具有数字电缆补偿功能的 TPS25810A-Q1 USB Type-C DFP 控制器和电源开关

1 特性

- 符合汽车应用 要求
- 具有符合 AEC-Q100 标准的下列特性：
  - 器件温度等级 T：环境工作温度范围为 –40°C 至 105°C
  - 器件人体模型 (HBM) 静电放电 (ESD) 分类等级 2
  - 器件组件充电模式 (CDM) ESD 分类等级 C4B
- 兼容 USB Type-C 版本 1.2 的下行数据端口 (DFP) 控制器
- 连接器连接或断开的检测
- 配置通道 (CC) STD、1.5A、3A 电流能力通告
- 超高速极性的确定
- V_{BUS} 应用和放电
- V_{CONN} 应用于电子标记电缆
- 音频和调试附件的识别
- 端口未连接时，I_{DDQ} 的典型值为 0.7μA
- 三个输入电源选项
  - IN1：USB 充电电源
  - IN2：V_{CONN} 电源
  - AUX：器件电源
- 电源唤醒可保证系统冬眠 (S4) 和关闭 (S5) 功耗状态下的低功耗
- 34mΩ（典型值）高侧金属氧化物半导体场效应晶体管 (MOSFET)
- 固定 3.4A I_{LIM} (±7.1%)
- 数字电缆补偿，I_{OUT} ≥ 1.95A
- 封装：20 引脚 WQFN (3mm × 4mm) (1)

2 应用

- 汽车信息娱乐系统
- 汽车后座 USB 充电

3 说明

TPS25810A-Q1 器件一款 USB Type-C 下行端口 (DFP) 控制器，集成了一个额定电流为 3A 的 USB 电源开关。此器件通过监测 Type-C 配置通道 (CC) 线路来发现连接的 USB 设备。如果连接了上行端口 (UFP) 器件，它会向 V_{BUS} 供电，并将可选 V_{BUS} 拉电流能力通过直通 CC 线路传达给 UFP。如果使用电子标记电缆连接了 UFP，它还会将 V_{CONN} 电源施加于电缆 CC 引脚。当连接 Type-C 音频附件或调试附件时，TPS25810A-Q1 可以识别并报告此连接。

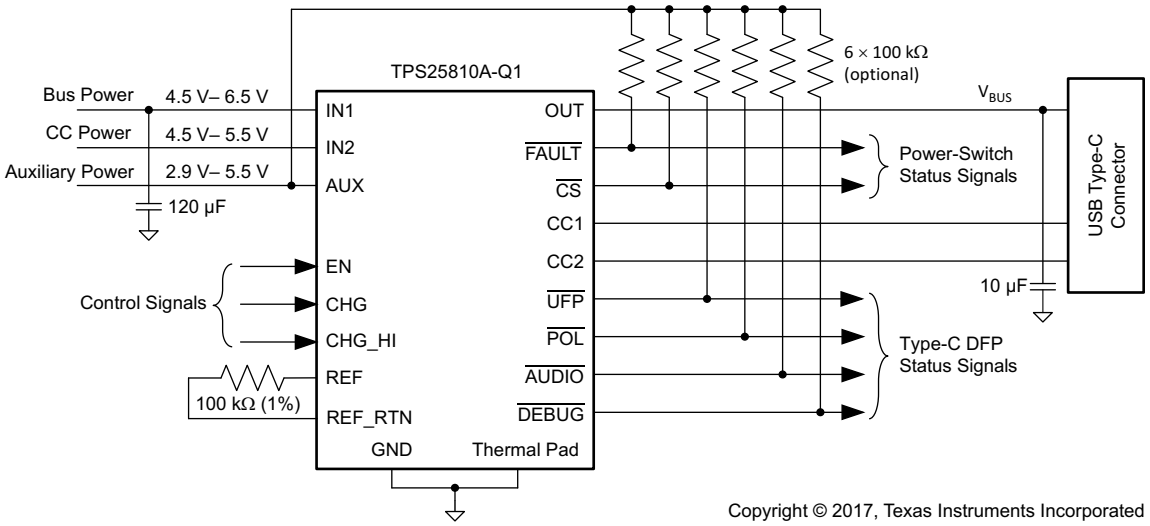
器件信息(1)

器件型号	封装	封装尺寸（标称值）
TPS25810A-Q1	超薄四方扁平无引线 (WQFN) (20)	3.00mm x 4.00mm

(1) 要了解所有可用封装，请参阅数据表末尾的可订购产品附录。

(1) CC 引脚符合 IEC-61000-4-2 标准

简化原理图



Copyright © 2017, Texas Instruments Incorporated

目录

1 特性 ... 1
2 应用 ... 1
3 说明 ... 1
4 修订历史记录 ... 2
5 说明 （续） ... 3
6 **Pin Configuration and Functions** ... 4
7 **Specifications** ... 5
7.1 Absolute Maximum Ratings ... 5
7.2 ESD Ratings ... 5
7.3 Recommended Operating Conditions ... 5
7.4 Thermal Information ... 6
7.5 Electrical Characteristics ... 6
7.6 Switching Characteristics ... 8
7.7 Typical Characteristics ... 10
8 **Detailed Description** ... 12
8.1 Overview ... 12
8.2 Functional Block Diagram ... 14
8.3 Feature Description ... 14
8.4 Device Functional Modes ... 22
9 **Application and Implementation** ... 24
9.1 Application Information ... 24
9.2 Typical Applications ... 24
10 **Power Supply Recommendations** ... 29
11 **Layout** ... 30
11.1 Layout Guidelines ... 30
11.2 Layout Example ... 31
12 器件和文档支持 ... 32
12.1 器件支持 ... 32
12.2 文档支持 ... 32
12.3 接收文档更新通知 ... 32
12.4 社区资源 ... 32
12.5 商标 ... 32
12.6 静电放电警告 ... 32
12.7 Glossary ... 32
13 机械、封装和可订购信息 ... 32

4 修订历史记录

日期	修订版本	注
2017 年 4 月	*	初始发行版

版权 © 2017, Texas Instruments Incorporated

5 说明（续）

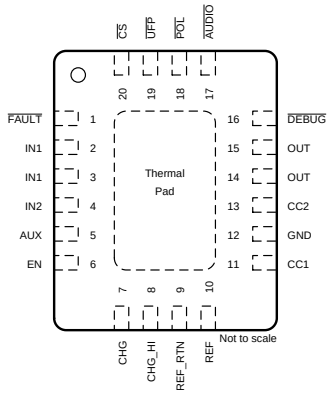
未连接 USB 负载时，TPS25810A-Q1 器件从 AUX 引脚上消耗的电流小于 0.7μA（典型值）。未连接 UFP 时，此器件通过使用 $\overline{\text{UFP}}$ 输出对 5V 高功率电源进行禁用，从而可在 S4 和 S5 系统功耗状态下进一步实现系统节能。在此模式下，该器件能够由电压较低 (3.3V) 的辅助电源 (AUX) 供电运行，该电源通常在低功耗状态（S4 和 S5）下为系统微控制器供电。

TPS25810A-Q1 器件集成了一个 34mΩ 电源开关，且无论 Type-C 电流通告级别为何，均具有固定的 3.4A 电流限制。$\overline{\text{FAULT}}$ 输出在开关处于过流和过热条件时发出信号。$\overline{\text{CS}}$ 输出用于实现负载电流大于 1.95A 的数字电缆补偿。电缆补偿也称为线路压降补偿，是将 USB 电源的电压降补偿到 UFP 负载的一种手段。

Copyright © 2017, Texas Instruments Incorporated

6 Pin Configuration and Functions

TPS25810A-Q1 RVC Package
20-Pin WQFN With Exposed Thermal Pad
Top View



Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
$\overline{\text{AUDIO}}$	17	O	Open-drain logic output that asserts when a Type-C audio accessory is identified on the CC lines
AUX	5	I	Auxiliary input supply. Connect to an always-alive system rail to use the power-wake feature. Short to IN1 and IN2 if only one supply is used.
CC1	11	I/O	Analog input/output that connects to the Type-C receptacle CC1 pin
CC2	13	I/O	Analog input/output that connects to the Type-C receptacle CC2 pin.
CHG	7	I	Charge-logic input to select between standard USB (500 mA for a Type-C receptacle supporting only USB 2.0, and 900 mA for Type-C receptacle supporting USB 3.1) or a Type-C current-sourcing ability.
CHG_HI	8	I	High-charge logic input to select between 1.5-A and 3-A Type-C current sourcing capability. Valid when CHG is set to Type-C current.
$\overline{\text{CS}}$	20	O	Open-drain output enabling digital cable compensation when load current is greater than 1.95 A, nominal.
$\overline{\text{DEBUG}}$	16	O	Open-drain logic output that asserts when a Type-C debug accessory is identified on the CC lines
EN	6	I	Enable logic input. Turns the device on and off
$\overline{\text{FAULT}}$	1	O	Fault event indicator. Open-drain logic output that asserts low to indicate a current-limit or thermal-shutdown event due to overtemperature.
GND	12	—	Power ground
IN1	2, 3	I	V_{BUS} input supply. Internal power switch connects IN1 to OUT.
IN2	4	I	V_{CONN} input supply. Internal power switch connects IN2 to CC1 or CC2. Short to IN1 if only one supply is used.
OUT	14, 15	O	Power switch output
$\overline{\text{POL}}$	18	O	Polarity open-drain logic output that signals which Type-C CC pin is connected to the CC line. This gives the information needed to multiplex the super-speed lines. Asserted when the CC2 pin is connected to the CC line in the cable.
REF	10	I	Analog input used to generate the internal current reference. Connect a 1% or better, 100-ppm, 100-kΩ resistor between this pin and REF_RTN.
REF_RTN	9	I	Precision signal-reference return. Connect to the REF pin via a 100-kΩ, 1% resistor.
$\overline{\text{UFP}}$	19	O	Open-drain logic output that asserts when a Type-C UFP is identified on the CC lines.
Thermal pad	—	—	Thermal pad on the bottom of the package. The thermal pad is internally connected to GND and is used to heat-sink the device to the circuit board. Connect the thermal pad to the GND plane.

Copyright © 2017, Texas Instruments Incorporated

7 Specifications

7.1 Absolute Maximum Ratings

over operating ambient temperature range, voltages are with respect to GND (unless otherwise noted) (1)

		MIN	MAX	UNIT
Pin voltage, V	$\overline{AUDIO}$, AUX, CC1, CC2, CHG, CHG_HI, $\overline{CS}$, $\overline{DEBUG}$, EN, $\overline{FAULT}$, IN1, IN2, OUT, $\overline{POL}$, REF, $\overline{UFP}$,	–0.3	7	V
	REF_RTN		Internally connected to GND	V
Pin positive source current, I_{SRC}	CC1, CC2, OUT, REF		Internally limited	A
Pin positive sink current, I_{SNK}	OUT (while applying V_{BUS})		5	A
	CC1, CC2 (while applying V_{CONN})		1	A
	$\overline{AUDIO}$, $\overline{CS}$, $\overline{DEBUG}$, $\overline{FAULT}$, $\overline{POL}$, $\overline{UFP}$		Internally limited	mA
Operating junction temperature, T_J		–40	180	°C
Storage temperature range, T_{stg}		–65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

7.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$ (1)	Electrostatic discharge	Human-body model (HBM), per per AEC Q100-002 (2)	±2 000	V
		Charged-device model (CDM), per per AEC Q100-011	±500	
		61000-4-2 contact discharge, CC1 and CC2 (3) IEC	±8 000	
		IEC 61000-4-2 air discharge, CC1 and CC2 (3)	±15 000	

(1) Electrostatic discharge (ESD) to measure device sensitivity and immunity to damage caused by assembly line electrostatic discharges into the device.
(2) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.
(3) Surges per IEC61000-402, 1999 applied between CC1, CC2 and output ground of the TPS25810EVM-745.

7.3 Recommended Operating Conditions

Voltages are with respect to GND (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V_{IN}	Supply voltage	IN1	4.5		6.5	V
		IN2	4.5		5.5	
		AUX	2.9		5.5	
V_I	Input voltage	CHG, CHG_HI, EN	0		5.5	V
V_{IH}	High-level input voltage	CHG, CHG_HI, EN	1.17			V
V_{IL}	Low-level voltage	CHG, CHG_HI, EN			0.63	V
V_{PU}	Pullup voltage	Used on $\overline{AUDIO}$, $\overline{CS}$, $\overline{DEBUG}$, $\overline{FAULT}$, $\overline{POL}$, $\overline{UFP}$,	0		5.5	V
I_{SRC}	Positive source current	OUT			3	A
		CC1 or CC2 when supplying V_{CONN}			250	mA
I_{SNK}	Positive sink current (10 ms moving average)	$\overline{AUDIO}$, $\overline{CS}$, $\overline{DEBUG}$, $\overline{FAULT}$, $\overline{POL}$, $\overline{UFP}$			10	mA
I_{SNK_PULSE}	Positive repetitive pulse sink current	$\overline{AUDIO}$, $\overline{CS}$, $\overline{DEBUG}$, $\overline{FAULT}$, $\overline{POL}$, $\overline{UFP}$			Internally limited	mA
R_{REF}	Reference resistor		98	100	102	kΩ
T_J	Operating junction temperature		–40		125	°C

Copyright © 2017, Texas Instruments Incorporated

7.4 Thermal Information

THERMAL METRIC(1)		TPS25810A-Q1 RVC (WQFN) 20 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	39.3	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	43.4	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	13	°C/W
ψ_{JT}	Junction-to-top characterization parameter	0.7	°C/W
ψ_{JB}	Junction-to-board characterization parameter	13	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	4.2	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, SPRA953.

7.5 Electrical Characteristics

–40°C ≤ T_J ≤ 125°C, 4.5 V ≤ V_{IN1} ≤ 6.5 V, 4.5 V ≤ V_{IN2} ≤ 5.5 V, 2.9 V ≤ V_{AUX} ≤ 5.5 V; V_{EN} = V_{CHG} = V_{CHG_HI} = V_{AUX}, R_{REF} = 100 kΩ. Typical values are at 25°C. All voltages are with respect to GND. I_{OUT} and I_{OS} defined as positive out of the indicated pin (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OUT – POWER SWITCH						
$r_{DS(on)}$	On-resistance(1)	T_J = 25°C, I_{OUT} = 3 A		34	37	mΩ
		–40°C ≤ T_J ≤ 85°C, I_{OUT} = 3 A		34	46	
		–40°C ≤ T_J ≤ 125°C, I_{OUT} = 3 A		34	55	
I_{REV}	OUT to IN reverse leakage current	V_{OUT} = 6.5 V, V_{IN1} = V_{EN} = 0 V, –40°C ≤ T_J ≤ 85°C, I_{REV} is current out of IN1 pin		0	3	µA
OUT – CURRENT LIMIT						
I_{OS}	Short-circuit current limit (1)		3.16	3.4	3.64	A
		R_{REF} = 10 Ω			7	
OUT – DISCHARGE						
	Discharge resistance	V_{OUT} = 4 V, UFP signature removed from CC lines, time < t_{w_DCHG}	400	500	600	Ω
	Bleed discharge resistance	V_{OUT} = 4 V, No UFP signature on CC lines, time > t_{w_DCHG}	100	150	250	kΩ
REF						
V_O	Output voltage		0.78	0.8	0.82	V
I_{OS}	Short circuit current	R_{REF} = 10 Ω	9.5		15.3	µA
$\overline{FAULT}$						
V_{OL}	Output low voltage	$I_{\overline{FAULT}}$ = 1 mA			350	mV
I_{OFF}	Off-state leakage	$V_{\overline{FAULT}}$ = 5.5 V			1	µA
$\overline{CS}$						
V_{OL}	Output low voltage	$I_{\overline{CS}}$ = 1 mA			350	mV
I_{OFF}	Off-state leakage	$V_{\overline{CS}}$ = 5.5 V			1	µA
I_{TH}	OUT sourcing, rising threshold current for load detect		1.8	1.95	2.1	A
	Hysteresis(2)			125		mA

(1) Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.
(2) These parameters are provided for reference only and do not constitute part of TI's published specifications for purposes of TI's product warranty.

