









TPS2062-1, TPS2065-1, TPS2066-1

#### ZHCSRS4B -FEBRUARY 2007 - REVISED JUNE 2024

# 限流、配电开关

## 1 特性

- 输出放电功能
- 70mΩ 高侧 MOSFET
- 1A 持续电流
- 过热和短路保护
- 精确电流限制

(最小值 1.1A,最大值 1.9A)

- 工作电压范围: 2.7V 至 5.5V
- 0.6ms 典型上升时间
- 欠压锁定
- 抗尖峰脉冲故障报告 (OC)
- 上电期间无 OC 尖峰脉冲
- 1μA 最大待机电源电流
- 环境温度范围: -40°C 至 85°C
- ESD 保护

## 2 应用

- 高容性负载
- 短路保护

# 3 说明

TPS206x-1 配电开关适用于可能具有高容性负载和发 生短路的应用。此器件包含 70mΩ N 沟道 MOSFET 电 源开关,适用于需要在单个封装内包含多个电源开关的 配电系统。每个开关由一个逻辑使能输入控制。栅极驱 动由一个内部电荷泵提供,此电荷泵设计用于控制电源 开关上升时间和下降时间以大大减少切换期间的电流涌 入。电荷泵无需外部组件并可在低至 2.7V 的电源电压 下工作。

这些开关提供放电功能,可对存储在输出电容器中的输 出电压进行受控放电。

当输出负载超过限流阈值或者短路出现时,该器件通过 切换至恒定电流模式,并通过将过流 (OCx) 下拉至逻 辑输出低电平来将输出电流限制在安全水平上。当持续 重过载和短路增加了开关内的功率耗散时,将引起结温 上升,这时一个过热保护电路将关闭此开关以避免器件 损坏。一旦器件充分冷却,此器件将自动从热关断中恢 复。内部电路确保此开关在有效输入电压出现前保持关 闭状态。这个配电开关设计用于将电流限值的典型值设 定在 1.5A 上。

				GENERAL S	WITCH CATALO	G		
33 mΩ, single	TPS201xA TPS202x TPS203x	0.2 A – 2 A 0.2 A – 2 A 0.2 A – 2 A	80 mΩ, dual	TPS2042B 500 mA TPS2052B 500 mA TPS2046 250 mA TPS2056 250 mA TPS2062 1 A TPS2060 1 A TPS2060 1.5 A TPS2064 1.5 A	80 mΩ, dual	80 mΩ, triple	80 mΩ, quad	80 mΩ, quad
80 mΩ, single	TPS2014 TPS2015 TPS2041B TPS2051B TPS2045 TPS2055 TPS2061 TPS2065	600 mA 1 A 500 mA 500 mA 250 mA 250 mA 1 A	260 mΩ IN1 OUT 1.3 Ω	TPS2100/1 IN1 500 mA IN2 10 mA TPS2102/3/4/5 IN1 500 mA IN2 100 mA	TPS2082 500 mA TPS2090 250 mA TPS2091 250 mA TPS2092 250 mA	TPS2053B 500 mA TPS2047 250 mA TPS2057 250 mA	TPS2044B 500 mA TPS2054B 500 mA	TPS2085 500 mA TPS2086 500 mA TPS2087 500 mA TPS2095 250 mA TPS2096 250 mA TPS2097 250 mA

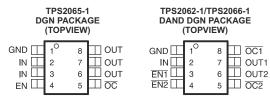


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# **4 Pin Configuration and Functions**



		PIN		I/O	DESCRIPTION
NAME	TPS2065-1	TPS2062-1	TPS2066-1	1/0	DESCRIPTION
EN	4	-	-	ı	Enable input, logic high turns on power switch
EN1	-	3	-	I	Enable input, logic low turns on channel 1
EN2	-	4	-	I	Enable input, logic high turns on channel 2
EN1	-	-	3	I	Enable input, logic high turns on channel 1
EN2	-	-	4	I	Enable input, logic high turns on channel 2
GND	1	1	1		Ground connection
IN	2, 3	2	2	I	Input voltage; connect a 0.1 $\mu$ F or greater ceramic capacitor from IN to GND as close to the IC as possible
<u>oc</u>	5	-	-	0	Active-low open-drain output, asserted during over-current
OC1	-	8	8	0	Active-low open-drain output, asserted during over-current for channel 1
OC2		5	5	0	Active-low open-drain output, asserted during over-current for channel 2
OUT	6, 7, 8	-	-	0	Power-switch output
OUT1	-	7	7	0	Power-switch output for channel 1
OUT2	-	6	6	0	Power-switch output for channel 2

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## **5 Specifications**

## 5.1 Absolute Maximum Ratings

over operating free-air temperature range unless otherwise noted(1)

		UNIT
Input voltage range, V <sub>I(IN)</sub> <sup>(2)</sup>	- 0.3 V to 6 V	
Output voltage range, V <sub>O(OUT)</sub> <sup>(2)</sup> , V <sub>O(OUTx)</sub>	-0.3 V to 6 V	
Input voltage range, $V_{I(\overline{EN})},V_{I(EN)},V_{I(\overline{ENx})},V_{I(ENx)}$	- 0.3 V to 6 V	
Voltage range, V <sub>I( OC)</sub> , V <sub>I( OCx)</sub>	- 0.3 V to 6 V	
Continuous output current, I <sub>O(OUT)</sub> , I <sub>O(OUTx)</sub>		Internally limited
Operating virtual junction temperature range,	Г <sub>Ј</sub>	-40°C to 125°C
Storage temperature range, T <sub>stg</sub>		- 65°C to 150°C
Electrostatic discharge (ESD) protection	Human body model MIL-STD-883C	2 kV
Liectiostatic discharge (ESD) protection	Charge device model (CDM)	500 V

<sup>(1)</sup> Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## **5.2 Recommended Operating Conditions**

	MIN	MAX	UNIT
Input voltage, V <sub>I(IN)</sub>	2.7	5.5	V
Input voltage, V <sub>I(EN)</sub> , V <sub>I(ENx)</sub> , V <sub>I(ENx)</sub> , V <sub>I(ENx)</sub>	0	5.5	V
Continuous output current, I <sub>O(OUT)</sub> , I <sub>O(OUTx)</sub>	0	1	А
Steady state current through discharge. Device disabled, measured through output pin(s)		8	mA
Operating virtual junction temperature, T <sub>J</sub>	-40	125	°C

#### 5.3 Thermal Information

	THERMAL METRIC <sup>(1)</sup>	D (SOIC) 8 PINS	DGN (HVSSOP) 8 PINS	UNIT
R <sub>0</sub> JA	Junction-to-ambient thermal resistance	119.3	53.6	
R <sub>0</sub> JC(top)	Junction-to-case (top) thermal resistance	67.6	58.7	
R <sub>0</sub> JB	Junction-to-board thermal resistance	59.6	35.5	90000
ψ ЈТ	Junction-to-top characterization parameter	20.3	2.7	°C/W°
ψ ЈВ	Junction-to-board characterization parameter	59.1	35.3	
R <sub>θ JC(bot)</sub>	Junction-to-case (bottom) thermal resistance	N/A	6.7	

<sup>(1)</sup> For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report.

## **5.4 Electrical Characteristics**

over recommended operating junction temperature range,  $V_{I(IN)} = 5.5 \text{ V}$ ,  $I_O = 1 \text{ A}$ ,  $V_{I(/ENx)} = 0 \text{ V}$ , or  $V_{I(ENx)} = 5.5 \text{ V}$  (unless otherwise noted)

	PARAMETER	TEST CONDITIONS <sup>(1)</sup>	MIN	TYP	MAX	UNIT				
POWER S	POWER SWITCH									
r <sub>DS(on)</sub>	Static drain-source on-state resistance, 5-V operation and 3.3-V operation	$V_{I(IN)}$ = 5 V or 3.3 V, $I_O$ = 1 A, $-40^{\circ}C \leqslant T_J \leqslant 125^{\circ}C$		70	135	mΩ				
	Static drain-source on-state resistance, 2.7-V operation <sup>(2)</sup>	$V_{I(IN)}$ = 2.7 V, $I_O$ = 1 A, -40°C $\leqslant$ $T_J \leqslant 125°C$		75	150	mΩ				

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<sup>(2)</sup> All voltages are with respect to GND.



