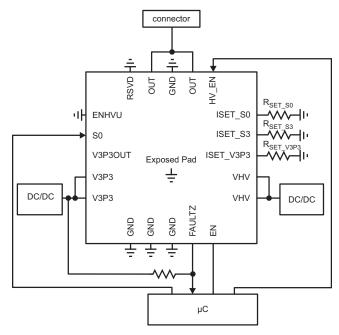


图 1. 典型应用

说明

TPS22981 是一款限流型电源多路复用器,此器件可从一个低电压电源(3V到3.6V)或者一个高电压电源(4.5V到19.8V)提供到一个外设的连接。由数字控制信号选择所需的输出。

高电压 (VHV) 和低电压 (V3P3) 开关电流限值由外部电阻设定。一旦达到电流限值,TPS22981 将会控制此开关以保持此电流限值。

当此高电压电源不在时,TPS22981 将保持到低电压 电源输出的连接。 高电压线路和高压使能信号出现 时,此高压开关将会与低电压开关一同打开,直到在低 压开关上检测到一个反向电流,从而实现最小下降电流 和击穿电流情况下的低压电源至高压电源的无缝转换。

为了避免 VHV 连接到 V3P3 连接切换期间的电流回流, TPS22981 将会中断 VHV 连接,将此输出放电至3.3V,然后连接至 V3P3。 当负载出现时,在返回3.3V 之前,此输出将转变到 0V。

TPS22981 采用 4mm x 4mm x 1mm 四方扁平无引线 (QFN) 封装。



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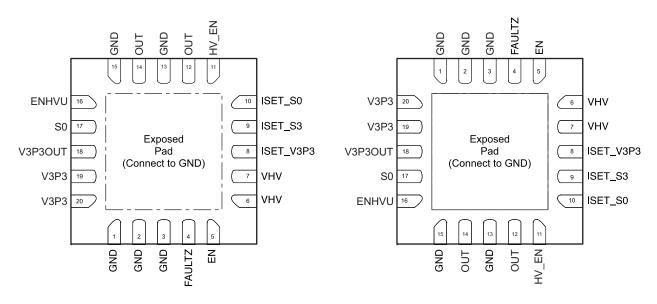
These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ORDERING INFORMATION

PART NUMBER	PACKAGE MARKING	PACKAGE	DEVICE SPECIFIC FEATURES
TPS22981RGPR	PS22981	RGP	Tape and Reel

Top View/Footprint

Bottom View



Package Size: 4mm x 4mm x 1mm height Pad Pitch: 0.5mm

DISSIPATION RATINGS

PACKAGE	THERMAL RESISTANCE ⁽¹⁾ θ _{JA}	POWER RATING ⁽¹⁾ T _A = 25°C	POWER RATING ⁽¹⁾ T _A = 70°C	DERATING FACTOR ABOVE $^{(2)}$ $T_A = 25^{\circ}C$
RGP	39.3°C/W	2.16W	1.02W	25.4mW/°C

⁽¹⁾ Simulated with high-K board

Maximum power dissipation is a function of T_{J(max)}, θ_{JA} and T_A. The maximum allowable power dissipation at any allowable ambient temperature is P_D = (T_{J(max)} - T_A) / θ_{JA}.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted)(1)

		VALUE	UNIT
	Input voltage range on V3P3 (VDD) ⁽²⁾	-0.3 to 3.6	
	Input voltage range on EN, HV_EN, ENHVU, ISET_V3P3, ISET_S0, ISET_S3, S0 ⁽²⁾	-0.3 to V3P3+0.3	
	Output voltage range on FAULTZ	-0.3 to V3P3+0.3	
V_{I}	Input voltage range on VHV ⁽²⁾	-0.3 to 20	V
	Output voltage range at OUT ⁽²⁾	-0.3 to 20	
	Voltage range between VHV and OUT (V _{VHV} –V _{OUT})	-7 to 20	
	Output voltage range at V3P3OUT ⁽²⁾	-0.3 to V3P3+0.3	
T _A	Operating ambient temperature range ⁽³⁾	-40 to 85	°C
T _{J (MAX)}	Maximum operating junction temperature	110	°C
T _{stg}	Storage temperature range	-65 to 150	°C
_	Charge Device Model (JESD 22 C101)	500	V
ESD Rating	Human Body Model (JESD 22 A114)	2	kV

⁽¹⁾ Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute—maximum—rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

·		·	-	MIN	MAX	UNIT
V _{3P3}	Cumply valtage res	Sunnly voltage range				
V_{HV}	Supply voltage rai	Supply voltage range				V
I _{LIM3P3OUT}	V3P3OUT Switch current range			0	500	mA
V_{IH}	Input logic high	EN, HV_EN, ENHVU, S0	V3P3	-0.6	V3P3	V
V _{IL}	Input logic low	EN, HV_EN, ENHVU, S0		0	0.6	V
R _{SET_V3P3}	3.3V switch curren	t limit set resistance	2	26.7	402	kΩ
R _{SET_S0}	VHV switch curren	t limit in S0 mode set resistance	2	26.7	402	kΩ
R _{SET_S3}	VHV switch curren	t limit in S3 mode set resistance	2	26.7	402	kΩ
R _{FAULTZ}	FAULTZ pull-up re	sistance to V3P3		30		kΩ

⁽²⁾ All voltage values are with respect to network ground terminal.