Copyright © 2017, Texas Instruments Incorporated

Electrical Characteristics (continued)

–40°C ≤ T_J ≤ 125°C, 4.5 V ≤ V_{IN1} ≤ 6.5 V, 4.5 V ≤ V_{IN2} ≤ 5.5 V, 2.9 V ≤ V_{AUX} ≤ 5.5 V; V_{EN} = V_{CHG} = V_{CHG_HI} = V_{AUX}, R_{REF} = 100 kΩ. Typical values are at 25°C. All voltages are with respect to GND. I_{OUT} and I_{OS} defined as positive out of the indicated pin (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
CC1, CC2 – V_{CONN} POWER SWITCH						
$r_{DS(on)}$	On-resistance	T_J = 25°C, I_{OUT} = 250 mA		365	420	mΩ
		–40°C ≤ T_J ≤ 85°C, I_{OUT} = 250 mA		365	530	
		–40°C ≤ T_J ≤ 125°C, I_{OUT} = 250 mA		365	600	
CC1, CC2 – V_{CONN} POWER SWITCH – CURRENT LIMIT						
I_{OS}	Short-circuit current limit(1)		300	355	410	mA
		R_{REF} = 10 Ω			800	
CC1, CC2 – CONNECT MANAGEMENT – DANGLING ELECTRONICALLY MARKED CABLE MODE						
I_{SRC}	Sourcing current on the pass-through CC Line	0 V ≤ V_{CCx} ≤ 1.5 V	64	80	96	µA
	Sourcing current on the Ra CC line	0 V ≤ V_{CCx} ≤ 1.5 V	64	80	96	
CC1, CC2 – CONNECT MANAGEMENT – ACCESSORY MODE						
I_{SRC}	CCx sourcing current (CC2 – audio, CC1-debug)	0 V ≤ V_{CCx} ≤ 1.5 V	64	80	96	µA
	CCx sourcing current (CC1 – audio, CC2-debug) (2)	0 V ≤ V_{CCx} ≤ 1.5 V		0		
CC1, CC2 – CONNECT MANAGEMENT – UFP MODE						
I_{SRC}	Sourcing current with either IN1 or IN2 in UVLO	0 V ≤ V_{CCx} ≤ 1.5 V V_{IN1} < $V_{TH_UVLO_IN1}$ or V_{IN2} < $V_{TH_UVLO_IN2}$	64	80	96	µA
I_{SRC}	Sourcing current	V_{CHG} = 0 V and V_{CHG_HI} = 0 V 0 V ≤ V_{CCx} ≤ 1.5 V	75	80	85	µA
		V_{CHG} = V_{AUX} and V_{CHG_HI} = 0 V 0 V ≤ V_{CCx} ≤ 1.5 V	170	180	190	
		V_{CHG} = V_{AUX} and V_{CHG_HI} = V_{AUX} 0 V ≤ V_{CCx} ≤ 2.45 V	312	330	348	
$\overline{UFP}$, $\overline{POL}$, $\overline{AUDIO}$, $\overline{DEBUG}$						
V_{OL}	Output low voltage	I_{SNK_PIN} = 1 mA			250	mV
I_{OFF}	Off-state leakage	V_{PIN} = 5.5 V			1	µA
EN, CHG, CHG_HI – LOGIC INPUTS						
V_{TH}	Rising threshold voltage			0.925	1.15	V
V_{TH}	Falling threshold voltage		0.65	0.875		V
	Hysteresis(2)			50		mV
I_{IN}	Input current	V_{EN} = 0 V or 6.5 V	–0.5		0.5	µA
OVERTEMPERATURE SHUTDOWN						
T_{TH_OTSD2}	Rising threshold temperature for device shutdown		155			°C
	Hysteresis(2)			20		°C
T_{TH_OTSD1}	Rising threshold temperature for OUT/ V_{CONN} switch shutdown in current limit		135			°C
	Hysteresis(2)			20		°C
IN1						
$V_{TH_UVLO_IN1}$	Rising threshold voltage for UVLO		3.9	4.1	4.3	V
	Hysteresis(2)			100		mV
$I_{IN1(DIS)}$	Disabled supply current	V_{EN} = 0 V, –40°C ≤ T_J ≤ 85°C			1	µA
$I_{IN1(CC_OPEN)}$	Enabled supply current with CC lines open	–40°C ≤ T_J ≤ 85°C			1	µA

Copyright © 2017, Texas Instruments Incorporated

Electrical Characteristics (continued)

–40°C ≤ T_J ≤ 125°C, 4.5 V ≤ V_{IN1} ≤ 6.5 V, 4.5 V ≤ V_{IN2} ≤ 5.5 V, 2.9 V ≤ V_{AUX} ≤ 5.5 V; V_{EN} = V_{CHG} = V_{CHG_HI} = V_{AUX}, R_{REF} = 100 kΩ. Typical values are at 25°C. All voltages are with respect to GND. I_{OUT} and I_{OS} defined as positive out of the indicated pin (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{IN1(Ra)}$	Enabled supply current with accessory or dangling electronically marked cable signature on CC lines				2	µA
$I_{IN1(Rd)}$	Enabled supply current with UFP attached	V_{CHG} = 0 V, or V_{CHG} = V_{AUX} and V_{CHG_HI} = 0 V		75	100	µA
				85	110	
IN2						
$V_{TH_UVLO_IN2}$	Rising threshold voltage for UVLO		3.9	4.1	4.3	V
	Hysteresis(2)			100		mV
$I_{IN2(DIS)}$	Disabled supply current	V_{EN} = 0 V, –40°C ≤ T_J ≤ 85°C			1	µA
$I_{IN2(CC_OPEN)}$	Enabled supply current with CC lines open	–40°C ≤ T_J ≤ 85°C			1	µA
$I_{IN2(Ra)}$	Enabled supply current with accessory or dangling electronically marked cable signature on CC lines				2	µA
$I_{IN2(Rd)}$	Enabled supply current with UFP signature on CC lines (Includes IN current that provides the CC output current to the UFP Rd resistor)	V_{CHG} = 0 V, 0 V ≤ V_{CCx} ≤ 1.5 V		98	110	µA
		V_{CHG} = V_{IN} and V_{CHG_HI} = 0 V, 0 V ≤ V_{CCx} ≤ 1.5 V		198	215	
		0 V ≤ V_{CCx} ≤ 2.45 V		348	373	
AUX						
$V_{TH_UVLO_AUX}$	Rising threshold voltage for UVLO		2.65	2.75	2.85	V
	Hysteresis(2)			100		mV
$I_{AUX(DIS)}$	Disabled supply current	V_{EN} = 0 V, –40°C ≤ T_J ≤ 85°C			1	µA
$I_{AUX(CC_OPEN)}$	Enabled internal supply current with CC lines open	–40°C ≤ T_J ≤ 85°C		0.7	3	µA
$I_{AUX(Ra)}$	Enabled supply current with accessory or dangling active cable signature on CC lines			140	185	µA
$I_{AUX(Rd_noIN)}$	Enabled supply current with UFP termination on CC lines and with either IN1 or IN2 in UVLO	V_{IN1} < $V_{TH_UVLO_IN1}$ or V_{IN2} < $V_{TH_UVLO_IN2}$		145	190	µA
$I_{AUX(Rd)}$	Enabled supply current with UFP termination on CC lines			55	82	µA

7.6 Switching Characteristics

–40°C ≤ T_J ≤ 125°C, 4.5 V ≤ V_{IN1} ≤ 6.5 V, 4.5 V ≤ V_{IN2} ≤ 5.5 V, 2.9 V ≤ V_{AUX} ≤ 5.5 V; V_{EN} = V_{CHG} = V_{CHG_HI} = V_{AUX}, R_{REF} = 100 kΩ. Typical values are at 25°C. All voltages are with respect to GND. I_{OUT} and I_{OS} defined as positive out of the indicated pin (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
OUT – POWER SWITCH						
t_r	Output-voltage rise time	V_{IN1} = 5 V, C_L = 1 µF, R_L = 100 Ω (measured from 10% to 90% of final value)	1.2	1.8	2.5	ms
t_f	Output-voltage fall time		0.35	0.55	0.75	ms
t_{on}	Output-voltage turnon time	V_{IN1} = 5 V, C_L = 1 µF, R_L = 100 Ω	2.5	3.5	5	ms
t_{off}	Output-voltage turnoff time		2	3	4.5	ms
OUT – CURRENT LIMIT						
t_{ios}	Current-limit response time to short circuit	V_{IN1} – V_{OUT} = 1 V, R_L = 10 mΩ, see Figure 1		1.5	4	µs

Copyright © 2017, Texas Instruments Incorporated

Switching Characteristics (continued)

–40°C ≤ T_J ≤ 125°C, 4.5 V ≤ V_{IN1} ≤ 6.5 V, 4.5 V ≤ V_{IN2} ≤ 5.5 V, 2.9 V ≤ V_{AUX} ≤ 5.5 V; V_{EN} = V_{CHG} = V_{CHG_HI} = V_{AUX}, R_{REF} = 100 kΩ. Typical values are at 25°C. All voltages are with respect to GND. I_{OUT} and I_{OS} defined as positive out of the indicated pin (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$\overline{\text{FAULT}}$						
t_{DEGA}	Asserting deglitch time due to overcurrent		5.5	8.2	10.7	ms
$t_{DEGA(OC)}$	Asserting deglitch time due to overtemperature in current limit(1)			0		ms
$t_{DEGA(OT)}$	Deasserting deglitch time		5.5	8.2	10.7	ms
$\overline{\text{CS}}$						
t_{DEGA}	Asserting deglitch time		5.5	8.2	10.7	ms
t_{DEGD}	Deasserting deglitch time		5.5	8.2	10.7	ms
OUT – DISCHARGE						
	R_{DCHG} discharge time	V_{OUT} = 1 V, time I_{SNK_OUT} > 1 mA after UFP signature removed from CC lines	39	65	96	ms
CC1, CC2 - V_{CONN} POWER SWITCH						
t_r	Output-voltage rise time	V_{IN2} = 5 V, C_L = 1 µF, R_L = 100 Ω (measured from 10% to 90% of final value)	0.15	0.25	0.35	ms
t_f	Output-voltage fall time		0.18	0.22	0.26	ms
t_{on}	Output-voltage turnon time	V_{IN2} = 5 V, C_L = 1 µF, R_L = 100 Ω	1	1.5	2	ms
t_{off}	Output-voltage turnoff time		0.3	0.4	0.55	ms
CC1, CC2 – V_{CONN} POWER SWITCH – CURRENT LIMIT						
t_{res}	Current-limit response time to short circuit	V_{IN2} – V_{CONN} = 1 V, R = 10 mΩ, see Figure 1		1	3	µs
$\overline{\text{UFP}}$, $\overline{\text{POL}}$, $\overline{\text{AUDIO}}$, $\overline{\text{DEBUG}}$						
t_{DEGR}	Asserting deglitch time		100	150	200	ms
t_{DEGF}	Deasserting deglitch time		7.9	12.5	17.7	ms

(1) These parameters are provided for reference only and do not constitute part of TI's published specifications for purposes of TI's product warranty.