# 5.4 Electrical Characteristics (续)

over recommended operating junction temperature range,  $V_{I(IN)} = 5.5 \text{ V}$ ,  $I_O = 1 \text{ A}$ ,  $V_{I(/ENx)} = 0 \text{ V}$ , or  $V_{I(ENx)} = 5.5 \text{ V}$  (unless otherwise noted)

	PARAMETER		TEST CONDITIONS	(1)	MIN	TYP	MAX	UNIT	
t <sub>r</sub> <sup>(2)</sup>	Diag time, output	V <sub>I(IN)</sub> = 5.5 V				0.6	1.5		
ŀr (−)	Rise time, output	V <sub>I(IN)</sub> = 2.7 V	C = 1 :: F B = F O	T. = 25°C		0.4	1	mo	
t <sub>f</sub> <sup>(2)</sup>	Fall time, output	V <sub>I(IN)</sub> = 5.5 V	$C_L = 1 \mu F, R_L = 5 \Omega,$	1j = 25 C	0.05		0.5	ms	
<b>L</b> f <sup>(−</sup> /	Fall time, output	V <sub>I(IN)</sub> = 2.7 V			0.05		0.5		
ENABLE I	INPUT EN OR EN						'		
V <sub>IH</sub>	High-level input voltage	2.7 V ≤ V <sub>I(IN)</sub> ≤5.	.5 V		2			V	
V <sub>IL</sub>	Low-level input voltage	$2.7 \text{ V} \leqslant \text{V}_{\text{I(IN)}} \leqslant 5$	5.5 V				0.8	V	
l <sub>l</sub>	Input current	$V_{I(ENx)} = 0 \text{ V or } 5.5$	5 V, V <sub>I(ENx)</sub> = 0 V or 5.5 V	1	-0.5		0.5	μА	
t <sub>on</sub> <sup>(2)</sup>	Turnon time	C <sub>L</sub> = 100 µ F, R <sub>L</sub> =	= 5 Ω				3		
t <sub>off</sub> <sup>(2)</sup>	Turnoff time	C <sub>L</sub> = 100 µ F, R <sub>L</sub> =	= 5 Ω				10	ms	
CURRENT	T LIMIT			l.					
	01 1 1 1 1 1	V <sub>I(INI)</sub> = 5 V, OUT o	V <sub>I(IN)</sub> = 5 V, OUT connected to GND,		1.1	1.5	1.9		
los	Short-circuit output current	device enabled int	to short-circuit	$-40$ °C $\leq$ T <sub>J</sub> $\leq$ 125°C	1.1	1.5	2.1	Α	
l <sub>OC</sub> (2) (4)	Overcurrent trip threshold	V <sub>I(IN)</sub> = 5 V, curren OUT	it ramp (≤ 100 A/s) on	TPS2062-1, TPS2065-1		2.4	3	Α	
SUPPLY C	CURRENT (TPS2065-1)								
		No load on OUT, \	No load on OUT, $V_{I(\overline{ENx})} = 5.5 \text{ V}$ ,			0.5	1		
Supply cur	rrent, low-level output	or $V_{I(ENx)} = 0 V$				0.5	10	μ <b>A</b>	
		No load on OUT, \	No load on OUT, $V_{I(\overline{ENx})} = 0 \text{ V}$ ,			43	60		
Supply cur	rrent, high-level output	or $V_{I(ENx)} = 5.5 \text{ V}$				43	70	μ A	
Reverse le	eakage current	V <sub>I(OUTx)</sub> = 5.5 V, IN	V <sub>I(OUTx)</sub> = 5.5 V, IN = ground <sup>(2)</sup>			0		μ <b>А</b>	
SUPPLY C	CURRENT (TPS2062-1)								
0 1		No load on OUT, \	$V_{1(FNx)} = 5.5 \text{ V},$	T <sub>J</sub> = 25°C		0.5	1		
Supply cur	rrent, low-level output	or $V_{I(ENx)} = 0 V$	(2.00)	$-40$ °C $\leq$ T <sub>J</sub> $\leq$ 125°C		0.5	20	μ <b>Α</b>	
0 1		No load on OUT, \	$V_{I(\overline{FNx})} = 0 \text{ V},$	T <sub>J</sub> = 25°C		50	70	— μА	
Supply cur	rrent, high-level output	or $V_{I(ENx)} = 5.5 \text{ V}$	( 2.00)	$-40$ °C $\leq$ T <sub>J</sub> $\leq$ 125°C		50	90		
Reverse le	eakage current	V <sub>I(OUTx)</sub> = 5.5 V, IN	N = ground <sup>(2)</sup>	T <sub>J</sub> = 25°C		0.2		μ <b>А</b>	
SUPPLY C	CURRENT (TPS2066-1)	1							
Cummbe com	reant love lovel output	No load on OUT, \	$J_{I(\overline{ENx})} = 5.5 \text{ V},$	T <sub>J</sub> = 25°C		0.5	1	4	
ouppiy cur	rrent, low-level output	or $V_{I(ENx)} = 0 V$	or $V_{I(ENX)} = 0 V$			0.5	20	μ A	
O l		No load on OUT, \	$V_{I(\overline{FNx})} = 0 \text{ V},$	T <sub>J</sub> = 25°C		95	120	20	
Supply cur	rrent, high-level output	or $V_{I(ENx)} = 5.5 \text{ V}$	or V <sub>I(ENx)</sub> = 5.5 V			95	120	μ A	
Reverse le	eakage current	V <sub>I(OUTx)</sub> = 5.5 V, IN	N = ground <sup>(2)</sup>	T <sub>.1</sub> = 25°C		0.2		μ <b>Α</b>	

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## 5.4 Electrical Characteristics (续)

over recommended operating junction temperature range,  $V_{I(IN)} = 5.5 \text{ V}$ ,  $I_O = 1 \text{ A}$ ,  $V_{I(/ENx)} = 0 \text{ V}$ , or  $V_{I(ENx)} = 5.5 \text{ V}$  (unless otherwise noted)

PARAMETER	TEST CONDITIONS <sup>(1)</sup>	MIN	TYP	MAX	UNIT
UNDERVOLTAGE LOCKOUT (TPS2062-1 and T	PS2065-1)				
Low-level input voltage, IN		2		2.5	V
Hysteresis, IN	T <sub>J</sub> = 25°C		75		mV
UNDERVOLTAGE LOCKOUT (TPS2066-1)	·	•			
Low-level input voltage, IN		2		2.6	V
Hysteresis, IN	T <sub>J</sub> = 25°C		75		mV
OVERCURRENT OC1 and OC2	·	•			
Output low voltage, V <sub>OL(OCx)</sub>	$I_{O(\overline{OCx})} = 5 \text{ mA}$			0.4	V
Off-state current <sup>(2)</sup>	V <sub>O( OCx)</sub> = 5 V or 3.3 V			1	μА
ŌC deglitch <sup>(2)</sup>	OCx assertion or deassertion	4	8	15	ms
Discharge resistance	V <sub>CC</sub> = 5 V, disabled, I <sub>O</sub> = 1 mA		100		Ω
THERMAL SHUTDOWN(3)				'	
Thermal shutdown threshold <sup>(2)</sup>		135			°C
Recovery from thermal shutdown <sup>(2)</sup>		125			°C
Hysteresis <sup>(2)</sup>			10		°C

<sup>(1)</sup> Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

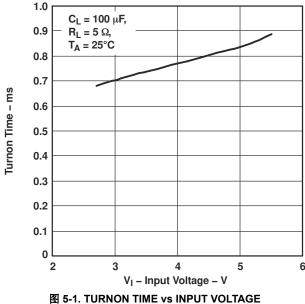
<sup>(2)</sup> Not tested in production, specified by design.

<sup>(3)</sup> The thermal shutdown only reacts under overcurrent conditions.

<sup>(4)</sup> TPS2066-1 does not have overcurrent trip threshold. Current limit is defined by I<sub>OS</sub>. See 节 7.7 for more details.



