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具有双向控制通道的 DS90UB903Q/DS90UB904Q 10 - 43MHz 18 位彩色 平面显示器-链路 (FPD-Link) III 串化器和解串器

查询样品: DS90UB903Q-Q1, DS90UB904Q-Q1

特性

- 10MHz 至 43MHz 输入并行端口时钟 (PCLK) 支持
- 210Mbps 至 903Mbps 数据吞吐量
- 单个差分对互连
- 具有 I²C 支持的双向控制接口通道
- 具有 DC 平衡编码的嵌入式时钟以支持 AC 耦合互连
- 能够驱动长达 10 米的屏蔽双绞线
- I²C 兼容串行接口
- 单个硬件器件寻址引脚
- 多达 4 个通用输入 (GPI) / 输出 (GPO)
- LOCK(锁定)输出报告,以及 AT-SPEED BIST(全速内置自检)诊断特性以验证连接完整性
- 集成端接电阻器
- 1.8V 或 3.3V 兼容并行数据接口
- 1.8V 单电源
- 符合 ISO 10605 静电放电 (ESD) 以及 IEC 61000-4-2 ESD 标准
- 汽车应用级产品:符合 AEC-Q100 2 级要求
- 温度范围: -40°C 至 +105°C
- 解串器上无需基准时钟
- 可编程接收均衡
- 电磁干扰 (EMI) / 电磁兼容性 (EMC) 迁移
 - DES 可编程展频 (SSCG) 输出
 - DES 接收器交错输出

应用范围

- 汽车显示系统
 - 中央信息显示屏
 - 导航显示屏
 - 后座娱乐系统
 - 触控显示屏

说明

DS90UB903Q/DS90UB904Q 芯片组为一个单个差分 对上的数据传输提供了具有高速正向通道和一个双向控 制通道的 FPD-Link III 接口。 DS90UB903Q/904Q 在 高速正向通道和双向控制通道数据路径上都包含差分信 令。 此串化器/解串器对针对图形主机控制器与显示模 块间的直接连接。 这个芯片组非常适合于将视频数据 驱动至要求 18 位色深(RGB666 + HS, VS 和 DE) 的显示屏以及双向控制通道总线。 主传输转换一个单 个高速串行数据流上的 21 位数据,连同一个从 I²C 端 口上接受控制信息的独立低延迟双向控制通道传输。 使用德州仪器 (TI) 的嵌入时钟技术可实现一个单个差 分对上的透明全双工通信,从而在两个方向上携带非对 称双向控制通道信息。 这个单个串行数据流通过消除 并行数据与时钟路径间的偏差,简化了印刷电路板 (PCB) 走线和电缆上的宽数据总线传输。 这样,通过 限制数据路径的宽度,大大节省了系统成本,相应地减 少了 PCB 层数、电缆宽度以及连接器尺寸和引脚数

此外,解串器输入提供均衡控制来补偿较长距离介质所造成的损耗。 內部 DC 均衡编码/解码被用来支持 AC 耦合互连。

此串化器采用 40 引脚超薄型四方扁平无引线 (WQFN) 封装,而解串器采用 48 引脚 WQFN 封装。

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Typical Application Diagram

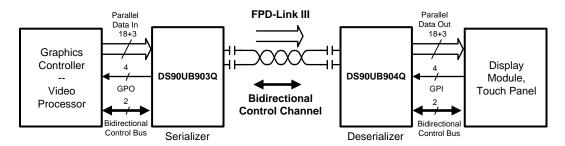


Figure 1. Typical Application Circuit

Block Diagrams

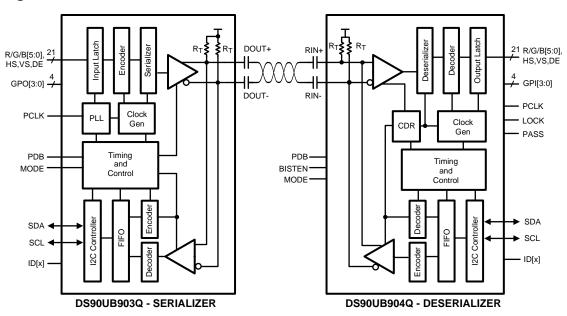


Figure 2. Block Diagram

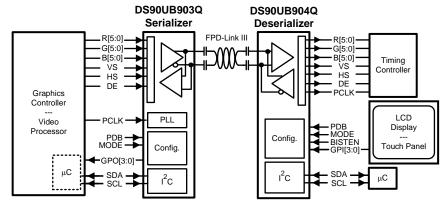
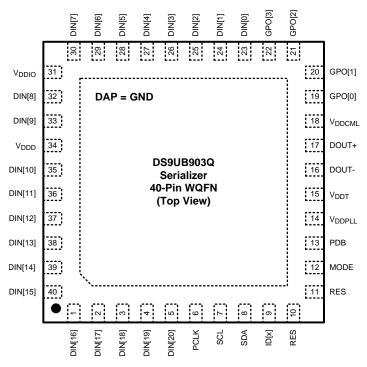


Figure 3. Application Block Diagram



DS90UB903Q Pin Diagram



Serializer - DS90UB903Q 40 Pin WQFN (Top View) See Package Number RTA0040A

DS90UB903Q SERIALIZER PIN DESCRIPTIONS

Pin Name	Pin No.	I/O, Type	Description
	ALLEL INTERFAC		
DIN[20:0]	5, 4, 3, 2, 1, 40, 39, 38, 37, 36, 35, 33, 32, 30, 29, 28, 27, 26, 25, 24, 23	Inputs, LVCMOS w/ pull down	Parallel data inputs.
PCLK	6	Input, LVCMOS w/ pull down	Pixel Clock Input Pin. Strobe edge set by TRFB control register.
GENERAL PUR	POSE OUTPUT (GPO)	
GPO[3:0]	22, 21, 20, 19	Output, LVCMOS	General-purpose output pins can be used to control and respond to various commands.
BIDIRECTIONA	L CONTROL BUS	- I ² C COMPATIB	LE
SCL	7	Input/Output, Open Drain	Clock line for the bidirectional control bus communication SCL requires an external pull-up resistor to V _{DDIO} .
SDA	8	Input/Output, Open Drain	Data line for the bidirectional control bus communication SDA requires an external pull-up resistor to V _{DDIO} .
MODE	12	Input, LVCMOS w/ pull down	I ² C Mode select MODE = L, Master mode (default); Device generates and drives the SCL clock line. Device is connected to slave peripheral on the bus. (Serializer initially starts up in Standby mode and is enabled through remote wakeup by Deserializer) MODE = H, Slave mode; Device accepts SCL clock input and attached to an I ² C controller master on the bus. Slave mode does not generate the SCL clock, but uses the clock generated by the Master for the data transfers.
ID[x]	9	Input, analog	Device ID Address Select Resistor to Ground and 10 k Ω pull-up to 1.8V rail. See Table 3
CONTROL AND	CONFIGURATIO	N	



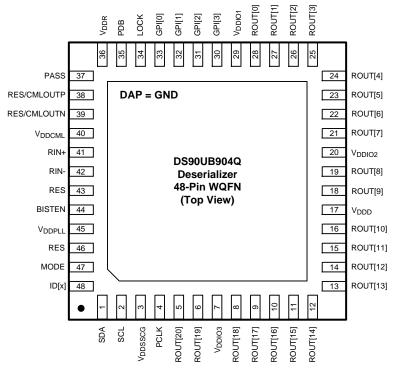


DS90UB903Q SERIALIZER PIN DESCRIPTIONS (continued)

Pin Name	Pin No.	I/O, Type	Description
PDB	13	Input, LVCMOS w/ pull down	Power down Mode Input Pin. PDB = H, Serializer is enabled and is ON. PDB = L, Serailizer is in Power Down mode. When the Serializer is in Power Down, the PLL is shutdown, and IDD is minimized. Programmed control register data are NOT retained and reset to default values
RES	10, 11	Input, LVCMOS w/ pull down	Reserved. This pin MUST be tied LOW.
FPD-LINK III INT	ERFACE		
DOUT+	17	Input/Output, CML	Non-inverting differential output, bidirectional control channel input. The interconnect must be AC Coupled with a 100 nF capacitor.
DOUT-	16	Input/Output, CML	Inverting differential output, bidirectional control channel input. The interconnect must be AC Coupled with a 100 nF capacitor.
POWER AND GR	OUND		
VDDPLL	14	Power, Analog	PLL Power, 1.8V ±5%
VDDT	15	Power, Analog	Tx Analog Power, 1.8V ±5%
VDDCML	18	Power, Analog	CML & Bidirectional Channel Driver Power, 1.8V ±5%
VDDD	34	Power, Digital	Digital Power, 1.8V ±5%
VDDIO	31	Power, Digital	Power for I/O stage. The single-ended inputs and SDA, SCL are powered from V_{DDIO} . V_{DDIO} can be connected to a 1.8V ±5% or 3.3V ±10%
VSS	DAP	Ground, DAP	DAP must be grounded. DAP is the large metal contact at the bottom side, located at the center of the WQFN package. Connected to the ground plane (GND) with at least 16 vias.



DS90UB904Q Pin Diagram



Deserializer - DS90UB904Q 48 Pin WQFN (Top View) See Package Number RHS0048A

DS90UB904Q DESERIALIZER PIN DESCRIPTIONS

Pin Name	Pin No.	I/O, Type	Description
LVCMOS PARA	LLEL INTERFAC	E	
ROUT[20:0]	5, 6, 8, 9, 10, 11, 12, 13, 14, 15, 16, 18, 19, 21, 22, 23, 24, 25, 26, 27, 28	Outputs, LVCMOS	Parallel data outputs.
PCLK	4	Output, LVCMOS	Pixel Clock Output Pin. Strobe edge set by RRFB control register.
GENERAL PURI	POSE INPUT (GP	PI)	
GPI[3:0]	30, 31, 32, 33	Input, LVCMOS	General-purpose input pins can be used to control and respond to various commands.
BIDIRECTIONAL	CONTROL BUS	- I ² C COMPATIB	LE
SCL	2	Input/Output, Open Drain	Clock line for the bidirectional control bus communication SCL requires an external pull-up resistor to V _{DDIO} .
SDA	1	Input/Output, Open Drain	Data line for bidirectional control bus communication SDA requires an external pull-up resistor to V _{DDIO} .
MODE	47	Input, LVCMOS w/ pull up	I ² C Mode select MODE = L, Master mode; Device generates and drives the SCL clock line, where required such as Read. Device is connected to slave peripheral on the bus. MODE = H, Slave mode (default); Device accepts SCL clock input and attached to an I ² C controller master on the bus. Slave mode does not generate the SCL clock, but uses the clock generated by the Master for the data transfers.
ID[x]	48	Input, analog	Device ID Address Select Resistor to Ground and 10 k Ω pull-up to 1.8V rail. See Table 4.
CONTROL AND	CONFIGURATIO	N	



DS90UB904Q DESERIALIZER PIN DESCRIPTIONS (continued)

Pin Name	Pin No.	I/O, Type	Description
PDB	35	Input, LVCMOS w/ pull down	Power down Mode Input Pin. PDB = H, Deserializer is enabled and is ON. PDB = L, Deserializer is in Power Down mode. When the Deserializer is in Power Down. Programmed control register data are NOT retained and reset to default values.
LOCK	34	Output, LVCMOS	LOCK Status Output Pin. LOCK = H, PLL is Locked, outputs are active LOCK = L, PLL is unlocked, ROUT and PCLK output states are controlled by OSS_SEL control register. May be used as Link Status.
RES	38, 39, 43, 46	-	Reserved. Pins 38, 39: Route to test point or leave open if unused. See also FPD-LINK III INTERFACE pin description section. Pin 46: This pin MUST be tied LOW. Pin 43: Leave pin open.
BIST MODE			
BISTEN	44	Input, LVCMOS w/ pull down	BIST Enable Pin. BISTEN = H, BIST Mode is enabled. BISTEN = L, BIST Mode is disabled.
PASS	37	Output, LVCOMS	PASS Output Pin for BIST mode. PASS = H, ERROR FREE Transmission PASS = L, one or more errors were detected in the received payload. Leave Open if unused. Route to test point (pad) recommended.
FPD-LINK III IN	TERFACE	1	
RIN+	41	Input/Output, CML	Non-inverting differential input, bidirectional control channel output. The interconnect must be AC Coupled with a 100 nF capacitor.
RIN-	42	Input/Output, CML	Inverting differential input, bidirectional control channel output. The interconnect must be AC Coupled with a 100 nF capacitor.
CMLOUTP	38	Output, CML	Non-inverting CML Output Monitor point for equalized differential signal. Test port is enabled via control registers.
CMLOUTN	39	Output, CML	Inverting CML Output Monitor point for equalized differential signal. Test port is enabled via control registers.
POWER AND G	ROUND		
VDDSSCG	3	Power, Digital	SSCG Power, 1.8V ±5% Power supply must be connected regardless if SSCG function is in operation.
VDDIO1/2/3	29, 20, 7	Power, Digital	LVCMOS I/O Buffer Power, The single-ended outputs and control input are powered from V_{DDIO} . V_{DDIO} can be connected to a 1.8V $\pm 5\%$ or 3.3V $\pm 10\%$
VDDD	17	Power, Digital	Digital Core Power, 1.8V ±5%
VDDR	36	Power, Analog	Rx Analog Power, 1.8V ±5%
VDDCML	40	Power, Analog	Bidirectional Channel Driver Power, 1.8V ±5%
VDDPLL	45	Power, Analog	PLL Power, 1.8V ±5%
VSS	DAP	Ground, DAP	DAP must be grounded. DAP is the large metal contact at the bottom side, located at the center of the WQFN package. Connected to the ground plane (GND) with at least 16 vias.