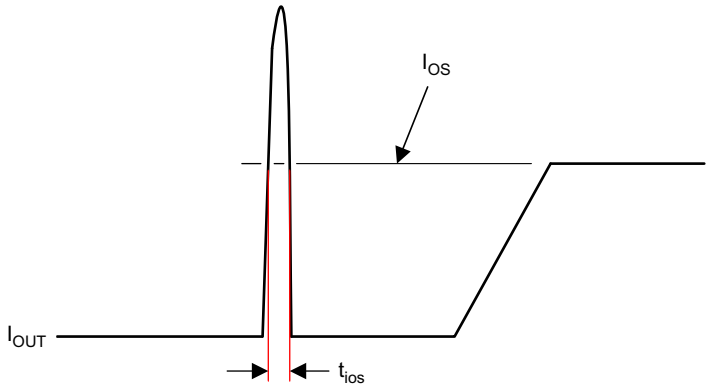


Figure 1. Output Short-Circuit Timing Diagram

Copyright © 2017, Texas Instruments Incorporated

7.7 Typical Characteristics

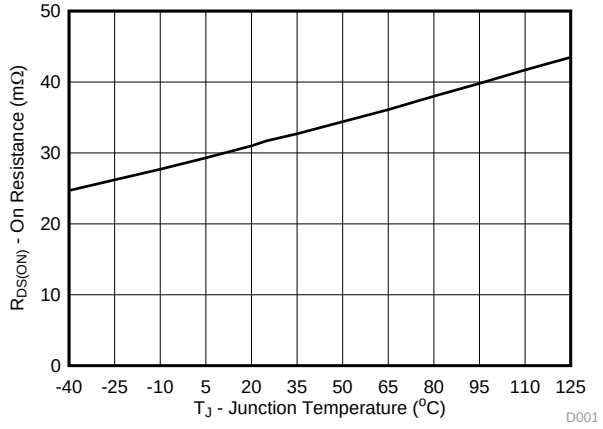


Figure 2. V_{BUS} Current-Limiting Switch On-Resistance vs Temperature

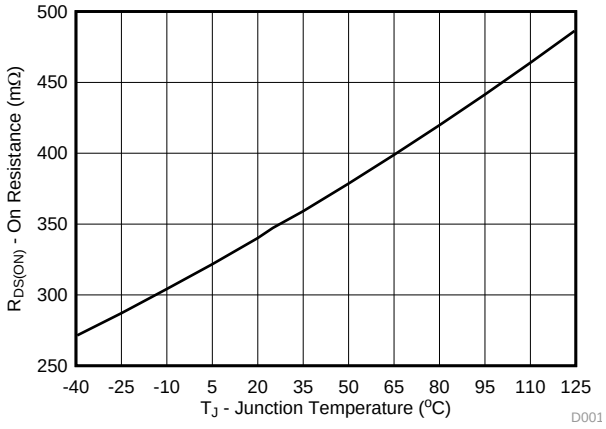


Figure 3. V_{CONN} Current-Limiting Switch On-Resistance vs Temperature

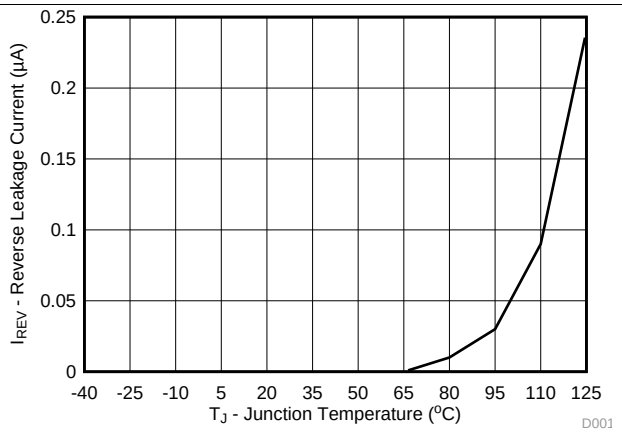


Figure 4. OUT Reverse Leakage Current vs Temperature

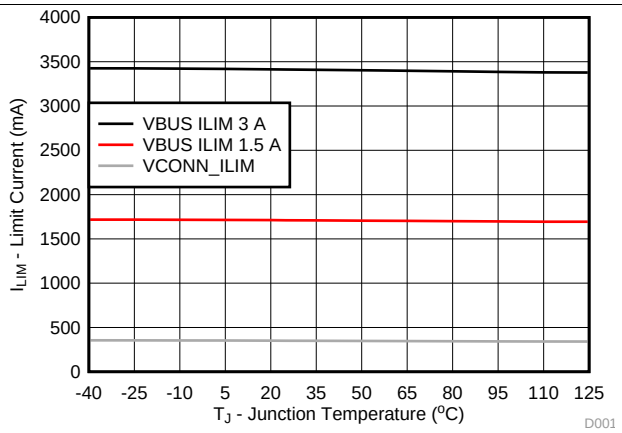


Figure 5. I_{LIM} for V_{BUS} and V_{CONN} vs Temperature

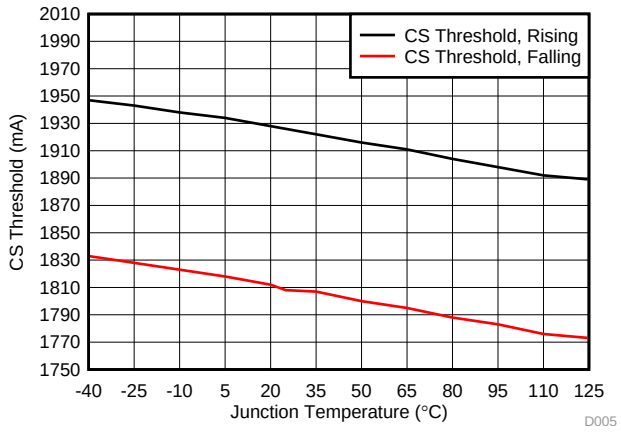


Figure 6. $\overline{CS}$ Threshold vs Temperature

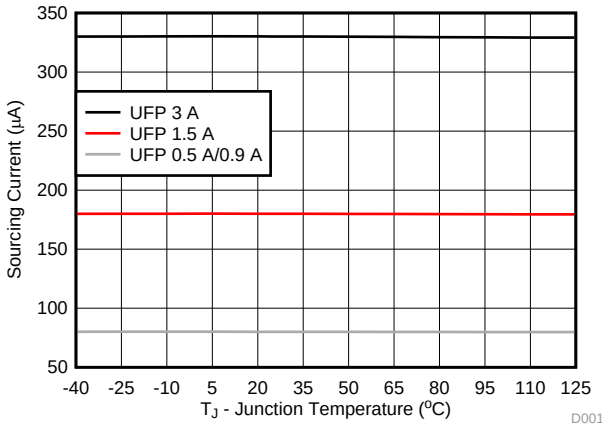


Figure 7. CC Sourcing Current to UFP vs Temperature

Copyright © 2017, Texas Instruments Incorporated

Typical Characteristics (continued)

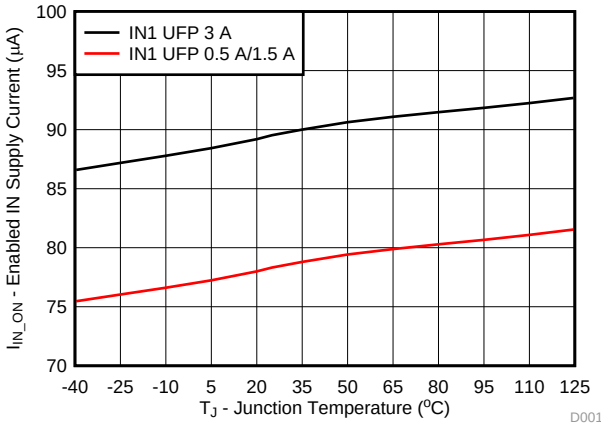


Figure 8. IN1 Current With UFP vs Temperature

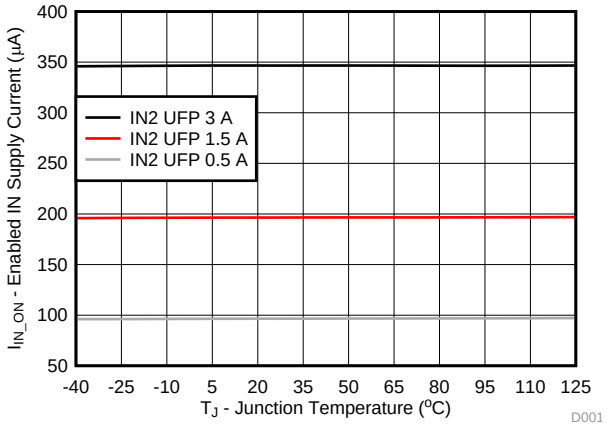
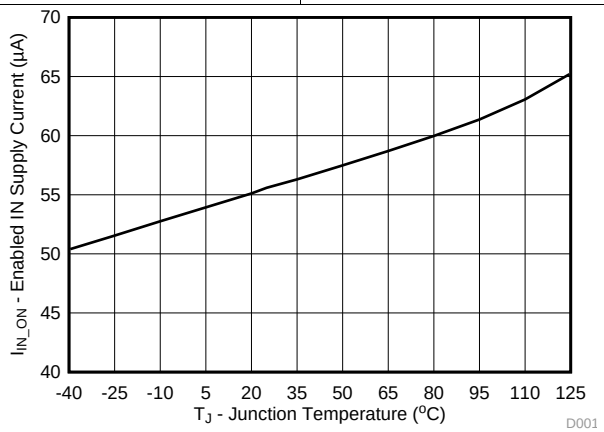


Figure 9. IN2 Current With UFP vs Temperature



V_{AUX} = 5 V

Figure 10. AUX Current With UFP vs Temperature

8 Detailed Description

8.1 Overview

The TPS25810A-Q1 device is a highly integrated USB Type-C™ downstream-facing port (DFP) controller, developed with a built-in power switch for the new USB Type-C connector and cable. The device provides all of the functionality needed to support a USB Type-C DFP in a system where USB power delivery (PD) source capabilities (for example, V_{BUS} > 5 V) are not implemented. It is designed to be compliant with the Type-C specification, revision 1.2.

8.1.1 USB Type-C Basic

For a detailed description of the Type-C specification, see the USB-IF Web site to download the latest released version. Some of the basic concepts of the Type-C specification that pertain to understanding the operation of the TPS25810A-Q1 (DFP device) are described as follows.

USB Type-C removes the need for different plug and receptacle types for host and device functionality. The Type-C receptacle replaces both Type-A and Type-B receptacles because the Type-C cable is pluggable in either direction between host and device. A host-to-device logical relationship is maintained via the configuration channel (CC). Optionally, hosts and devices can be either providers or consumers of power when USB PD communication is used to swap roles.

All USB Type-C ports operate in one of the following three data modes:

- Host mode: the port can only be host (provider of power).
- Device mode: the port can only be device (consumer of power).
- Dual-role mode: the port can be either host or device.

Port types:

- DFP (downstream facing port): Host
- UFP (upstream facing port): Device
- DRP (dual-role port): Host or device

Valid DFP-to-UFP connections:

- Table 1 describes valid DFP-to-UFP connections.
- Host-to-host and device-to-device have no functions.

Table 1. DFP-to-UFP Connections

	HOST-MODE PORT	DEVICE-MODE PORT	DUAL-ROLE PORT
Host-mode port	No function	Works	Works
Device-mode port	Works	No function	Works
Dual-role port	Works	Works	Works(1)

(1) This may be automatic or manually driven.

8.1.2 Configuration Channel

The function of the configuration channel (CC) is to detect connections and configure the interface across the USB Type-C cables and connectors.

Functionally, the configuration channel serves the following purposes:

- Detect connection to the USB ports
- Resolve cable orientation and twist connections to establish USB data-bus routing
- Establish DFP and UFP roles between two connected ports
- Discover and configure power: USB Type-C current modes or USB power delivery
- Discover and configure optional alternate and accessory modes
- Enhance flexibility and ease of use

Typical flow of DFP-to-UFP configuration is shown in Figure 11:

Copyright © 2017, Texas Instruments Incorporated

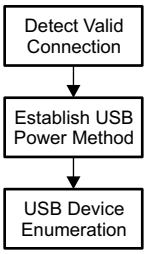


Figure 11. Flow of DFP-to-UFP Configuration

8.1.3 Detecting a Connection

DFPs and DRPs fulfill the role of detecting a valid connection over USB Type-C. Figure 12 shows a DFP-to-UFP connection made with Type-C cable. As shown in Figure 12, the detection concept is based on being able to detect terminations in the product that has been attached. A pullup and pulldown termination model is used. A pullup termination can be replaced by a current source.

- In the DFP-to-UFP connection, the DFP monitors both CC pins for a voltage lower than the unterminated voltage.
- A UFP advertises Rd on both of its CC pins (CC1 and CC2).
- A powered cable advertises Ra on only one of the CC pins of the plug. Ra is used to inform the source to apply V_{CONN}.
- An analog audio device advertises Ra on both CC pins of the plug, which identifies it as an analog audio device. V_{CONN} is not applied on either CC pin in this case.

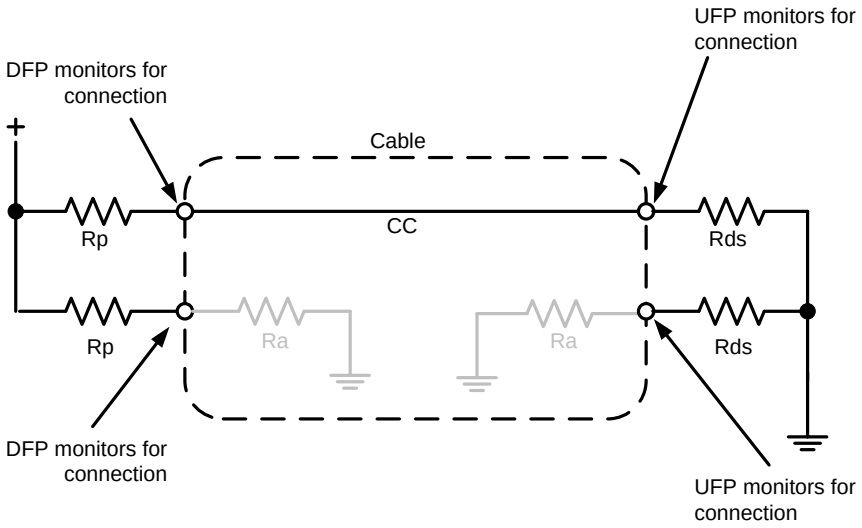
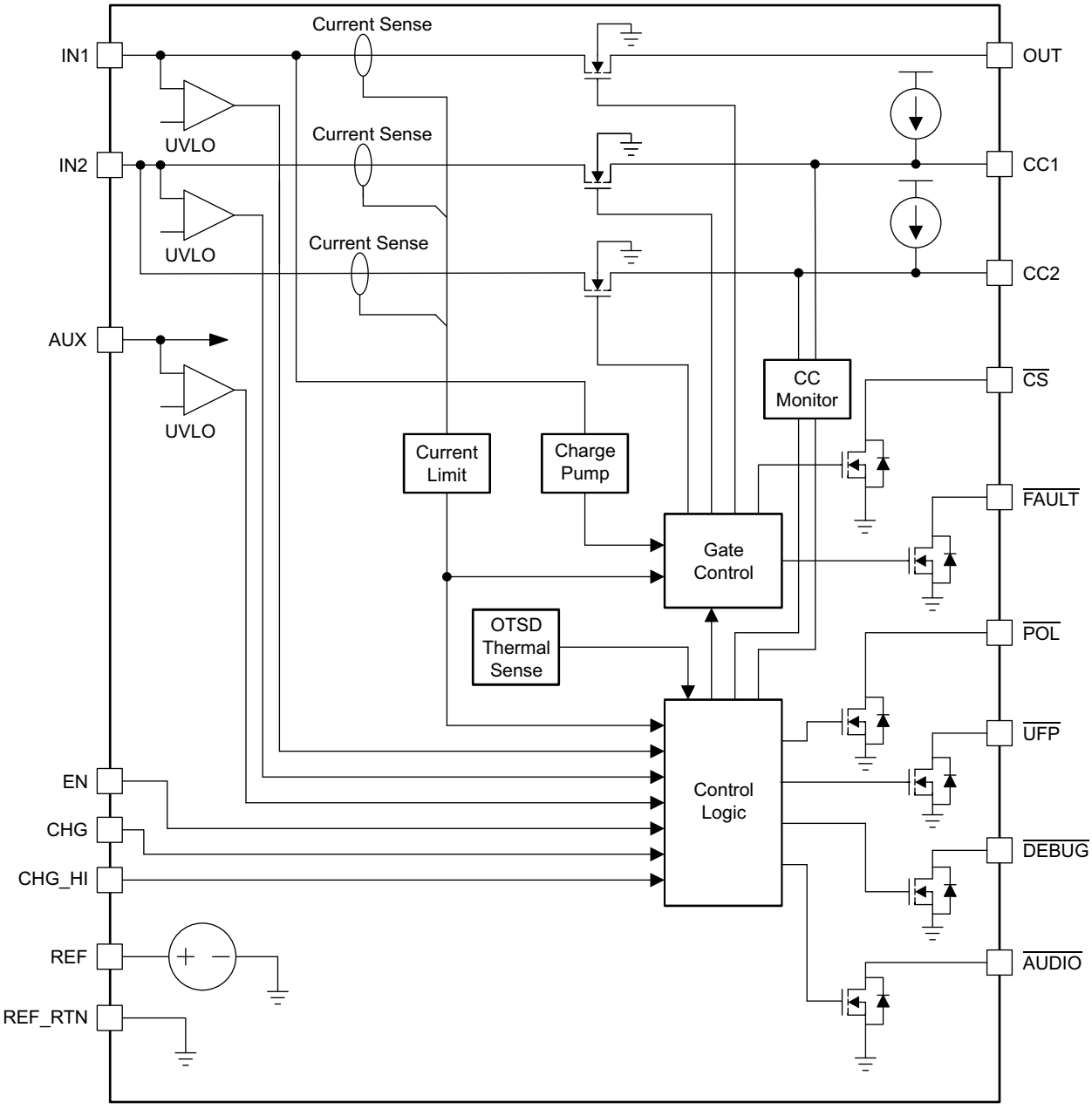


Figure 12. DFP-to-UFP Connection

8.2 Functional Block Diagram



Copyright © 2017, Texas Instruments Incorporated

8.3 Feature Description

TheTPS25810A-Q1 device is a DFP Type-C port controller with integrated power switches for V_{CONN} and V_{BUS}. It does not support BC 1.2 charging modes inherently, because it does not interact with USB D+ and D– data lines. The TPS25810A-Q1 device can be used in conjunction with a BC 1.2 controller like the TPS2514A-Q1 device to support BC1.2 and Type-C charging modes in a single Type-C DFP port. See the TPS25810 EVM User's Guide and *Application and Implementation* section of this data sheet for more details. The TPS25810A-Q1 device can be used in a USB 2.0 only or in a USB 3.1 port implementation. When used in a USB 3.1 port, the $\overline{POL}$ pin can control an external super-speed MUX to handle the Type-C flippable feature.