## **5.5 Typical Characteristics**



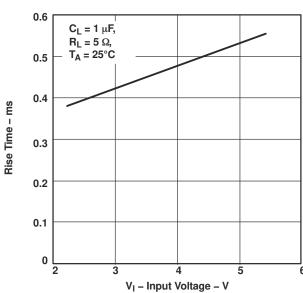


图 5-3. RISE TIME vs INPUT VOLTAGE

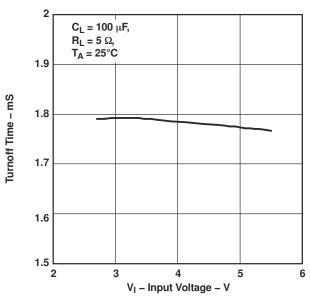


图 5-2. TURNOFF TIME vs INPUT VOLTAGE

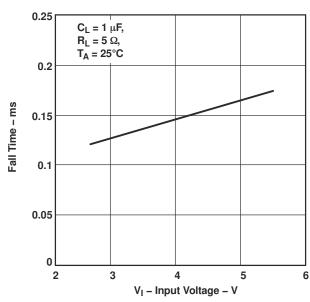


图 5-4. FALL TIME vs INPUT VOLTAGE



## 5.5 Typical Characteristics (continued)

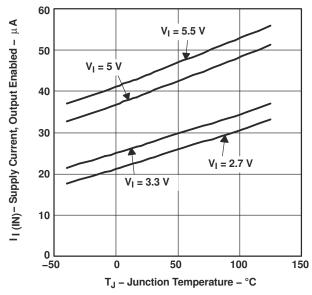


图 5-5. TPS2061, TPS2065-1 SUPPLY CURRENT, OUTPUT ENABLED vs JUNCTION TEMPERATURE

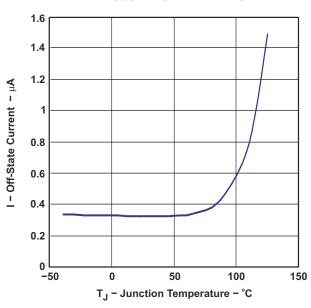


图 5-7. TPS2065-1 SUPPLY CURRENT, OUTPUT DISABLED vs JUNCTION TEMPERATURE

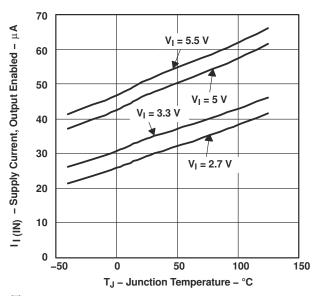


图 5-6. TPS2062-1, TPS2066-1 SUPPLY CURRENT, OUTPUT ENABLED vs JUNCTION TEMPERATURE

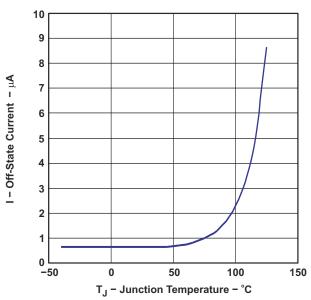


图 5-8. TPS2062-1, TPS2066-1 SUPPLY CURRENT, OUTPUT DISABLED vs JUNCTION TEMPERATURE

## 5.5 Typical Characteristics (continued)

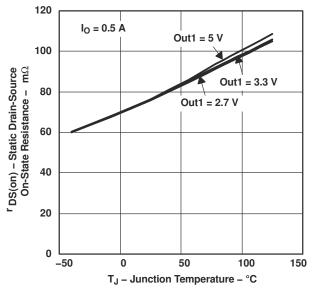


图 5-9. STATIC DRAIN-SOURCE ON-STATE RESISTANCE vs JUNCTION TEMPERATURE

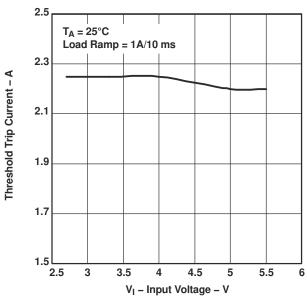


图 5-11. THRESHOLD TRIP CURRENT vs INPUT VOLTAGE

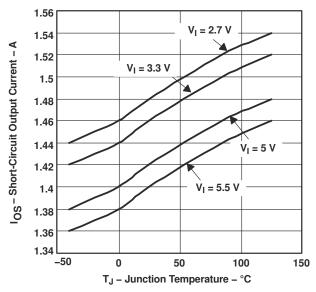


图 5-10. SHORT-CIRCUIT OUTPUT CURRENT vsJUNCTION TEMPERATURE

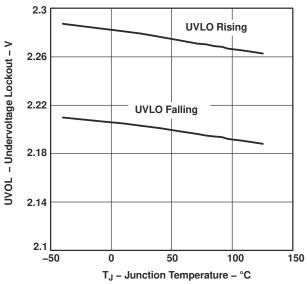


图 5-12. UNDERVOLTAGE LOCKOUT vs JUNCTION TEMPERATURE



# **5.5 Typical Characteristics (continued)**

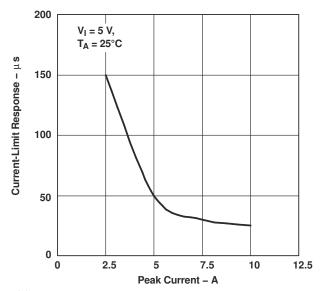
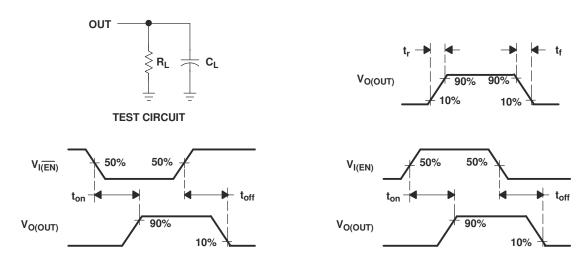


图 5-13. CURRENT-LIMIT RESPONSE vs PEAK CURRENT



## **6 Parameter Measurement Information**



**VOLTAGE WAVEFORMS** 

图 6-1. Test Circuit and Voltage Waveforms

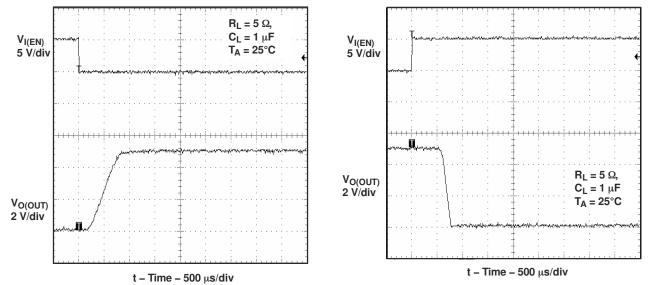
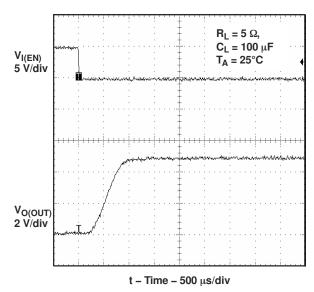


图 6-2. Turnon Delay and Rise Time With 1-µF Load

图 6-3. Turnoff Delay and Fall Time With 1- $\mu$ F Load

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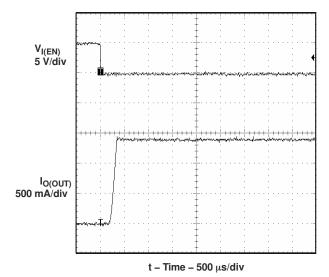




V<sub>I(EN)</sub> 5 V/div  $R_L = 5 \Omega$ V<sub>O(OUT)</sub> 2 V/div  $C_L = 100 \ \mu F$ T<sub>A</sub> = 25°C t – Time – 500  $\mu$ s/div

图 6-4. Turnon Delay and Rise Time With 100-µF Load

图 6-5. Turnoff Delay and Fall Time With 100-µF Load



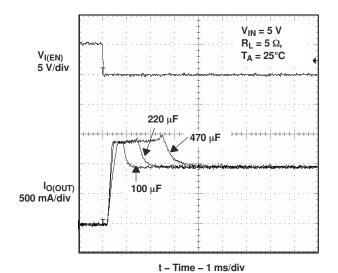


图 6-6. Short-Circuit Current, Device Enabled Into **Short** 

图 6-7. Inrush Current With Different Load Capacitance

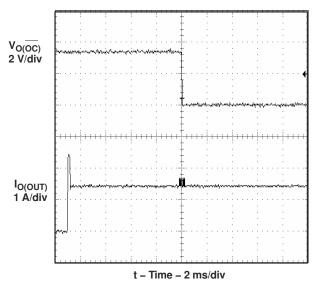


图 6-8.  $2-\Omega$  Load Connected to Enabled Device

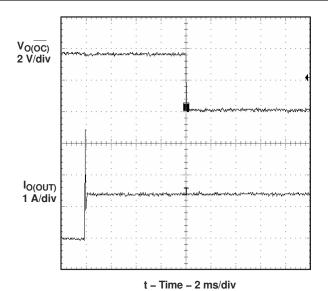


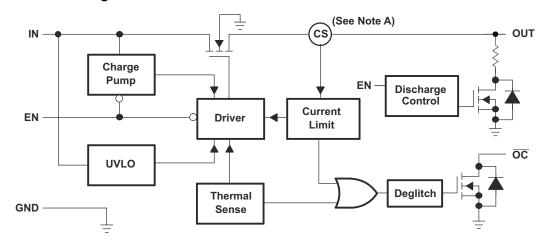
图 6-9. 1- $\Omega$  Load Connected to Enabled Device

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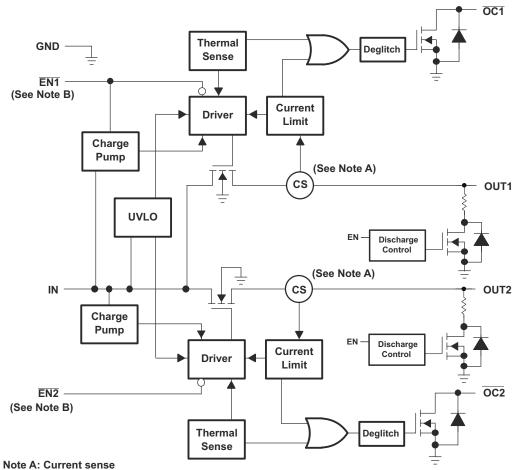
## 7 Detailed Description

## 7.1 Functional Block Diagram



Note A: Current sense

图 7-1. FUNCTIONAL BLOCK DIAGRAM (TPS2065-1)



Note B: Active low (ENx) for TPS2062. Active high (ENx) for TPS2066

图 7-2. (TPS2062-1 and TPS2066-1)

**FUNCTIONAL BLOCK DIAGRAM** 

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#### 7.2 Power Switch

The power switch is an N-channel MOSFET with a low on-state resistance. Configured as a high-side switch, the power switch prevents current flow from OUT to IN and IN to OUT when disabled. The power switch supplies a minimum current of 1 A.