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

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Absolute Maximum Ratings (1)(2)(3)

Supply Voltage – V _{DDn} (1.8V)		-0.3V to +2.5V
Supply Voltage – V _{DDIO}		-0.3V to +4.0V
LVCMOS Input Voltage I/O Voltage		-0.3V to + (VDDIO + 0.3V)
CML Driver I/O Voltage (V _{DD})		$-0.3V$ to $+(V_{DD} + 0.3V)$
CML Receiver I/O Voltage (V _{DD})		$-0.3V$ to $(V_{DD} + 0.3V)$
Junction Temperature		+150°C
Storage Temperature		−65°C to +150°C
Maximum Package Power Dissipation Capacity		$1/\theta_{JA}$ °C/W above +25°
Package Derating		
Lead WQFN	θ _{JA} (based on 16 thermal vias)	30.7 °C/W
40 Lead WQFN	θ_{JC} (based on 16 thermal vias)	6.8 °C/W
48 Lead WQFN	θ _{JA} (based on 16 thermal vias)	26.9 °C/W
48 Lead WQFN	θ_{JC} (based on 16 thermal vias)	4.4 °C/W
ESD Rating (IEC 61000-4-2)		$R_D = 330\Omega, C_S = 150pF$
Air Discharge (DOUT+, DOUT-, RIN+, RIN-)		≥±25 kV
Contact Discharge (DOUT+, DOUT-, RIN+, RIN-)		≥±10 kV
ESD Rating (ISO10605)		$R_D = 330\Omega, C_S = 150/330pF$
ESD Rating (ISO10605)		$R_D = 2K\Omega, C_S = 150/330pF$
Air Discharge (DOUT+, DOUT-, RIN+, RIN-)		≥±15 kV
Contact Discharge (DOUT+, DOUT-, RIN+, RIN-)		≥±10 kV
ESD Rating (HBM)		≥±8 kV
ESD Rating (CDM)		≥±1 kV
ESD Rating (MM)		≥±250 V

^{(1) &}quot;Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur, including inoperability and degradation of device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional; the device should not be operated beyond such conditions.

Recommended Operating Conditions⁽¹⁾

		Min	Nom	Max	Units
Supply Voltage (V _{DDn})		1.71	1.8	1.89	V
LVCMOS Supply Voltage (V _{DDIO}) OR		1.71	1.8	1.89	V
LVCMOS Supply Voltage (V _{DDIO})		3.0	3.3	3.6	V
	V _{DDn} (1.8V)			25	mVp-p
Supply Noise	V _{DDIO} (1.8V)			25	mVp-p
	V _{DDIO} (3.3V)			25 mVp-p 25 mVp-p 50 mVp-p	mVp-p
Operating Free Air Temperature (T _A)	Operating Free Air Temperature (T _A)		+25	+105	°C
PCLK Clock Frequency		10		43	MHz

⁽¹⁾ Supply noise testing was done with minimum capacitors (as shown on Figure 37 and Figure 38) on the PCB. A sinusoidal signal is AC coupled to the VDDn (1.8V) supply with amplitude = 25 mVp-p measured at the device VDDn pins. Bit error rate testing of input to the Ser and output of the Des with 10 meter cable shows no error when the noise frequency on the Ser is less than 1 MHz. The Des on the other hand shows no error when the noise frequency is less than 750 kHz.

⁽²⁾ For soldering specifications: see product folder at www.ti.com

⁽³⁾ If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/Distributors for availability and specifications.



Electrical Characteristics (1)(2)(3)

Symbol	Parameter	Condi	tions	Min	Тур	Max	Units
LVCMOS	DC SPECIFICATIONS 3	3.3V I/O (SER INPUTS, DES	OUTPUTS, GPI, GPO, CO	NTROL INP	UTS AND O	UTPUTS)	
V _{IH}	High Level Input Voltage	V _{IN} = 3.0V to 3.6V	2.0		V _{IN}	V	
V_{IL}	Low Level Input Voltage	$V_{IN} = 3.0V \text{ to } 3.6V$		GND		0.8	V
I _{IN}	Input Current	$V_{IN} = 0V \text{ or } 3.6V, V_{IN} = 3.0$	V to 3.6V	-20	±1	+20	μΑ
V _{OH}	High Level Output Voltage	$V_{\rm DDIO} = 3.0 \text{V to } 3.6 \text{V}, I_{\rm OH} =$	- −4 mA	2.4		V _{DDIO}	V
V _{OL}	Low Level Output Voltage	$V_{DDIO} = 3.0V$ to 3.6V, $I_{OL} =$	+4 mA	GND		0.4	V
l _{os}	Output Short Circuit	V _{OUT} = 0V	Serializer GPO Outputs		-24		mA
	Current	V _{OUT} = 0 V	Deserializer LVCMOS Outputs		-39		ША
l _{oz}	TRI-STATE Output Current	$PDB = 0V,$ $V_{OUT} = 0V \text{ or } V_{DD}$	LVCMOS Outputs	-20	±1	+20	μΑ
LVCMOS	DC SPECIFICATIONS 1	.8V I/O (SER INPUTS, DES	OUTPUTS, GPI, GPO, CO	NTROL INP	UTS AND O	UTPUTS)	
V _{IH}	High Level Input Voltage	V _{IN} = 1.71V to 1.89V		0.65 V _{IN}		V _{IN} +0.3	V
V_{IL}	Low Level Input Voltage	V _{IN} = 1.71V to 1.89V	GND		0.35 V _{IN}	V	
I _{IN}	Input Current	$V_{IN} = 0V \text{ or } 1.89V, V_{IN} = 1.38V$	-20	±1	+20	μΑ	
V _{OH}	High Level Output Voltage	$V_{DDIO} = 1.71V$ to 1.89V, I_{OH}	_H = −4 mA	V _{DDIO} - 0.45		V _{DDIO}	V
V _{OL}	Low Level Output Voltage	$V_{DDIO} = 1.71V \text{ to } 1.89V$ $I_{OL} = +4 \text{ mA}$	Deserializer LVCMOS Outputs	GND		0.45	V
l _{os}	Output Short Circuit	V _{OUT} = 0V	Serializer GPO Outputs		-11		mA
	Current	VOUT - UV	Deserializer LVCMOS Outputs		-20		ША
l _{OZ}	TRI-STATE Output Current	$PDB = 0V,$ $V_{OUT} = 0V \text{ or } V_{DD}$	LVCMOS Outputs	-20	±1	+20	μΑ
CML DRI	VER DC SPECIFICATIO	NS (DOUT+, DOUT-)					
V _{OD}	Output Differential Voltage	$R_T = 100\Omega$ (Figure 8)		268	340	412	mV
ΔV _{OD}	Output Differential Voltage Unbalance	$R_L = 100\Omega$			1	50	mV
V _{OS}	Output Differential Offset Voltage	$R_L = 100\Omega$ (Figure 8)		V _{DD (MIN)} - V _{OD (MAX)}	V _{DD} - V _{OD}	V _{DD (MAX)} - V _{OD (MIN)}	V
ΔV _{OS}	Offset Voltage Unbalance	$R_L = 100\Omega$			1	50	mV
I _{os}	Output Short Circuit Current	DOUT+/- = 0V			-27		mA
R _T	Differential Internal Termination Resistance	Differential across DOUT+	and DOUT-	80	100	120	Ω

⁽¹⁾ The Electrical Characteristics tables list ensured specifications under the listed Recommended Operating Conditions except as otherwise modified or specified by the Electrical Characteristics Conditions and/or Notes. Typical specifications are estimations only and are not ensured.

⁽²⁾ Current into device pins is defined as positive. Current out of a device pin is defined as negative. Voltages are referenced to ground except VOD, ΔVOD, VTH and VTL which are differential voltages.

⁽³⁾ Typical values represent most likely parametric norms at 1.8V or 3.3V, T_A = +25°C, and at the Recommended Operation Conditions at the time of product characterization and are not ensured.

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Electrical Characteristics⁽¹⁾⁽²⁾⁽³⁾ (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Condition	ons	Min	Тур	Max	Units
V _{TH}	Differential Threshold High Voltage	(Figure 40)				+90	m)/
V_{TL}	Differential Threshold Low Voltage	(Figure 10)		-90			mV
V _{IN}	Differential Input Voltage Range	RIN+ - RIN-		180			mV
I _{IN}	Input Current	$V_{IN} = V_{DD}$ or 0V, $V_{DD} = 1.89$	J	-20	±1	+20	μΑ
R _T	Differential Internal Termination Resistance	Differential across RIN+ and	RIN-	80	100	120	Ω
SER/DES S	SUPPLY CURRENT *DI	GITAL, PLL, AND ANALOG	/DD				
I _{DDT}	Serializer (Tx) VDDn Supply Current (includes load current)	$R_T = 100\Omega$ WORST CASE pattern (Figure 5)	VDDn = 1.89V PCLK = 43 MHz		62	90	mA
		$R_T = 100\Omega$ RANDOM PRBS-7 pattern	Default Registers		55		
VDI Cur	Serializer (Tx) VDDIO Supply Current (includes load current)	$R_T = 100\Omega$ WORST CASE pattern (Figure 5)	VDDIO = 1.89V PCLK = 43 MHz Default Registers		2	5	A
			VDDIO = 3.6V PCLK = 43 MHz Default Registers		7	15	- mA
I _{DDTZ}			$V_{DDn} = 1.89V$		370	775	
I _{DDIOTZ}	Serializer (Tx) Supply Current Power-down	PDB = 0V; All other LVCMOS Inputs = 0V	$V_{DDIO} = 1.89V$		55	125	μΑ
			$V_{DDIO} = 3.6V$		65	135	
I _{DDR}	Deserializer (Rx) VDDn Supply Current (includes load	V _{DDn} = 1.89V, C _L = 8 pF WORST CASE Pattern (Figure 5)	PCLK = 43 MHz SSCG[3:0] = ON Default Registers		60	96	
	current)	V _{DDn} = 1.89V, C _L = 8 pF RANDOM PRBS-7 Pattern	PCLK = 43 MHz Default Registers		53		mA
I _{DDIOR}	Deserializer (Rx) VDDIO Supply Current (includes lead	V _{DDIO} = 1.89V, C _L = 8 pF WORST CASE Pattern (Figure 5)	PCLK = 43 MHz Default Registers		21	32	mA
	Current (includes load current)	V _{DDIO} = 3.6V, C _L = 8 pF WORST CASE Pattern	PCLK = 43 MHz Default Registers		49	83	
I _{DDRZ}	Deserializer (Rx)		$V_{DDn} = 1.89V$		42	400	μA
I _{DDIORZ}	Supply Current	PDB = 0V; All other LVCMOS Inputs = 0V	V _{DDIO} = 1.89V		8	40	
	Power-down	$V_{DDIO} = 3.6V$			350	800	

Recommended Serializer Timing for PCLK⁽¹⁾

Symbol	Parameter	Conditions	Min	Тур	Max	Units
t _{TCP}	Transmit Clock Period	10 MHz – 43 MHz	23.3	Т	100	ns
t _{TCIH}	Transmit Clock Input High Time		0.4T	0.5T	0.6T	ns
t _{TCIL}	Transmit Clock Input Low Time		0.4T	0.5T	0.6T	ns
t _{CLKT}	PCLK Input Transition Time (Figure 11)		0.5		3	ns
f _{OSC}	Internal oscillator clock source			25		MHz

⁽¹⁾ Recommended Input Timing Requirements are input specifications and not tested in production.