Copyright © 2017, Texas Instruments Incorporated

Feature Description (continued)

8.3.1 Configuration Channel Pins CC1 and CC2

Each device has two pins, CC1 and CC2, that serve to detect an attachment to the port and to resolve cable orientation. These pins are also used to establish the current broadcast to a valid UFP, configure V_{CONN}, and detect attachment of a debug or audio-adapter accessory.

Table 2 lists the response to various attachments to its port.

Table 2. TPS25810A-Q1 Response

TPS25810A-Q1 TYPE-C PORT	CC1	CC2	TPS25810A-Q1 RESPONSE(1)					
			OUT	V_{CONN} on CC1 or CC2	$\overline{POL}$	$\overline{UFP}$	$\overline{AUDIO}$	$\overline{DEBUG}$
Nothing attached	OPEN	OPEN	OPEN	NO	Hi-Z	Hi-Z	Hi-Z	Hi-Z
UFP connected	Rd	OPEN	IN1	NO	Hi-Z	LOW	Hi-Z	Hi-Z
	OPEN	Rd	IN1	NO	LOW	LOW	Hi-Z	Hi-Z
Powered cable, no UFP connected	OPEN	Ra	OPEN	NO	Hi-Z	Hi-Z	Hi-Z	Hi-Z
	Ra	OPEN	OPEN	NO	Hi-Z	Hi-Z	Hi-Z	Hi-Z
Powered cable, UFP connected	Rd	Ra	IN1	CC2	Hi-Z	LOW	Hi-Z	Hi-Z
	Ra	Rd	IN1	CC1	LOW	LOW	Hi-Z	Hi-Z
Debug accessory connected	Rd	Rd	OPEN	NO	Hi-Z	Hi-Z	Hi-Z	LOW
Audio-adapter accessory connected	Ra	Ra	OPEN	NO	Hi-Z	Hi-Z	LOW	Hi-Z

(1) $\overline{POL}$, $\overline{UFP}$, $\overline{AUDIO}$, and $\overline{DEBUG}$ are open-drain outputs; pull high with 100 kΩ to AUX when used. Tie to GND or leave open when not used.

8.3.2 Current Capability Advertisement and Overload Protection

The TPS25810A-Q1 device supports all three Type-C current advertisements as defined by the USB Type-C standard. Current broadcast to a connected UFP is controlled by the CHG and CHG_HI pins. For each broadcast level, the device protects itself from a UFP that draws current in excess of the USB Type-C current advertisement of that port by setting the current limit as shown in Table 3.

Table 3. USB Type-C Current Advertisement

CHG	CHG_HI	CC CAPABILITY BROADCAST	CURRENT LIMIT (TYP)	$\overline{CS}$ THRESHOLD (TYP)
0	0	STD	3.4 A	1.95 A
0	1	STD	3.4 A	1.95 A
1	0	1.5 A	3.4 A	1.95 A
1	1	3 A	3.4 A	1.95 A

Under OUT overload conditions, an internal OUT current-limit regulator limits the output current to the selected I_{LIM} based on CHG and CHG_HI selection. In applications where V_{CONN} is supplied via CC1 or CC2, separate fixed current-limit regulators protect these pins from overload at the level indicated in the *Electrical Characteristics* table. When an overload condition is present, the device maintains a constant output current, with the output voltage determined by (I_{OS} × R_{LOAD}). Two possible overload conditions can occur. The first overload condition occurs when either: 1) input voltage is first applied, enable is true, and a short circuit is present (load which draws I_{OUT} > I_{OS}), or 2) input voltage is present and the TPS25810A-Q1 device is enabled into a short circuit. The output voltage is held near zero potential with respect to ground and the TPS25810A-Q1 device ramps the output current to I_{OS}. Both limit the current to I_{OS} until the overload condition is removed or the device begins to thermal cycle. This is demonstrated in Figure 23 where the device was enabled into a short, and subsequently cycles current off and on as the thermal protection engages.

Copyright © 2017, Texas Instruments Incorporated

The second condition is when an overload occurs while the device is enabled and fully turned on. The device responds to the overload condition within time t_{ios} (see Figure 1) when the specified overload (per *Electrical Characteristics*) is applied. The response speed and shape vary with the overload level, input circuit, and rate of application. The current-limit response can be either simply settling to I_{OS} or turnoff and controlled return to I_{OS}. Similar to the previous case, the TPS25810A-Q1 device limits the current to I_{OS} until the overload condition is removed or the device begins to thermal cycle.

The TPS25810A-Q1 device thermal cycles if an overload condition is present long enough to activate thermal limiting in any of the above cases. This is due to the relatively large power dissipation [(V_{IN} – V_{OUT}) × I_{OS}] driving the junction temperature up. The device turns off when the junction temperature exceeds 135°C (minimum) while in current limit. The device remains off until the junction temperature cools 20°C and then restarts. The current-limit profile is shown in Figure 13.

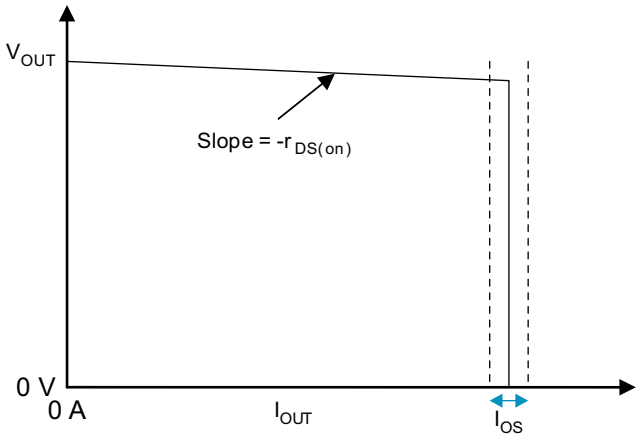


Figure 13. Current-Limit Profile

8.3.3 Undervoltage Lockout (UVLO)

The undervoltage lockout (UVLO) circuit disables the power switch until the input voltage reaches the UVLO turnon threshold. Built-in hysteresis prevents unwanted on-off cycling due to input voltage droop during turnon.

8.3.3.1 Device Power Pins (IN1, IN2, AUX, OUT, and GND)

The device has multiple input power pins: IN1, IN2 and AUX. IN1 is connected to OUT by the internal power FET and serves as the supply for the Type-C charging current. IN2 is the supply for V_{CONN} and ties directly between the V_{CONN} power switch on its input and CC1 or CC2 on its output. AUX, the auxiliary input supply, provides power to the device. See the *Functional Block Diagram*.

In the simplest implementation where multiple supplies are not available, IN1, IN2, and AUX can be tied together. However, in mobile systems (battery powered) where system power savings is paramount, IN1 and IN2 can be powered by the high-power dc-dc supply (>3-A capability), and AUX can be connected to the low-power supply that typically powers the system microcontroller when the system is in the hibernate or sleep power state. Unlike IN1 and IN2, AUX can operate directly from a 3.3-V supply commonly used to power the microcontroller when the system is put in low-power mode. Ceramic bypass capacitors close to the device from the INx and AUX pins to GND are recommended to alleviate bus transients.

The recommended operating voltage range for IN1 and IN2 is 4.5 V to 5.5 V, whereas AUX can be operated from 2.9 V to 5.5 V. However IN1, the high-power supply, can operate up to 6.5 V. This higher input voltage affords a larger IR loss budget in systems where a long cable harness is used, and results in high IR losses with 3-A charging current. Increasing IN1 beyond 5.5 V enables longer cable and board trace lengths between the device and the Type-C receptacle while meeting the USB specification for V_{BUS} ≥ 4.75 V at the connector.

Figure 14 illustrates the point. In this example IN1 is at 5 V, which restricts the IR loss budget from the dc-dc converter to the connector to 250 mV.

Copyright © 2017, Texas Instruments Incorporated

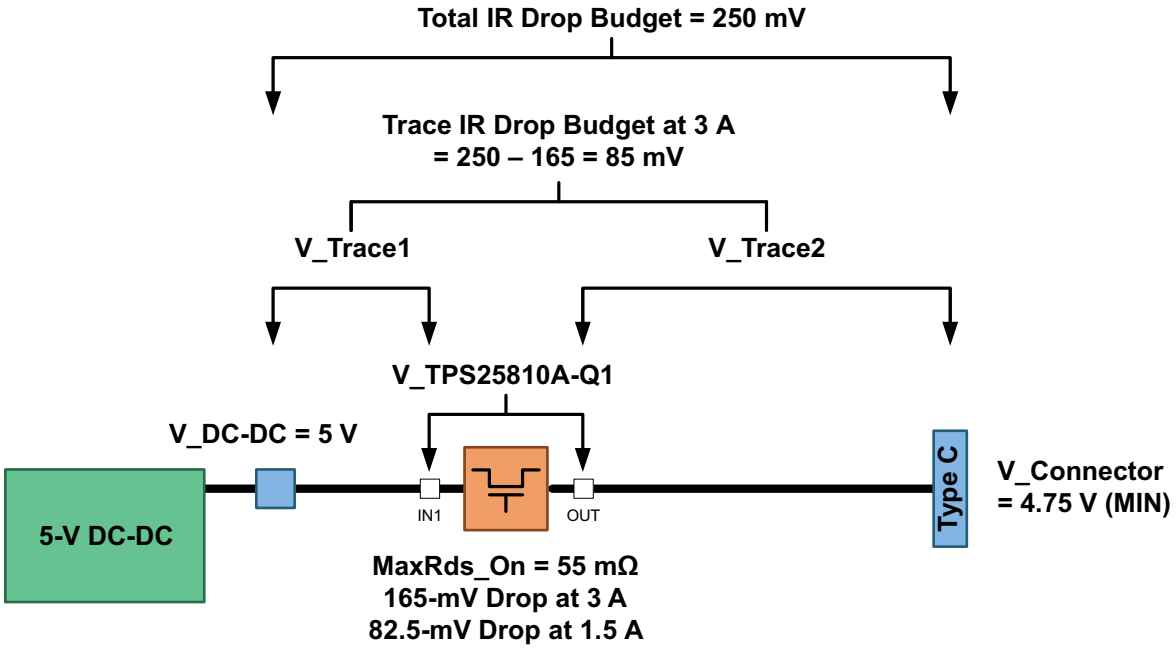


Figure 14. Total IR Loss Budget

8.3.3.2 FAULT Response

The $\overline{\text{FAULT}}$ pin is an open-drain output asserted low when the device OUT current exceeds its programmed value and the overtemperature threshold (T_{TH_OTSD1}) is crossed. See the *Electrical Characteristics* for overcurrent and overtemperature values. The $\overline{\text{FAULT}}$ signal remains asserted until the fault condition is removed and the device resumes normal operation. An internal deglitch circuit eliminates false overcurrent-fault reporting.

Connect $\overline{\text{FAULT}}$ with a pullup resistor to AUX. $\overline{\text{FAULT}}$ can be left open or tied to GND when not used.

8.3.3.3 Thermal Shutdown

The device has two internal overtemperature shutdown thresholds, T_{TH_OTSD1} and T_{TH_OTSD2}, to protect the internal FET from damage and assist with overall safety of the system. T_{TH_OTSD2} is greater than T_{TH_OTSD1}. $\overline{\text{FAULT}}$ is asserted low to signal a fault condition when the device temperature exceeds T_{TH_OTSD1} and the current-limit switch is disabled. However, when T_{TH_OTSD2} is exceeded, all open-drain outputs are left open and the device is disabled such that minimum power is dissipated. The device attempts to power up when the die temperature decreases by 20°C.

8.3.3.4 REF

A 100-kΩ (1% or better recommended) resistor is connected from this pin to REF_RTN. The REF pin sets the reference current required to bias the internal circuitry of the device. The overload current-limit tolerance and CC currents depend upon the accuracy of this resistor. Using a ±1% or better low-temperature-coefficient resistor yields the best current-limit accuracy and overall device performance.

8.3.3.5 Audio Accessory Detection

The USB Type-C specification defines an audio-adapter decode state which allows implementation of an analog USB Type-C to 3.5-mm headset adapter. An audio accessory device is detected when both CC1 and CC2 pins detect V_{Ra} voltage (when pulled to ground by an Ra resistor). The open-drain $\overline{\text{AUDIO}}$ pin is asserted low to indicate the detection of such a device.

Table 4. Audio Accessory Detection

CC1	CC2	$\overline{\text{AUDIO}}$	STATE
Ra	Ra	Asserted (pulled low)	Audio-adapter accessory connected

Copyright © 2017, Texas Instruments Incorporated

Platforms supporting the audio accessory function can be triggered by the $\overline{\text{AUDIO}}$ pin to enable accessory mode circuits to support the audio function. When the Ra pulldown is removed from the CC2 pin, $\overline{\text{AUDIO}}$ is deasserted or pulled high. The TPS25810A-Q1 device monitors the CC2 pin for audio device detach. When this function is not needed (for example in a data-less port), $\overline{\text{AUDIO}}$ can be tied to GND or left open.

8.3.3.6 Debug Accessory Detection

The Type-C spec supports an optional debug-accessory mode, used for debug only and not to be used for communicating with commercial products. When the TPS25810A-Q1 device detects V_{Rd} voltage on both CC1 and CC2 pins (when pulled to ground by an Rd resistor), it asserts $\overline{\text{DEBUG}}$ low. With $\overline{\text{DEBUG}}$ asserted, the system can enter debug mode for factory testing or a similar functional mode. $\overline{\text{DEBUG}}$ deasserts or pulls high when Rd is removed from CC1. The CC1 pin is monitored for debug-accessory detach.