## 7.3 Charge Pump

An internal charge pump supplies power to the driver circuit and provides the necessary voltage to pull the gate of the MOSFET above the source. The charge pump operates from input voltages as low as 2.7 V and requires little supply current.

#### 7.4 Driver

The driver controls the gate voltage of the power switch. To limit large current surges and reduce the associated electromagnetic interference (EMI) produced, the driver incorporates circuitry that controls the rise times and fall times of the output voltage.

## 7.5 Enable (ENx or ENx)

The logic enable disables the power switch and the bias for the charge pump, driver, and other circuitry to reduce the supply current. The supply current is reduced to less than 1  $\mu$  A when a logic high is present on  $\overline{\text{ENx}}$ , or when a logic low is present on ENx. A logic zero input on  $\overline{\text{ENx}}$ , or a logic high input on ENx restores bias to the drive and control circuits and turns the switch on. The enable input is compatible with both TTL and CMOS logic levels.

#### 7.6 Current Sense

A sense FET monitors the current supplied to the load. The sense FET measures current more efficiently than conventional resistance methods. When an overload or short circuit is encountered, the current-sense circuitry sends a control signal to the driver. The driver in turn reduces the gate voltage and drives the power FET into its saturation region, which switches the output into a constant-current mode and holds the current constant while varying the voltage on the load.

#### 7.7 Overcurrent

A sense FET is employed to check for overcurrent conditions. Unlike current-sense resistors, sense FETs do not increase the series resistance of the current path. When an overcurrent condition is detected, the device maintains a constant output current and reduces the output voltage accordingly. Complete shutdown occurs only if the fault is present long enough to activate thermal limiting.

There are two kinds of current limit profiles for the TPS206x-1 devices.

The TPS2062-1 and TPS2065-1 have an output I vs V characteristic similar to the plot labeled **Current Limit** with **Peaking** in  $\boxed{8}$  7-3. This type of limiting can be characterized by two parameters, the overcurrent trip threshold ( $I_{OC}$ ), and the short-circuit output current threshold ( $I_{OS}$ ).

The TPS2066-1 has an output I vs V characteristic similar to the plot labeled **Flat Current Limit** in  $\boxtimes$  7-3. This type of limiting can be characterized by one parameters, the short circuit current ( $I_{OS}$ ).

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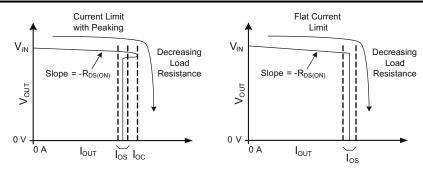


图 7-3. Current Limit Profiles

#### 7.7.1 Overcurrent Conditions (TPS2062-1 and TPS2065-1)

Three possible overload conditions can occur for the TPS2062-1 and TPS2065-1. In the first condition, the output has been shorted before the device is enabled or before  $V_{I(IN)}$  has been applied (see  $\boxtimes$  6-6 through  $\boxtimes$  6-9). The TPS2062-1 and TPS2065-1 senses the short and immediately switches into a constant-current output.

In the second condition, a short or an overload occurs while the device is enabled. At the instant the overload occurs, high currents may flow for a short period of time before the current-limit circuit can react. After the current-limit circuit has tripped (reached the overcurrent trip threshold  $(I_{OC})$ ), the device switches into constant-current mode and current is limited at the short-circuit output current threshold  $(I_{OS})$ .

In the third condition, the load has been gradually increased beyond the recommended operating current. The current is permitted to rise until the overcurrent trip threshold ( $I_{OC}$ ) is reached or until the thermal limit of the device is exceeded. The TPS2062-1 and TPS2065-1 are capable of delivering current up to the current-limit threshold without damaging the device. Once the overcurrent trip threshold ( $I_{OC}$ ) has been reached, the device switches into the constant-current mode current is limited at the short-circuit output current threshold ( $I_{OS}$ ).

#### 7.7.2 Overcurrent Conditions (TPS2066-1)

Three possible overload conditions can occur for the TPS2066-1. In the first condition, the output has been shorted before the device is enabled or before  $V_{I(IN)}$  has been applied. The TPS2066-1 senses the short and immediately switches into a constant-current output.

In the second condition, a short or an overload occurs while the device is enabled. At the instant the overload occurs, high currents may flow for a short period of time before the current-limit circuit can react. After the short-circuit output current threshold ( $I_{OS}$ ) is reached, the device switches into constant-current mode.

In the third condition, the load has been gradually increased beyond the recommended operating current. After the short-circuit output current threshold ( $I_{OS}$ ) is reached, the device switches into constant-current mode.

## 7.8 Overcurrent ( OCx)

The  $\overline{OCx}$  open-drain output is asserted (active low) when an overcurrent or overtemperature condition is encountered. The output remains asserted until the overcurrent or overtemperature condition is removed. A 10-ms deglitch circuit prevents the  $\overline{OCx}$  signal from oscillation or false triggering. If an overtemperature shutdown occurs, the  $\overline{OCx}$  is asserted instantaneously.

#### 7.9 Thermal Sense

The TPS206x-1 implements a thermal sensing to monitor the operating temperature of the power distribution switch. In an overcurrent or short-circuit condition the junction temperature rises. When the die temperature rises to approximately  $140^{\circ}$ C due to overcurrent conditions, the internal thermal sense circuitry turns off the switch, thus preventing the device from damage. Hysteresis is built into the thermal sense, and after the device has cooled approximately 10 degrees, the switch turns back on. The switch continues to cycle off and on until the fault is removed. The open-drain false reporting output ( $\overline{OCx}$ ) is asserted (active low) when an overtemperature shutdown or overcurrent occurs.



## 7.10 Undervoltage Lockout

A voltage sense circuit monitors the input voltage. When the input voltage is below approximately 2 V, a control signal turns off the power switch.

## 7.11 Discharge Function

When the device is disabled (when enable is deasserted or during power-up power-down when  $V_I < UVLO$ ) the discharge function is active. The discharge function offers a resistive discharge path for the external storage capacitor. The discharge function is suitable only to discharge filter capacitors for limited time and cannot dissipate steady state currents greater than 8 ma.

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## 8 Application and Implementation

#### 备注

以下应用部分中的信息不属于 TI 器件规格的范围, TI 不担保其准确性和完整性。TI 的客户应负责确定器件是否适用于其应用。客户应验证并测试其设计,以确保系统功能。

#### 8.1 Application Information

#### 8.1.1 Power-Supply Considerations

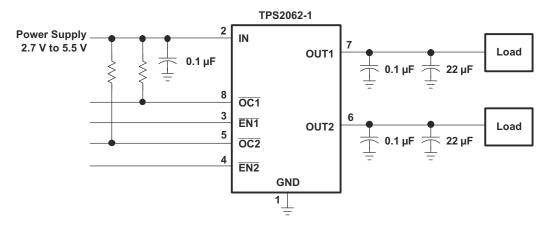


图 8-1. Typical Application

A 0.01-  $\mu$  F to 0.1-  $\mu$  F ceramic bypass capacitor between IN and GND, close to the device, is recommended. Placing a high-value electrolytic capacitor on the output pin(s) is recommended when the output load is heavy. This precaution reduces power-supply transients that may cause ringing on the input. Additionally, bypassing the output with a 0.01-  $\mu$  F to 0.1-  $\mu$  F ceramic capacitor improves the immunity of the device to short-circuit transients.

## 8.1.2 OC Response

The  $\overline{\text{OCx}}$  open-drain output is asserted (active low) when an overcurrent or overtemperature shutdown condition is encountered after a 10-ms deglitch timeout. The output remains asserted until the overcurrent or overtemperature condition is removed. Connecting a heavy capacitive load to an enabled device can cause a momentary overcurrent condition; however, no false reporting on  $\overline{\text{OCx}}$  occurs due to the 10-ms deglitch circuit. The TPS206x-1 is designed to eliminate false overcurrent reporting. The internal overcurrent deglitch eliminates the need for external components to remove unwanted pulses.  $\overline{\text{OCx}}$  is not deglitched when the switch is turned off due to an overtemperature shutdown.