⁽³⁾ In applications where high power dissipation and/or poor package thermal resistance is present, the maximum ambient temperature may have to be derated. Maximum ambient temperature [TA(max)] is dependent on the maximum operating junction temperature [TJ(max)], the maximum power dissipation of the device in the application [PD(max)], and the junction-to-ambient thermal resistance of the part/package in the application (MJA), as given by the following equation: TA(max) = TJ(max) – (MJA × PD(max))

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ELECTRICAL CHARACTERISTICS

Unless otherwise noted the specification applies over the V_{DD} range and operating junction temp $-40^{\circ}\text{C} \le T_{J} \le 85^{\circ}\text{C}$. Typical values are for $V_{3P3} = 3.3\text{V}$, $V_{HV} = 15\text{V}$, and $T_{J} = 25^{\circ}\text{C}$.

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
POWER SU	JPPLIES AND CURRENTS						
V _{3P3}	V3P3 Input voltage range		3	3.3	3.6	V	
V _{HV}	VHV Input voltage range		4.5		19.8	V	
VHVACT	Active quiescent current from VHV	HV_EN = 1, EN = 1			150	μA	
VHVSD	Shutdown leakage current from VHV	HV_EN = 0, EN = 0 or 1			60	μA	
DDACT		EN = 1, HV_EN = 0			500	μA	
DDACTHV	Active quiescent current from V3P3	EN = 1, HV_EN = 1			500	μA	
DDSD	Shutdown quiescent current from V3P3	EN = 0, OUT = 0 V			30	μA	
DIS	OUT Discharge current	EN = 1, V _{HV} = 5V, HV_EN = 1→0	5		10	mA	
	511 511 51111111 00 001	V = 0 V			1		
IN	HV_EN, EN, ENHVU, S0, S3 Input pin leakage	V = V3P3			1	μA	
SWITCH AI	ND RESISTANCE CHARACTERISTICS		-		"		
R _{SHV}	VHV Switch resistance	V _{HV} = 5V to 18V, I _{VHV} = 0.9A			250	mΩ	
R _{S3P3}	V3P3 Switch resistance	V _{3P3} = 3.3 V, I _{V3P3} = 0.9 A			125	mΩ	
R _{S3P3BYP}	V3P3 Bypass switch resistance	V _{3P3} = 3.3 V, I _{V3P3} = 500 mA			500	mΩ	
V _{OLFAULTZ}	FAULTZ VOL	I _{FAULTZ} = 250 μA			0.6	V	
VOLTAGE	THRESHOLDS		-		"		
		VHV Input falling	3.6	4			
V_{HVUVLO}	VHV Under voltage lockout	VHV Input rising		4	4.3	V	
_		V3P3 Input falling	1.8	2.25			
V _{3P3UVLO}	V3P3 Under voltage lockout	V3P3 Input rising		2.25	2.5	V	
V _{FAULTZVAL}	V3P3 Voltage for valid FAULTZ	EN = 1	1.8			V	
THERMAL	SHUTDOWN						
T _{SD}	Shutdown temperature		110	120	130	°C	
T _{SDHYST}	Shutdown hysteresis			10		°C	
CURRENT	LIMIT						
		(4)	1				
		$R_{SFT, SO, 3} = 402 k\Omega^{(1)}$	80	100	120		
I _{LIMHV}	VHV Switch current limit state S0 or S3	$R_{SET_S0,3} = 402 \text{ k}\Omega^{(1)}$ $R_{SET_S0,3} = 80.6 \text{ k}\Omega^{(1)}$	80 446	100 496	120 546	mA	
I_{LIMHV}	VHV Switch current limit state S0 or S3	$R_{SET_S0,3} = 80.6 \text{ k}\Omega^{(1)}$	446	496	546	mA	
	VHV Switch current limit state S0 or S3 Maximum VHV switch current limit	$R_{SET_S0,3} = 80.6 \text{ k}\Omega^{(1)}$ $R_{SET_S0,3} = 26.7 \text{ k}\Omega^{(1)}$				mA A	
		$\begin{aligned} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \end{aligned}$	446 1423	496 1498	546 1573		
I _{LIMVHVMAX}		$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \end{split}$	446 1423 1.8	496 1498 2.4	546 1573 3.1		
LIMVHVMAX	Maximum VHV switch current limit	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \end{split}$	446 1423 1.8 80	496 1498 2.4 100 496	546 1573 3.1 120 546	A	
I _{LIMVHVMAX}	Maximum VHV switch current limit	$\begin{split} R_{SET_S0,3} &= 80.6 \; k\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \; k\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \; \Omega \\ R_{SET_V3P3} &= 402 \; k\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \; k\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \; k\Omega^{(1)} \end{split}$	446 1423 1.8 80 446	496 1498 2.4 100	546 1573 3.1 120	A	
LIMVHVMAX LIM3P3	Maximum VHV switch current limit V3P3 Switch current limit	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \end{split}$	446 1423 1.8 80 446 1423	496 1498 2.4 100 496 1498	546 1573 3.1 120 546 1573	A mA	