If the debug-accessory mode is not used, tie $\overline{\text{DEBUG}}$ to GND or leave it open.

Table 5. Debug Accessory Detection

CC1	CC2	$\overline{\text{POL}}$	STATE
Rd	Rd	Asserted (pulled low)	Debug accessory connected

8.3.3.7 Plug Polarity Detection

Reversible Type-C plug orientation is reported by the $\overline{\text{POL}}$ pin when a UFP is connected. However, when no UFP is attached $\overline{\text{POL}}$ remains deasserted, irrespective of cable plug orientation. Table 6 describes the $\overline{\text{POL}}$ state based on which of the device CC pins detects V_{Rd} from an attached UFP pulldown.

Table 6. Plug Polarity Detection

CC1	CC2	$\overline{\text{POL}}$	STATE
Rd	Open	Hi-Z	UFP connected
Open	Rd	Asserted (pulled low)	UFP connected with reverse plug orientation

Copyright © 2017, Texas Instruments Incorporated

Figure 15 shows an example implementation which uses the $\overline{POL}$ terminal to control the SEL terminal on the HD3SS3212 device. The HD3SS3212 device provides switching on the differential channels between Port B and Port C to Port A, depending on cable orientation. For details on the HD3SS3212 device, see *HD3SS3212x Two-Channel Differential 2:1/1:2 USB3.1 Mux/Demux*.

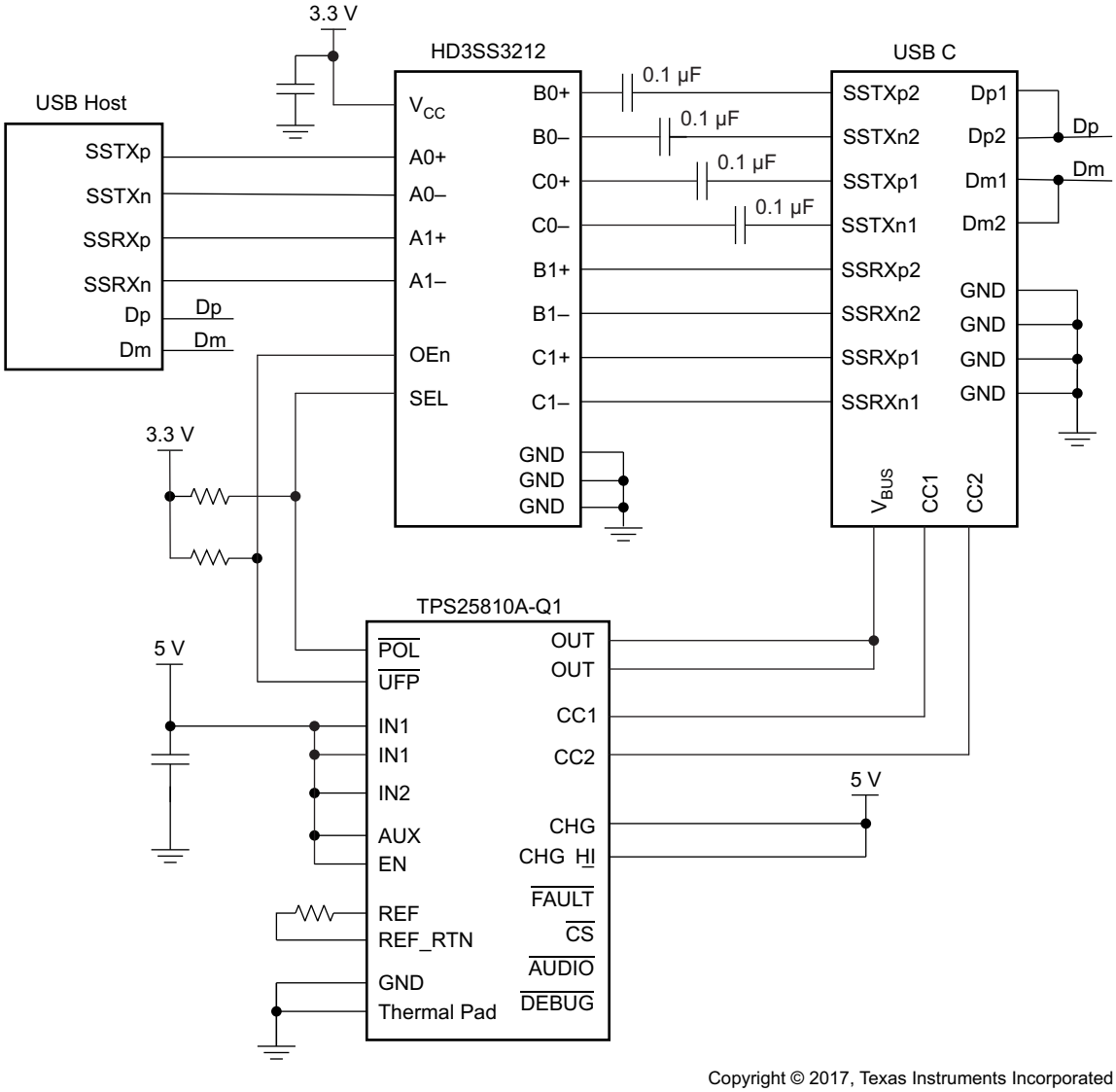


Figure 15. Example Implementation

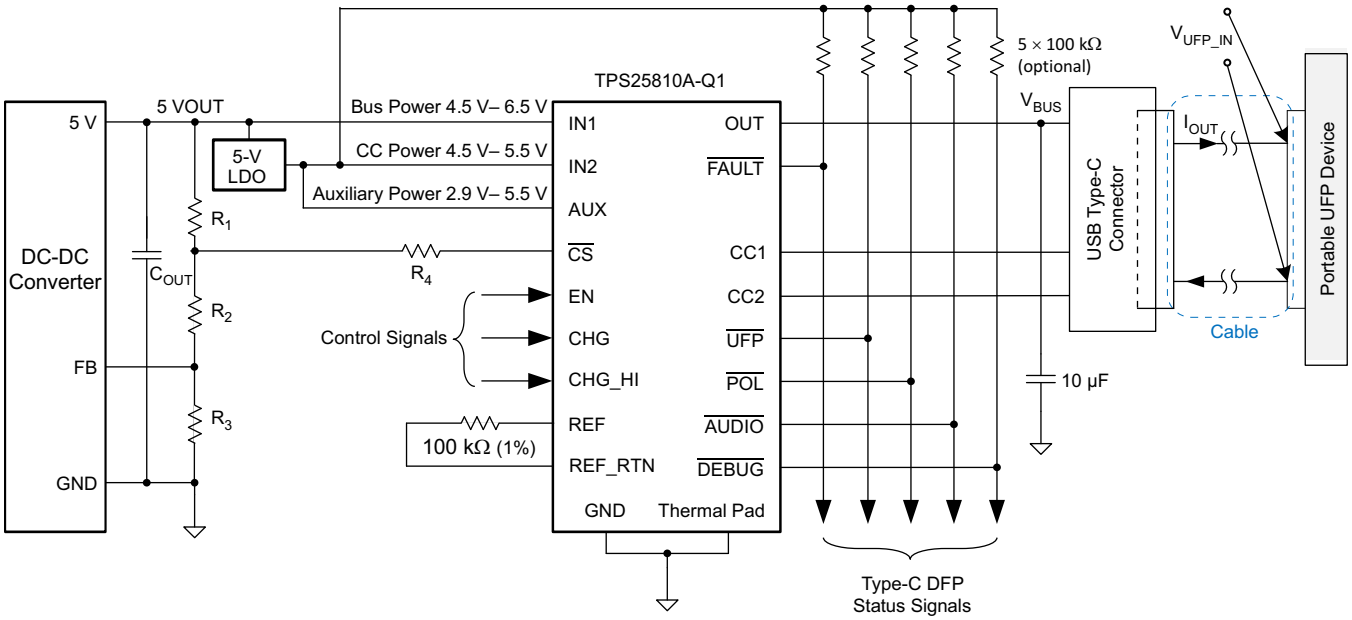
8.3.3.8 Device Enable Control

The logic enable pin (EN) controls the power switch and device supply current. The supply current is reduced to less than 1 µA when a logic low is present on EN. The EN pin provides a convenient way to turn on or turn off the device while it is powered. The enable input threshold has built-in hysteresis. When this pin is pulled high, the device is turned on or enabled. When the device is disabled (EN pulled low) the internal FETs tied to IN1 and IN2 are disconnected, all open-drain outputs are left open (Hi-Z), and the monitor block for CC1 and CC2 is turned off. The EN terminal should not be left floating.

8.3.3.9 Cable Compensation ($\overline{CS}$)

The TPS25810A-Q1 device monitors the current to a UFP, and if the load current exceeds 1.95 A (typ), the $\overline{CS}$ pin asserts. This can be useful for implementing a digital droop-compensation scheme by altering the feedback resistor ratio of the IN1 power source.

Figure 16 shows a USB charging design using the TPS25810A-Q1 device. The 5-V (typical) nominal output of the USB power supply, designated 5 V_{OUT} herein, is often a dc-dc converter in automotive applications. V_{UFP_IN} refers to the voltage across the inside contacts of the USB connector of a UFP device. Official USB specifications should be consulted for the most up-to-date requirements. For illustration purposes, it is assumed the minimum and maximum voltages allowed for V_{UFP_IN} are 4 V and 5.25 V, respectively. In general, when V_{UFP_IN} is 5 V, the UFP draws optimum current and requires the minimum amount of time to recharge its battery.



Copyright © 2017, Texas Instruments Incorporated

Figure 16. TPS25810A-Q1 Charging System Schematic

In a practical system, there are voltage drops from the dc-dc output, 5 V_{OUT}, to V_{UFP_IN} which include the on-resistance of the TPS25810A-Q1 device power switch, USB cabling and connector contact resistances. Under rated UFP load current, these drops can be several hundred millivolts, decreasing V_{UFP_IN} below the optimal 5-V level. In addition, as V_{UFP_IN} decreases below 5 V, most modern UFPs decrease their load current to prevent possible overload conditions and to maintain V_{UFP_IN} above 4 V. Lower-than-optimum load current increases the time required to recharge the UFP battery. For example, in Figure 16, assuming that the loss resistance is 113 mΩ (includes 79 mΩ of USB cable resistance and 34 mΩ of power switch resistance) and 5 V_{OUT} is 5 V, the input voltage of UFP (V_{UFP_IN}) is about 4.66 V at 3 A. The TPS25810A-Q1 device provides the $\overline{CS}$ pin to report high-charging-current conditions and increase the 5 V_{OUT} voltage as shown in Figure 17

Copyright © 2017, Texas Instruments Incorporated

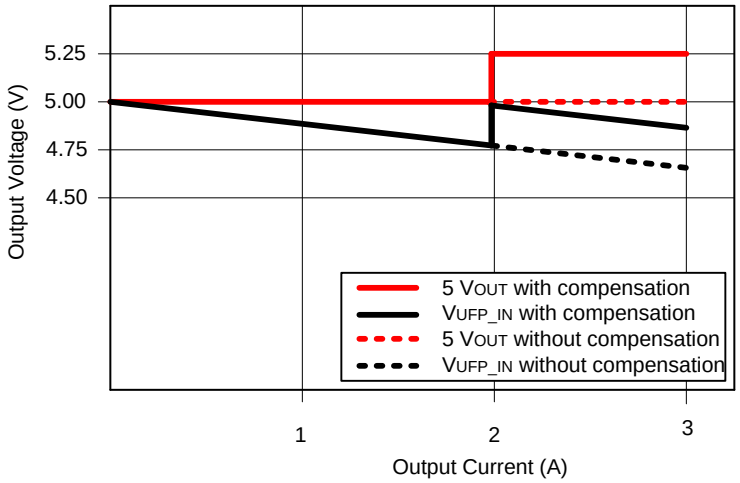


Figure 17. TPS25810A-Q1 $\overline{CS}$ Function

Equation 1 through Equation 4 refer to Figure 16

The power supply output voltage is calculated in Equation 1.

$$5V_{OUT} = \frac{(R_1 + R_2 + R_3) \times V_{FB}}{R_3} \quad (1)$$

5 V_{OUT} and V_{FB} are known. If R_3 is given and R_1 is fixed, R_2 can be calculated. The 5 V_{OUT} voltage change with compensation is shown in Equation 2 and Equation 3.

$$\Delta V = \frac{(R_2 + R_3) \times R_1 \times V_{FB}}{R_3 \times R_4} \quad (2)$$

$$\Delta V = \left(\frac{5V_{OUT}}{V_{FB}} - \frac{R_1}{R_3} \right) \frac{R_1 \times V_{FB}}{R_4} \quad (3)$$

If R_1 is less than R_3, then Equation 3 can be simplified as Equation 4.

$$\Delta V \approx \frac{5V_{OUT} \times R_1}{R_4} \quad (4)$$

8.3.3.10 Power Wake

The power-wake feature offers the mobile-systems designer a way to save on system power when no UFP is attached to the Type-C port. See Figure 18. To enable power wake, the $\overline{UFP}$ pins from any combination of two TPS25810A-Q1 devices are tied together (each with its own 100-kΩ pullup) to the enable pin of a 5-V, 6-A dc-dc buck converter. When no UFP is detected on both Type-C ports, the EN pin of the dc-dc converter is pulled high, thereby disabling it. Because the TPS25810A-Q1 device is powered by an always-on 3.3-V LDO, turning off the supply to IN1 and IN2 does not affect its operation in the detach state. Anytime a UFP is detected on either port, the corresponding $\overline{UFP}$ pin is pulled low, enabling the dc-dc converter to provide charging current to the attached UFP. Turning off the high-power dc-dc converter when ports are unattached saves on system power. This method can save a significant amount of power, because the TPS25810A-Q1 device requires only 0.7 µA (typical) via the AUX pin when no UFP device is connected.