Serializer Switching Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
t _{LHT}	CML Low-to-High Transition Time	$R_L = 100\Omega$ (Figure 6)		150	330	ps
t _{HLT}	CML High-to-Low Transition Time	$R_L = 100\Omega$ (Figure 6)		150	330	ps
t _{DIS}	Data Input Setup to PCLK	Carializas Data Isparta (Figure 40)	2.0			ns
t _{DIH}	Data Input Hold from PCLK	Serializer Data Inputs (Figure 12)	2.0			ns
t _{PLD}	Serializer PLL Lock Time	$R_L = 100\Omega^{(1)(2)}$		1	2	ms
t _{SD}	Serializer Delay	$R_T = 100\Omega$, PCLK = 10–43 MHz Register 0x03h b[0] (TRFB = 1) (Figure 14)	6.386T + 5	6.386T + 12	6.386T + 19.7	ns
t _{JIND}	Serializer Output Deterministic Jitter	Serializer output intrinsic deterministic jitter . Measured (cycle-cycle) with PRBS-7 test pattern PCLK = 43 MHz ⁽³⁾⁽⁴⁾		0.13		UI
t _{JINR}	Serializer Output Random Jitter	Serializer output intrinsic random jitter (cycle-cycle). Alternating-1,0 pattern. PCLK = 43 MHz ⁽³⁾⁽⁴⁾		0.04		UI
t _{JINT}	Peak-to-peak Serializer Output Jitter	Serializer output peak-to-peak jitter includes deterministic jitter, random jitter, and jitter transfer from serializer input. Measured (cycle-cycle) with PRBS-7 test pattern. PCLK = 43 MHz ⁽³⁾⁽⁴⁾		0.396		UI
λ_{STXBW}	Serializer Jitter Transfer Function -3 dB Bandwidth	PCLK = 43 MHz, Default Registers (Figure 20) ⁽³⁾		1.90		MHz
δ_{STX}	Serializer Jitter Transfer Function (Peaking)	PCLK = 43 MHz, Default Registers (Figure 20) ⁽³⁾		0.944		dB
δ_{STXf}	Serializer Jitter Transfer Function (Peaking Frequency)	PCLK = 43 MHz, Default Registers (Figure 20) ⁽³⁾		500		kHz

- (1) t_{PLD} and t_{DDLT} is the time required by the serializer and deserializer to obtain lock when exiting power-down state with an active PCLK
- (2) Specification is ensured by design.
- (3) Typical values represent most likely parametric norms at 1.8V or 3.3V, T_A = +25°C, and at the Recommended Operation Conditions at the time of product characterization and are not ensured.
- (4) UI Unit Interval is equivalent to one ideal serialized data bit width. The UI scales with PCLK frequency.

Deserializer Switching Characteristics

Symbol	Parameter	Conditions	Pin/Freq.	Min	Тур	Max	Units
t _{RCP}	Receiver Output Clock Period	$t_{RCP} = t_{TCP}$	PCLK	23.3	Т	100	ns
t _{PDC}	PCLK Duty Cycle	Default Registers SSCG[3:0] = OFF	PCLK	45	50	55	%
t _{CLH}	LVCMOS Low-to-High Transition Time	V _{DDIO} : 1.71V to 1.89V or 3.0 to 3.6V, C _L = 8 pF (lumped load) Default Registers (Figure 16) ⁽¹⁾		1.3	2.0	2.8	
t _{CHL}	LVCMOS High-to-Low Transition Time		PCLK	1.3	2.0	2.8	ns
t _{CLH}	LVCMOS Low-to-High Transition Time	V _{DDIO} : 1.71V to 1.89V or 3.0 to 3.6V,	Deserializer ROUTn Data Outputs	1.6	2.4	3.3	
t _{CHL}	LVCMOS High-to-Low Transition Time	C _L = 8 pF (lumped load) Default Registers (Figure 16) ⁽¹⁾		1.6	2.4	3.3	ns
t _{ROS}	ROUT Setup Data to PCLK	V _{DDIO} : 1.71V to 1.89V or		0.38T	0.5T		
t _{ROH}	ROUT Hold Data to PCLK	3.0V to 3.6V, C _L = 8 pF (lumped load) Default Registers	Deserializer ROUTn Data Outputs	0.38T	0.5T		ns

⁽¹⁾ Specification is ensured by characterization and is not tested in production.



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Deserializer Switching Characteristics (continued)

Symbol	Parameter	Conditions	Pin/Freq.	Min	Тур	Max	Units
t _{DD}	Deserializer Delay	Default Registers Register 0x03h b[0] (RRFB = 1) (Figure 17)	10 MHz-43 MHz	4.571T + 8	4.571T + 12	4.571T + 16	ns
t _{DDLT}	Deserializer Data Lock Time	(Figure 15) ⁽²⁾	10 MHz-43 MHz			10	ms
t _{RJIT}	Receiver Input Jitter Tolerance	(Figure 19, Figure 21) ⁽³⁾⁽⁴⁾	43 MHz		0.53		UI
t _{RCJ}	Bassiner Clask litter	PCLK	10 MHz		300	550	
Re	Receiver Clock Jitter	$SSCG[3:0] = OFF^{(1)(5)}$	43 MHz		120	250	ps
t _{DPJ}	Deserializer Period Jitter	PCLK	10 MHz		425	600	
	Deserranzer Period Jiller	$SSCG[3:0] = OFF^{(1)(6)}$	43 MHz		320	480	ps
t _{DCCJ}	Deserializer Cycle-to-Cycle Clock	PCLK	10 MHz		320	500	
	Jitter	$SSCG[3:0] = OFF^{(1)(7)}$	43 MHz		300	500	ps
fdev	Spread Spectrum Clocking Deviation Frequency	LVCMOS Output Bus	20 MHz-43 MHz		±0.5% to ±2.0%		%
fmod	Spread Spectrum Clocking Modulation Frequency	SSC[3:0] = ON (Figure 22)	20 MHz-43 MHz		9 kHz to 66 kHz		kHz

 t_{PLD} and t_{DDLT} is the time required by the serializer and deserializer to obtain lock when exiting power-down state with an active PCLK UI – Unit Interval is equivalent to one ideal serialized data bit width. The UI scales with PCLK frequency.

 t_{RJIT} max (0.61UI) is limited by instrumentation and actual t_{RJIT} of in-band jitter at low frequency (<2 MHz) is greater 1 UI. t_{DCJ} is the maximum amount of jitter measured over 30,000 samples based on Time Interval Error (TIE).

t_{DPJ} is the maximum amount the period is allowed to deviate measured over 30,000 samples.

t_{DCCJ} is the maximum amount of jitter between adjacent clock cycles measured over 30,000 samples.



Bidirectional Control Bus AC Timing Specifications (SCL, SDA) - I²C Compliant

Over recommended supply and temperature ranges unless otherwise specified. See Figure 4.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
RECOMM	ENDED INPUT TIMING REQUIREMENTS	(1)		•		
f _{SCL}	SCL Clock Frequency		>0		100	kHz
t _{LOW}	SCL Low Period		4.7			μs
t _{HIGH}	SCL High Period		4.0			μs
t _{HD:STA}	Hold time for a start or a repeated start condition		4.0			μs
t _{SU:STA}	Set Up time for a start or a repeated start condition		4.7			μs
t _{HD:DAT}	Data Hold Time	f _{SCL} = 100 kHz	0		3.45	μs
t _{SU:DAT}	Data Set Up Time		250			ns
t _{SU:STO}	Set Up Time for STOP Condition		4.0			μs
t _r	SCL & SDA Rise Time				1000	ns
t _f	SCL & SDA Fall Time				300	ns
C _b	Capacitive load for bus				400	pF
SWITCHII	NG CHARACTERISTICS ⁽²⁾			•		
f _{SCL}	SCL Clock Frequency	Serializer MODE = 0 - R/W Register 0x05 = 0x40'h		100		kHz
		Deserializer MODE = 0 - READ Register 0x06 b[6:4] = 0x00'h				KHZ
+	SCL Low Period	Serializer MODE = 0 - R/W Register 0x05 = 0x40'h	4.7			116
t _{LOW}	SCL Low Period	Deserializer MODE = 0 - READ Register 0x06 b[6:4] = 0x00'h	4.7			μs
t	SCL High Period	Serializer MODE = 0 - R/W Register 0x05 = 0x40'h	4.0			116
t _{HIGH}	SCL High Fellou	Deserializer MODE = 0 - READ Register 0x06 b[6:4] = 0x00'h	4.0			μs
t _{HD:STA}	Hold time for a start or a repeated start condition	Serializer MODE = 0 Register 0x05 = 0x40'h	4.0			μs
t _{SU:STA}	Set Up time for a start or a repeated start condition	Serializer MODE = 0 Register 0x05 = 0x40'h	4.7			μs
t _{HD:DAT}	Data Hold Time		0		3.45	μs
t _{SU:DAT}	Data Set Up Time		250			ns
t _{SU:STO}	Set Up Time for STOP Condition	Serializer MODE = 0	4.0			μs
t _f	SCL & SDA Fall Time				300	ns
t _{BUF}	Bus free time between a stop and start condition	Serializer MODE = 0	4.7			μs
		Serializer MODE = 1		1		
t _{TIMEOUT}	NACK Time out	Deserializer MODE = 1 Register 0x06 b[2:0]=111'b		25		ms

¹⁾ Recommended Input Timing Requirements are input specifications and not tested in production.

⁽²⁾ Specification is ensured by design.



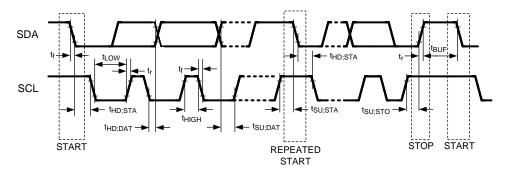


Figure 4. Bidirectional Control Bus Timing

Bidirectional Control Bus DC Characteristics (SCL, SDA) - I²C Compliant

Over recommended supply and temperature ranges unless otherwise specified.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
V _{IH}	Input High Level	SDA and SCL	0.7 x V _{DDIO}		V _{DDIO}	V
V_{IL}	Input Low Level Voltage	SDA and SCL	GND		0.3 x V _{DDIO}	V
V_{HY}	Input Hysteresis	SDA and SCL		>50		mV
l _{OZ}	TRI-STATE Output Current	PDB = 0V, V _{OUT} = 0V or V _{DD}	-20	±1	+20	μΑ
I _{IN}	Input Current	SDA or SCL, Vin = V _{DDIO} or GND	-20	±1	+20	μΑ
C _{IN}	Input Pin Capacitance			<5		pF
V _{OL}	Low Lovel Output Voltage	SCL and SDA, $V_{DDIO} = 3.0V$ $I_{OL} = 1.5 \text{ mA}$			0.36	V
	Low Level Output Voltage	SCL and SDA, $V_{DDIO} = 1.71V$ $I_{OL} = 1 \text{ mA}$			0.36	V