LIMVHVMAX LIM3P3 LIM3P3MAX REV3P3	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0.0 \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0.0 \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4	546 1573 3.1 120 546 1573 3.1	A mA	
LIMVHVMAX LIM3P3 LIM3P3MAX REV3P3 TV3P3RC	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4 40	546 1573 3.1 120 546 1573 3.1 85	A mA A mA μs	
LIMYHVMAX LIM3P3 LIM3P3MAX REV3P3 TV3P3RC TVHVSC	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time VHV Switch short circuit response time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \end{split}$ $V_{OUT} &= V_{3P3} \rightarrow V_{3P3} + 20 \text{ mV}$ $C_{OUT} &\leq 20 \text{ pF}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4	546 1573 3.1 120 546 1573 3.1 85	A mA A mA µs	
LIM3P3 LIM3P3MAX REV3P3 TV3P3RC TVHVSC TV3P3SC	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time VHV Switch short circuit response time V3P3 Switch short circuit response time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4 40	546 1573 3.1 120 546 1573 3.1 85	A mA A mA μs	
LIMVHVMAX LIM3P3 LIM3P3MAX REV3P3 TV3P3RC TVHVSC TV3P3SC TRANSITIC	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time VHV Switch short circuit response time V3P3 Switch short circuit response time V3P3 Switch short circuit response time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \\ V_{OUT} &= V_{3P3} {\longrightarrow} V_{3P3} + 20 \text{ mV} \\ C_{OUT} &\leq 20 \text{ pF} \\ C_{OUT} &\leq 20 \text{ pF} \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4 40	546 1573 3.1 120 546 1573 3.1 85 100	A mA µs µs µs	
LIM3P3 LIM3P3MAX LIM3P3MAX REV3P3 TV3P3RC TVHVSC TV3P3SC TRANSITIC T3P3OFF	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time VHV Switch short circuit response time V3P3 Switch short circuit response time VHV to V3P3 Off time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \\ V_{OUT} &= V_{3P3} \rightarrow V_{3P3} + 20 \text{ mV} \\ C_{OUT} &\leq 20 \text{ pF} \\ C_{OUT} &\leq 20 \text{ pF} \\ \\ C_{OUT} &= 1.1 \text{ µF, EN} = 1, \text{ HV_EN} = 1 \rightarrow 0 \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4 40	546 1573 3.1 120 546 1573 3.1 85 100	A mA A mA µs µs µs	
ILIMYHVMAX ILIM3P3 ILIM3P3MAX IREV3P3 TV3P3RC TVHVSC TV3P3SC TRANSITIC T3P3OFF T0-3.3V	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time VHV Switch short circuit response time V3P3 Switch short circuit response time V3P3 Switch short circuit response time V3P3 Switch short circuit response time ON DELAYS VHV to V3P3 Off time OV to 3.3V Ramp time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \ \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \\ V_{OUT} &= V_{3P3} \rightarrow V_{3P3} + 20 \text{ mV} \\ C_{OUT} &\leq 20 \text{ pF} \\ C_{OUT} &\leq 20 \text{ pF} \\ \\ C_{OUT} &\leq 20 \text{ pF} \\ \\ C_{OUT} &\leq 20 \text{ pF} \\ \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4 40	546 1573 3.1 120 546 1573 3.1 85 100	A mA A mA µs µs µs ms ms	
ILIMHV ILIMVHVMAX ILIM3P3 ILIM3P3MAX IREV3P3 TV3P3RC TVHVSC TV3P3SC TRANSITIC T3P3OFF T0-3.3V T3.3V-VHV TVHV-3.3V	Maximum VHV switch current limit V3P3 Switch current limit Maximum V3P3 switch current limit V3P3 Switch reverse current limit V3P3 Switch reverse current response time VHV Switch short circuit response time V3P3 Switch short circuit response time VHV to V3P3 Off time	$\begin{split} R_{SET_S0,3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_S0,3} &= 0 \Omega \\ R_{SET_V3P3} &= 402 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 80.6 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 26.7 \text{ k}\Omega^{(1)} \\ R_{SET_V3P3} &= 0\Omega \\ \\ V_{OUT} &= V_{3P3} \rightarrow V_{3P3} + 20 \text{ mV} \\ C_{OUT} &\leq 20 \text{ pF} \\ C_{OUT} &\leq 20 \text{ pF} \\ \\ C_{OUT} &= 1.1 \text{ µF, EN} = 1, \text{ HV_EN} = 1 \rightarrow 0 \end{split}$	446 1423 1.8 80 446 1423 1.8	496 1498 2.4 100 496 1498 2.4 40	546 1573 3.1 120 546 1573 3.1 85 100	A mA A mA µs µs µs	