Copyright © 2017, Texas Instruments Incorporated

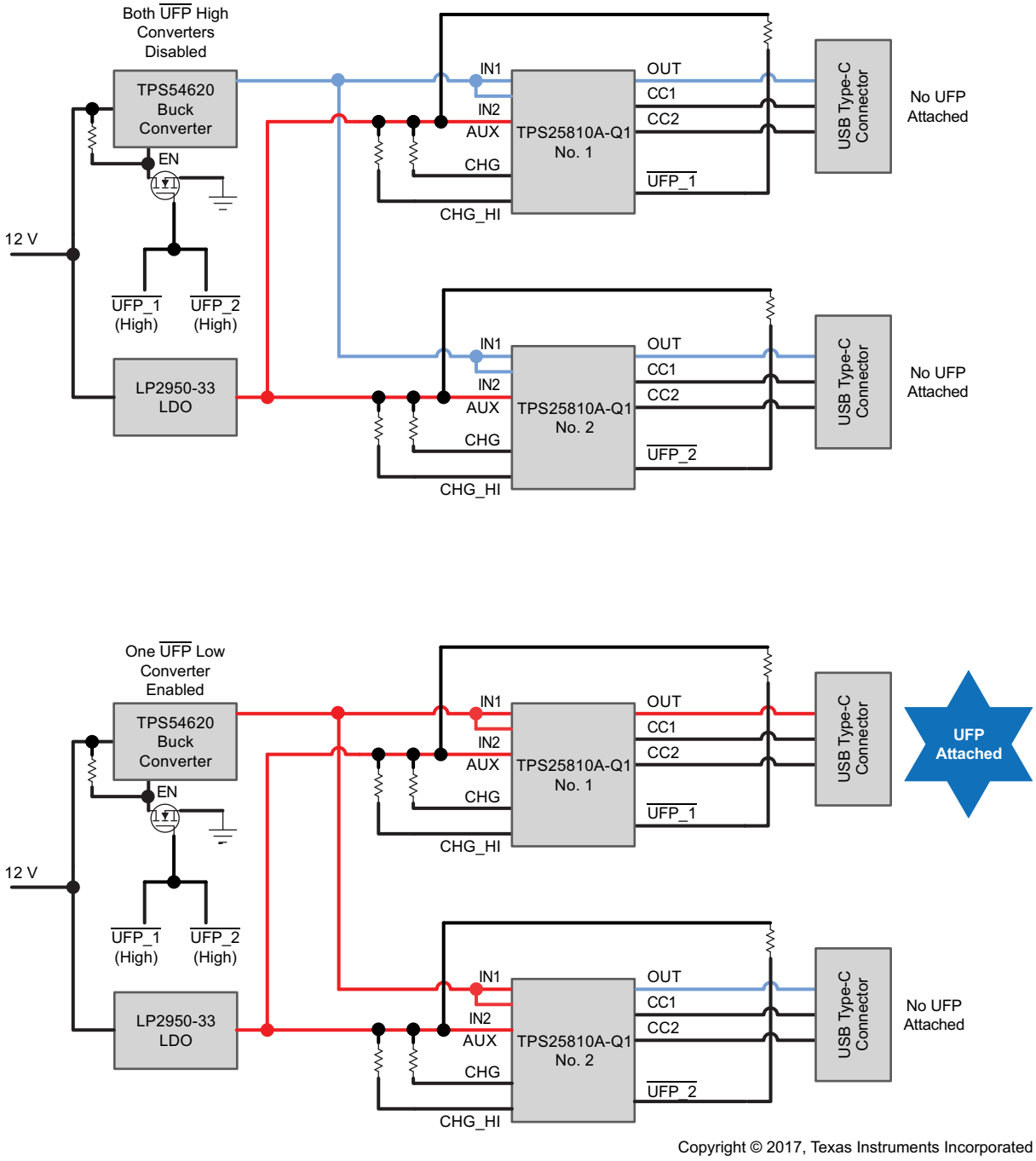


Figure 18. Power-Wake Implementation

8.4 Device Functional Modes

The TPS25810A-Q1 device is a Type-C controller with integrated power switches that supports all Type-C functions in a downstream facing port. The device manages current advertisement and protection for a connected UFP and active cable. Each device starts its operation by monitoring the AUX bus. When V_{AUX} exceeds the undervoltage-lockout threshold, the device samples the EN pin. A high level on this pin enables the device, and normal operation begins. Having successfully completed its start-up sequence, the device now

Copyright © 2017, Texas Instruments Incorporated

Device Functional Modes (continued)

actively monitors its CC1 and CC2 pins for attachment to a UFP. When a UFP is detected on either the CC1 or CC2 pin, the internal MOSFET starts to turn on after the required deglitch time is met. The internal MOSFET starts conducting and allows current to flow from IN1 to OUT. If Ra is detected on the other CC pin (not connected to the UFP), V_{CONN} is applied to allow current to flow from IN2 to the CC pin connected to Ra. For a complete listing of various device operational modes, see Table 2.

Copyright © 2017, Texas Instruments Incorporated

9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The TPS25810A-Q1 device is a Type-C DFP controller that supports all Type-C DFP required functions. It applies power to V_{BUS} when a UFP attach is detected and removes power when it detects the UFP is detached. The device exposes its identity via its CC pin, advertising its current capability based on the CHG and CHG_HI pin settings. The TPS25810A-Q1 device also limits its advertised current internally and provides robust protection to a fault on the system V_{BUS} power rail.

After a connection is established, either device is capable of providing V_{CONN} to power circuits in the cable plug on the CC pin that is not connected to the CC wire in the cable. V_{CONN} is internally current-limited and has its own supply pin, IN2. Apart from providing charging current to a UFP, the TPS25810A-Q1 device also supports audio and debug accessory modes.

The following design procedure can be used to implement a full-featured Type-C DFP.

NOTE

BC 1.2 is not supported in the TPS25810A-Q1 device. To support BC 1.2 with Type-C charging modes in a single Type-C connector, a dedicated charging port (DCP) controller something like a TPS2514A-Q1 device must be used.

9.2 Typical Applications

9.2.1 Type-C DFP Port Implementation Without BC 1.2 Support

Figure 19 shows a minimal Type-C DFP implementation capable of supporting 5-V and 3-A charging.

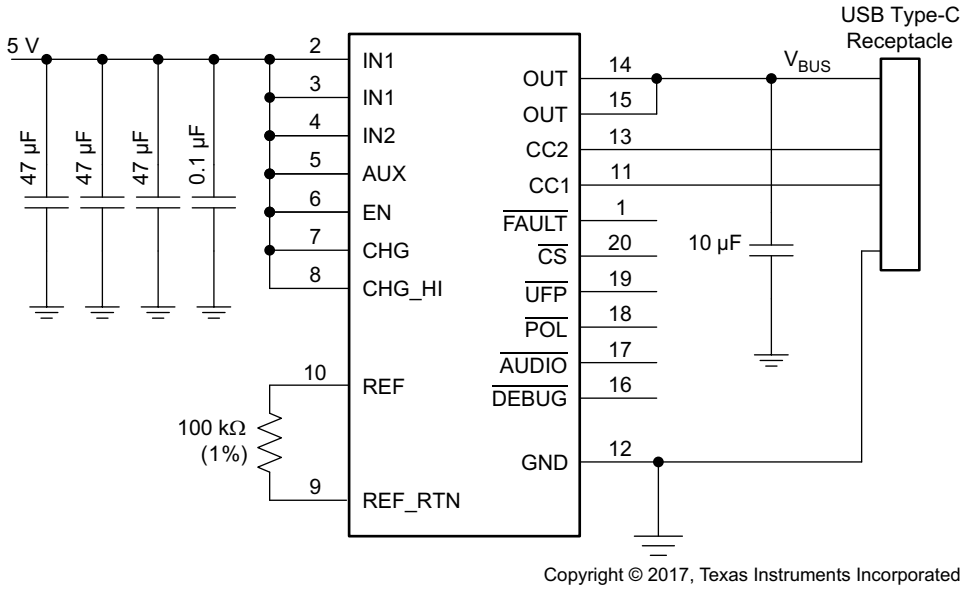


Figure 19. Type-C DFP Port Implementation Without BC 1.2 Support

Copyright © 2017, Texas Instruments Incorporated

Typical Applications (continued)

9.2.1.1 Design Requirements

9.2.1.1.1 Input and Output Capacitance

Input and output capacitance improves the performance of the device. The actual capacitance should be optimized for the particular application. For all applications, a 0.1-μF or greater ceramic bypass capacitor between INx and GND is recommended as close to the device as possible for local noise decoupling.

All protection circuits, including those of the TPS25810A-Q1 device, have the potential for input voltage overshoots and output voltage undershoots. Input voltage overshoots can be caused by either of two effects. The first cause is an abrupt application of input voltage in conjunction with input power-bus inductance and input capacitance when the INx pin is high-impedance (before turnon). Theoretically, the peak voltage is 2 times the applied voltage. The second cause is due to the abrupt reduction of output short-circuit current when the device turns off and energy stored in the input inductance drives the input voltage high. Input voltage droops may also occur with large load steps and as the output is shorted. Applications with large input inductance (for instance, connecting the evaluation board to the bench power supply through long cables) may require large input capacitance to prevent the voltage overshoot from exceeding the absolute maximum voltage of the device.

The fast current-limit speed of the TPS25810A-Q1 device to hard output short circuits isolates the input bus from faults. However, ceramic input capacitance in the range of 1 μF to 22 μF adjacent to the input aids in both response time and limiting the transient seen on the input power bus. Momentary input transients to 6.5 V are permitted. Output voltage undershoot is caused by the inductance of the output power bus just after a short has occurred and the device has abruptly reduced the OUT current. Energy stored in the inductance drives the OUT voltage down, and potentially negative, as it discharges. An application with large output inductance (such as from a cable) benefits from the use of a high-value output capacitor to control voltage undershoot.

When implementing a USB-standard application, 120-μF minimum output capacitance is required. Typically, a 150-μF electrolytic capacitor is used, which is sufficient to control voltage undershoots. Because in Type-C applications, DFP is a cold socket when no UFP is attached, the output capacitance should be placed at the INx pin versus the OUT pin, as is done in USB Type-A ports. It is also recommended to put a 10-μF ceramic capacitor on the OUT pin for better voltage bypass.

9.2.1.2 Detailed Design Procedure

The TPS25810A-Q1 device supports up to three different input voltages, based on the application. In the simplest implementation, all input pins are tied to a single voltage source set to 5 V, as shown in Figure 19. However, it is recommended to set a slightly higher (100 mV to 200 mV) input voltage, when possible, to compensate for IR loss from the source to the Type-C connector.

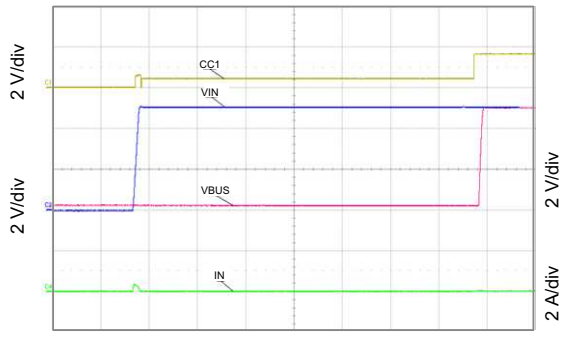
Other design considerations are listed as follows:

- Place at least 120 μF of bypass capacitance close to the INx pins rather than the OUT pin, as Type-C is a cold-socket connector.
- A 10-μF bypass capacitor is recommended to be placed near a Type-C receptacle V_{BUS} pin to handle load transients.
- Depending on the maximum current-level advertisement supported by the Type-C port in the system, set the CHG and CHG_HI levels accordingly. Advertisement of 3 A is shown in Figure 19.
- The EN, CHG, and CHG_HI pins can be tied directly to GND or V_{AUX} without a pullup resistor.
  - CHG and CHG_HI can also be dynamically controlled by a microcontroller to change the current advertisement level to the UFP.
- When an open-drain output of the TPS25810A-Q1 device is not used, it can be left open or tied to GND.
- Use a 1% 100-kΩ resistor to connect between the REF and REF_RTN pins, placing it close to the device pin and keeping it isolated from switching noise on the board.

Copyright © 2017, Texas Instruments Incorporated

Typical Applications (continued)

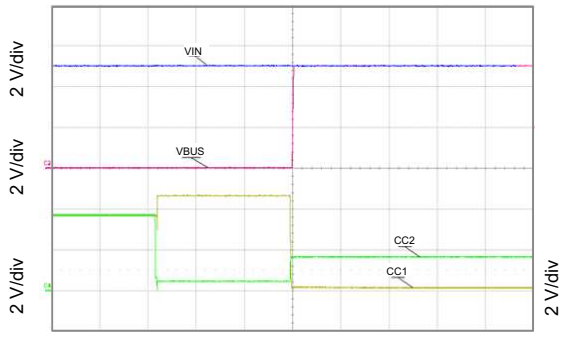
9.2.1.3 Application Curves



Time 20 ms/div

Basic start-up: IN1 = IN2 = AUX = EN = CHG = CHG_HI = 5 V
CC1 = Rd CC2 = open

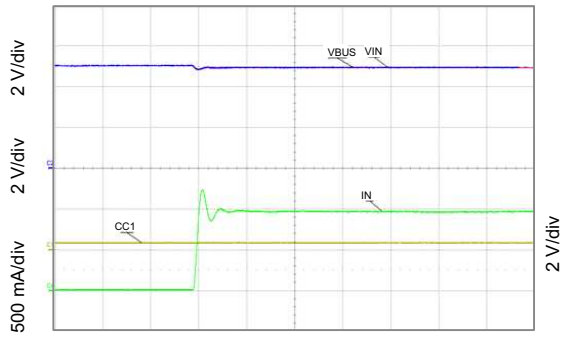
Figure 20. Basic Start-Up



Time 50 ms/div

IN1 = IN2 = AUX = EN = CHG = CHG_HI = 5 V
CC1 = open CC2 = open → Rd

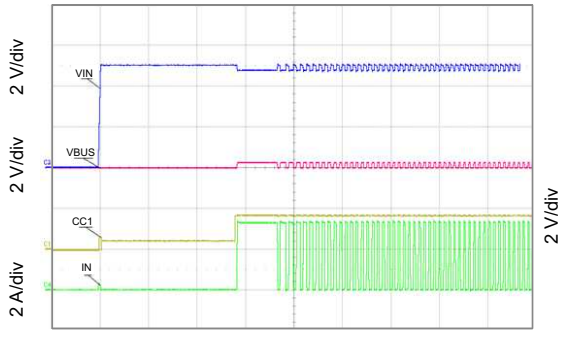
Figure 21. Start-Up



Time 200 µs/div

IN1 = IN2 = AUX = EN = 5 V; CHG = CHG_HI = 0 V
CC1 = open CC2 = Rd OUT = open → 5 Ω

