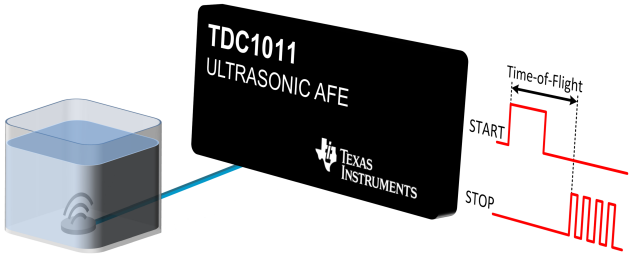
TDC1011 面向液位感测、浓度感测应用的单通道超声波感测模拟前端 (AFE)

1 特性

- 测量范围：高达 8ms
- 工作电流：1.8µA (2 SPS)
- 发送器通道 TX
  - 支持单传感器应用
  - 可编程激励：31.25kHz 至 4MHz，多达 31 个脉冲
- 接收器通道 RX
  - STOP 逐周期抖动：50ps$_{RMS}$
  - 低噪声、可编程增益放大器
  - 可访问外部滤波器的信号链设计
  - 针对回声质检的可编程阈值比较器
  - 针对较长 TOF 测量的可编程低功耗模式
- 温度测量
  - 2 个 PT1000/500 RTD 接口
  - RTD 间的匹配精度为 0.02°C$_{RMS}$
- 工作温度范围：-40°C 至 125°C

2 应用

- 不同材料箱中的各项测量：
  - 液位
  - 液体鉴别/浓度



3 说明

TDC1011 是一款全集成超声波感测模拟前端 (AFE)，常用于汽车、工业、医疗和消费品市场中的液位、液体鉴别/浓度以及接近传感/远距感测应用。 与 MSP430/C2000 MCU、电源、无线网络和源代码配套使用时，TI 可提供完整的超声波感测解决方案。

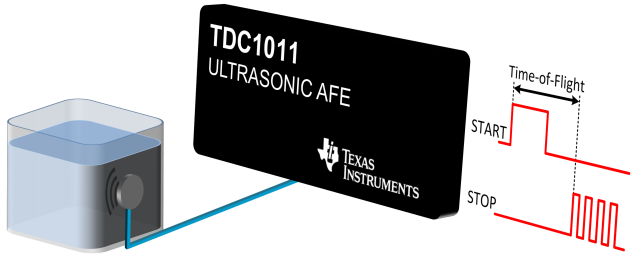
TI 的超声波 AFE 可编程且具有灵活性，可适应广泛的应用和终端设备。 TDC10111 可针对多种发射脉冲和频率、增益和信号阈值进行配置，以便在多种传感器频率（31.25kHz 至 4MHz）和 Q 系数下使用。 同样，接收路径可编程设定，因此在更远的距离/更大的箱体尺寸范围内也能够检测到通过多种介质传播的超声波。

TDC1011 可选择不同的工作模式，并且针对低功耗进行了优化，这使得它成为电池供电应用的理想选择。 低噪声放大器和比较器产生的抖动极低，可实现皮秒级分辨率和精度。

器件信息(1)

器件型号	封装	封装尺寸（标称值）
TDC1011	TSSOP (PW-28)	9.70mm x 4.40mm

(1) 要了解所有可用封装，请见数据表末尾的可订购产品附录。



An IMPORTANT NOTICE at the end of this data sheet addresses availability, warranty, changes, use in safety-critical applications, intellectual property matters and other important disclaimers. PRODUCTION DATA.

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4 修订历史记录

日期	修订版本	注释
2015 年 7 月	*	首次发布。

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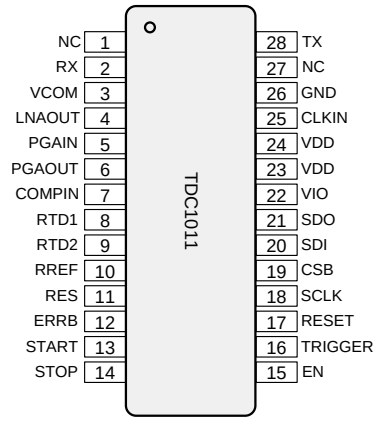
5 Pin Configuration and Functions

TSSOP
(PW) 28 Pin
Top View



Pin Functions

PIN		TYPE(1)	DESCRIPTION
NAME	NO.		
NC	1		No Connect (leave floating)
RX	2	I	Receive input
VCOM	3	P	Output common mode voltage bias
LNAOUT	4	O	Low noise amplifier output (for ac decoupling capacitor)
PGAIN	5	I	Programmable gain amplifier input
PGAOUT	6	O	Programmable gain amplifier output
COMPIN	7	I	Echo qualification and zero-crossing detector input
RTD1	8	O	Resistance temperature detector channel 1
RTD2	9	O	Resistance temperature detector channel 2
RREF	10	O	Reference resistor for temperature measurement
RES	11	I	Reserved (connect to GND)
ERRB	12	O	Error flag (open drain)
START	13	O	Start pulse output
STOP	14	O	Stop pulse output
EN	15	I	Enable (active high; when low the TDC1011 is in SLEEP mode)
TRIGGER	16	I	Trigger input
RESET	17	I	Reset (active high)
SCLK	18	I	Serial clock for the SPI interface
CSB	19	I	Chip select for the SPI interface (active low)
SDI	20	I	Serial data input for the SPI interface
SDO	21	O	Serial data output for the SPI interface
VIO	22	P	Positive I/O supply
VDD	23, 24	P	Positive supply; all VDD supply pins must be connected to the supply. Place a 100-nF bypass capacitor to ground in close proximity to the pin.
CLKIN	25	I	Clock input
GND	26	G	Negative supply
NC	27		No Connect (leave floating)
TX	28	O	Transmit output

(1) G = Ground, I = Input, O = Output, P = Power

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6 Specifications

6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted) (1)(2)

		MIN	MAX	UNIT
V_{DD}	Analog supply voltage, VDD pins	–0.3	6.0	V
V_{IO}	I/O supply voltage (V_{IO} must always be lower than or equal to V_{DD} supply)	–0.3	6.0	V
V_I	Voltage on any analog input pin(3)	–0.3	V_{DD} + 0.3	V
V_I	Voltage on any digital input pin(3)	–0.3	V_{IO} + 0.3	V
I_I	Input current at any pin		5	mA
T_J	Operating junction temperature	–40	125	°C
T_{stg}	Storage temperature range	–65	150	°C

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
(2) All voltages are with respect to ground, unless otherwise specified.
(3) When the input voltage at a pin exceeds the power supplies, the current at that pin must not exceed 5 mA and the voltage (V_I) at that pin must not exceed 6.0 V.

6.2 ESD Ratings

				VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC A100-002(1)		±2000	V
		Charged-device model (CDM), per AEC Q100-011	All pins	±500	
			Corner pins (1, 14, 15 and 28)	±750	

(1) AEC Q100-002 indicates HBM stressing is done in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V_{DD}	Analog supply voltage, VDD pins	2.7	5.5	V
V_{IO}	Digital supply voltage, (V_{IO} must always be lower than or equal to V_{DD} supply)	1.8	V_{DD}	V
V_I	Voltage on any analog input pin	GND	V_{DD}	V
V_I	Voltage on any digital input pin	GND	V_{IO}	V
f_{CLKIN}	Operating frequency	0.06	16	MHz
T_J	Operating junction temperature	–40	125	°C

6.4 Thermal Information(1)

THERMAL METRIC		TDC1011 TSSOP PW (28 PINS)	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	83.5	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	29.9	
$R_{\theta JB}$	Junction-to-board thermal resistance	40.8	
ψ_{JT}	Junction-to-top characterization parameter	2.4	
ψ_{JB}	Junction-to-board characterization parameter	40.3	

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, SPRA953.

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6.5 Electrical Characteristics

The electrical ratings specified in this section apply to all specifications in this document, unless otherwise noted. These specifications are interpreted as conditions that do not degrade the device parametric or functional specifications for the life of the product containing it. T_A = 25°C, V_{DD} = V_{IO} = 3.7 V, V_{COM} = V_{CM} = V_{DD} / 2, C_{VCOM} = 10 nF (unless otherwise noted).