⁽¹⁾ Equation 1 is used to calculate the required resistance for a given minimum I_{LIM}. The nearest 1% resistance is chosen and the corresponding I_{LIM} variance is shown.



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FUNCTIONAL BLOCK DIAGRAM

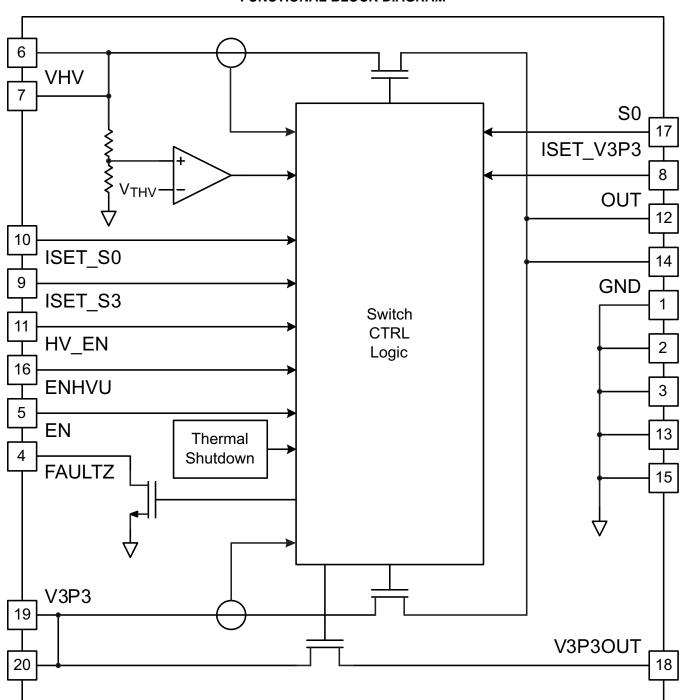


Figure 2. Functional Block Diagram



PIN FUNCTIONS

PIN		DESCRIPTION				
NO.	NAME	DESCRIFTION				
1, 2, 3, 13, 15	GND	Device ground. All GND pins must be connected to board ground.				
4 FAULTZ		Fault condition output. This pin is an open drain pull-down indicating a fault condition. Place a pull-up resistance (R _{FAULTZ}) between this pin and V3P3. Float pin or tie pin to GND if unused.				
5	EN	Device active-high enable.				
6, 7	VHV	High voltage power supply input. See the Input Inductive Bounce at Short Circuit section for more information.				
8 ISET_V3P3 Sets the current limit for V3P3. Place resistor between this pin and GND. See Equation 3 value.		Sets the current limit for V3P3. Place resistor between this pin and GND. See Equation 3 to calculate resistor value.				
9 ISET_S3		Sets the current limit for VHV in S3 mode. Place resistor between this pin and GND. See Equation 1 to calculate resistor value.				
10	ISET_S0	Sets the current limit for VHV in S0 mode. Place resistor between this pin and GND. See Equation 2 to calculate resistor value.				
11	HV_EN	Active-high voltage output enable.				
12, 14	OUT	Power output. Place a minimum of 1µF capacitor as close to this pin as possible.				
16	ENHVU	Enable VHV UVLO control of device enable. When asserted high, both V3P3 and VHV must be present for device enable. When low, only V3P3 must be present for device enable.				
17	S0	When this pin is asserted, the device is put in S0 mode. Otherwise the device operates in S3 mode.				
18	V3P3OUT	$3.3V$ bypass output. When ENHVU is low, this path is enabled by EN and the V3P3 UVLO. When ENHVU is high, this path is enabled by EN and both the V3P3 UVLO and the VHV UVLO. Place a minimum of $0.1\mu F$ capacitor as close to this pin as possible.				
19, 20	V3P3	3.3V power supply input. Place a minimum of 0.1µF capacitor as close to this pin as possible.				
EP	GND	Exposed pad must be connected to device GND.				