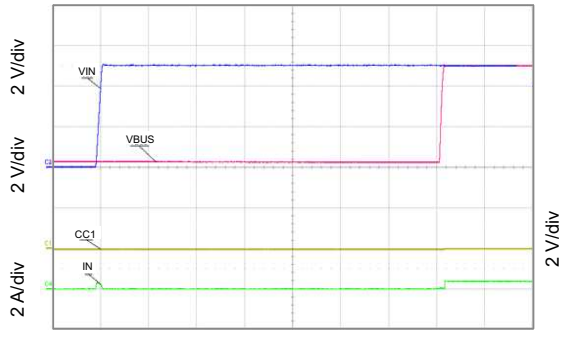
Figure 22. Load Step



Time 50 ms/div

IN1 = IN2 = AUX = EN = CHG = CHG_HI = 5 V
CC1 = Rd CC2 = open OUT = shorted

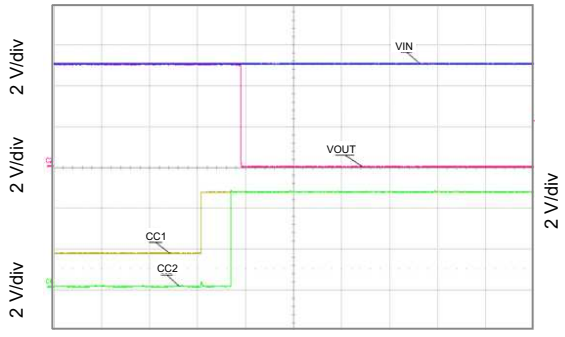
Figure 23. Hot-Plug to Short



Time 20 ms/div

IN1 = IN2 = AUX = EN = CHG = CHG_HI = 5 V
CC1 = short CC2 = Rd

Figure 24. Short On CC1



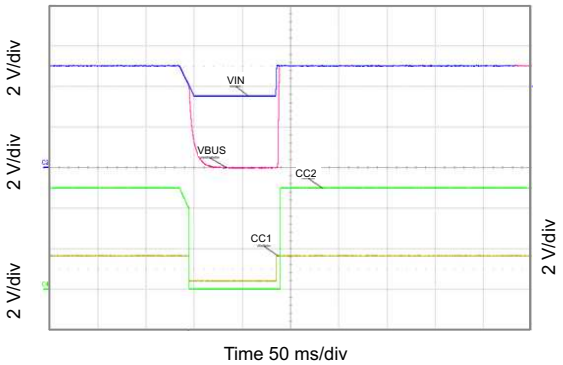
Time 20 ms/div

IN1 = IN2 = AUX = EN = CHG = CHG_HI = 5 V
CC1 = Rd → open CC2 = open

Figure 25. Remove Rd

Copyright © 2017, Texas Instruments Incorporated

Typical Applications (continued)



V_{IN} 5 V → 3.5 V (100 ms) → 5 V (1 V/ms) CC1 = Rd
IN1 = IN2 = AUX = EN = CHG = CHG_HI = 5 V CC2 = Ra

Figure 26. Brown-Out Test

Copyright © 2017, Texas Instruments Incorporated

Typical Applications (continued)

9.2.2 Type-C DFP Port Implementation With BC 1.2 (DCP Mode) Support

Figure 27 shows a Type-C DFP implementation capable of supporting 5-V, 3-A charging in a Type-C port that is also able to support charging of legacy devices when used with a Type-C µB cable assembly for charging phones and handheld devices equipped with a µB connector.

This implementation requires the use of a TPS2514A-Q1 device, a USB dedicated charging-port (DCP) controller with auto-detect feature to charge not only BC 1.2-compliant handheld devices but also popular phones and tablets that incorporate their own propriety charging algorithm. See *TPS2513A-Q1, TPS2514A-Q1 USB Dedicated Charging Port Controller* for more details.

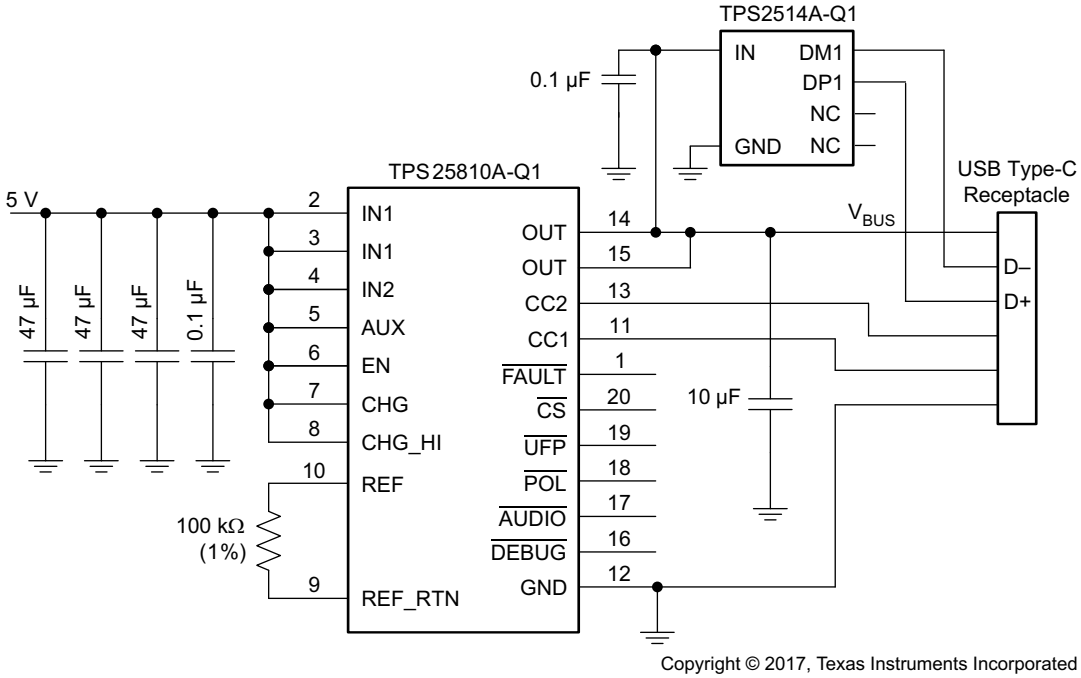


Figure 27. Type-C DFP Port Implementation With BC 1.2 (DCP Mode) Support

9.2.2.1 Design Requirements

See *Design Requirements* for the design requirements.

9.2.2.2 Detailed Design Procedure

See *Detailed Design Procedure* for the detailed design procedure.

9.2.2.3 Application Curves

See *Application Curves* for the application curves.

Copyright © 2017, Texas Instruments Incorporated

10 Power Supply Recommendations

The device has three power supply inputs. IN1, which is directly connected to OUT via the power MOSFET, is tied to the V_{BUS} pin in the Type-C receptacle. IN2 has a current-limiting switch and is multiplexed either to the CC1 or CC2 pin in the Type-C receptacle, depending on cable plug polarity. AUX is the device supply. In most applications, all three supplies are tied together. In a special implementation like power wake, IN1 and IN2 are tied to a single supply, whereas AUX is powered by a supply that is always ON and can be as low as 2.9 V.

USB Specification Revisions 2.0 and 3.1 require V_{BUS} voltage at the connector to be between 4.75 V and 5.5 V. Depending on layout and routing from the supply to the connector, the voltage drop on V_{BUS} must be tightly controlled. Locate the input supply close to the device. For all applications, a 10-μF or greater ceramic bypass capacitor between OUT and GND is recommended, located as close to the Type-C connector of the device as possible for local noise decoupling. The power supply should be rated higher than the current limit setting to avoid voltage droops during overcurrent and short-circuit conditions.

Copyright © 2017, Texas Instruments Incorporated

11 Layout

11.1 Layout Guidelines

Layout best practices as they apply to the TPS25810A-Q1 device are listed as follows.

- For all applications, a 10-µF ceramic capacitor is recommended near the Type-C receptacle and another 120-µF ceramic capacitor close to the IN1 pin.
  - The optimum placement of the 120-µF capacitor is closest to the IN1 and GND pins of the device.
  - Care must be taken to minimize the loop area formed by the bypass capacitor connection, the IN1 pin, and the GND pin of the device. See Figure 28 for a PCB layout example.
- High-current-carrying power-path connections to the device should be as short as possible and should be sized to carry at least twice the full-load current.
  - Have the input and output traces as short as possible. The most common cause of voltage loss failure in USB power delivery is the resistance associated with the V_{BUS} trace. Trace length, maximum current being supplied for normal operation, and total resistance associated with the V_{BUS} trace must be taken into account while budgeting for voltage loss.
  - For example, a power-carrying trace that supplies 3 A, at a distance of 20 inches, 0.1-in. wide, with 2-oz. copper on the outer layer has a total resistance of approximately 0.046 Ω and voltage loss of 0.14 V. The same trace at 0.05 in. wide has a total resistance of approximately 0.09 Ω and voltage loss of 0.28 V.
  - Make power traces as wide as possible.
- The resistor attached to the REF pin of the device has several requirements:
  - It is recommended to use a 1% 100-kΩ low-temperature-coefficient resistor.
  - It should be connected to the REF and REF_RTN pins (pins 9 and pin 10, respectively).
  - The REF_RTN pin should be isolated from the GND plane. See Figure 28.
  - The trace routing between the REF and REF_RTN pins of the device should be as short as possible to reduce parasitic effects on current-limit and current-advertisement accuracy. These traces should not have any coupling to switching signals on the board.
- Locate all TPS25810A-Q1 pullup resistors for open-drain outputs close to their connection pin. Pullup resistors should be 100 kΩ.
  - When a particular open-drain output is not used or needed in the system, leave the associated pin open or tied to GND.
- Keep the CC lines close to the same length.
- Thermal considerations:
  - When properly mounted, the thermal-pad package provides significantly greater cooling ability than an ordinary package. To operate at rated power, the thermal pad must be soldered to the board GND plane directly under the device. The thermal pad is at GND potential and can be connected using multiple vias to inner-layer GND. Other planes, such as the bottom side of the circuit board, can be used to increase heat sinking in higher-current applications. See *PowerPad™ Thermally Enhanced Package* and *PowerPAD™ Made Easy* for more information on using this thermal pad package.
  - Obtaining acceptable performance with alternate layout schemes is possible; however, the layout example in the following section has been shown to produce good results and is intended as a guideline.
- ESD considerations:
  - The TPS25810A-Q1 device has built-in ESD protection for CC1 and CC2. Keep trace length to a minimum from the Type-C receptacle to the TPS25810A-Q1 device on CC1 and CC2.
  - A 10-µF output capacitor should be placed near the Type-C receptacle.
  - See the TPS25810EVM-745 evaluation module for an example of a double-layer board that passes IEC61000-4-2 testing.
  - Do not create stubs or test points on the CC lines. Keep the traces short if possible, and use minimal vias along the traces [1–2 inches (2.54 cm–5.08 cm) or less].
  - See *ESD Protection Layout Guide* for additional information.
  - Have a dedicated ground plane layer, if possible, to avoid differential voltage buildup.

Copyright © 2017, Texas Instruments Incorporated

11.2 Layout Example

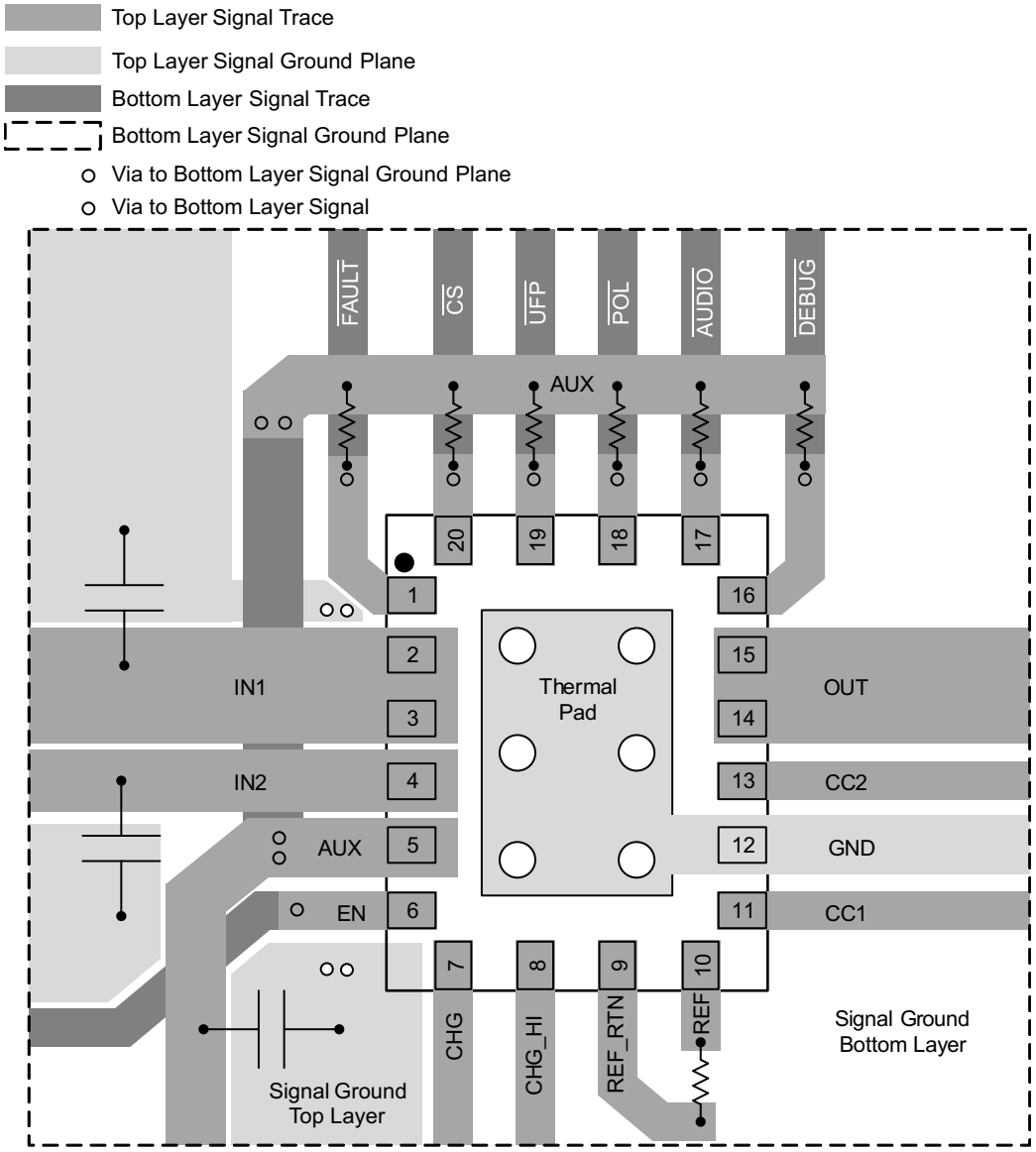


Figure 28. Layout Example

12 器件和文档支持

12.1 器件支持

12.1.1 Third-Party Products Disclaimer

TI'S PUBLICATION OF INFORMATION REGARDING THIRD-PARTY PRODUCTS OR SERVICES DOES NOT CONSTITUTE AN ENDORSEMENT REGARDING THE SUITABILITY OF SUCH PRODUCTS OR SERVICES OR A WARRANTY, REPRESENTATION OR ENDORSEMENT OF SUCH PRODUCTS OR SERVICES, EITHER ALONE OR IN COMBINATION WITH ANY TI PRODUCT OR SERVICE.