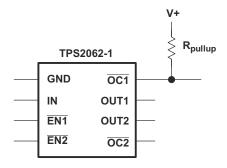


图 8-2. Typical Circuit for the OC Pin

#### 8.1.3 Power Dissipation and Junction Temperature

The low on-resistance on the N-channel MOSFET allows the small surface-mount packages to pass large currents. The thermal resistances of these packages are high compared to those of power packages; it is good design practice to check power dissipation and junction temperature. Begin by determining the r<sub>DS(on)</sub> of the Nchannel MOSFET relative to the input voltage and operating temperature. As an initial estimate, use the highest per switch can be calculated by:

$$P_D = r_{DS(on)} \times I^2$$

Multiply this number by the number of switches being used. This step renders the total power dissipation from the N-channel MOSFETs.

Finally, calculate the junction temperature:

$$T_J = P_D \times R_{\theta JA} + T_A$$

Where:

T<sub>A</sub>= Ambient temperature °C

 $R_{\theta JA}$  = Thermal resistance

P<sub>D</sub> = Total power dissipation based on number of switches being used.

Compare the calculated junction temperature with the initial estimate. If they do not agree within a few degrees, repeat the calculation, using the calculated value as the new estimate. Two or three iterations are generally sufficient to get a reasonable answer.

#### 8.1.4 Thermal Protection

Thermal protection prevents damage to the IC when heavy-overload or short-circuit faults are present for extended periods of time. The TPS206x-1 implements a thermal sensing to monitor the operating junction temperature of the power distribution switch. In an overcurrent or short-circuit condition, the junction temperature rises due to excessive power dissipation. Once the die temperature rises to approximately 140°C due to overcurrent conditions, the internal thermal sense circuitry turns the power switch off, thus preventing the power switch from damage. Hysteresis is built into the thermal sense circuit, and after the device has cooled approximately 10°C, the switch turns back on. The switch continues to cycle in this manner until the load fault or input power is removed. The OCx open-drain output is asserted (active low) when an overtemperature shutdown or overcurrent occurs.

#### 8.1.5 Undervoltage Lockout (UVLO)

An undervoltage lockout ensures that the power switch is in the off state at power up. Whenever the input voltage falls below approximately 2 V, the power switch is quickly turned off. The UVLO facilitates the design of hot-insertion systems where it is not possible to turn off the power switch before input power is removed. The UVLO also keeps the switch from being turned on until the power supply has reached at least 2 V, even if the switch is enabled. On reinsertion, the power switch is turned on, with a controlled rise time to reduce EMI and voltage overshoots.

#### 8.1.6 Universal Serial Bus (USB) Applications

The universal serial bus (USB) interface is a 12-Mb/s, or 1.5-Mb/s, multiplexed serial bus designed for low-tomedium bandwidth PC peripherals (for example, keyboards, printers, scanners, and mice). The four-wire USB interface is conceived for dynamic attach-detach (hot plug-unplug) of peripherals. Two lines are provided for differential data, and two lines are provided for 5-V power distribution.

USB data is a 3.3-V level signal, but power is distributed at 5 V to allow for voltage drops in cases where power is distributed through more than one hub across long cables. Each function must provide its own regulated 3.3 V from the 5-V input or its own internal power supply.

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The USB specification defines the following five classes of devices, each differentiated by power-consumption requirements:

- Hosts/self-powered hubs (SPH)
- Bus-powered hubs (BPH)
- Low-power, bus-powered functions
- · High-power, bus-powered functions
- Self-powered functions

SPHs and BPHs distribute data and power to downstream functions. The TPS206x-1 has higher current capability than required by one USB port; so, it can be used on the host side and supplies power to multiple downstream ports or functions.

#### 8.1.7 Host/Self-Powered and Bus-Powered Hubs

Hosts and SPHs have a local power supply that powers the embedded functions and the downstream ports (see 8-3). This power supply must provide from 5.25 V to 4.75 V to the board side of the downstream connection under full-load and no-load conditions. Hosts and SPHs are required to have current-limit protection and must report overcurrent conditions to the USB controller. Typical SPHs are desktop PCs, monitors, printers, and stand-alone hubs.

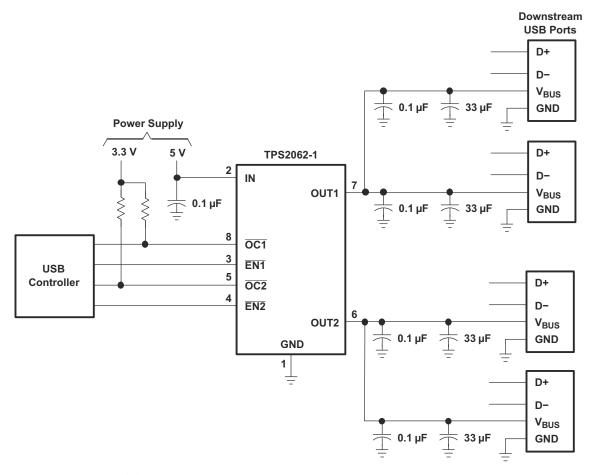


图 8-3. Typical Four-Port USB Host / Self-Powered Hub

BPHs obtain all power from upstream ports and often contain an embedded function. The hubs are required to power up with less than one unit load. The BPH usually has one embedded function, and power is always available to the controller of the hub. If the embedded function and hub require more than 100 mA on power up,

the power to the embedded function may need to be kept off until enumeration is completed. This can be accomplished by removing power or by shutting off the clock to the embedded function. Power switching the embedded function is not necessary if the aggregate power draw for the function and controller is less than one unit load. The total current drawn by the bus-powered device is the sum of the current to the controller, the embedded function, and the downstream ports, and it is limited to 500 mA from an upstream port.

#### 8.1.8 Low-Power Bus-Powered and High-Power Bus-Powered Functions

Both low-power and high-power bus-powered functions obtain all power from upstream ports; low-power functions always draw less than 100 mA; high-power functions must draw less than 100 mA at power up and can draw up to 500 mA after enumeration. If the load of the function is more than the parallel combination of 44  $\Omega$  and 10  $\mu$ F at power up, the device must implement inrush current limiting (see 8-4). With TPS206x-1, the internal functions draw more than 500 mA, which fits the needs of some applications such as motor driving circuits.

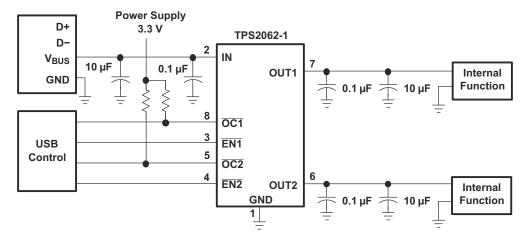


图 8-4. High-Power Bus-Powered Function

#### 8.1.9 USB Power-Distribution Requirements

USB can be implemented in several ways, and, regardless of the type of USB device being developed, several power-distribution features must be implemented.

- Hosts/SPHs must:
  - Current-limit downstream ports
  - Report overcurrent conditions on USB V<sub>BUS</sub>
- · BPHs must:
  - Enable/disable power to downstream ports
  - Power up at <100 mA</li>
  - Limit inrush current (<44  $\Omega$  and 10  $\mu$  F)
- · Functions must:
  - Limit inrush currents
  - Power up at <100 mA</li>

The feature set of the TPS206x-1 allows them to meet each of these requirements. The integrated current-limiting and overcurrent reporting is required by hosts and self-powered hubs. The logic-level enable and controlled rise times meet the need of both input and output ports on bus-powered hubs, as well as the input ports for bus-powered functions (see 8-5).



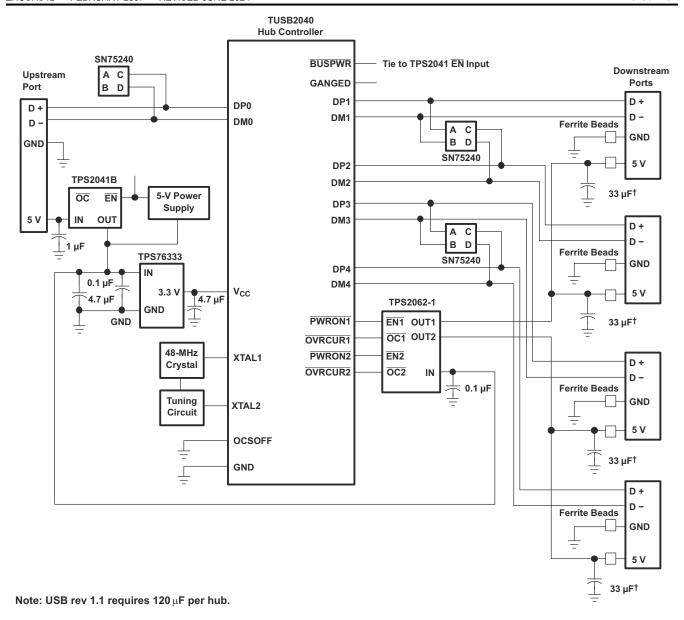


图 8-5. Hybrid Self / Bus-Powered Hub Implementation

#### 8.1.10 Generic Hot-Plug Applications

In many applications it may be necessary to remove modules or pc boards while the main unit is still operating. These are considered hot-plug applications. Such implementations require the control of current surges seen by the main power supply and the card being inserted. The most effective way to control these surges is to limit and slowly ramp the current and voltage being applied to the card, similar to the way in which a power supply normally turns on. Due to the controlled rise times and fall times of the TPS206x-1, these devices can be used to provide a softer start-up to devices being hot-plugged into a powered system. The UVLO feature of the TPS206x-1 also ensures that the switch is off after the card has been removed, and that the switch is off during the next insertion. The UVLO feature insures a soft start with a controlled rise time for every insertion of the card or module.