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
TRANSMITTER SIGNAL PATH (TX)							
$V_{OUT(TX)}$	Output voltage swing	f_{out} = 1 MHz, R_L = 75 Ω to V_{CM}	HIGH		V_{DD} – 0.32		V
			LOW		0.32		V
$I_{OUT(TX)}$	Output drive current	f_{out} = 1 MHz, R_L = 75 Ω to V_{CM}			22		mA_{RMS}
$f_{OUT(TX)}$	Output TX frequency	f_{CLKIN} = 8 MHz, divide-by-2 (programmable; see *Transmitter Signal Path*)			4		MHz
RECEIVER SIGNAL PATH (RX)							
Δt_{STOP}	STOP cycle-to-cycle jitter	LNA capacitive feedback, G_{PGA} = 6 dB, f_{IN} = 1 MHz, V_{IN} = 100 mV_{PP}, C_{VCOM} = 1 µF and Figure 14			50		ps_{RMS}
LNA							
G_{LNA}	LNA gain	Capacitive feedback, C_{IN} = 300 pF, f_{IN} = 1 MHz, R_L = 100 kΩ to V_{CM}, C_{VCOM} = 1 µF			20		dB
en_{LNA}	LNA input referred noise density	Capacitive feedback, C_{IN} = 300 pF, f = 1 MHz, V_{DD} = 3.1 V, V_{IN} = V_{CM}, R_L = ∞, C_{VCOM} = 1 µF			2		$nV/\sqrt{Hz}$
$V_{IN(LNA)}$	Input voltage range	Resistive feedback, R_L = 1 kΩ to V_{CM}, C_{VCOM} = 1 µF	HIGH		V_{CM} + (V_{CM} – 0.24) / (G_{LNA})		V
			LOW		V_{CM} – (V_{CM} – 0.24) / (G_{LNA})		V
$V_{OUT(LNA)}$	Output voltage range	Resistive feedback, R_L = 1 kΩ to V_{CM}, C_{VCOM} = 1 µF	HIGH		V_{DD} – 0.24		V
			LOW		GND + 0.24		V
SR_{LNA}	Slew rate(1)	Resistive feedback, R_L = 1 kΩ to V_{CM}, 100mV step, C_{VCOM} = 1 µF			9		V/µs
BW_{LNA}	–3-dB bandwidth	Capacitive feedback, C_{IN} = 300 pF, R_L= 100 kΩ to V_{CM}, C_{VCOM} = 1 µF			5		MHz
$V_{OS(LNA)}$	LNA input offset voltage	Resistive mode, V_{IN} = V_{CM}, R_L = ∞			±320		µV
VCOM							
V_{COM}	VCOM output voltage	C_{VCOM} = 1 µF			V_{CM}		V
	VCOM output error				0.5%		
PGA							
$V_{IN(PGA)}$	PGA input range	R_L = 100 kΩ to V_{CM}, C_L = 10 pF to GND	HIGH		V_{CM} + (V_{CM} – 0.06) / (G_{PGA})		V
			LOW		V_{CM} – (V_{CM} – 0.06) / (G_{PGA})		V
G_{PGAMIN}	PGA min gain	DC, R_L = ∞, C_L = 10 pF			0		dB
G_{PGAMAX}	PGA max gain				21		dB
ΔG_{PGA}	PGA gain step size				3		dB
$G_{E(PGA)}$	PGA gain error	DC, G_{PGA} = 0 dB, R_L = ∞, C_L = 10 pF			5%		
TCG_{PGA}	PGA gain temperature coefficient	DC, G_{PGA} = 0 dB, R_L = ∞, C_L = 10 pF			170		ppm/°C
en_{PGA}	PGA input referred noise density	G_{PGA} = 21 dB, f = 1 MHz, V_{DD} = 3.1V, V_{IN} = V_{CM}, R_L = ∞, C_{VCOM} = 1 µF			3.1		$nV/\sqrt{Hz}$
$V_{OUT(PGA)}$	Output range	R_L = 100 kΩ to V_{CM}, C_L = 10 pF to GND	HIGH		V_{DD} – 0.06		V
			LOW		60		mV
BW_{PGA}	–3-db bandwidth	G_{PGA} = 21 dB, R_L = 100 kΩ to V_{CM}, C_L = 10 pF, C_{VCOM} = 1 µF			5		MHz
SR_{PGA}	Slew rate(1)	G_{PGA} = 21 dB, R_L = 100 kΩ to V_{CM}, C_L = 10 pF, C_{VCOM} = 1 µF			12.5		V/µs

(1) The slew rate is measured from 10% to 90% and is represented by the average of the rising and falling slew rates.

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Electrical Characteristics (continued)

The electrical ratings specified in this section apply to all specifications in this document, unless otherwise noted. These specifications are interpreted as conditions that do not degrade the device parametric or functional specifications for the life of the product containing it. T_A = 25°C, V_{DD} = V_{IO} = 3.7 V, V_{COM} = V_{CM} = V_{DD} / 2, C_{VCOM} = 10 nF (unless otherwise noted).

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ZERO CROSS COMPARATOR						
$V_{OS(COMP)}$	Input offset voltage (2)	Referred to V_{COM}		±115		µV
en_{COMP}	Zero crossing comparator input referred noise (2)	1 MHz		5		nV/√Hz
$HYST_{COMP}$	Hysteresis (2)	Referred to V_{COM}		-10		mV
THRESHOLD DETECTOR						
V_{THDET}	Threshold level	ECHO_QUAL_THLD = 0h, V_{COM} referred		–35		mV
		ECHO_QUAL_THLD = 7h, V_{COM} referred		–1.5		V
TEMPERATURE SENSOR INTERFACE (3)						
T_{ERROR}	Temperature measurement accuracy	R_{REF} = 1 kΩ, PT1000 range: –40 to 125°C (4)		1		°C
		R_{REF} = 1 kΩ, PT1000 range: –15°C to 85°C (4)		0.5		°C
	Relative accuracy	R_{REF} = 1 kΩ, R_{RTD1} = R_{RTD2} = 1.1 kΩ		0.02		$°C_{RMS}$
TG_E	Gain error			5.8		m°C/°C
POWER SUPPLY						
I_{DD}	VDD supply current	Sleep (EN = CLKIN = TRIGGER = low)		0.61		µA
		Continuous receive mode, LNA and PGA bypassed		2.8	3	mA
		Continuous receive mode, LNA and PGA active		6.2	7.5	mA
		Temp. measurement only (PT1000 mode) (5)		370	400	µA
		Temp. measurement (PT500 mode) (6)		500	540	µA
I_{IO}	VIO supply sleep current (2)	Sleep (EN = CLKIN = TRIGGER = low)		2		nA
DIGITAL INPUT/OUTPUT CHARACTERISTICS						
V_{IL}	Input logic low threshold				0.2 × V_{IO}	V
V_{IH}	Input logic high threshold		0.8 × V_{IO}			V
V_{OL}	Output logic low threshold	SDO pin, 100-µA current			0.2	V
		SDO pin, 1.85-mA current			0.4	V
		START and STOP pins, 100-µA current			0.5	V
		START and STOP pins, 1.85-mA current			0.6	V
		ERRB pin, 100-µA current			0.2	V
		ERRB pin, 1.85-mA current			0.4	V
V_{OH}	Output logic high threshold	SDO pin, 100-µA current	V_{IO} – 0.2			V
		SDO pin, 1.85-mA current	V_{IO} – 0.6			V
		START and STOP pins, 100-µA current	V_{IO} – 0.5			V
		START and STOP pins, 1.85-mA current	V_{IO} – 0.6			V
		ERRB pin, 0-µA current	V_{IO} – 0.2			V
I_{OMAX}	Maximum output current for SDO, START and STOP			1.85		mA

(2) Specified by design.
(3) With ideal external components. For more detail see *Temp Sensor Measurement* section.
(4) PT1000 RTD approximate resistance: 800 Ω ≡ –52°C, 931 Ω ≡ –18°C, 1.10 kΩ ≡ 26°C, 1.33 kΩ ≡ 86°C and 1.48 kΩ ≡ 125°C.
(5) Specified currents include 120µA which flows through the RTD sensor in PT1000 mode (TEMP_RTD_SEL = 0).
(6) Specified currents include 240µA which flows through the RTD sensor in PT500 mode (TEMP_RTD_SEL = 1).