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APPLICATION INFORMATION

TYPICAL APPLICATION

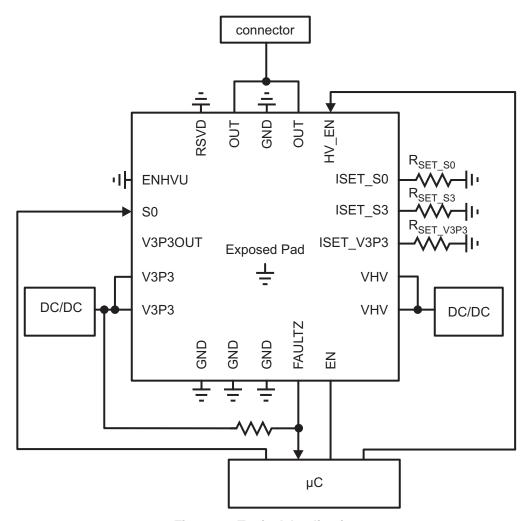


Figure 3. Typical Application

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CURRENT LIMIT

Figure 4 shows a simplified view of the TPS22981 current limit function. Both the high voltage supply current limit and the V3P3 supply current limit are adjustable by external resistors

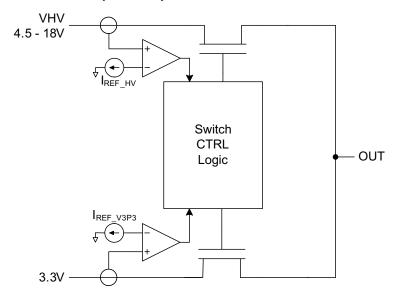


Figure 4. Simplified Current Limit Diagram

The current I_{REF_HV} and I_{REF_V3P3} that set the current limit threshold are set with three external resistors as shown in Figure 5. When the TPS22981 is passing the V3P3 voltage, the current limit is set by R_{SET_V3P3} . The VHV path has two modes that allow setting two different current limits. The S0 pin determines which current limit is used. When S0 is asserted high, R_{SET_S0} sets the current limit. When S0 is low, R_{SET_S3} sets the current limit. This allows the system to have two separate VHV current limits for different modes such as active and sleep.

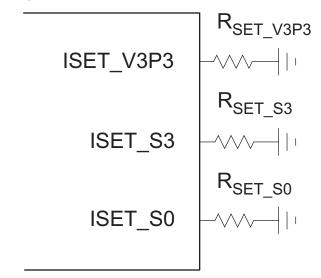
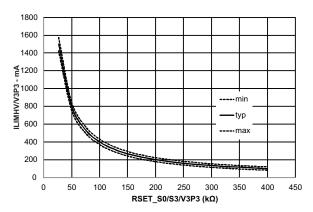


Figure 5. External R_{SET} Resistance to set Current Limits

CURRENT LIMIT THRESHOLD



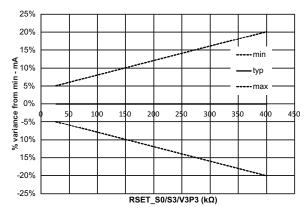


Figure 6. I_{LIM} vs R_{SET} for VHV and V3P3

Figure 7. % Variance from min I_{LIM} vs R_{SET}

Figure 6 shows the minimum, typical, and maximum current limit for either supply versus its corresponding R_{SET} value. Equation 1 is used to determine the RSET needed to set a typical I_{LIM} for a given supply and mode. Figure 7 shows the percent variation from the typical I_{LIM} value to the minimum and maximum I_{LIM} values.

$$R_{SET} = \frac{40 \text{ k}\Omega \times Amps}{I_{LIMTYP}}$$
(1)

Where

 R_{SET} = external resistor used to set the current limit for V3P3, VHV (S0), or VHV (S3), and I_{LIMTYP} = typical current limit for V3P3, VHV (S0), or VHV (S3) set by the external R_{SET} resistor.

Each resistor is placed between the corresponding ISET pin and GND, as shown in Figure 5, providing a minimum current limit between 100mA and 1.5A. For a given R_{SET} the minimum current limit and the maximum current limit are determined by Equation 2 and Equation 3.

$$I_{LIMMIN} = \frac{38429}{R_{SET}} - 0.0161 A$$

$$I_{LIMMAX} = \frac{41571}{R_{SET}} + 0.0161 A$$
(2)

MAXIMUM CURRENT LIMIT THRESHOLD

The TPS22981 has a maximum current limit I_{LIMVHVMAX} and I_{LIM3P3MAX}. This prevents excessive current in the case of an ISET pin being shorted to ground.