AC Timing Diagrams and Test Circuits

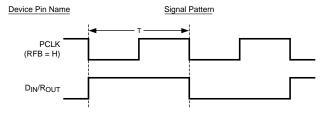


Figure 5. "Worst Case" Test Pattern

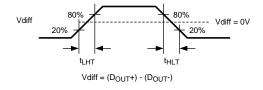


Figure 6. Serializer CML Output Load and Transition Times



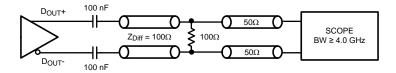


Figure 7. Serializer CML Output Load and Transition Times

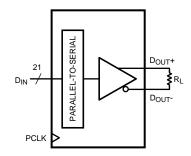


Figure 8. Serializer VOD DC Diagram

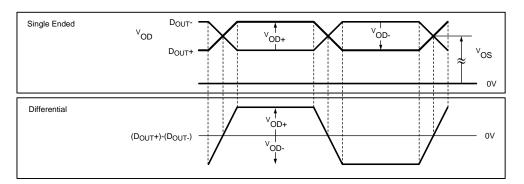


Figure 9. Serializer VOD DC Diagram

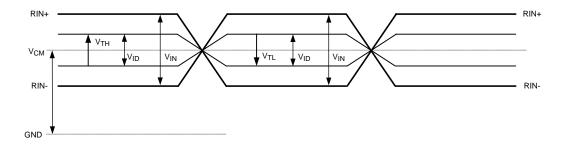


Figure 10. Differential VTH/VTL Definition Diagram

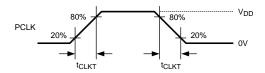


Figure 11. Serializer Input Clock Transition Times



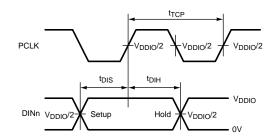


Figure 12. Serializer Setup/Hold Times

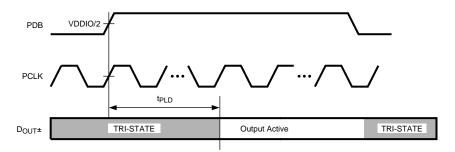


Figure 13. Serializer Data Lock Time

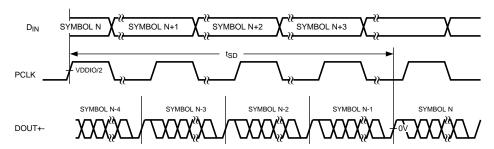


Figure 14. Serializer Delay

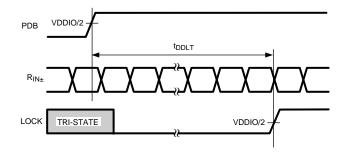


Figure 15. Deserializer Data Lock Time

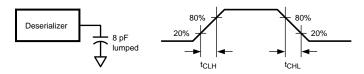


Figure 16. Deserializer LVCMOS Output Load and Transition Times



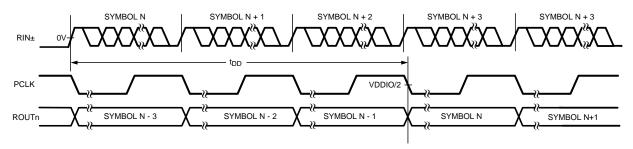


Figure 17. Deserializer Delay

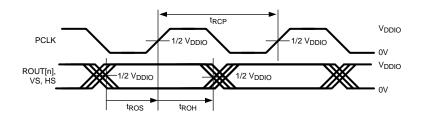


Figure 18. Deserializer Output Setup/Hold Times

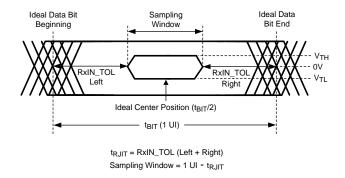


Figure 19. Receiver Input Jitter Tolerance

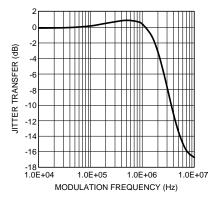


Figure 20. Typical Serializer Jitter Transfer Function Curve at 43 MHz



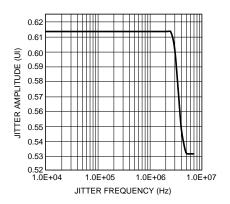


Figure 21. Typical Deserializer Input Jitter Tolerance Curve at 43 MHz

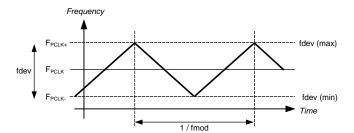


Figure 22. Spread Spectrum Clock Output Profile



Table 1. DS90UB903Q Control Registers

Addr (Hex)	Name	Bits	Field	R/W	Default	Description
0	I ² C Device ID	7:1	DEVICE ID	RW	0xB0'h	7-bit address of Serializer; 0x58'h (1011_000X'b) default
0	I ² C Device ID		SER ID SEL	IXVV	OXDOTI	0: Device ID is from ID[x] 1: Register I ² C Device ID overrides ID[x]
		7:3	RESERVED		0x00'h	Reserved
1	Reset	2	STANDBY	RW	0	Standby mode control. Retains control register data. Supported only when MODE = 0 0: Enabled. Low-current Standby mode with wake-up capability. Suspends all clocks and functions. 1: Disabled. Standby and wake-up disabled
		1	DIGITAL RESET0	RW	0 self clear	1: Resets the device to default register values. Does not affect device I ² C Bus or Device ID
		0	DIGITAL RESET1	RW	0 self clear	1: Digital Reset, retains all register values
2	Reserved	7:0	RESERVED		0x20'h	Reserved
	Reserved	7:6	RESERVED		11'b	Reserved
	VDDIO Control	5	VDDIO CONTOL	RW	1	Auto V _{DDIO} detect Allows manual setting of VDDIO by register. 0: Disable 1: Enable (auto detect mode)
	VDDIO Mode	4	VDDIO MODE	RW	1	VDDIO voltage set Only used when VDDIOCONTROL = 0 0: 1.8V 1: 3.3V
3	I ² C Pass-Through	3	I ² C PASS- THROUGH	RW	1	I ² C Pass-Through 0: Disabled 1: Enabled
	RESERVED	2	RESERVED		0	Reserved
	PCLK_AUTO	1	PCLK_AUTO	RW	1	Switch over to internal 25 MHz Oscillator clock in the absence of PCLK 0: Disable 1: Enable
	TRFB	0	TRFB	RW	1	Pixel Clock Edge Select: 0: Parallel Interface Data is strobed on the Falling Clock Edge. 1: Parallel Interface Data is strobed on the Rising Clock Edge.
4	RESERVED	7:0	RESERVED		0x80'h	Reserved
5	I ² C Bus Rate	7:0	I ² C BUS RATE	RW	0x40'h	I^2C SCL frequency is determined by the following: $f_{SCL} = 6.25$ MHz / Register value (in decimal) 0x40'h = ~100 kHz SCL (default) Note: Register values <0x32'h are NOT supported.
6	DES ID	7:1	DES DEV ID	RW	0xC0'h	Deserializer Device ID = 0x60'h (1100_000X'b) default
		0	RESERVED			Reserved
7	Slave ID	7:1	SLAVE DEV ID	RW	0x00'h	Slave Device ID. Sets remote slave I ² C address.
<u>'</u>	Slave ID	0	RESERVED			Reserved
8	Reserved	7:0	RESERVED		0x00'h	Reserved
9	Reserved	7:0	RESERVED		0x01'h	Reserved
Α	Reserved	7:0	RESERVED		0x00'h	Reserved
В	Reserved	7:0	RESERVED		0x00'h	Reserved

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Table 1. DS90UB903Q Control Registers (continued)

Addr (Hex)	Name	Bits	Field	R/W	Default	Description
	Reserved	7:3	RESERVED		0x00'h	Reserved
С	PCLK Detect	2	PCLK DETECT	R	0	1: Valid PCLK detected 0: Valid PCLK not detected
C	Reserved	3	RESERVED		0	Reserved
	Cable Link Detect Status	0	LINK DETECT	R	0	0: Cable link not detected 1: Cable link detected
D	Reserved	7:0	RESERVED		0x11'h	Reserved
Е	Reserved	7:0	RESERVED		0x01'h	Reserved
F	Reserved	7:0	RESERVED		0x03'h	Reserved
10	Reserved	7:0	RESERVED		0x03'h	Reserved
11	Reserved	7:0	RESERVED		0x03'h	Reserved
12	Reserved	7:0	RESERVED		0x03'h	Reserved
			GPCR[7]			0: LOW
			GPCR[6]			1: HIGH
			GPCR[5]			
13	General Purpose	7:0	GPCR[4]	RW	0x00'h	
13	Control Reg	7.0	GPCR[3]	NVV	0.00011	
			GPCR[2]			
			GPCR[1]			
			GPCR[0]			

Table 2. DS90UB904Q Control Registers

Addr (Hex)	Name	Bits	Field	R/W	Default	Description
0	I ² C Device ID	7:1	DEVICE ID	RW	0xC0'h	7-bit address of Deserializer; 0x60h (1100_000X) default
U	T C Device ID	0	DES ID SEL			0: Device ID is from ID[x] 1: Register I ² C Device ID overrides ID[x]
		7:3	RESERVED		0x00'h	Reserved
1	Reset	2	REM_WAKEUP	RW	0	Remote Wake-up Select 1: Enable Generate remote wakeup signal automatically wake-up the Serializer in Standby mode 0: Disable Puts the Serializer in Standby mode
		1	DIGITALRESET0	RW	0 self clear	1: Resets the device to default register values. Does not affect device I ² C Bus or Device ID
		0	DIGITALRESET1	RW	0 self clear	1: Digital Reset, retains all register values



Table 2. DS90UB904Q Control Registers (continued)

Addr (Hex)	Name	Bits	Field	R/W	Default	Description	
	RESERVED	7:6	RESERVED		00'b	Reserved	
	Auto Clock	5	AUTO_CLOCK	RW	0	Output PCLK or Internal 25 MHz Oscillator clock O: Only PCLK when valid PCLK present	
	OSS Select	4	OSS_SEL	RW	0	Output Sleep State Select 0: Outputs = TRI-STATE, when LOCK = L 1: Outputs = LOW, when LOCK = L	
2	SSCG	3:0	SSCG		0000'b	SSCG Select 0000: Normal Operation, SSCG OFF (default) 0001: fmod (kHz) PCLK/2168, fdev ±0.50% 0010: fmod (kHz) PCLK/2168, fdev ±1.00% 0011: fmod (kHz) PCLK/2168, fdev ±1.50% 0100: fmod (kHz) PCLK/2168, fdev ±2.00% 0101: fmod (kHz) PCLK/1300, fdev ±0.50% 0101: fmod (kHz) PCLK/1300, fdev ±1.00% 0111: fmod (kHz) PCLK/1300, fdev ±1.50% 1000: fmod (kHz) PCLK/1300, fdev ±0.50% 1001: fmod (kHz) PCLK/1300, fdev ±0.50% 1001: fmod (kHz) PCLK/868, fdev ±0.50% 1010: fmod (kHz) PCLK/868, fdev ±1.50% 1100: fmod (kHz) PCLK/868, fdev ±1.50% 1100: fmod (kHz) PCLK/868, fdev ±2.00% 1101: fmod (kHz) PCLK/868, fdev ±0.50% 1110: fmod (kHz) PCLK/868, fdev ±0.50% 1111: fmod (kHz) PCLK/850, fdev ±1.50%	
	RESERVED	7:6	RESERVED		11'b	Reserved	
	VDDIO Control	5	VDDIO CONTROL	RW	1	Auto voltage control 0: Disable 1: Enable (auto detect mode)	
	VDDIO Mode	4	VDDIO MODE	RW	0	VDDIO voltage set Only used when VDDIOCONTROL = 0 0: 1.8V 1: 3.3V	
3	I ² C Pass-Through	3	I ² C PASS- THROUGH	RW	1	I ² C Pass-Through Mode 0: Disabled 1: Enabled	
	Auto ACK	2	AUTO ACK	RW	0	0: Disable 1: Enable	
	RESERVED	1	RESERVED		0	Reserved	
	RRFB	0	RRFB	RW	1	Pixel Clock Edge Select 0: Parallel Interface Data is strobed on the Falling Clock Edge 1: Parallel Interface Data is strobed on the Rising Clock Edge.	
4	EQ Control	7:0	EQ	RW	0x00'h	EQ Gain 00'h = ~0.0 dB 01'h = ~4.5 dB 03'h = ~6.5 dB 07'h = ~7.5 dB 0F'h = ~8.0 dB 1F'h = ~11.0 dB 3F'h = ~12.5 dB FF'h = ~14.0 dB	
5	RESERVED	7:0	RESERVED		0x00'h	Reserved	



Table 2. DS90UB904Q Control Registers (continued)

RESERVED 7 RESERVED 0 Reserved	Addr (Hex)	Name	Bits	Field	R/W	Default	Description
SCL Prescale 6-4 SCL_PRESCALE RW O00b O00 - 100 kHz SCL (default) O00 - 100 kHz SCL (default) O00 - 100 kHz SCL (default) O00 - 100 kHz SCL O00 kHz SCL O00 - 100 kHz SCL O00 kHz SCL		RESERVED	7	RESERVED		0	Reserved
Remote NACK 3 REM_NACK_TIME RW 1 In slave mode (MODE = 1) if bit is set the I ^C C core will automatically timeout when no acknowledge condition was detected. Enable		SCL Prescale	6:4	SCL_PRESCALE	RW	000'b	from a slave device (MODE = 0) 000 : ~100 kHz SCL (default) 001 : ~125 kHz SCL 101 : ~11 kHz SCL 110 : ~33 kHz SCL 111 : ~50 kHz SCL
Remote NACK 2:0 NACK_TIMEOUT RW 1111b 0000: 2.0 ms 0001: 3.6 ms 001: 3.6 ms 001: 3.6 ms 001: 3.6 ms 001: 3.1 Ms 100: 14.4 ms 100: 14.6 ms 111: 25.0 ms 25.0	6	Remote NACK	3		RW	1	In slave mode (MODE = 1) if bit is set the I ² C core will automatically timeout when no acknowledge condition was detected. 1: Enable
SER ID		Remote NACK	2:0	NACK_TIMEOUT	RW	111'b	000: 2.0 ms 001: 5.2 ms 010: 8.6 ms 011: 11.8 ms 100: 14.4 ms 101: 18.4 ms 110: 21.6 ms
B	7	SER ID	7:1	SER DEV ID	RW	0xB0'h	
B			0	RESERVED			Reserved
10	0	IDIOI Indov	7:1	ID[0] INDEX	RW	0x00'h	Target slave Device ID slv_id0 [7:1]
Description	O	ib[0] index	0	RESERVED			Reserved
A ID[2] Index 7:1 ID[2] INDEX RW 0x00'h Reserved Target slave Device ID slv_id2 [7:1] Reserved Target slave Device ID slv_id3 [7:1] Reserved Target slave Device ID slv_id3 [7:1] Reserved Target slave Device ID slv_id3 [7:1] Reserved Target slave Device ID slv_id4 [7:1] Reserved Target slave Device ID slv_id5 [7:1] Reserved Target slave Device ID slv_id5 [7:1] Reserved Target slave Device ID slv_id5 [7:1] Reserved Target slave Device ID slv_id6 [7:1] Reserved Target slave Device ID slv_id6 [7:1] Reserved Target slave Device ID slv_id6 [7:1] Reserved Target slave Device ID slv_id7 [7:1] Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id2 [7:1] Reserved Alias to match Device ID slv_id2 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Alias to match Device ID slv_id4 [7:1	0	ID[1] Indov	7:1	ID[1] INDEX	DW	0v00'h	Target slave Device ID slv_id1 [7:1]
A	9	ID[1] Index	0	RESERVED	RW	0,0011	Reserved
B	^	IDIOI Index	7:1	ID[2] INDEX	DW	0,0016	Target slave Device ID slv_id2 [7:1]
B	A	ID[2] Index	0	RESERVED	KVV	UXUU II	Reserved
C ID[4] Index 7:1 ID[4] INDEX RW 0x00'h Reserved Target slave Device ID slv_id4 [7:1] Reserved Target slave Device ID slv_id5 [7:1] Reserved Target slave Device ID slv_id6 [7:1] Reserved Target slave Device ID slv_id6 [7:1] Reserved Target slave Device ID slv_id6 [7:1] Reserved Target slave Device ID slv_id7 [7:1] Reserved Alias to match Device ID slv_id7 [7:1] Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id2 [7:1] Reserved Alias to match Device ID slv_id2 [7:1] Reserved Alias to match Device ID slv_id2 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Alias to match Device ID slv_id4 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Alias to match Device ID slv_id4 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Alias to m	D	ID[3] Indov	7:1	ID[3] INDEX	DW/	0v00'h	Target slave Device ID slv_id3 [7:1]
Description of the image of t	В	iD[3] index	0	RESERVED	KVV	UXUU II	Reserved
D	0	IDIAI I. J.	7:1	ID[4] INDEX	DW	0001	Target slave Device ID slv_id4 [7:1]
D	C	ID[4] Index	0	RESERVED	RW	0x00'h	Reserved
Reserved Reserved	-	IDIEI I. J.	7:1	ID[5] INDEX	DW	0001	Target slave Device ID slv_id5 [7:1]
F ID[7] Index 0 RESERVED RW 0x00'h Reserved	D	ID[5] Index	0	RESERVED	RW	UXUU'n	Reserved
Total ID[7] Index Total ID[7] ID[7	_	IDIOI Index	7:1	ID[6] INDEX	DW	00015	Target slave Device ID slv_id6 [7:1]
Total Process of the content of th	E	ID[6] Index	0	RESERVED	RW	UXUU'n	Reserved
10	_		7:1	ID[7] INDEX	514		Target slave Device ID slv_id7 [7:1]
10	F	ID[7] Index	0	RESERVED	RW	0x00'h	
11 ID[1] Match 7:1 ID[1] MATCH RW 0x00'h Reserved Alias to match Device ID slv_id1 [7:1] Reserved Alias to match Device ID slv_id2 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Alias to match Device ID slv_id4 [7:1] Reserved Alias to match			7:1	ID[0] MATCH			Alias to match Device ID slv_id0 [7:1]
Total Match	10	ID[0] Match	0		RW	0x00'h	
11			7:1	ID[1] MATCH			Alias to match Device ID slv_id1 [7:1]
12 ID[2] Match 7:1 ID[2] MATCH RW 0x00'h Reserved Reserved	11	ID[1] Match	0		RW	0x00'h	Reserved
12 ID[2] Match 0 RESERVED RW 0x00'h Reserved 13 ID[3] Match 7:1 ID[3] MATCH 0 RESERVED RW 0x00'h Reserved Alias to match Device ID slv_id3 [7:1] Reserved 14 ID[4] Match 7:1 ID[4] MATCH RW 0x00'h Reserved Alias to match Device ID slv_id4 [7:1]							Alias to match Device ID slv id2 [7:1]
13 ID[3] Match 7:1 ID[3] MATCH RW 0x00'h Alias to match Device ID slv_id3 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] Reserved Alias to match Device ID slv_id4 [7:1] RW 0x00'h Alias to match Device ID slv_id4 [7:1] RW Alias to match Device ID slv_id4	12	ID[2] Match			RW	0x00'h	
13 ID[3] Match 0 RESERVED RW 0x00'h Reserved 14 ID[4] Match 7:1 ID[4] MATCH RW 0x00'h Reserved Alias to match Device ID slv_id4 [7:1]							
14 ID[4] Match 7:1 ID[4] MATCH RW 0x00'h Alias to match Device ID slv_id4 [7:1]	13	ID[3] Match			RW	0x00'h	
14 IDI4I Match RW 0x00'h							
	14	ID[4] Match	0	RESERVED	RW	0x00'h	Reserved