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6.6 Timing Requirements

T_A = 25°C, V_{DD} = V_{IO} = 3.7 V and f_{SCLK} = 1 MHz (unless otherwise noted).

		MIN	NOM	MAX	UNIT
f_{SCLK}	Serial clock frequency			26	MHz
t_1	High period, SCLK	16			ns
t_2	Low period, SCLK	16			ns
t_3	Set-up time, nCS to SCLK	10			ns
t_4	Set-up time, SDI to SCLK	12			ns
t_5	Hold time, SCLK to SDI	12			ns
t_6	SCLK transition to SDO valid time	16			ns
t_7	Hold time, SCLK transition to nCS rising edge	10			ns
t_8	nCS inactive	17			ns
t_9	Hold time, SCLK transition to nCS falling edge	10			ns
t_r / t_f	Signal rise and fall times(1)		1.8		ns

(1) The slew rate is measured from 10% to 90% and is represented by the average of the rising and falling slew rates.

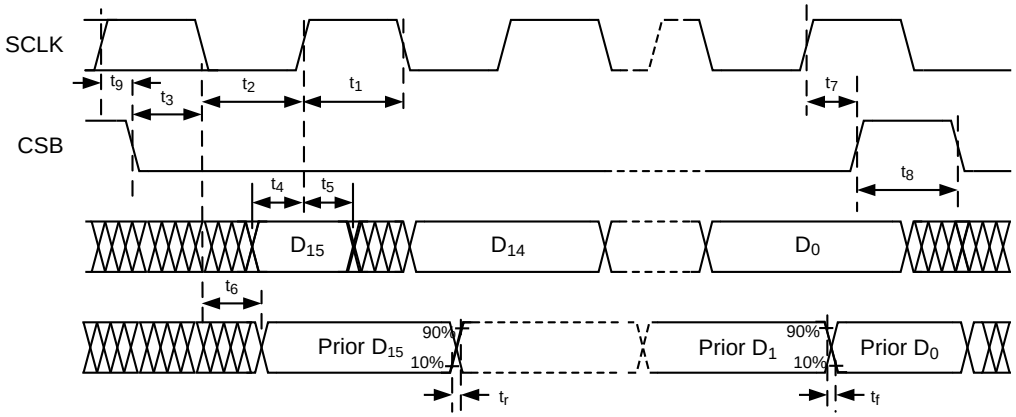

Figure 1. SPI Timing Diagram

6.7 Switching Characteristics

T_A = 25°C, V_{DD} = V_{IO} = 3.7 V, f_{CLKIN} = 8 MHz.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
START, STOP, ENABLE, RESET, CLOCKIN, TRIGGER, ERR						
PW_{START}	Pulse width for START signal	TX_FREQ_DIV = 2h, NUM_TX = 1		1		µs
		TX_FREQ_DIV = 2h, NUM_TX = 2		2		µs
		TX_FREQ_DIV = 2h, NUM_TX ≥ 3		3		µs
t_r / $t_{f\ START}$	Rise/fall time for START signal	20% to 80%, 20-pF load		0.25		ns
t_r / $t_{f\ STOP}$	Rise/fall time for STOP signal	20% to 80%, 20-pF load		0.25		ns
f_{CLKIN}	Maximum CLKIN input frequency			16		MHz
t_r / $t_{f\ CLKIN}$	CLKIN input rise/fall time(1)	20% to 80%		10		ns
t_r / $t_{f\ TRIG}$	TRIGGER input rise/fall time(1)	20% to 80%		10		ns
t_{EN_TRIG}	Enable to trigger wait time(1)			50		ns
t_{RES_TRIG}	Reset to trigger wait time(1)	TX_FREQ_DIV = 2h (see *TX/RX Measurement Sequencing and Timing*)		3.05		µs

(1) Specified by design.

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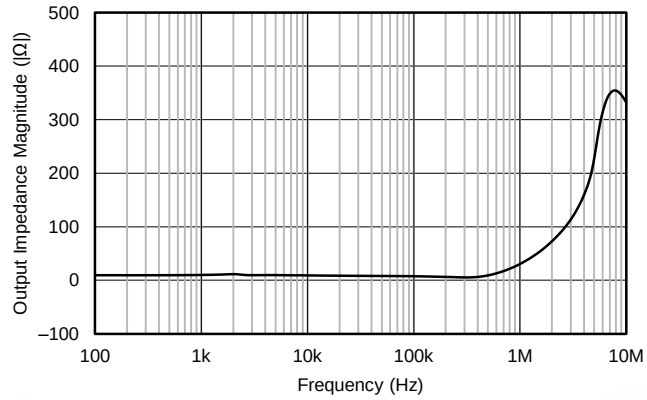
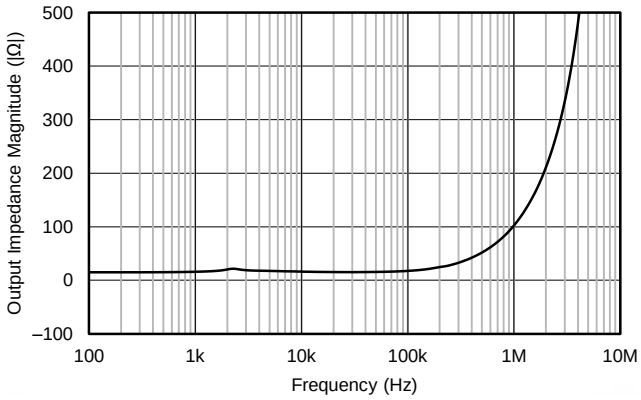
6.8 Typical Characteristics

At T_A = 25°C, unless otherwise noted.



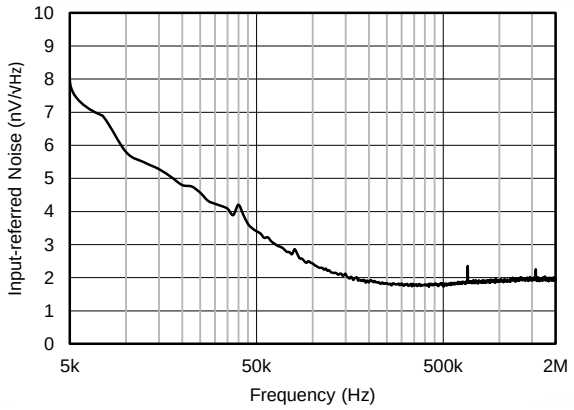
V_{DD} = V_{IO} = 3.7V | Capacitive Feedback Mode | R_L = 1kΩ

Figure 2. LNA Z_{OUT} vs Frequency



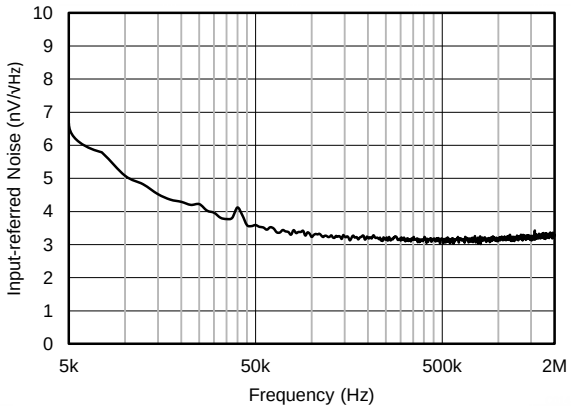
V_{DD} = V_{IO} = 3.7V | Gain of 21dB | R_L = 1kΩ