TRANSITION DELAYS

Output transitions of the TPS22981 voltages are shown in Figure 8. When the device transitions from V_{HV} to V_{3P3} at the output, the power switches both turn off until the output falls to near the V_{3P3} voltage. During this time, a discharge current of I_{DIS} pulls OUT down. If a load is also pulling current from OUT, the output will drop to near 0V due to the switch off time of T_{3P3OFF} .

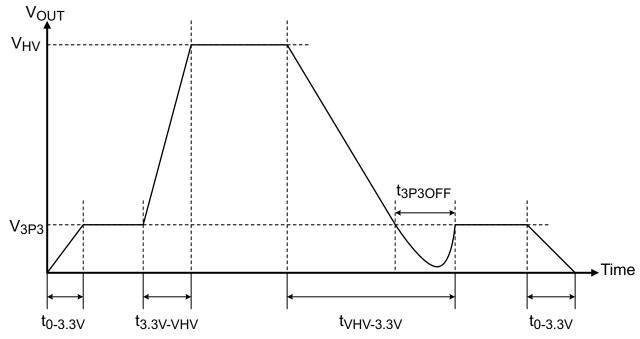


Figure 8. Output Voltage Transitions (Timing transitions are 10% to > 90%)

DIGITAL CONTROL SIGNALS

The voltage at OUT is controlled by two input digital logic signals, EN and HV_EN. HV_EN controls the state of the VHV switch and EN controls the state of V3P3 switch. Table 1 lists the possible output states given the conditions of the digital logic signals and the device is not in UVLO. See Table 2 for a more complete description including both UVLO conditions.

Table 1. Output state of OUT Given the States EN and HV_EN

EN	HV_EN	OUT
0	0	OPEN
0	1	OPEN
1	0	V3P3
1	1	VHV

Figure 9 shows possible combinations of EN and HV_EN controlling OUT of the TPS22981.

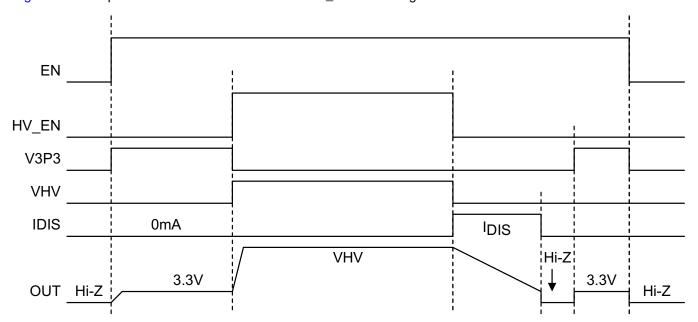


Figure 9. Logic Waveforms Displaying the Transition Between VHV and V3P3



OVER-CURRENT LIMIT AND SHORT CIRCUIT PROTECTION

When the load at OUT attempts to draw more current than the limit set by the external R_{SET} resistors for the V3P3 switch and VHV switch (for both S0 and S3 modes), the device will operate in a constant current mode while lowering the output voltage. Figure 10 shows the delay, t_{LIM} , which occurs from the instance an overcurrent fault is detected until the output current is lowered to I_{LIMHV} tolerances for VHV or I_{LIM3V3} tolerances for V3P3 shown in Figure 6. Figure 11 shows the response time versus a resistance shorted across the output.

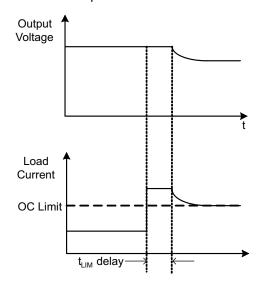


Figure 10. Overcurrent Output Response



Figure 11. Overcurrent Response Time vs Short Resistance

All short circuit conditions are treated as over-current conditions. In the event of a short circuit, the device will limit the output current to the corresponding R_{SET} value and continue to do so until thermal shutdown is encountered or the short circuit condition is removed.

REVERSE CURRENT PROTECTION

Reverse current protection for the V3P3 supply to OUT triggers at I_{REV3P3} causing the V3P3 supply switch to open. When the HV_EN signal is not asserted and reverse current protection is triggered, a discharge current source is turned on to bring the output voltage to near the V3P3 voltage.

REVERSE CURRENT BLOCKING

The VHV switch blocks reverse current flow from OUT to VHV when the switch is off.

THERMAL SHUTDOWN

The device enters thermal shutdown when junction temperature reaches T_{SD} . The device will resume previous state on power up once the junction temperature has dropped by 10C. Connect thermal vias to the exposed GND pad underneath the device package for improved thermal diffusion.