Table 2. DS90UB904Q Control Registers (continued)

Addr (Hex)	Name	Bits	Field	R/W	Default	Description
45	IDIGI Martala	7:1	ID[5] MATCH	DW	0001	Alias to match Device ID slv_id5 [7:1]
15	ID[5] Match	0	RESERVED	RW	0x00'h	Reserved
40	IDICI Matak	7:1	ID[6] MATCH	DW	00015	Alias to match Device ID slv_id6 [7:1]
16	ID[6] Match	0	RESERVED	RW	0x00'h	Reserved
4.7	IDIZI Matak	7:1	ID[7] MATCH	DW	00016	Alias to match Device ID slv_id [7:1]
17	ID[7] Match	0	RESERVED	RW	0x00'h	Reserved
18	RESERVED	7:0	RESERVED		0x00'h	Reserved
19	RESERVED	7:0	RESERVED		0x01'h	Reserved
1A	RESERVED	7:0	RESERVED		0x00'h	Reserved
1B	RESERVED	7:0	RESERVED		0x00'h	Reserved
	RESERVED	7:3	RESERVED		0x00'h	Reserved
	RESERVED	2	RESERVED		0	Reserved
1C	Signal Detect Status	1		R	0	O: Active signal not detected 1: Active signal detected
	LOCK Pin Status	0		R	0	0: CDR/PLL Unlocked 1: CDR/PLL Locked
1D	Reserved	7:0	RESERVED		0x17'h	Reserved
1E	Reserved	7:0	RESERVED		0x07'h	Reserved
1F	Reserved	7:0	RESERVED		0x01'h	Reserved
20	Reserved	7:0	RESERVED		0x01'h	Reserved
21	Reserved	7:0	RESERVED		0x01'h	Reserved
22	Reserved	7:0	RESERVED		0x01'h	Reserved
23	General Purpose Control Reg	7:0	GPCR[7] GPCR[6] GPCR[5] GPCR[4] GPCR[3] GPCR[2] GPCR[1] GPCR[0]	RW	0x00'h	0: LOW 1: HIGH
24	BIST	0	BIST_EN	RW	0	BIST Enable 0: Normal operation 1: Bist Enable
25	BIST_ERR	7:0	BIST_ERR	R	0x00'h	Bist Error Counter
26	Remote Wake Enable	7:6	REM_WAKEUP_ EN	RW	00'b	11: Enable remote wake up mode 00: Normal operation mode Other values are NOT supported
		5:0	RESERVED	RW	0	Reserved
27	BCC	7:6	BCC	RW	00'b	11: Normal operation mode
	BCC	5:0	RESERVED		0	Reserved
		7:5	RESERVED		0	Reserved
3F	CMLOUT Config	4	CMLOUT P/N Enable	RW	1	1: Disabled (Default) 0: Enabled
		3:0	RESERVED		0	Reserved



FUNCTIONAL DESCRIPTION

The DS90UB903Q/904Q FPD-Link III chipset is intended for video display applications. The Serializer/Deserializer chipset operates from a 10 MHz to 43 MHz pixel clock frequency. The DS90UB903Q transforms a 21-bit wide parallel LVCMOS data bus along with a bidirectional control bus into a single high-speed differential pair. The high-speed serial bit stream contains an embedded clock and DC-balance information which enhances signal quality to support AC coupling. The DS90UB904Q receives the single serial data stream and converts it back into a 21-bit wide parallel data bus together with the bidirectional control channel data bus.

The control channel function of the DS90UB903Q/904Q provides bidirectional communication between the host processor and display. The integrated control channel transfers data simultaneously over the same differential pair used for video data interface. This interface offers advantages over other chipsets by eliminating the need for additional wires for programming and control. The control supports I²C port. The bidirectional control channel offers asymmetrical communication and is not dependent on video blanking intervals.

DISPLAY APPLICATION

The DS90UB903Q/904Q chipset is intended for interface between a host (graphics processor) and a Display. It supports a 21 bit parallel video bus for 18-bit color depth (RGB666) display format. In a RGB666 configuration, 18 color bits (R[5:0], G[5:0], B[5:0]), Pixel Clock (PCLK) and three control bits (VS, HS and DE) are supported across the serial link.

The DS90UB903Q Serializer accepts a 21-bit parallel data bus along with a bidirectional control bus. The parallel data and bidirectional control channel information is converted into a single differential link. The integrated bidirectional control channel bus supports I2C compatible operation for controlling auxiliary data transport to and from host processor and display module. The DS90UB904Q Deserializer extracts the clock/control information from the incoming data stream and reconstructs the 21-bit data with control channel data.

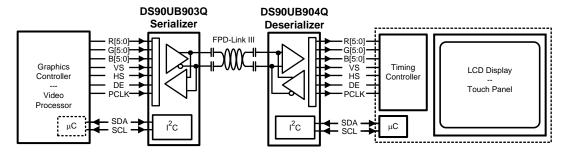


Figure 23. Typical Display System Diagram

SERIAL FRAME FORMAT

The DS90UB903Q/904Q chipset will transmit and receive a pixel of data in the following format:

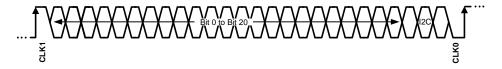


Figure 24. Serial Bitstream for 28-bit Symbol

The High Speed Forward Channel is a 28-bit symbol composed of 21 bits of data containing video data & control information transmitted from Serializer to Deserializer. CLK1 and CLK0 represent the embedded clock in the serial stream. CLK1 is always HIGH and CLK0 is always LOW. This data payload is optimized for signal transmission over an AC coupled link. Data is randomized, balanced and scrambled.

The bidirectional control channel data is transferred along with the high-speed forward data over the same serial link. This architecture provides a full duplex low speed forward channel across the serial link together with a high speed forward channel without the dependence of the video blanking phase.



DESCRIPTION OF BIDIRECTIONAL CONTROL BUS AND I2C MODES

The I²C compatible interface allows programming of the DS90UB903Q, DS90UB904Q, or an external remote device (such as a display) through the bidirectional control channel. Register programming transactions to/from the DS90UB903Q/904Q chipset are employed through the clock (SCL) and data (SDA) lines. These two signals have open-drain I/Os and both lines must be pulled-up to VDDIO by external resistor. Figure 4 shows the timing relationships of the clock (SCL) and data (SDA) signals. Pull-up resistors or current sources are required on the SCL and SDA busses to pull them high when they are not being driven low. A logic zero is transmitted by driving the output low. A logic high is transmitted by releasing the output and allowing it to be pulled-up externally. The appropriate pull-up resistor values will depend upon the total bus capacitance and operating speed. The DS90UB903Q/904Q I²C bus data rate supports up to 100 kbps according to I²C specification.

To start any data transfer, the DS90UB903Q/904Q must be configured in the proper I²C mode. Each device can function as an I²C slave proxy or master proxy depending on the mode determined by MODE pin. The Ser/Des interface acts as a virtual bridge between Master Controller Unit (MCU) and the remote device. When the MODE pin is set to High, the device is treated as a slave proxy; acts as a slave on behalf of the remote slave. When addressing a remote peripheral or Serializer/Deserializer (not wired directly to the MCU), the slave proxy will forward any byte transactions sent by the Master controller to the target device. When MODE pin is set to Low, the device will function as a master proxy device; acts as a master on behalf of the I²C master controller. Note that the devices must have complementary settings for the MODE configuration. For example, if the Serializer MODE pin is set to High then the Deserializer MODE pin must be set to Low and vice-versa.

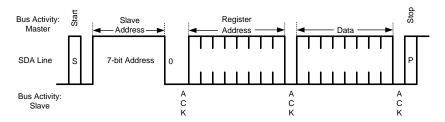


Figure 25. Write Byte

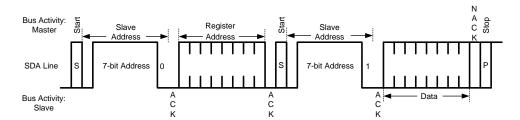


Figure 26. Read Byte

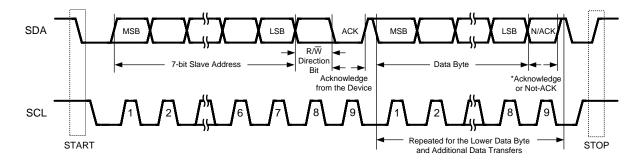


Figure 27. Basic Operation



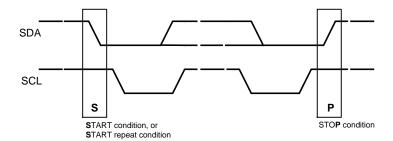


Figure 28. START and STOP Conditions



SLAVE CLOCK STRETCHING

In order to communicate and synchronize with remote devices on the I²C bus through the bidirectional control channel, slave clock stretching must be supported by the I²C master controller/MCU. The chipset utilizes bus clock stretching (holding the SCL line low) during data transmission; where the I²C slave pulls the SCL line low prior to the 9th clock of every I²C data transfer (before the ACK signal). The slave device will not control the clock and only stretches it until the remote peripheral has responded.

Any remote access involves the clock stretching period following the transmitted byte, prior to completion of the acknowledge bit. Since each byte transferred to the I²C slave must be acknowledged separately, the clock stretching will be done for each byte sent by the host controller. For remote accesses, the "Response Delay" shown is on the order of 12 µs (typical). See Application Note AN-2173 (SNLA131) for more details.

ID[X] ADDRESS DECODER

The ID[x] pin is used to decode and set the physical slave address of the Serializer/Deserializer (I 2 C only) to allow up to six devices on the bus using only a single pin. The pin sets one of six possible addresses for each Serializer/Deserializer device. The pin must be pulled to VDD (1.8V, NOT VDDIO)) with a 10 k Ω resistor and a pull down resistor (RID) of the recommended value to set the physical device address. The recommended maximum resistor tolerance is 0.1% worst case (0.2% total tolerance).