Figure 3. PGA Z_{OUT} vs Frequency



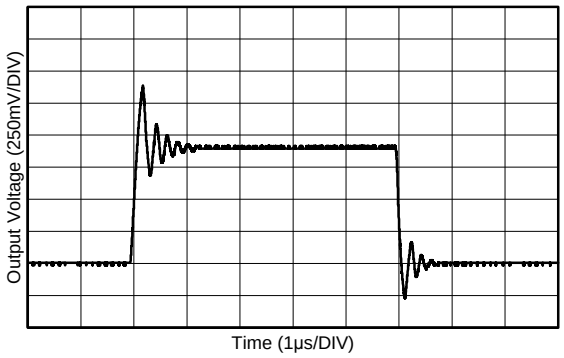
V_{DD} = V_{IO} = 3.1V | Capacitive Feedback Mode | R_L = ∞

Figure 4. LNA Input-referred Noise vs Frequency



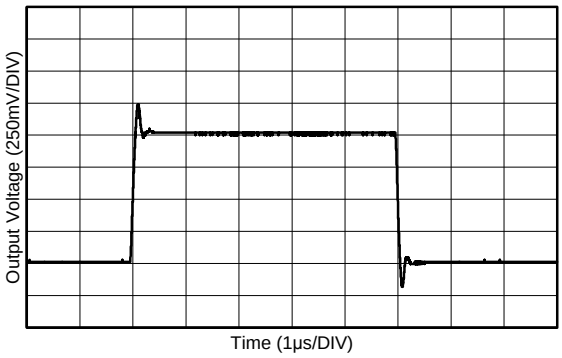
V_{DD} = V_{IO} = 3.7V | Gain of 21dB | R_L = ∞

Figure 5. PGA Input-referred Noise vs Frequency



V_{DD} = V_{IO} = 3.7V
V_{IN} = 100mV
Resistive Feedback Mode
R_L = 1kΩ
f_{IN} = 100kHz

Figure 6. LNA Response



V_{DD} = V_{IO} = 3.7V
V_{IN} = 100mV
Gain of 21dB
R_L = 100kΩ
f_{IN} = 100kHz

Figure 7. PGA Response

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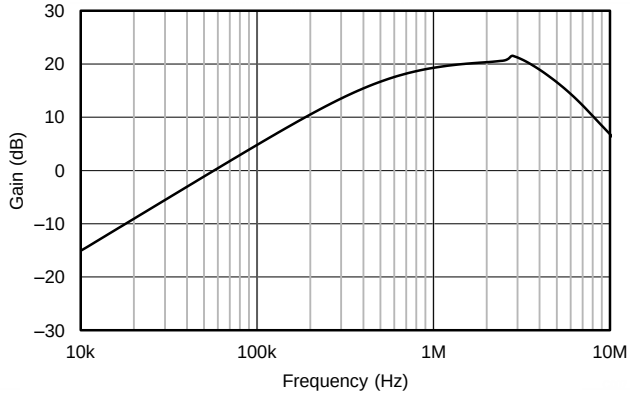
Typical Characteristics (continued)

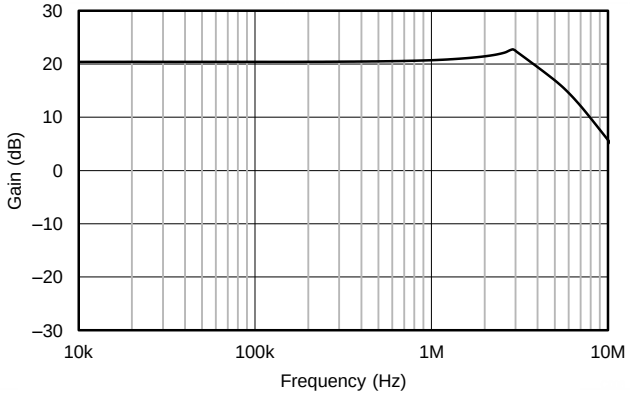
At T_A = 25°C, unless otherwise noted.



V_{DD} = V_{IO} = 3.7V | Capacitive Feedback Mode | R_L = 100kΩ

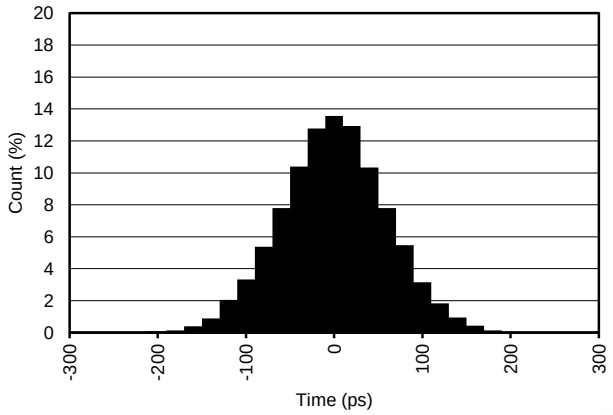
C_{IN} = 300 pF

Figure 8. LNA Gain vs Frequency



V_{DD} = V_{IO} = 3.7V | Gain of 21dB | R_L = 100kΩ

Figure 9. PGA Gain vs Frequency

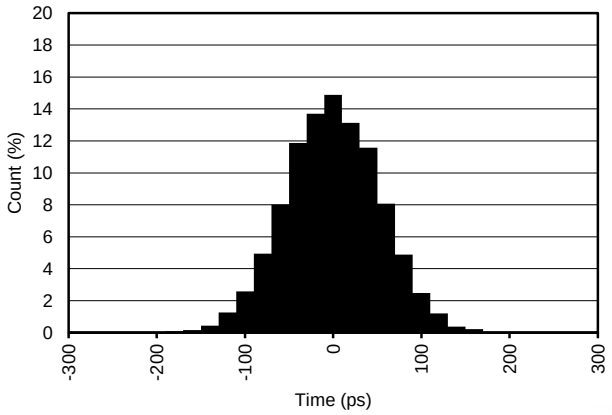


V_{DD} = V_{IO} = 5V
V_{IN} = 100mV

LNA Capacitive Feedback Mode (See Figure 14)

PGA Gain of 6dB
f_{IN} = 1MHz
Count >= 10000

Figure 10. RX Jitter Histogram

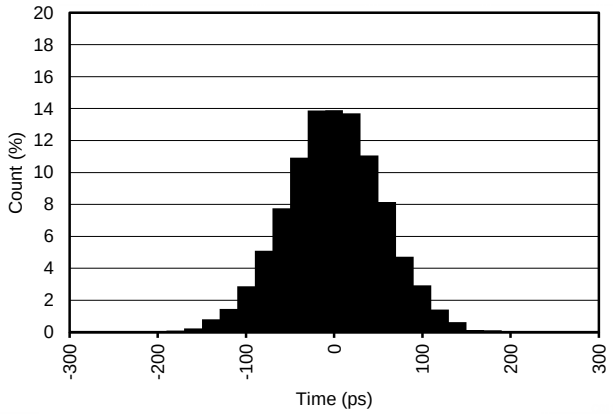


V_{DD} = V_{IO} = 3.7V
V_{IN} = 100mV
T_A = 25C°

LNA Capacitive Feedback Mode (See Figure 14)

PGA Gain of 6dB
f_{IN} = 1MHz
Count >= 10000

Figure 11. RX Jitter Histogram

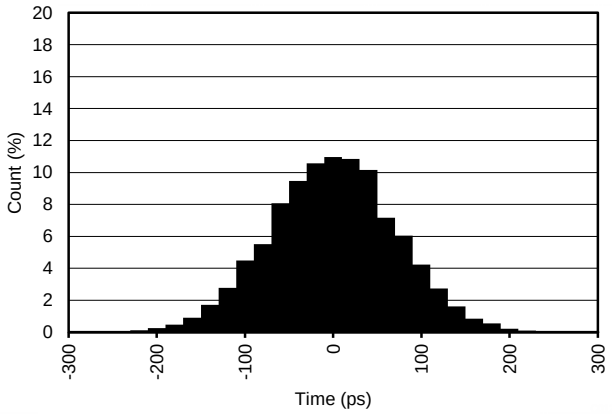


V_{DD} = V_{IO} = 3.7V
V_{IN} = 100mV
T_A = -40C°

LNA Capacitive Feedback Mode (See Figure 14)