UVLO and ENABLE

When ENHVU is low, the TPS22981 is enabled by the logical AND of the EN input, the V3P3 UVLO, and the Thermal Shutdown. When the V3P3 UVLO threshold has been crossed, the device is not in thermal shutdown, and the EN input is high, the device will enable. When the V3P3 UVLO triggers, regardless of the states of any digital logic controls, the device will open all switches.

ENHVU adds the VHV UVLO to the logical decision enabling the device. When ENHVU is high, the TPS22981 is enabled by the logical AND of the EN input, the V3P3 UVLO, the VHV UVLO, and the Thermal Shutdown. When both UVLO thresholds have been crossed, the device is not in thermal shutdown, and the EN input is high, the device will enable. When either UVLO triggers, regardless of the states of any digital logic controls, the device will open all switches. Table 2 shows the pin and voltage configurations for enabling the device. Note, a 1 for the UVLO columns means the device is in a UVLO condition.

Table 2. Device Enable Control (when in an under-voltage condition, UVLO = 1)

		•		_	
EN	ENHVU	HV_EN	V3P3 UVLO	VHV UVLO	OUT
0	Х	X	X	X	OPEN
1	Х	X	1	X	OPEN
1	1	X	X	1	OPEN
1	0	0	0	X	V3P3
1	1	0	0	0	V3P3
1	Х	1	0	0	VHV
1	0	1	0	1	V3P3



FAULTZ Output

The TPS22981 has an open-drain FAULTZ output. When the device is in a fault condition, the FAULTZ output will pull low. Connect FAULTZ through a pull-up resistance to V3P3. A Fault occurs during any of the following conditions.

- EN = 1 and V3P3 is in UVLO (device enabled and V3P3 is in an under-voltage condition)
- EN = 1 and in Thermal Shutdown condition
- EN = 1, HV_EN = 1, and VHV is in UVLO (device enabled, high voltage enabled, and VHV is in an under-voltage condition)

Table 3 shows these conditions and the resulting FAULTZ output. Note, when V3P3 is below the UVLO threshold, FAULTZ will be 0 when EN=1 or 1 when EN=0. However, when V3P3 falls below $V_{FAULTZVAL}$, the FAULTZ output is unknown.

Table 3. FAULTZ Output Conditions (when in an under-voltage condition, UVLO = 1)

EN	HV_EN	Thermal Shutdown	V3P3 UVLO	VHV UVLO	FAULTZ (Active Low)
0	X	X	X	X	1
1	Х	X	1	X	0
1	Х	Yes	0	X	0
1	0	No	0	1	1
1	1	No	0	1	0
1	Х	No	0	0	1

It is recommended that the pull-up resistance on FAULTZ be $100k\Omega$ and must be greater than or equal to $30k\Omega$.

INPUT INDUCTIVE BOUNCE AT SHORT CIRCUIT

When a significant inductance is seen at the VHV input, suddenly turning off large current through the device may produce a large enough inductive voltage bounce on the VHV pin to exceed the maximum safe operating condition and damage the TPS22981. To prevent this, reduce any inductance at the VHV input.

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PACKAGING INFORMATION

Orderable part number	Status	Material type	Package Pins	Package qty Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
TPS22981RGPR	Active	Production	QFN (RGP) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPR.A	Active	Production	QFN (RGP) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPR.B	Active	Production	QFN (RGP) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPRG4	Active	Production	QFN (RGP) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPRG4.A	Active	Production	QFN (RGP) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPRG4.B	Active	Production	QFN (RGP) 20	3000 LARGE T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPT	Active	Production	QFN (RGP) 20	250 SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPT.A	Active	Production	QFN (RGP) 20	250 SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981
TPS22981RGPT.B	Active	Production	QFN (RGP) 20	250 SMALL T&R	Yes	NIPDAU	Level-2-260C-1 YEAR	-40 to 85	PS22981

⁽¹⁾ Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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⁽²⁾ Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ RoHS values: Yes, No. RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

⁽⁴⁾ Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.



PACKAGE OPTION ADDENDUM

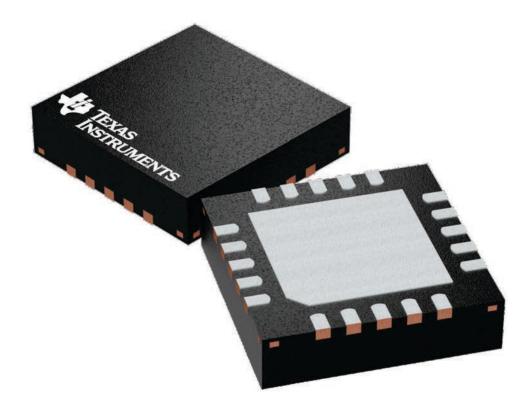
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and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