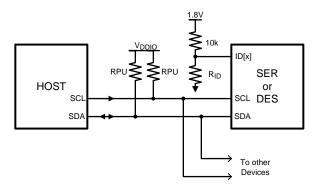


Figure 29. Bidirectional Control Bus Connection

Table 3. ID[x] Resistor Value – DS90UB903Q

ID[x] Resistor Value - DS90UB903Q Ser							
Resistor RID Ω (±0.1%)	Address 7'b ⁽¹⁾	Address 8'b 0 appended (WRITE)					
0, GND	7b' 101 1000 (h'58)	8b' 1011 0000 (h'B0)					
2.0k	7b' 101 1001 (h'59)	8b' 1011 0010 (h'B2)					
4.7k	7b' 101 1010 (h'5A)	8b' 1011 0100 (h'B4)					
8.2k	7b' 101 1011 (h'5B)	8b' 1011 0110 (h'B6)					
12.1k	7b' 101 1100 (h'5C)	8b' 1011 1000 (h'B8)					
39.0k	7b' 101 1110 (h'5E)	8b' 1011 1100 (h'BC)					

⁽¹⁾ Specification is ensured by design.



Table 4. ID[x] Resistor Value – DS90UB9040	Table 4.
--	----------

ID[x] Resistor Value - DS90UB904Q Des							
Resistor RID Ω (±0.1%)	Address 7'b ⁽¹⁾	Address 8'b 0 appended (WRITE)					
0, GND	7b' 110 0000 (h'60)	8b' 1100 0000 (h'C0)					
2.0k	7b' 110 0001 (h'61)	8b' 1100 0010 (h'C2)					
4.7k	7b' 110 0010 (h'62)	8b' 1100 0100 (h'C4)					
8.2k	7b' 110 0011 (h'63)	8b' 1101 0110 (h'C6)					
12.1k	7b' 110 0100 (h'64)	8b' 1101 1000 (h'C8)					
39.0k	7b' 110 0110 (h'66)	8b' 1100 1100 (h'CC)					

⁽¹⁾ Specification is ensured by design.

CAMERA MODE OPERATION

In Camera mode, I²C transactions originate from the Deserializer from the Master controller (Figure 30). The I²C slave core in the Deserializer will detect if a transaction is intended for the Serializer or a slave at the Serializer. Commands are sent over the bidirectional control channel to initiate the transactions. The Serializer will receive the command and generate an I²C transaction on its local I²C bus. At the same time, the Serializer will capture the response on the I²C bus and return the response as a command on the forward channel link. The Deserializer parses the response and passes the appropriate response to the Deserializer I²C bus.

To configure the devices for camera mode operation, set the Serializer MODE pin to Low and the Deserializer MODE pin to High. Before initiating any I²C commands, the Deserializer needs to be programmed with the target slave device addresses and Serializer device address. SER_DEV_ID Register 0x07h sets the Serializer device address and SLAVE_x_MATCH/SLAVE_x_INDEX registers 0x08h~0x17h set the remote target slave addresses. The slave address match registers must also be set. In slave mode the address register is compared with the address byte sent by the I²C master. If the addresses are equal to any of registers values, the I²C slave will acknowledge the transaction to the I²C master allowing reads or writes to target device.

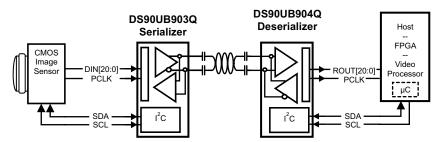


Figure 30. Typical Camera System Diagram

DISPLAY MODE OPERATION

In Display mode, I²C transactions originate from the controller attached to the Serializer. The I²C slave core in the Serializer will detect if a transaction targets (local) registers within the Serializer or the (remote) registers within the Deserializer or a remote slave connected to the I²C master interface of the Deserializer. Commands are sent over the forward channel link to initiate the transactions. The Deserializer will receive the command and generate an I²C transaction on its local I²C bus. At the same time, the Deserializer will capture the response on the I²C bus and return the response as a command on the bidirectional control channel. The Serializer parses the response and passes the appropriate response to the Serializer I²C bus.

The physical device ID of the I^2C slave in the Serializer is determined by the analog voltage on the ID[x] input. It can be reprogrammed by using the SER_DEV_ID register and setting the bit . The device ID of the logical I^2C slave in the Deserializer is determined by programming the DES ID in the Serializer. The state of the ID[x] input on the Deserializer is used to set the device ID. The I^2C transactions between Ser/Des will be bridged between the host to the remote slave.



To configure the devices for display mode operation, set the Serializer MODE pin to High and the Deserializer MODE pin to Low. Before initiating any I²C commands, the Serializer needs to be programmed with the target slave device address and Serializer device address. DES_DEV_ID Register 0x06h sets the Deserializer device address and SLAVE_DEV_ID register 0x7h sets the remote target slave address. If the I²C slave address matches any of registers values, the I²C slave will acknowledge the transaction allowing read or write to target device. Note: In Display mode operation, registers 0x08h~0x17h on Deserializer must be reset to 0x00.

PROGRAMMABLE CONTROLLER

An integrated I²C slave controller is embedded in each of the DS90UB903Q Serializer and DS90UB904Q Deserializer. It must be used to access and program the extra features embedded within the configuration registers. Refer to Table 1 and Table 2 for details of control registers.

I²C PASS THROUGH

I²C pass-through provides an alternative means to independently address slave devices. The mode enables or disables I²C bidirectional control channel communication to the remote I²C bus. This option is used to determine whether or not an I²C instruction is to be transferred over to the remote I²C device. When enabled, the I²C bus traffic will continue to pass through and will be received by I²C devices downstream. If disabled, I²C commands will be excluded to the remote I²C device. The pass through function also provides access and communication to only specific devices on the remote bus. The feature is effective for both Camera mode and Display mode.

SYNCHRONIZING MULTIPLE LINKS

For applications requiring synchronization across multiple links, it is recommended to utilize the General Purpose Input/Output (GPI/GPO) pins to transmit control signals to synchronize slave peripherals together. To synchronize the peripherals properly, the system controller needs to provide a sync signal output. Note this form of synchronization timing relationship has a non-deterministic latency. After the control data is reconstructed from the birectional control channel, there will be a time variation of the GPI/GPO signals arriving at the different target devices (between the parallel links). The maximum latency delta (t1) of the GPI/GPO data transmitted across multiple links is 25 us.

Note: The user must verify that the timing variations between the different links are within their system and timing specifications.

The maximum time (t1) between the rising edge of GPI/GPO (i.e. sync signal) arriving at SER A and SER B is 25 us

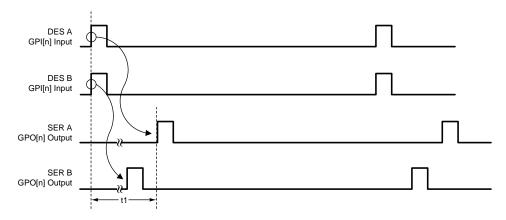


Figure 31. GPI/GPO Delta Latency

GENERAL PURPOSE I/O (GPI/GPO)

The DS90UB903Q/904Q has up to 4 GPO and 4 GPI on the Serializer and Deserializer respectively. The GPI/GPO maximum switching rate is up to 66 kHz for communication between Deserializer GPI to Serializer GPO.



AT-SPEED BIST (BISTEN, PASS)

An optional AT SPEED Built in Self Test (BIST) feature supports at speed testing of the high-speed serial and the bidirectional control channel link. Control pins at the Deserializer are used to enable the BIST test mode and allow the system to initiate the test and set the duration. A HIGH on PASS pin indicates that all payloads received during the test were error free during the BIST duration test. A LOW on this pin at the conclusion of the test indicates that one or more payloads were detected with errors.

The BIST duration is defined by the width of BISTEN. BIST starts when Deserializer LOCK goes HIGH and BISTEN is set HIGH. BIST ends when BISTEN goes LOW. Any errors detected after the BIST Duration are not included in PASS logic.

Note: AT-SPEED BIST is only available in the Camera mode and not the Display mode

The following diagram shows how to perform system AT SPEED BIST:

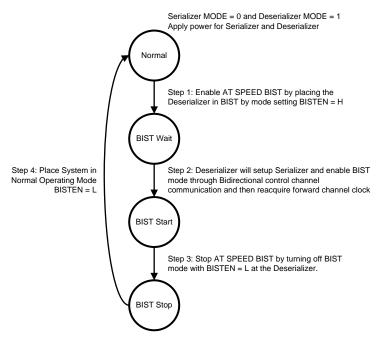


Figure 32. AT-SPEED BIST System Flow Diagram

Step 1: Place the Deserializer in BIST Mode.

Serializer and Deserializer power supply must be supplied. Enable the AT SPEED BIST mode on the Deserializer by setting the BISTEN pin High. The 904 GPI[1:0] pins are used to select the PCLK frequency of the on-chip oscillator for the BIST test on high speed data path.

Table 5. BIST Os	cillator	Frequency	Select
------------------	----------	-----------	--------

Des GPI[1:0]	Oscillator Source	min (MHz)	typ (MHz)	max (MHz)
00	External PCLK	10		43
01	Internal		50	
10	Internal		25	
11	Internal		12.5	



The Deserializer GPI[1:0] set to 00 will bypass the on-chip oscillator and an external oscillator to Serializer PCLK input is required. This allows the user to operate BIST under different frequencies other than the predefined ranges.

Step 2: Enable AT SPEED BIST by placing the Serializer into BIST mode.

Deserializer will communicate through the bidirectional control channel to configure Serializer into BIST mode. Once the BIST mode is set, the Serializer will initiate BIST transmission to the Deserializer.

Wait 10 ms for Deserializer to acquire lock and then monitor the LOCK pin transition from LOW to HIGH. At this point, AT SPEED BIST is operational and the BIST process has begun. The Serializer will start transfer of an internally generated PRBS data pattern through the high speed serial link. This pattern traverses across the interconnecting link to the Deserializer. Check the status of the PASS pin; a HIGH indicates a pass, a LOW indicates a fail. A fail will stay LOW for ½ a clock cycle. If two or more bits in the serial frame fail, the PASS pin will toggle ½ clock cycle HIGH and ½ clock cycle low. The user can use the PASS pin to count the number of fails on the high speed link. In addition, there is a defined SER and DES register that will keep track of the accumulated error count. The Serializer 903 GPO[0] pin will be assigned as a PASS flag error indicator for the bidirectional control channel link.

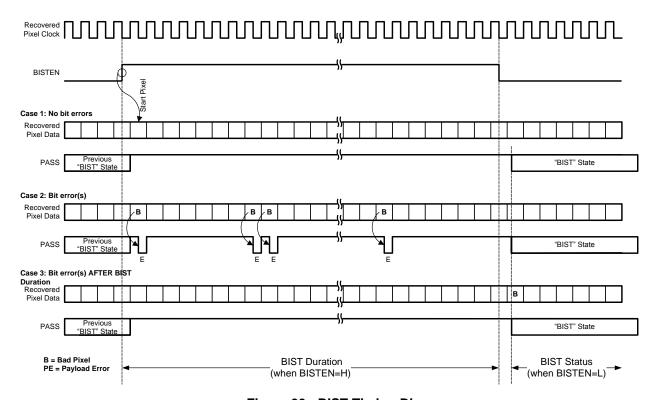


Figure 33. BIST Timing Diagram

Step 3: Stop at SPEED BIST by turning off BIST mode in the Deserializer to determine Pass/Fail.

To end BIST, the system must pull BISTEN pin of the Deserializer LOW. The BIST duration is fully defined by the BISTEN width and Deserializer LOCK is HIGH; thus the Bit Error Rate is determined by how long the system holds BISTEN HIGH.



```
\frac{\text{BIST Duration (s)}}{1 \text{ Pixel period (ns) x Total Bits}} = \text{BIST Duration (s) x } \frac{f_{\text{pixel}} \text{ (MHz)}}{Pixel} \times \text{Total Pixels Transmitted} = \text{Total Bits Transmitted}
\bullet \quad \text{Bit (Pixel) Error Rate} \\ \bullet \quad \text{(for passing BIST)} = \text{[Total Bits Transmitted x Bits/Pixel]}^{-1}
= \text{[Total Bits Transmitted x Bits/Pixel]}^{-1}
```

Figure 34. BIST BER Calculation

Step 4: Place system in Normal Operating Mode by disabling BIST at the Serializer.

Once Step 3 is complete, AT SPEED BIST is over and the Deserializer is out of BIST mode. To fully return to Normal mode, apply Normal input data into the Serializer.

Any PASS result will remain unless it is changed by a new BIST session or cleared by asserting and releasing PDB. The default state of PASS after a PDB toggle is HIGH.

It is important to note that AT SPEED BIST will only determine if there is an issue on the link that is not related to the clock and data recovery of the link (whose status is flagged with LOCK pin).

LVCMOS VDDIO OPTION

1.8V or 3.3V SER Inputs and DES Outputs are user seletable to provide compatibility with 1.8V and 3.3V system interfaces.

REMOTE WAKE UP (Camera Mode)

After initial power up, the Serializer is in a low-power Standby mode. The Deserializer (controlled by ECU/MCU) 'Remote Wake-up' register allows the Deserializer side to generate a signal across the link to remotely wake-up the Serializer. Once the Serializer detects the wake-up signal Serializer switches from Standby mode to active mode. In active mode, the Serializer locks onto PCLK input (if present), otherwise the on-chip oscillator is used as the input clock source. Note the MCU controller should monitor the Deserializer LOCK pin and confirm LOCK = H before performing any I²C communication across the link.