PGA Gain of 6dB
f_{IN} = 1MHz
Count >= 10000

Figure 12. RX Jitter Histrogam



V_{DD} = V_{IO} = 3.7V
V_{IN} = 100mV
T_A = 125C°

LNA Capacitive Feedback Mode (See Figure 14)

PGA Gain of 6dB
f_{IN} = 1MHz
Count >= 10000

Figure 13. RX Jitter Histogram

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7 Parameter Measurement Information

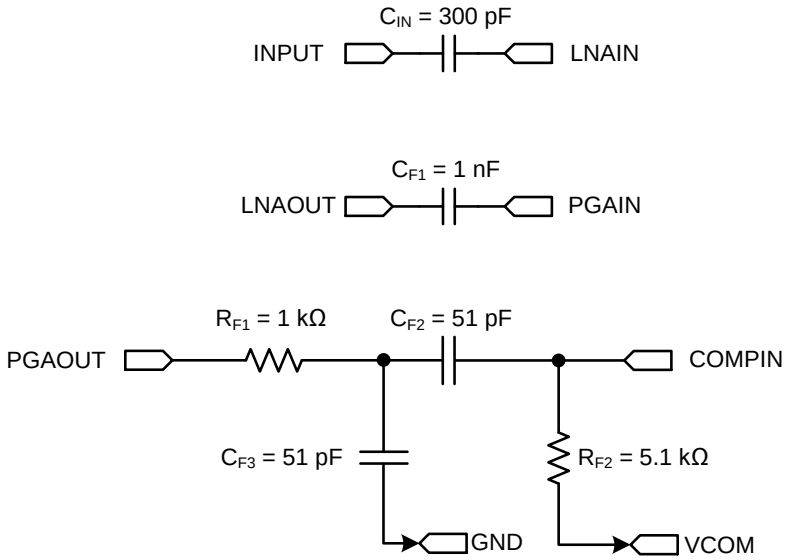

Figure 14. External Circuits for Jitter Measurement

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8 Detailed Description

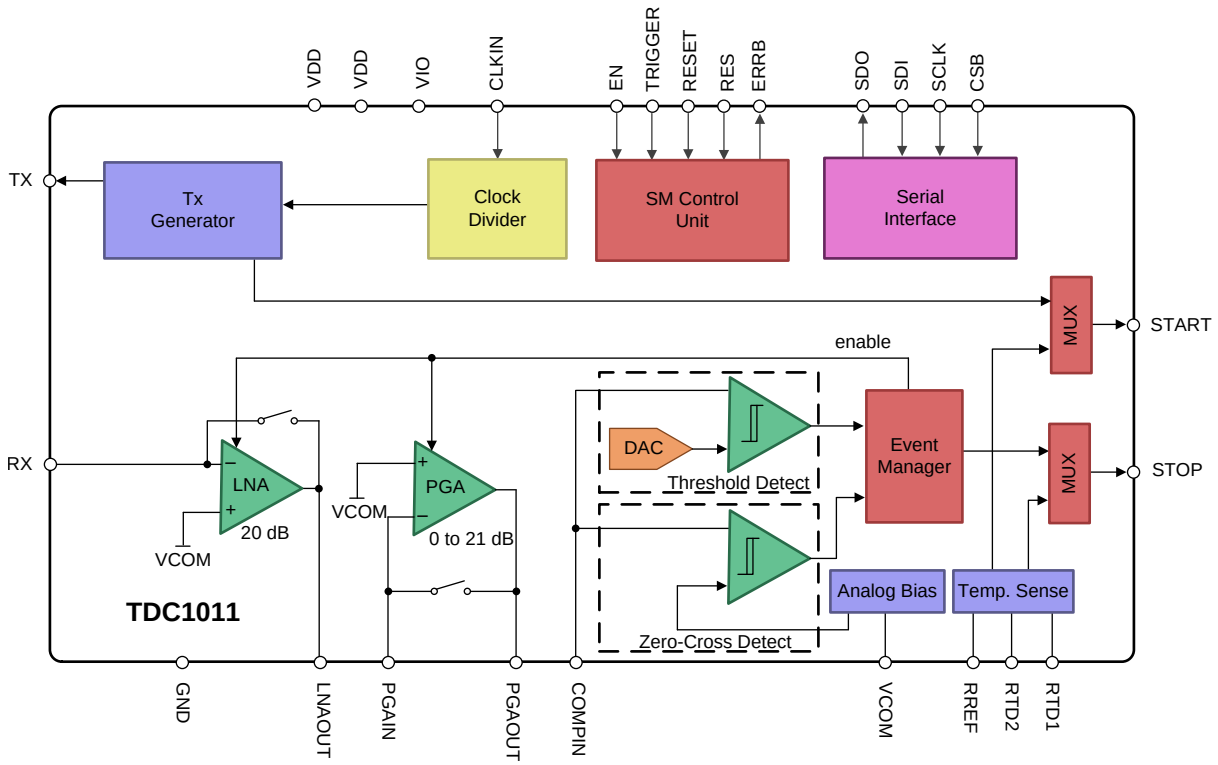
8.1 Overview

The main functional blocks of TDC1011 are the Transmit (TX) and the Receive (RX) Channels. The transmitter supports flexible settings for driving various ultrasonic transducers, and the receiver provides configurable blocks with a wide range of settings for signal conditioning in various applications. The receive signal chain consists of an LNA (Low Noise Amplifier), a PGA (Programmable Gain Amplifier), and two auto-zeroed comparators for echo qualification and STOP pulse generation.

A measurement cycle is initiated with a trigger signal on the TRIGGER pin of the device. After a trigger signal is asserted, an output pulse is generated on the START pin. This signal is used as the time reference to begin a TOF measurement. The transmitter generates programmable TX pulses, synchronous to the rising edge of the START pulse, to drive an ultrasonic transducer and generate an ultrasonic wave that is shot through an acoustic medium. The receiver detects the ultrasound wave that traveled through the medium and generates STOP signals. Whether the ultrasound wave is received directly or from a reflection will depend on the system configuration. The STOP signals are used by an external Time-to-Digital Converter (TDC), which functions as a very accurate stopwatch. The system must include a TDC to measure the TOF based on the interval between the START and STOP pulses. In some applications with medium-range accuracy requirements (ns range), a microcontroller can be used to measure the TOF duration. In applications with high-range accuracy requirements (ps range), TI recommends using the TDC7200 time-to-digital converter to measure the TOF duration.

In each application, the TDC1011 has to be configured by a serial interface (SPI) for the various application-specific parameters that are explained in the following sections.

8.2 Functional Block Diagram



8.3 Feature Description

8.3.1 Transmitter Signal Path

The Transmitter (TX) path consists of a Clock Divider block and a TX Generator block. The clock divider allows the TDC1011 to divide the clock source that is connected to the CLKIN pin down to the resonant frequency (f_R) of the transducer used. The clock divider allows division factors in powers of 2. The division factor of the clock divider can be programmed with the *TX_FREQ_DIV* field in the *CONFIG_0* register.

The TX Generator block can drive a transducer with a programmable number of TX pulses. The frequency of these pulses is defined as $f_{CLKIN}/(2^{TX_FREQ_DIV+1})$, and should match the f_R of the transducer. The number of pulses is configured by programming the *NUM_TX* field in the *CONFIG_0* register.

For example, for f_{CLKIN} = 8 MHz and TX_FREQ_DIV = 2h (divide by 8), the divided clock frequency is 1 MHz.

In addition to the programmable number of pulses, the TX Generator also provides options to introduce a 180º pulse shift at pulse position n or damping the last TX pulse. In some situations, damping can reduce the ringing of the transducer for very short TOF measurements. These features are further described in the *TRANSMIT Operation* section of the datasheet.