12.2 文档支持

12.2.1 相关文档

《PowerPad™ 热增强型封装》

《PowerPAD™ 速成》

《TPS25810EVM-745 用户指南》

《TPS25810 高压 DFP 保护》

12.2.2 相关链接

下面的表格中列出了快速访问链接。类别包括技术文档、支持和社区资源、工具和软件，以及立即购买的快速链接。

12.3 接收文档更新通知

要接收文档更新通知，请转至 TI.com 上的器件产品文件夹。单击右上角的*通知我* 进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

12.4 社区资源

下列链接提供到 TI 社区资源的连接。链接的内容由各个分销商“按照原样”提供。这些内容并不构成 TI 技术规范，并且不一定反映 TI 的观点；请参阅 TI 的 《使用条款》。

TI E2E™ 在线社区 ***TI 的工程师对工程师 (E2E) 社区。*** **此社区的创建目的在于促进工程师之间的协作。在 e2e.ti.com 中，您可以咨询问题、分享知识、拓展思路并与同行工程师一道帮助解决问题。**

设计支持 ***TI 参考设计支持*** **可帮助您快速查找有帮助的 E2E 论坛、设计支持工具以及技术支持的联系信息。**

12.5 商标

E2E is a trademark of Texas Instruments.
USB Type-C is a trademark of USB Implementers Forum, Inc..
All other trademarks are the property of their respective owners.

12.6 静电放电警告



这些装置包含有限的内置 ESD 保护。 存储或装卸时，应将导线一起截短或将装置放置于导电泡棉中，以防止 MOS 门极遭受静电损伤。

12.7 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是针对指定器件可提供的最新数据。本数据随时可能发生变更并且不对本文档进行修订，恕不另行通知。要获得这份数据表的浏览器版本，请查阅左侧的导航窗格。

版权 © 2017, Texas Instruments Incorporated

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TPS25810ATWRVCRQ1	Active	Production	WQFN (RVC) | 20	3000 | LARGE T&R	Yes	SN	Level-2-260C-1 YEAR	-40 to 125	25810AQ
TPS25810ATWRVCRQ1.A	Active	Production	WQFN (RVC) | 20	3000 | LARGE T&R	Yes	SN	Level-2-260C-1 YEAR	-40 to 125	25810AQ

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

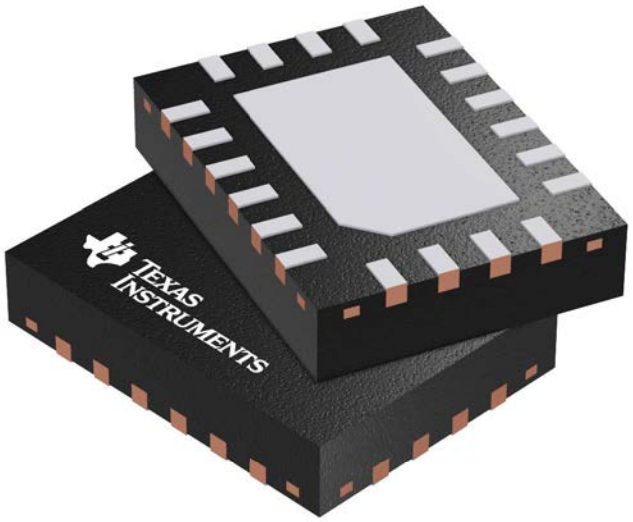
Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

GENERIC PACKAGE VIEW

RVC 20 WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



Images above are just a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.

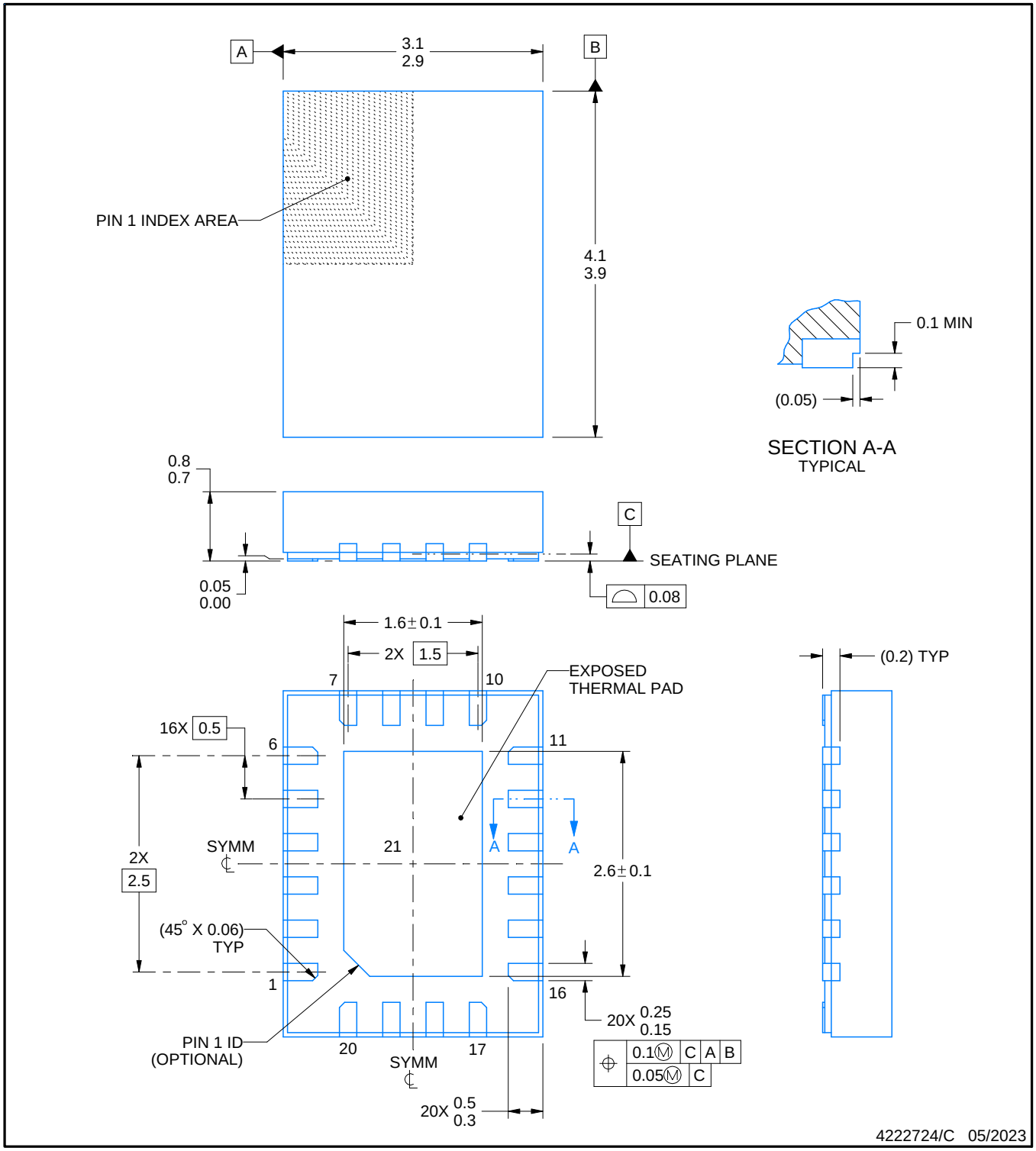
4209819/B

RVC0020B

PACKAGE OUTLINE

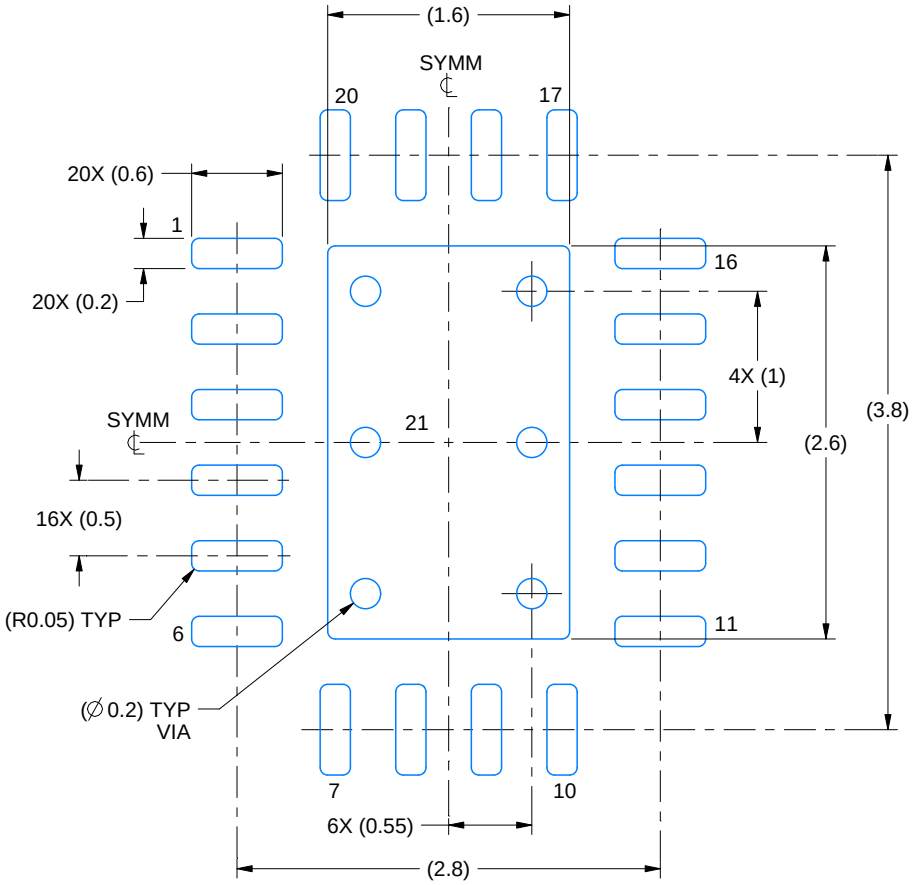
WQFN - 0.8 mm max height

PLASTIC QUAD FLATPACK - NO LEAD

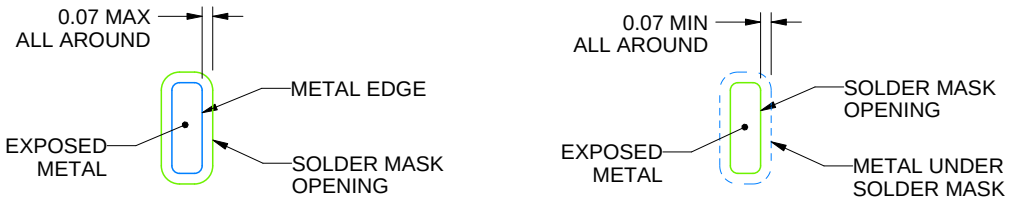


NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:20X



NON SOLDER MASK
DEFINED
(PREFERRED)

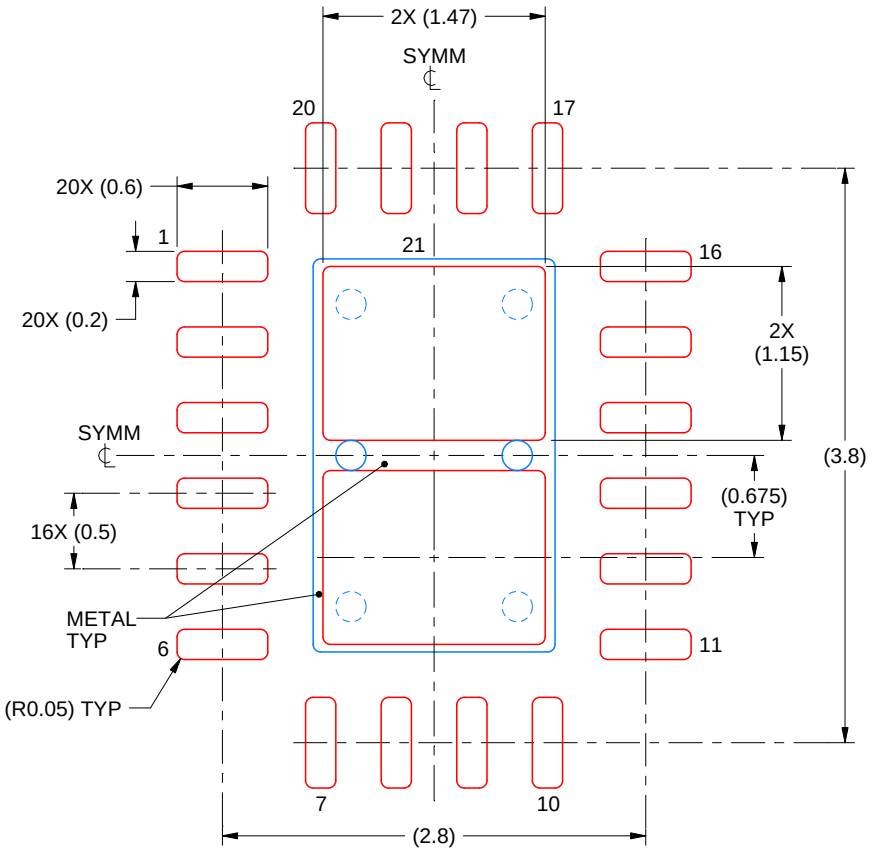
SOLDER MASK
DEFINED

SOLDER MASK DETAILS

4222724/C 05/2023

NOTES: (continued)

4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
5. Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



SOLDER PASTE EXAMPLE
BASED ON 0.1 mm THICK STENCIL

EXPOSED PAD
81.3% PRINTED SOLDER COVERAGE BY AREA
SCALE:20X

4222724/C 05/2023

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

重要通知和免责声明

TI“按原样”提供技术和可靠性数据（包括数据表）、设计资源（包括参考设计）、应用或其他设计建议、网络工具、安全信息和其他资源，不保证没有瑕疵且不做出任何明示或暗示的担保，包括但不限于对适销性、与某特定用途的适用性或不侵犯任何第三方知识产权的暗示担保。

这些资源可供使用 TI 产品进行设计的熟练开发人员使用。您将自行承担以下全部责任：(1) 针对您的应用选择合适的 TI 产品，(2) 设计、验证并测试您的应用，(3) 确保您的应用满足相应标准以及任何其他安全、安保法规或其他要求。

这些资源如有变更，恕不另行通知。TI 授权您仅可将这些资源用于研发本资源所述的 TI 产品的相关应用。严禁以其他方式对这些资源进行复制或展示。您无权使用任何其他 TI 知识产权或任何第三方知识产权。对于因您对这些资源的使用而对 TI 及其代表造成的任何索赔、损害、成本、损失和债务，您将全额赔偿，TI 对此概不负责。

TI 提供的产品受 TI 销售条款)、TI 通用质量指南 或 ti.com 上其他适用条款或 TI 产品随附的其他适用条款的约束。TI 提供这些资源并不会扩展或以其他方式更改 TI 针对 TI 产品发布的适用的担保或担保免责声明。 除非德州仪器 (TI) 明确将某产品指定为定制产品或客户特定产品，否则其产品均为按确定价格收入目录的标准通用器件。

TI 反对并拒绝您可能提出的任何其他或不同的条款。

版权所有 © 2025，德州仪器 (TI) 公司

最后更新日期：2025 年 10 月