For Remote Wake-up to function properly:

- The chipset needs to be configured in Camera mode: Serializer MODE = 0 and Deserializer MODE = 1
- Serializer expects remote wake-up by default at power on.
- Configure the control channel driver of the Deserializer to be in remote wake-up mode by setting Deserializer Register 0x26h = 0xC0h
- Perform remote wake-up on Serializer by setting Deserializer Register 0x01 b[2] = 1
- Return the control channel driver of the Deserializer to the normal operation mode by setting Deserializer Register 0x26h = 0x00h
- Configure the control channel driver of the Deserializer to be in normal operation mode by setting Deserializer Register 0x27h = 0xC0h.

Serializer can also be put into standby mode by programming the Deserializer remote wake-up control register 0x01 b[2] REM WAKEUP to 0.

POWERDOWN

The SER has a PDB input pin to ENABLE or Powerdown the device. The modes can be controlled by the host and is used to disable the Link to save power when the remote device is not operational. An auto mode is also available. In this mode, the PDB pin is tied High and the SER switches over to an internal oscillator when the PCLK stops or not present. When a PCLK starts again, the SER will then lock to the valid input PCLK and transmits the data to the DES. In powerdown mode, the high-speed driver outputs are static (High).

The DES has a PDB input pin to ENABLE or Powerdown the device. This pin can be controlled by the system and is used to disable the DES to save power. An auto mode is also available. In this mode, the PDB pin is tied High and the DES will enter powerdown when the serial stream stops. When the serial stream starts up again, the DES will lock to the input stream and assert the LOCK pin and output valid data. In powerdown mode, the Data and PCLK outputs are set by the OSS_SEL control register.



POWER UP REQUIREMENTS AND PDB PIN

It is required to delay and release the PDB input signal after VDD (VDDn and VDDIO) power supplies have settled to the recommended operating voltages. A external RC network can be connected to the PDB pin to ensure PDB arrives after all the VDD have stabilized.

SIGNAL QUALITY ENHANCERS

Des - Receiver Input Equalization (EQ)

The receiver inputs provided input equalization filter in order to compensate for loss from the media. The level of equalization is controlled via register setting. Note this function can be observed at the CMLOUTP/N test port enabled via the control registers.

EMI REDUCTION

Des - Receiver Staggered Output

The Receiver staggered outputs allows for outputs to switch in a random distribution of transitions within a defined window. Outputs transitions are distributed randomly. This minimizes the number of outputs switching simultaneously and helps to reduce supply noise. In addition it spreads the noise spectrum out reducing overall EMI.

Des Spread Spectrum Clocking

The DS90UB904Q parallel data and clock outputs have programmable SSCG ranges from 9 kHz–66 kHz and ±0.5%–±2% from 20 MHz to 43 MHz. The modulation rate and modulation frequency variation of output spread is controlled through the SSC control registers.

PIXEL CLOCK EDGE SELECT (TRFB/RRFB)

The TRFB/RRFB selects which edge of the Pixel Clock is used. For the SER, this register determines the edge that the data is latched on. If TRFB register is 1, data is latched on the Rising edge of the PCLK. If TRFB register is 0, data is latched on the Falling edge of the PCLK. For the DES, this register determines the edge that the data is strobed on. If RRFB register is 1, data is strobed on the Rising edge of the PCLK. If RRFB register is 0, data is strobed on the Falling edge of the PCLK.

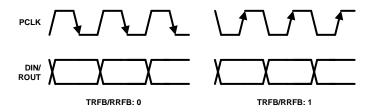


Figure 35. Programmable PCLK Strobe Select



APPLICATIONS INFORMATION

AC COUPLING

The SER/DES supports only AC-coupled interconnects through an integrated DC balanced decoding scheme. External AC coupling capacitors must be placed in series in the FPD-Link III signal path as illustrated in Figure 36.



Figure 36. AC-Coupled Connection

For high-speed FPD-Link III transmissions, the smallest available package should be used for the AC coupling capacitor. This will help minimize degradation of signal quality due to package parasitics. The I/O's require a 100 nF AC coupling capacitors to the line.

TYPICAL APPLICATION CONNECTION

Figure 37 shows a typical connection of the DS90UB903Q Serializer.

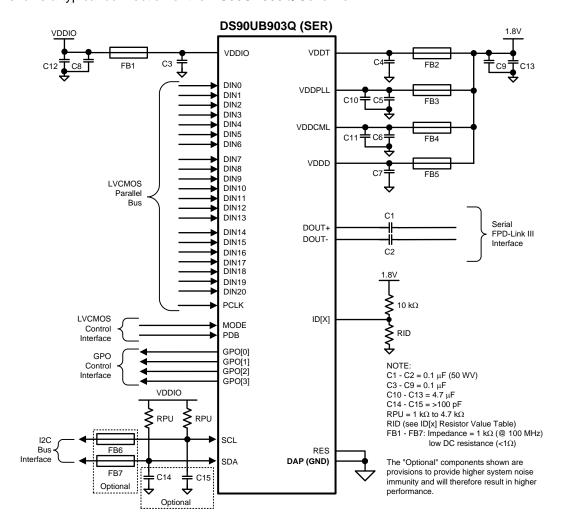


Figure 37. DS90UB903Q Typical Connection Diagram — Pin Control



Figure 38 shows a typical connection of the DS90UB904Q Deserializer.

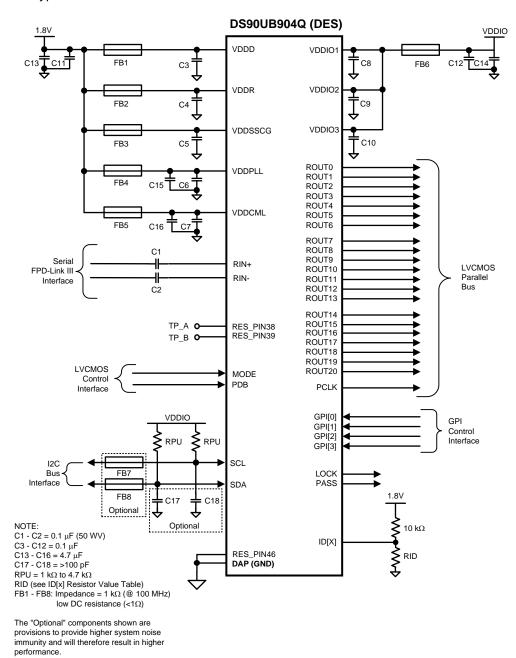


Figure 38. DS90UB904Q Typical Connection Diagram — Pin Control

TRANSMISSION MEDIA

The Ser/Des chipset is intended to be used over a wide variety of balanced cables depending on distance and signal quality requirements. The Ser/Des employ internal termination providing a clean signaling environment. The interconnect for FPD-Link III interface should present a differential impedance of 100 Ohms. Use of cables and connectors that have matched differential impedance will minimize impedance discontinuities. Shielded or un-shielded cables may be used depending upon the noise environment and application requirements. The chipset's optimum cable drive performance is achieved at 43 MHz at 10 meters length. The maximum signaling



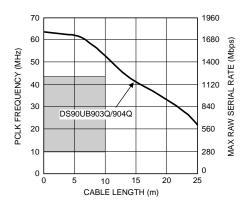
rate increases as the cable length decreases. Therefore, the chipset supports 50 MHz at shorter distances. Other cable parameters that may limit the cable's performance boundaries are: cable attenuation, near-end crosstalk and pair-to-pair skew. The maximum length of cable that can be used is dependant on the quality of the cable (gauge, impedance), connector, board (discontinuities, power plane), the electrical environment (e.g. power stability, ground noise, input clock jitter, PCLK frequency, etc.) and the application environment.

The resulting signal quality at the receiving end of the transmission media may be assessed by monitoring the differential eye opening of the CMLOUT P/N output. A differential probe should be used to measure across the termination resistor at the CMLOUT P/N pins.

For obtaining optimal performance, we recommend:

- · Use Shielded Twisted Pair (STP) cable
- 100Ω differential impedance and 24 AWG (or lower AWG) cable
- · Low skew, impedance matched
- Ground and/or terminate unused conductors

Figure 39 shows the Typical Performance Characteristics demonstrating various lengths and data rates using Rosenberger HSD and Leoni DACAR 538 Cable.



*Note: Equalization is enabled for cable lengths greater than 7 meters

Figure 39. Rosenberger HSD & Leoni DACAR 538 Cable Performance

PCB LAYOUT AND POWER SYSTEM CONSIDERATIONS

Circuit board layout and stack-up for the Ser/Des devices should be designed to provide low-noise power feed to the device. Good layout practice will also separate high frequency or high-level inputs and outputs to minimize unwanted stray noise pickup, feedback and interference. Power system performance may be greatly improved by using thin dielectrics (2 to 4 mils) for power / ground sandwiches. This arrangement provides plane capacitance for the PCB power system with low-inductance parasitics, which has proven especially effective at high frequencies, and makes the value and placement of external bypass capacitors less critical. External bypass capacitors should include both RF ceramic and tantalum electrolytic types. RF capacitors may use values in the range of 0.01 uF to 0.1 uF. Tantalum capacitors may be in the 2.2 uF to 10 uF range. Voltage rating of the tantalum capacitors should be at least 5X the power supply voltage being used.

Surface mount capacitors are recommended due to their smaller parasitics. When using multiple capacitors per supply pin, locate the smaller value closer to the pin. A large bulk capacitor is recommend at the point of power entry. This is typically in the 50uF to 100uF range and will smooth low frequency switching noise. It is recommended to connect power and ground pins directly to the power and ground planes with bypass capacitors connected to the plane with via on both ends of the capacitor. Connecting power or ground pins to an external bypass capacitor will increase the inductance of the path.

A small body size X7R chip capacitor, such as 0603, is recommended for external bypass. Its small body size reduces the parasitic inductance of the capacitor. The user must pay attention to the resonance frequency of these external bypass capacitors, usually in the range of 20-30 MHz. To provide effective bypassing, multiple capacitors are often used to achieve low impedance between the supply rails over the frequency of interest. At high frequency, it is also a common practice to use two vias from power and ground pins to the planes, reducing the impedance at high frequency.



Some devices provide separate power for different portions of the circuit. This is done to isolate switching noise effects between different sections of the circuit. Separate planes on the PCB are typically not required. Pin Description tables typically provide guidance on which circuit blocks are connected to which power pin pairs. In some cases, an external filter many be used to provide clean power to sensitive circuits such as PLLs.

Use at least a four layer board with a power and ground plane. Locate LVCMOS signals away from the differential lines to prevent coupling from the LVCMOS lines to the differential lines. Closely-coupled differential lines of 100 Ohms are typically recommended for differential interconnect. The closely coupled lines help to ensure that coupled noise will appear as common-mode and thus is rejected by the receivers. The tightly coupled lines will also radiate less.

Information on the WQFN style package is provided in Application Note: AN-1187 "Leadless Leadframe Package (LLP) Application Report" (literature number SNOA401).

INTERCONNECT GUIDELINES

See AN-1108 (SNLA008) and AN-905 (SNLA035) for full details.

- Use 100Ω coupled differential pairs
- Use the S/2S/3S rule in spacings
 - S = space between the pair
 - 2S = space between pairs
 - 3S = space to LVCMOS signal
- Minimize the number of Vias
- Use differential connectors when operating above 500Mbps line speed
- Maintain balance of the traces
- · Minimize skew within the pair

Additional general guidance can be found in the LVDS Owner's Manual - available in PDF format from the Texas Instruments web site at: www.ti.com/lyds



Revision History

04/16/2012

- Added CMLOUT P/N in DS90UB904Q Deserializer Pin Descriptions
- Added ESD CDM and ESD MM values
- Added 3.3V I/O VOH conditions: IOH = -4 mA
- Corrected 3.3V I/O VOL conditions: IOL = +4 mA
- Changed NSID DS90UB903/904QSQX to qty 2500
- Added "Only used when VDDIOCONTROL = 0" note for UB904 Register 0x03 bit[4] description
- Added Register 0x27 BCC in UB904 Register table
- Added Register 0x3F CML Output in UB904 Register table
- Updated SLAVE CLOCK STRETCHING in Functional Description section
- Updated REMOTE WAKE UP (Camera Mode) procedure in Functional Description section
- · Updated Des Receiver Input Equalization (EQ) in Functional Description section
- Updated TRANSMISSION MEDIA in Applications Information section

04/16/2013

Changed layout of National Data Sheet to TI format.





10-Dec-2020

PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead finish/ Ball material	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
							(6)				
DS90UB903QSQ/NOPB	ACTIVE	WQFN	RTA	40	1000	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 105	UB903QSQ	Samples
DS90UB903QSQE/NOPB	ACTIVE	WQFN	RTA	40	250	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 105	UB903QSQ	Samples
DS90UB903QSQX/NOPB	ACTIVE	WQFN	RTA	40	2500	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 105	UB903QSQ	Samples
DS90UB904QSQ/NOPB	ACTIVE	WQFN	RHS	48	1000	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 105	UB904QSQ	Samples
DS90UB904QSQE/NOPB	ACTIVE	WQFN	RHS	48	250	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 105	UB904QSQ	Samples
DS90UB904QSQX/NOPB	ACTIVE	WQFN	RHS	48	2500	RoHS & Green	SN	Level-3-260C-168 HR	-40 to 105	UB904QSQ	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.