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4 x 4, 0.5 mm pitch

VERY THIN QUAD FLATPACK

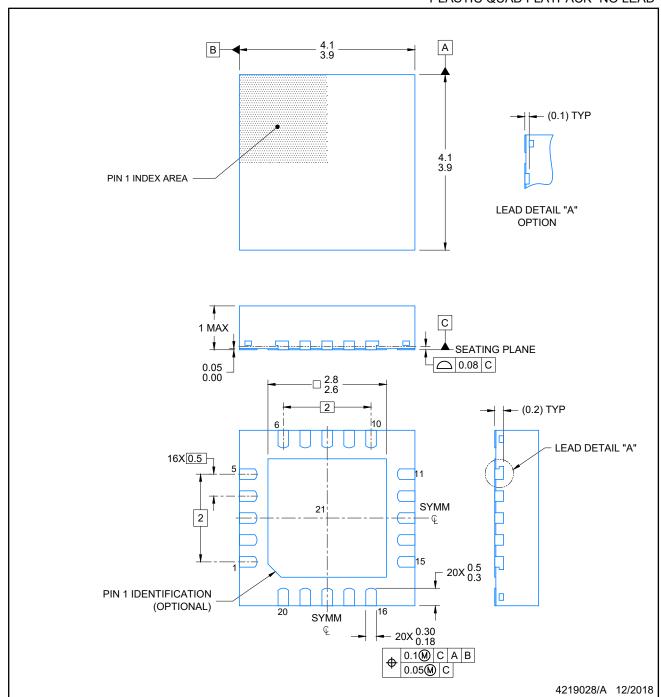


Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4224735/A



PLASTIC QUAD FLATPACK- NO LEAD

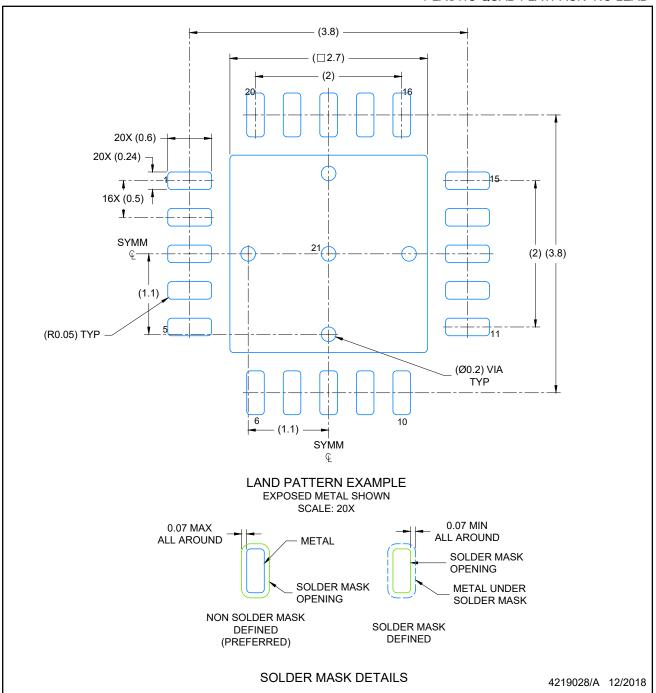


NOTES:

- All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for optimal thermal and mechanical performance.



PLASTIC QUAD FLATPACK- NO LEAD

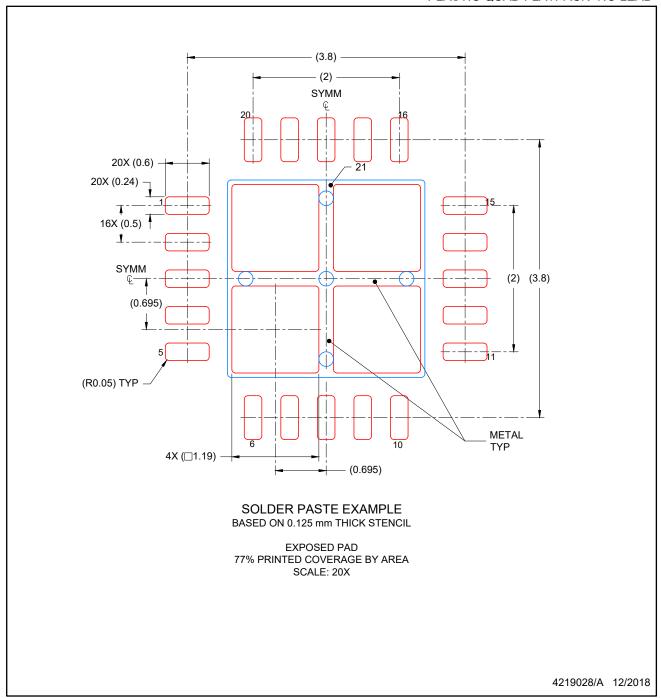


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK- NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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