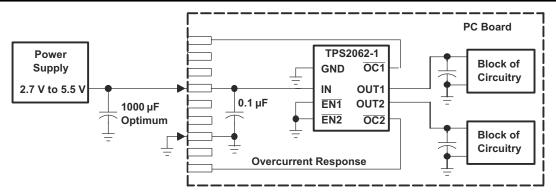


图 8-6. Typical Hot-Plug Implementation

By placing the TPS206x-1 between the  $V_{\text{CC}}$  input and the rest of the circuitry, the input power reaches these devices first after insertion. The typical rise time of the switch is approximately 1 ms, providing a slow voltage ramp at the output of the device. This implementation controls system surge currents and provides a hotplugging mechanism for any device.

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## 9 Device and Documentation Support

TI offers an extensive line of development tools. Tools and software to evaluate the performance of the device, generate code, and develop solutions are listed below.

#### 9.1 Device Support

## 9.2 Documentation Support

## 9.3 接收文档更新通知

要接收文档更新通知,请导航至 ti.com 上的器件产品文件夹。点击*通知* 进行注册,即可每周接收产品信息更改摘要。有关更改的详细信息,请查看任何已修订文档中包含的修订历史记录。

## 9.4 支持资源

TI E2E™中文支持论坛是工程师的重要参考资料,可直接从专家处获得快速、经过验证的解答和设计帮助。搜索现有解答或提出自己的问题,获得所需的快速设计帮助。

链接的内容由各个贡献者"按原样"提供。这些内容并不构成 TI 技术规范,并且不一定反映 TI 的观点;请参阅 TI 的使用条款。

#### 9.5 Trademarks

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ESD 的损坏小至导致微小的性能降级,大至整个器件故障。精密的集成电路可能更容易受到损坏,这是因为非常细微的参数更改都可能会导致器件与其发布的规格不相符。

#### 9.7 术语表

TI术语表本术语表列出并解释了术语、首字母缩略词和定义。

## 10 Revision History

注:以前版本的页码可能与当前版本的页码不同

Cha	anges from Revision A (March 2009) to Revision B (June 2024)	Page
• }	通篇更新了表格、图和交叉参考的编号格式	1
• }	删除了"功耗额定值"表	1
•	Thermal Information of new needle respin, TPS2066DGNR-1	4
• /	Added Thermal Information table	4
• (	Updated max UVLO value for TPS2066-1	4
• (	Updated max Supply current, high-level output for TPS2066-1	4
• (	Updated Overcurrent trip threshold to apply only to TPS2062-1 and TPS2065-1	4
• (	Updated section information	15
• /	Added 节 7.7.1	16
• /	Added 节 7.7.2	16

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# 11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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#### PACKAGING INFORMATION

Orderable part number	Status	Material type	Package   Pins	Package qty   Carrier	RoHS	Lead finish/	MSL rating/	Op temp (°C)	Part marking
	(1)	(2)			(3)	Ball material	Peak reflow		(6)
						(4)	(5)		
TPS2062D-1	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2062-1
TPS2062D-1.A	Active	Production	SOIC (D)   8	75   TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2062-1
TPS2062DR-1	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2062-1
TPS2062DR-1.A	Active	Production	SOIC (D)   8	2500   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2062-1
TPS2065DGN-1.A	Active	Production	HVSSOP (DGN)   8	80   TUBE	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2065
									-1
TPS2065DGNR-1	Active	Production	HVSSOP (DGN)   8	2500   LARGE T&R	Yes	NIPDAU   NIPDAUAG	Level-1-260C-UNLIM	-40 to 125	2065
									-1
TPS2065DGNR-1.A	Active	Production	HVSSOP (DGN)   8	2500   LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2065
									-1
TPS2066DGNR-1	Active	Production	HVSSOP (DGN)   8	2500   LARGE T&R	Yes	FULL NIPDAU	Level-1-260C-UNLIM	-40 to 125	2066
			, , ,	·		NIPDAU			-1
TPS2066DGNR-1.A	Active	Production	HVSSOP (DGN)   8	2500   LARGE T&R	Yes	FULL NIPDAU	Level-1-260C-UNLIM	-40 to 125	2066
									-1

<sup>(1)</sup> Status: For more details on status, see our product life cycle.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

<sup>(2)</sup> Material type: When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

<sup>(3)</sup> RoHS values: Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

<sup>(4)</sup> Lead finish/Ball material: Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

<sup>(5)</sup> MSL rating/Peak reflow: The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

<sup>(6)</sup> Part marking: There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

## **PACKAGE OPTION ADDENDUM**

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Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

#### OTHER QUALIFIED VERSIONS OF TPS2062-1, TPS2065-1, TPS2066-1:

• Automotive: TPS2062-Q1, TPS2065-Q1, TPS2066-Q1

NOTE: Qualified Version Definitions:

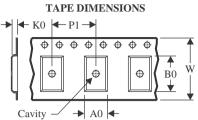
Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

# **PACKAGE MATERIALS INFORMATION**

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## TAPE AND REEL INFORMATION





	-
A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

#### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

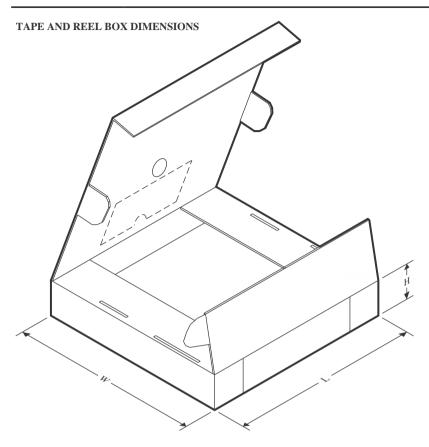


#### \*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS2062DR-1	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
TPS2065DGNR-1	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
TPS2065DGNR-1	HVSSOP	DGN	8	2500	330.0	12.4	5.3	3.3	1.3	8.0	12.0	Q1
TPS2066DGNR-1	HVSSOP	DGN	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1



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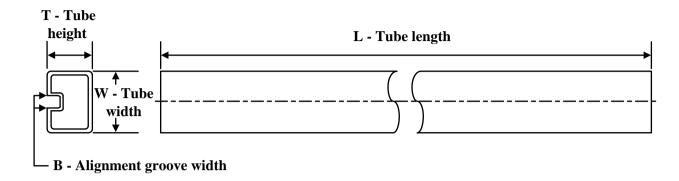
#### \*All dimensions are nominal

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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS2062DR-1	SOIC	D	8	2500	340.5	338.1	20.6
TPS2065DGNR-1	HVSSOP	DGN	8	2500	364.0	364.0	27.0
TPS2065DGNR-1	HVSSOP	DGN	8	2500	346.0	346.0	35.0
TPS2066DGNR-1	HVSSOP	DGN	8	2500	353.0	353.0	32.0

# **PACKAGE MATERIALS INFORMATION**

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## **TUBE**



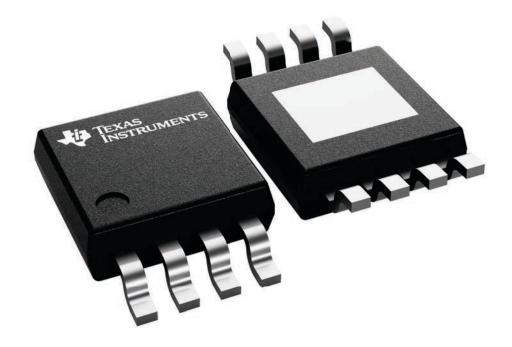
#### \*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (µm)	B (mm)
TPS2062D-1	D	SOIC	8	75	507	8	3940	4.32
TPS2062D-1.A	D	SOIC	8	75	507	8	3940	4.32
TPS2065DGN-1.A	DGN	HVSSOP	8	80	330	6.55	500	2.88
TPS2065DGN-1.A	DGN	HVSSOP	8	80	322	6.55	1000	3.01