PACKAGE OPTION ADDENDUM

10-Dec-2020

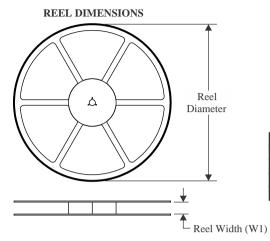
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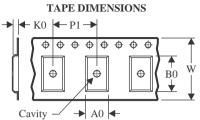
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PACKAGE MATERIALS INFORMATION

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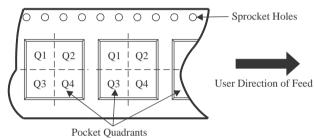
TAPE AND REEL INFORMATION





A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

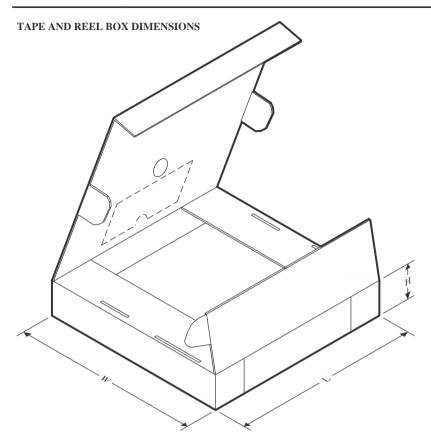


*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DS90UB903QSQ/NOPB	WQFN	RTA	40	1000	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q1
DS90UB903QSQE/NOPB	WQFN	RTA	40	250	178.0	16.4	6.3	6.3	1.5	12.0	16.0	Q1
DS90UB903QSQX/NOPB	WQFN	RTA	40	2500	330.0	16.4	6.3	6.3	1.5	12.0	16.0	Q1
DS90UB904QSQ/NOPB	WQFN	RHS	48	1000	330.0	16.4	7.3	7.3	1.3	12.0	16.0	Q1
DS90UB904QSQE/NOPB	WQFN	RHS	48	250	178.0	16.4	7.3	7.3	1.3	12.0	16.0	Q1
DS90UB904QSQX/NOPB	WQFN	RHS	48	2500	330.0	16.4	7.3	7.3	1.3	12.0	16.0	Q1



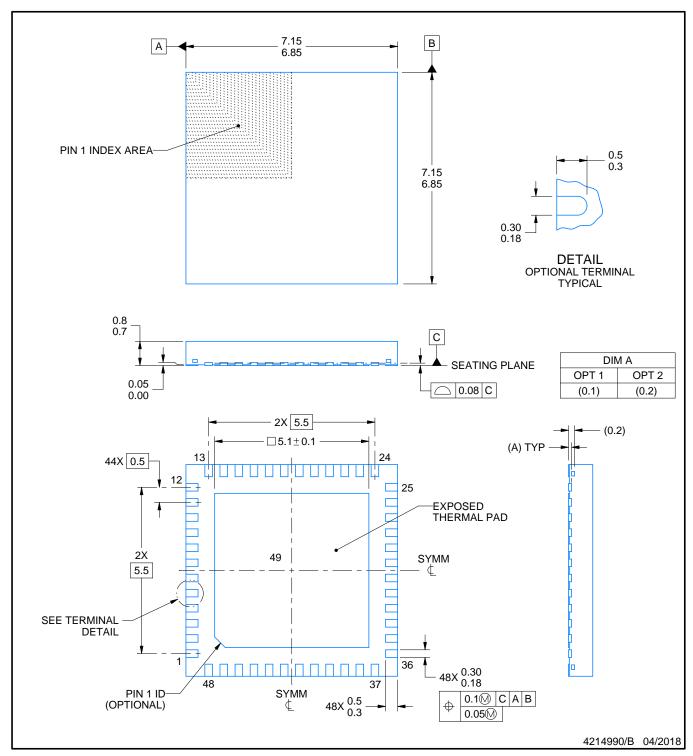
www.ti.com 9-Aug-2022



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DS90UB903QSQ/NOPB	WQFN	RTA	40	1000	356.0	356.0	35.0
DS90UB903QSQE/NOPB	WQFN	RTA	40	250	208.0	191.0	35.0
DS90UB903QSQX/NOPB	WQFN	RTA	40	2500	356.0	356.0	35.0
DS90UB904QSQ/NOPB	WQFN	RHS	48	1000	356.0	356.0	35.0
DS90UB904QSQE/NOPB	WQFN	RHS	48	250	208.0	191.0	35.0
DS90UB904QSQX/NOPB	WQFN	RHS	48	2500	356.0	356.0	35.0

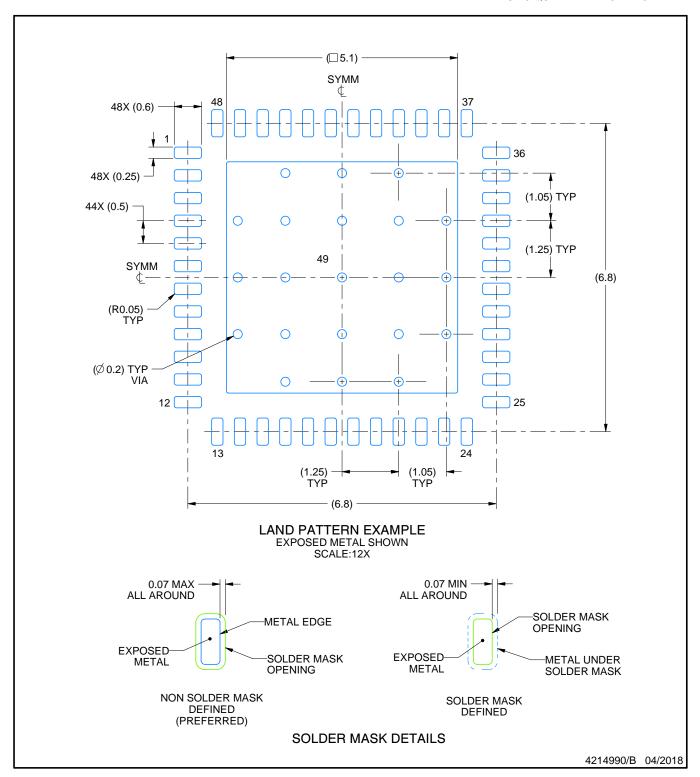




NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

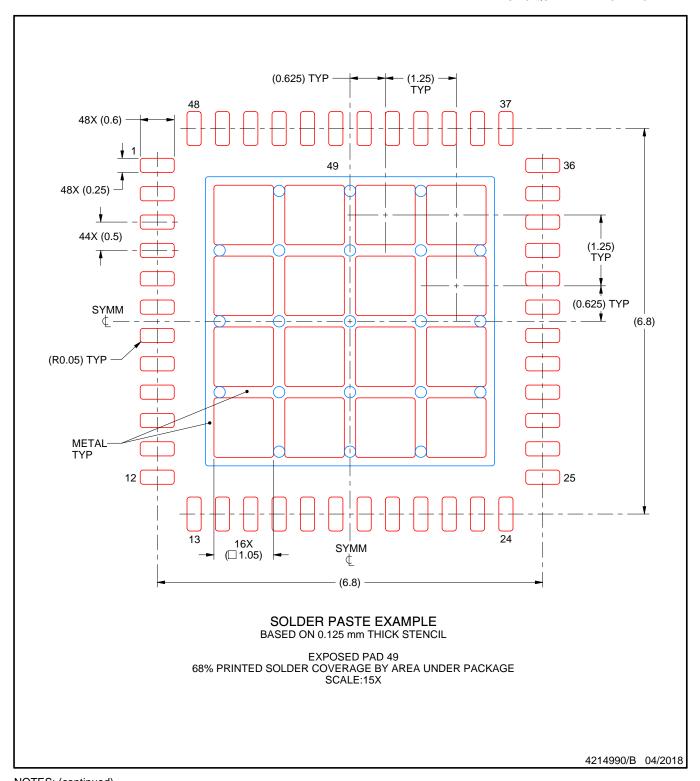




NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



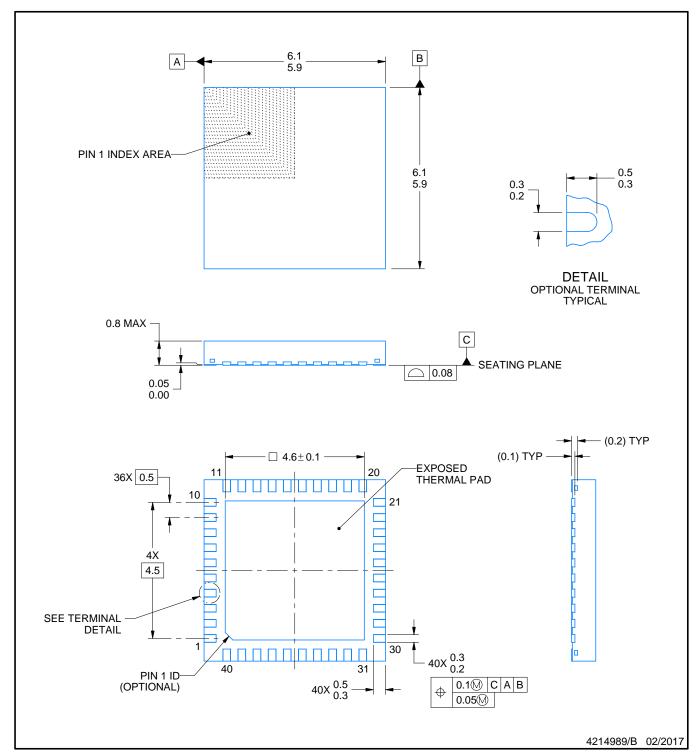


NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



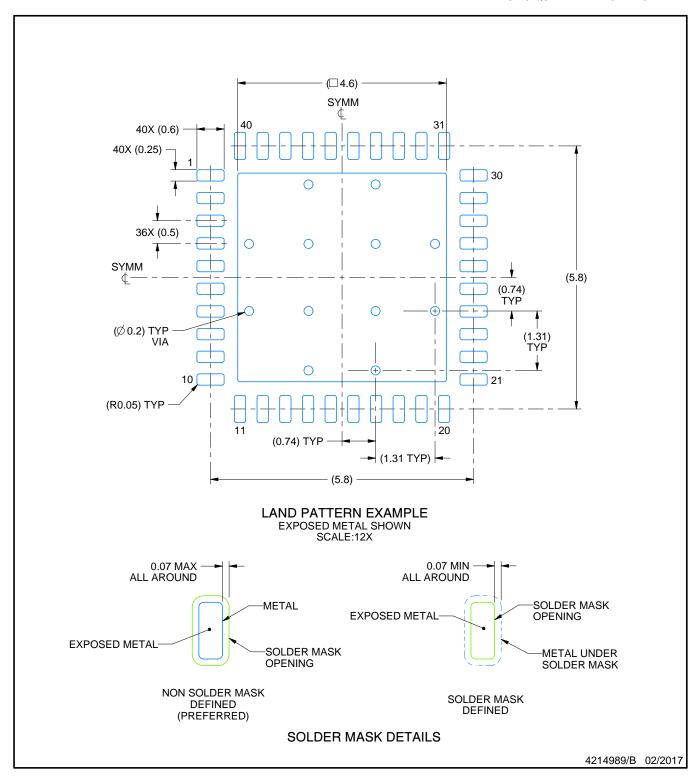




NOTES:

- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
 2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

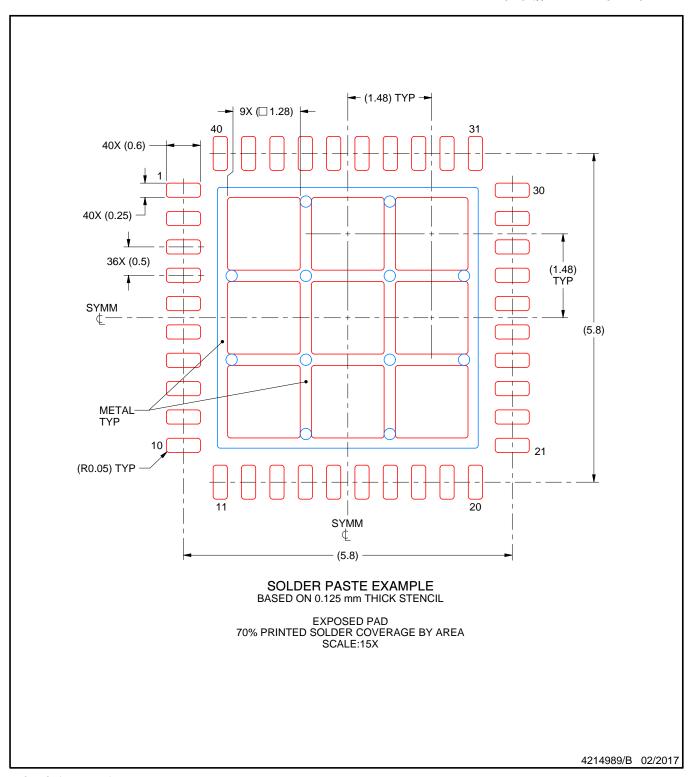




NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- 5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.





NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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