8.3.2 Receiver Signal Path

The RX signal path consists of an LNA, PGA, and a pair of comparators. The LNA and PGA provide the required amplification of the receive signal. The amplified receive signal is fed into a set of comparators which generate pulses on the STOP pin based on the programmed threshold levels. The block diagram for the receiver path can be seen in Figure 15.

If the 20-dB to 41-dB of gain provided by the TDC1011 is insufficient, additional gain can be added prior to the COMPIN pin. Likewise, with a strong received signal, if the gain from the LNA or PGA is not needed, they can be bypassed and the transducer signal can be directly connected to the COMPIN pin.

A band-pass filter centered on the transducer's response can be used between each stage of the receiver path to reduce the noise; note that the inputs of the LNA, PGA, and comparators should be biased to the VCOM pin's potential. The comparators connected to the COMPIN pin are used for echo qualification and generation of STOP pulses that correspond to the zero-crossings of the echo signal. The STOP pulses are used with a START pulse to calculate the TOF of the echo in the medium.

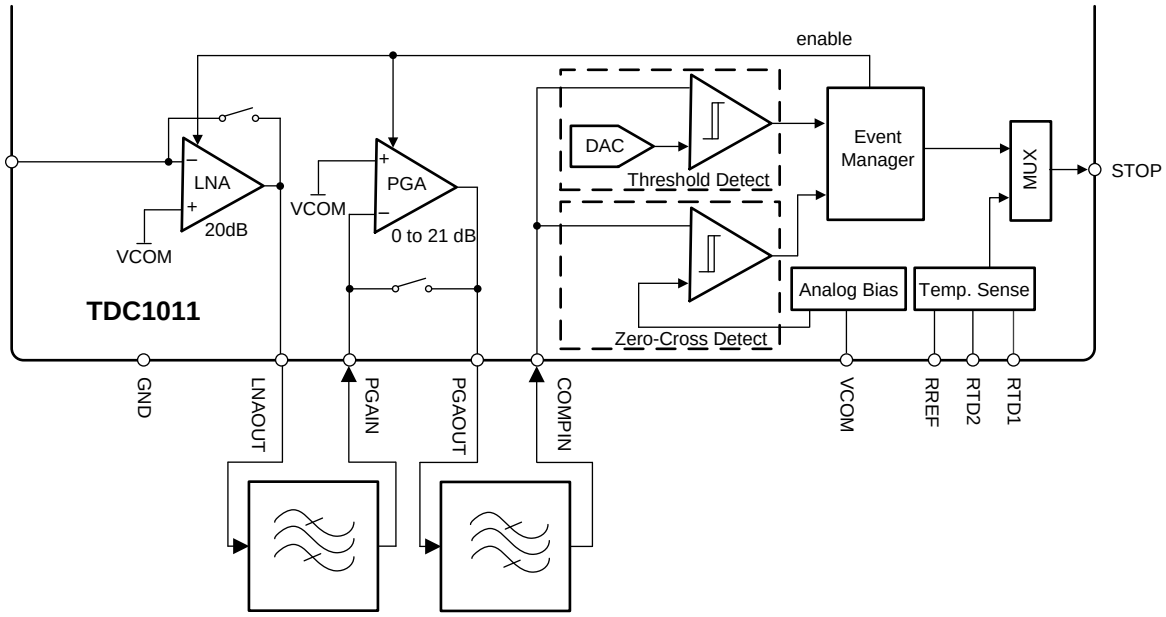


Figure 15. TDC1011 Receiver Path

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Feature Description (continued)

8.3.3 Low Noise Amplifier (LNA)

The LNA in the TDC1011's front-end limits the input-referred noise and ensures timing accuracy for the generated STOP pulses. The LNA is an inverting amplifier designed for a closed-loop gain of 20 dB with the aid of an external input capacitor or resistor, and it can be programmed for two feedback configurations. The band-pass configuration, referred to as capacitive feedback mode, must be combined with an input capacitor. The low-pass configuration, referred to as resistive feedback mode, must be combined with an input resistor. The recommended values for the input components are 300 pF and 900 Ω respectively.

The LNA can be configured in capacitive feedback mode for transducers with resonant frequencies in the order of a couple of MHz. This is done by clearing the *LNA_FB* bit in the TOF_1 register to 0. As shown in Figure 16, the external capacitor, C_{IN}, should be placed between the transducer and the input pin. This provides an in-band gain of C_{IN}/C_F, where C_F is the on-chip 30-pF feedback capacitor. Provided that C_{IN} = 300 pF, the in-band gain of the LNA circuit is:

$$\text{Gain}_{\text{in-band}} = \frac{C_{IN}}{C_F} = \frac{300\ \text{pF}}{30\ \text{pF}} = 10 \quad (1)$$

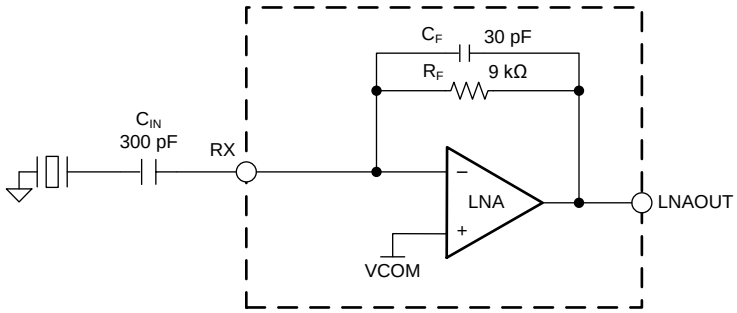

Figure 16. LNA Capacitive Feedback Configuration

The capacitive feedback configuration of the LNA has a band-pass frequency response. The high-pass corner frequency is set by the internal feedback components R_F (9 kΩ) and C_F (30 pF), and is approximately 590 kHz. The in-band gain is set by the capacitor ratio and the LNA's 50-MHz gain-bandwidth product sets the low-pass corner of the frequency response. For example, an in-band gain of 10 results in a bandpass response between 590 kHz and 5 MHz.

The LNA can be configured in resistive feedback mode for transducers with resonant frequencies in the order of a couple of hundreds of kHz. This is done by setting the *LNA_FB* bit in the TOF_1 register to 1. In this configuration, the internal feedback capacitor C_F is disconnected (see Figure 17), and the DC gain of the LNA circuit is determined by the ratio between the internal feedback resistor R_F (9 kΩ) and an external resistor R_{IN}. For R_{IN} = 900 Ω, the gain of the circuit is 10.

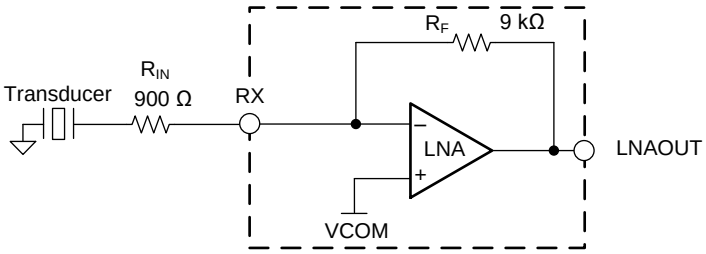

Figure 17. LNA Resistive Feedback Configuration

The LNA can be bypassed and disabled by writing a 1 to the *LNA_CTRL* bit in the TOF_1 register.

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Feature Description (continued)

8.3.4 Programmable Gain Amplifier (PGA)

The PGA, shown in Figure 18, is an inverting amplifier with an input resistance of R_{IN} = 500 Ω and a programmable feedback resistor R_{FB} that can be programmed to set a 0-dB to 21-dB gain in 3-dB steps. This can be done by programming the *PGA_GAIN* field in the TOF_1 register. The bandwidth of the PGA is scaled based on its programmed gain. The typical bandwidth of the PGA with a 100-kΩ load to VCM and a 10-pF capacitor to ground are listed in Table 1.

Table 1. Typical PGA Bandwidth

PGA_GAIN (Hex)	Gain (dB)	Bandwidth (MHz)
0h	0	19.0
1h	3	16.8
2h	6	14.4
3h	9	12.3
4h	12	10.0
5h	15	8.2
6h	18	6.6
7h	21	5.0

The PGA can be bypassed and disabled by writing a 1 to the *PGA_CTRL* bit in the TOF_1 register. The output of the PGA should not be loaded directly with capacitances greater than 10 pF.