3 x 3, 0.65 mm pitch

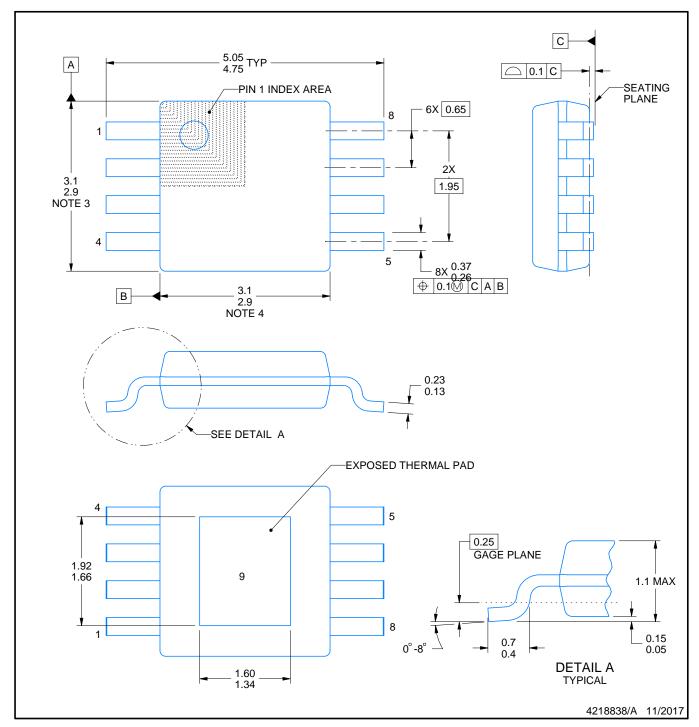
SMALL OUTLINE PACKAGE

This image is a representation of the package family, actual package may vary. Refer to the product data sheet for package details.



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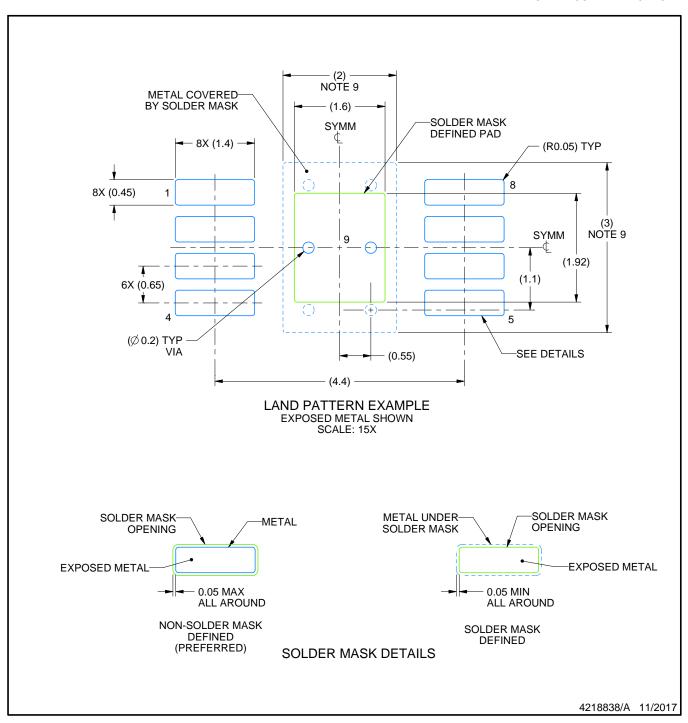
#### NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-187.

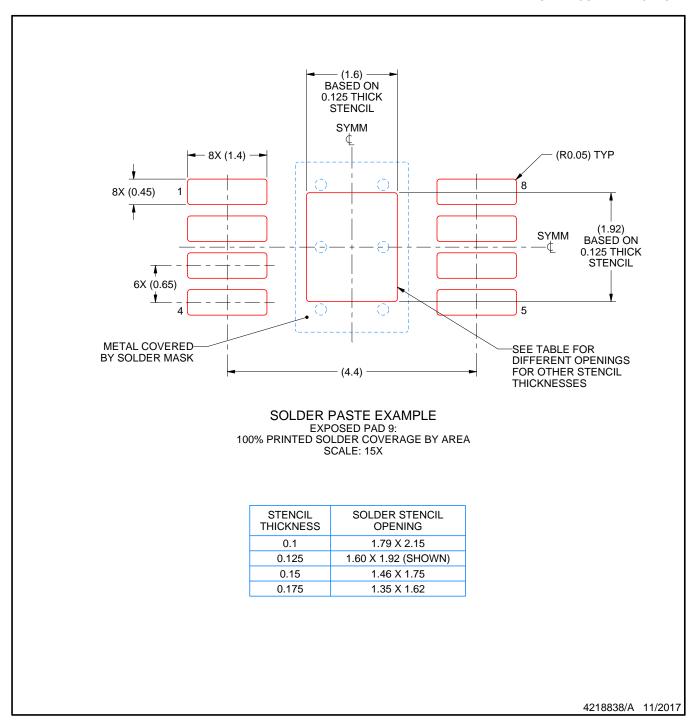




NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
- 9. Size of metal pad may vary due to creepage requirement.





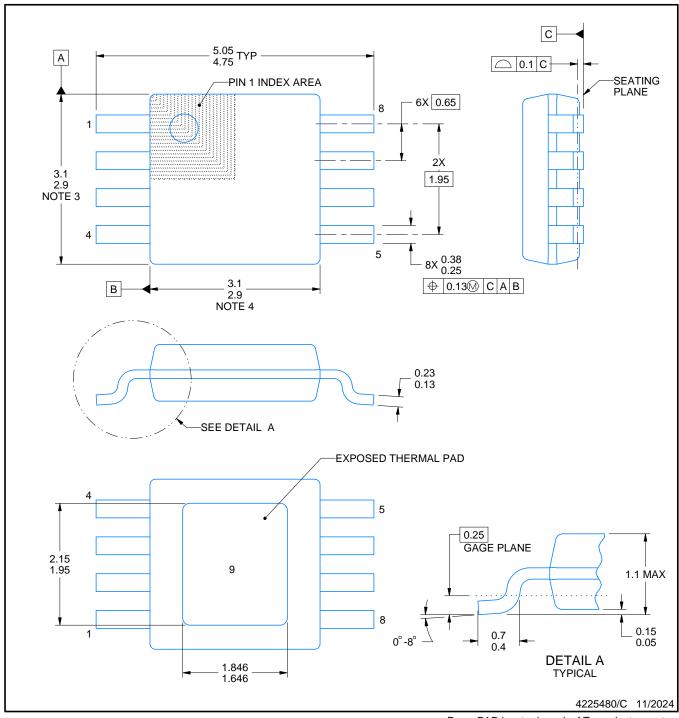
NOTES: (continued)

- 10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 11. Board assembly site may have different recommendations for stencil design.



# PowerPAD<sup>™</sup> HVSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



#### NOTES:

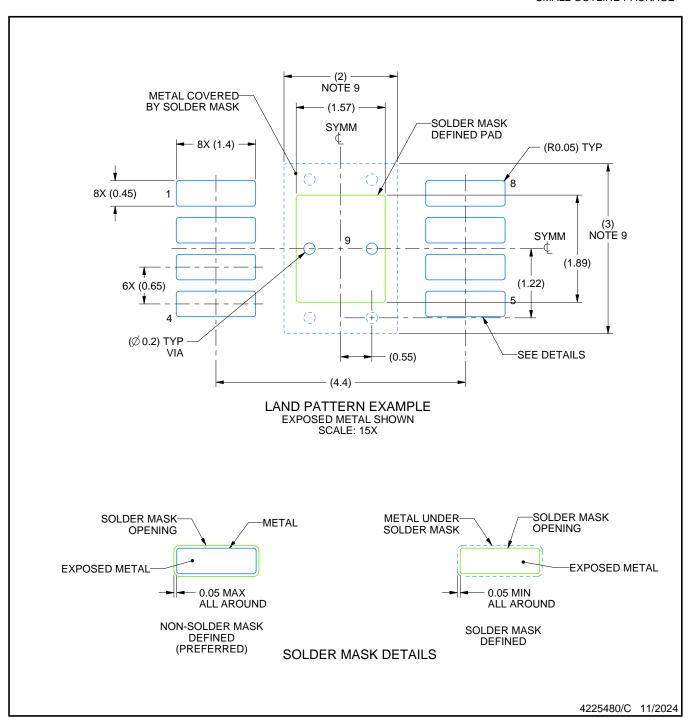
PowerPAD is a trademark of Texas Instruments.

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-187.