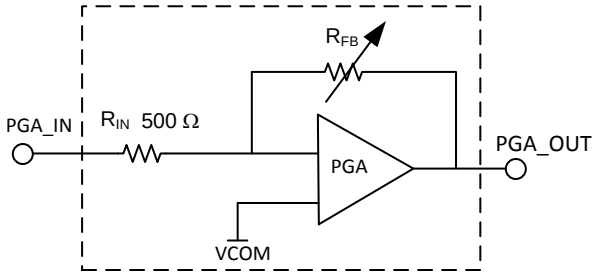

Figure 18. TDC1011 Programmable Gain Amplifier

8.3.5 Receiver Filters

It is recommended to place two filters in the RX path to minimize the receive path noise and obtain maximum timing accuracy. As shown in Figure 19, one filter is placed between the LNAOUT and the PGAIN pins, and another filter is placed between the PGAOUT and the COMPIN pins.

With an in-band gain of 10, the LNA has a bandwidth of 5 MHz. For most applications, a low-pass filter between the LNAOUT and PGAIN pins is sufficient.

As shown in Figure 19, the second filter stage can use a cascade of a low-pass filter (R_{F1} and C_{F3}) followed by a high-pass filter (C_{F2} and R_{F2}) referenced to VCOM. Design of the filter is straightforward. The R_{F1} and C_{F2} can be chosen first. A reasonable set of values for R_{F1} and C_{F2} could be: R_{F1} = 1 kΩ ± 10% and C_{F2} = 50pF ± 10%. Given the center frequency of interest to be f_C and the filter bandwidth to be f_B, the value of C_{F3} can be calculated as:

$$C_{F3} = \frac{1}{2\pi R_{F1}\left(f_C + f_B\right)} \quad (2)$$

R_{F2} and C_{F2} determine the high-pass corner of the filter. R_{F2} should be referenced to VCOM to maintain the DC bias level at the comparator input during the echo receive time. For values of R_{F2} larger than R_{F1}, there will be limited loading effect from the high-pass filter to the low-pass filter resulting in more accurate corner frequencies. The chosen values shown in the figure below result in a high-pass corner frequency of about 600 kHz and a low-pass corner frequency of about 3 MHz.

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More complex filters can be used; external gain is acceptable if the signal amplitude is too low. If the pass-band of the filter is wider than an octave, it is recommended to use a filter design which has linear group delay.

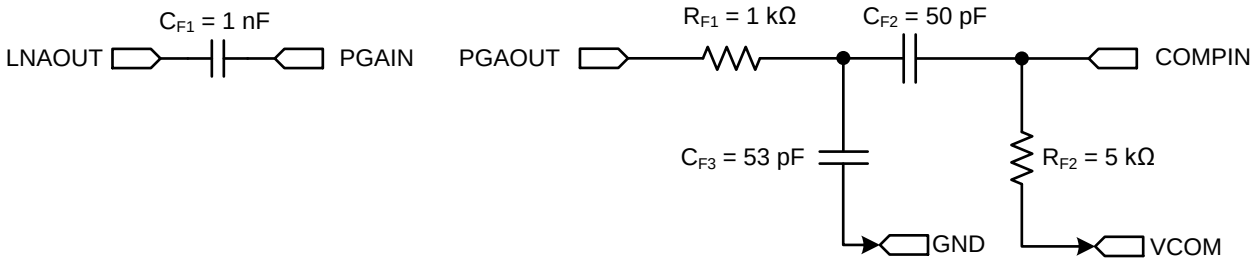

Figure 19. Filter for a 1-MHz Operation

8.3.6 Comparators for STOP Pulse Generation

The STOP pulse generation block of the TDC1011 contains two auto-zeroed comparators (a zero-cross detect and a threshold-detect comparator), a threshold setting DAC, and an event manager.

Comparator auto-zero periods occur at the beginning of every TOF receive cycle. During these periods, the comparator's input offset is stored in an internal 2.5-pF capacitor, and it is subtracted from the input signal during the echo processing phase. The duration of auto-zero period is configured with the *AUTOZERO_PERIOD* field located in the *CLOCK_RATE* register.

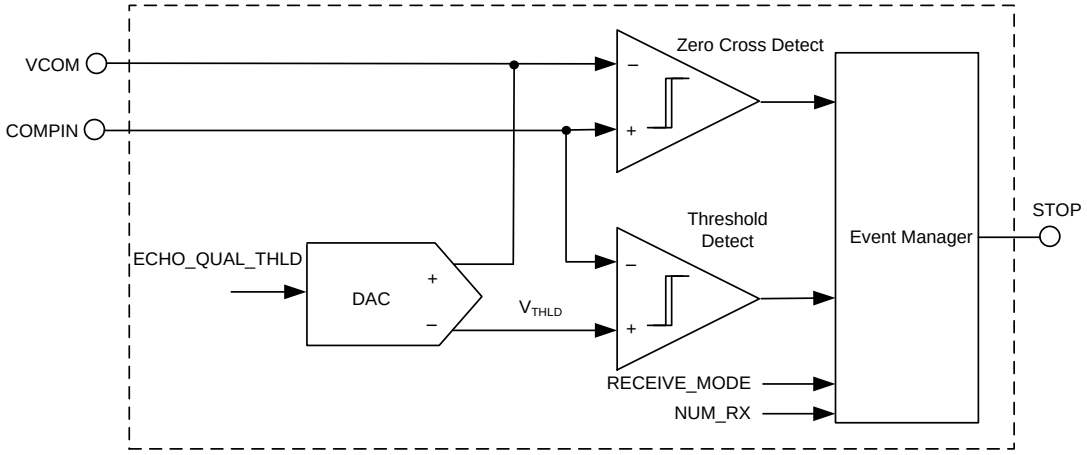

Figure 20. STOP Pulse Generation Circuit

8.3.6.1 Threshold Detector and DAC

The threshold detect comparator in Figure 20 compares the echo amplitude with a programmable threshold level (V_{THLD}) controlled by a DAC. The DAC voltage is set by *the ECHO_QUAL_THLD* field in register CONFIG_3 and provides 8 programmable threshold levels, V_{THLD}. The typical levels are summarized in Table 2:

Table 2. Echo Qualification Threshold Levels

ECHO_QUAL_THLD	0h	1h	2h	3h	4h	5h	6h	7h
Typical V_{THLD} (mV)	–35	–50	–75	–125	–220	–410	–775	–1500

8.3.6.2 Zero-cross Detect Comparator

The zero-cross detect comparator compares the amplified echo signal at COMPIN with its reference voltage, which is V_{COM}. As shown in Figure 21, the comparator produces a low-to-high transition when the amplitude of the echo signal rises above V_{COM}. The comparator produces a high-to-low transition when the echo amplitude falls below $V_{COM} – V_{HYST}$. The built-in negative-sided hysteresis of 10 mV in reference to V_{COM} ensures accurate zero-cross time instances associated with the rising edges of the echo signal and immunity of the comparator output to noise.