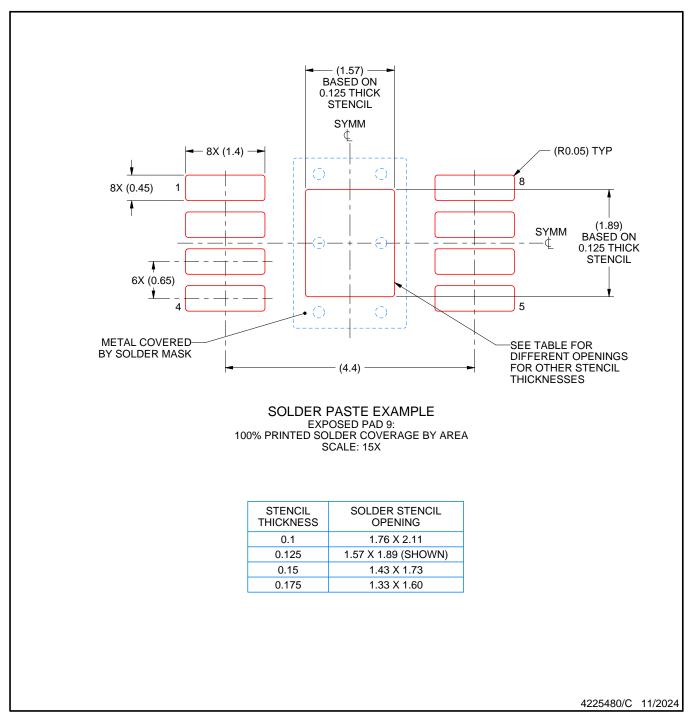




NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 8. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
- 9. Size of metal pad may vary due to creepage requirement.





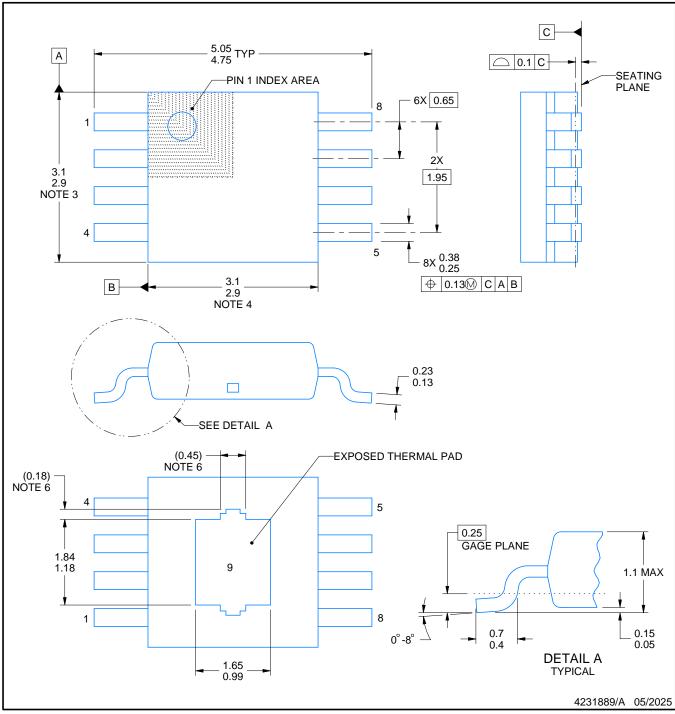
NOTES: (continued)

- 10. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 11. Board assembly site may have different recommendations for stencil design.



# PowerPAD<sup>™</sup> VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



#### NOTES:

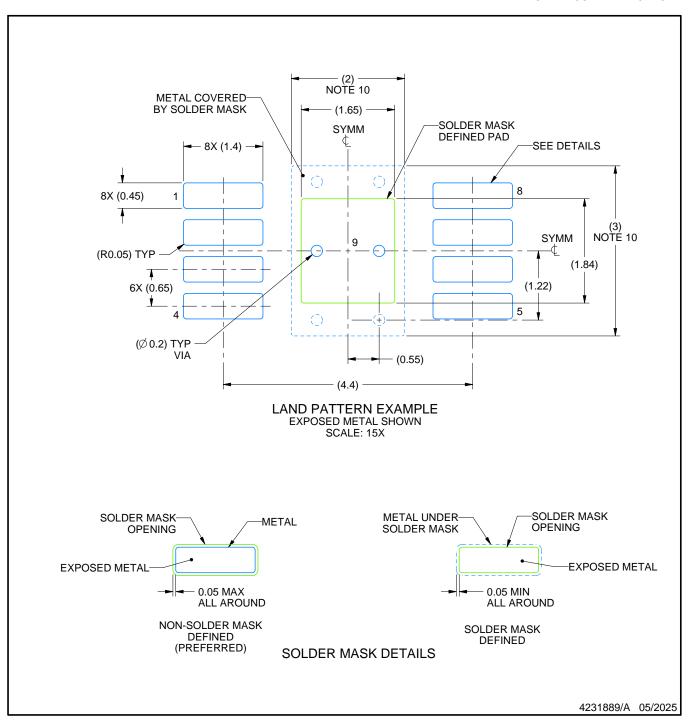
PowerPAD is a trademark of Texas Instruments.

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

  2. This drawing is subject to change without notice.

  3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not
- exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-187.
- 6. Features may differ or may not be present.

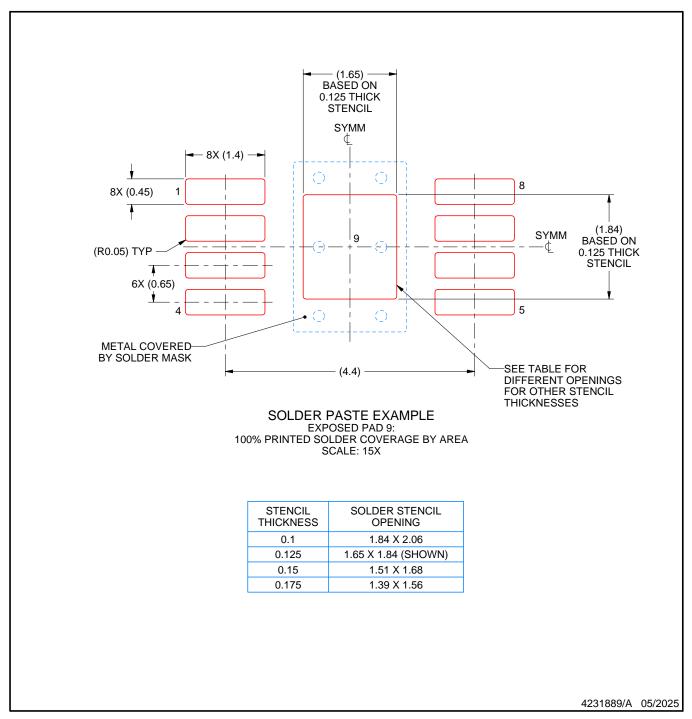




NOTES: (continued)

- 7. Publication IPC-7351 may have alternate designs.
- 8. Solder mask tolerances between and around signal pads can vary based on board fabrication site.
- 9. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.
- 10. Size of metal pad may vary due to creepage requirement.





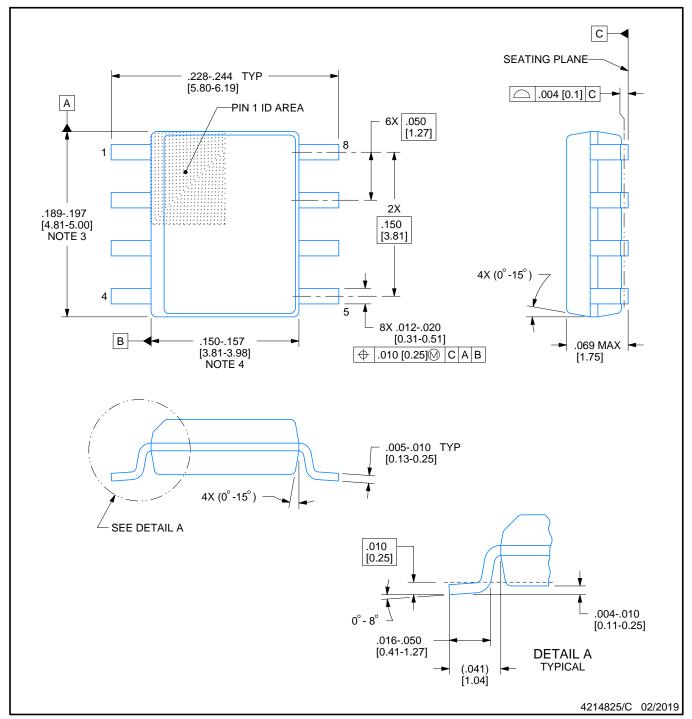
NOTES: (continued)

- 11. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 12. Board assembly site may have different recommendations for stencil design.





SMALL OUTLINE INTEGRATED CIRCUIT



## NOTES:

- 1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- 4. This dimension does not include interlead flash.
- 5. Reference JEDEC registration MS-012, variation AA.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.



SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

- 8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
- 9. Board assembly site may have different recommendations for stencil design.



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