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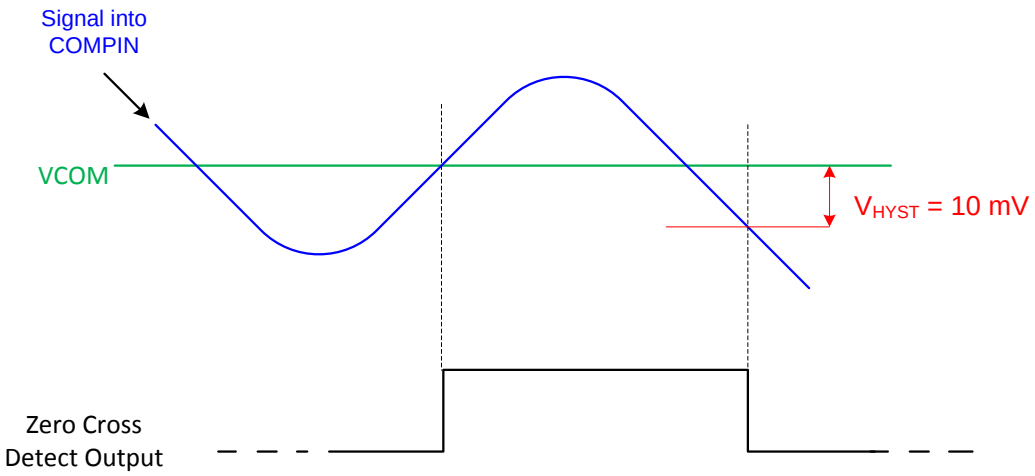

Figure 21. Zero-Cross Detector Output Signal

The output of the zero-cross detect comparator is passed to the event manager, where depending on the decision of the threshold-detect comparator.

8.3.6.3 Event Manager

The event manager is a digital state machine in the STOP pulse generation circuit of the TDC1011. The event manager controls the maximum number of STOP pulses to generate on the STOP pin and the receive mode for the STOP pulse generation. The number of STOP pulses is configured in the *NUM_RX* field in the CONFIG_1 register. The receive mode is selected in the RECEIVE_MODE bit of the CONFIG_4 register. See sections *Single Echo Receive Mode* and *Multiple Echo Receive Mode* for details about the receiver modes of the TDC1011.

An example for *NUM_RX* = 2h and *RECEIVE_MODE* = 0 is shown in Figure 22. When the echo signal amplitude exceeds values smaller than V_{THLD}, the threshold detect comparator indicates to the event manager to qualify the next zero-cross event as valid. When the qualified zero-cross is detected by the zero-cross detect comparator, the event manager passes the pulse to the STOP pin until the number of receive events programmed in *NUM_RX* is reached.

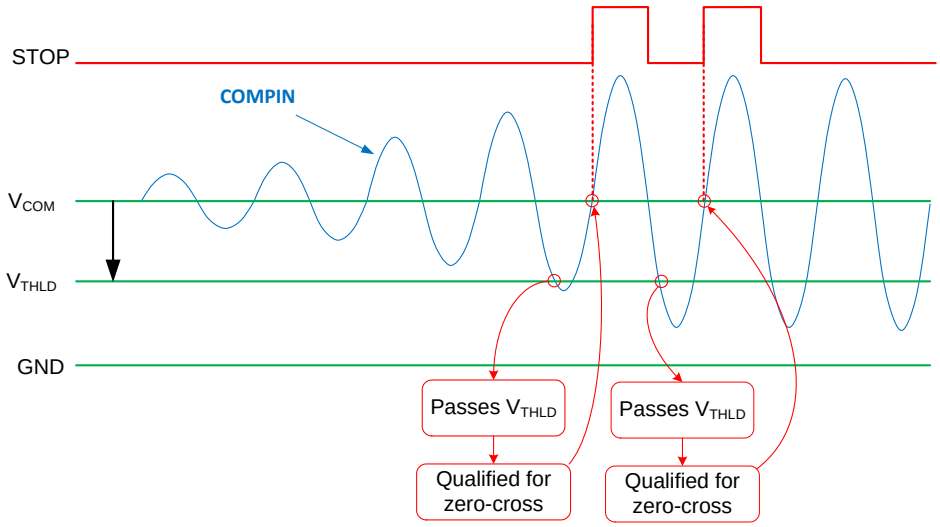

Figure 22. Signal Qualification, Zero-cross Detection and STOP Pulse Generation

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8.3.7 Common-mode Buffer (VCOM)

The output of the internal common-mode buffer is present at the VCOM pin. This pin should be bypassed to ground with a low-leakage 10-nF capacitor and it should not be loaded with more than 20 µA. The common-mode buffer can be disabled with the *VCOM_SEL* bit in the CONFIG_2 register. If disabled, an external reference voltage must be applied to the VCOM pin.

During a time-of-flight measurement, the common-mode reference will take approximately 16 µs to settle if starting from zero initial conditions. Using a larger capacitor will increase the settling time of the internal common-mode reference. The implications of a larger VCOM capacitor are further explored in the *Common-mode Reference Settling Time* section.

8.3.8 Temperature Sensor

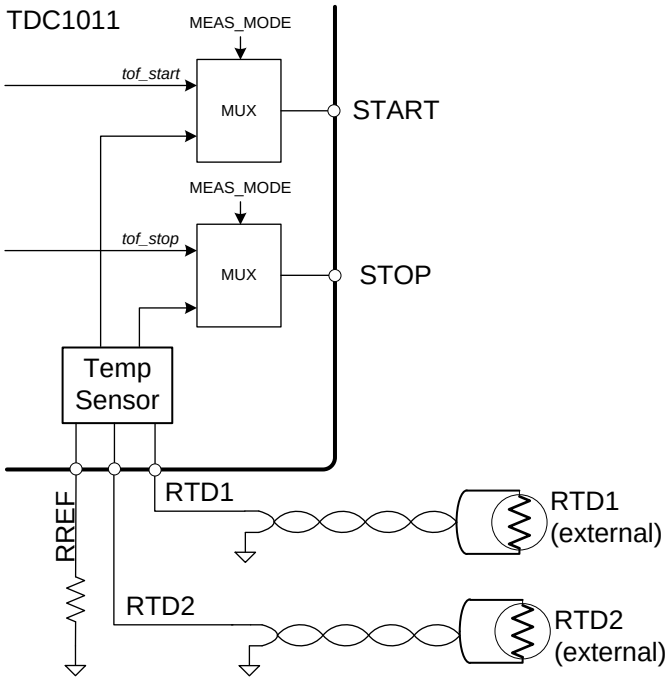

Figure 23. Temperature Sensor Measurement

Accurate measurements of level, and concentration may require compensation for the temperature dependency of the speed of sound in the medium. The TDC1011 provides two temperature sensor connections, allowing to measure up to two locations with RTDs, as shown in Figure 23.

The temperature sensor block supports PT1000 or PT500 sensors. The type of RTD used must be selected in the *TEMP_RTD_SEL* bit of the CONFIG_3 register. The system requires a temperature-stable external reference resistor (R_{REF}). If the RTD type is PT500, then R_{REF} should be 500 Ω. If the RTD type is PT1000, then R_{REF} should be 1 kΩ. The reference resistor needs to have either a low temperature coefficient or be calibrated for temperature shift.

The logic timing in a temperature measurement is controlled by the *TEMP_CLK_DIV* bit in the CONFIG_3 register. As shown in Figure 24, the external clock can be divided by 8 or by the value resulting from the *TX_FREQ_DIV* field configuration in the CONFIG_0 register. It is recommended to operate the temperature measurement block at frequencies of 1 MHz or less.

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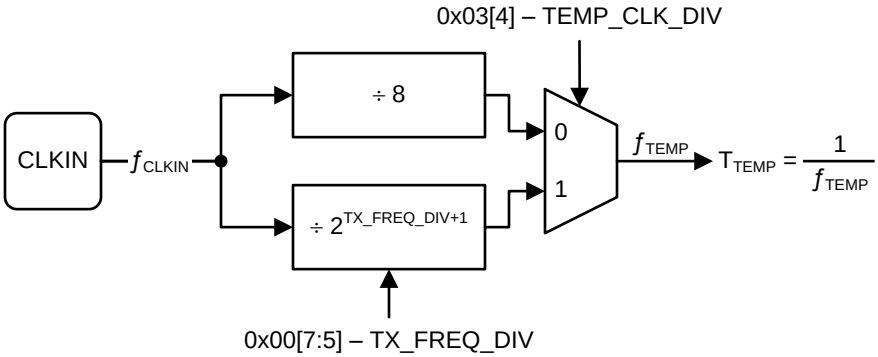

Figure 24. Timing Source for the Temperature Measurement

See the *TDC1011 Temperature Sensing Interface* application note listed in the Temperature Measurement with Multiple RTDs section for details about artifacts in the temperature measurement and how to manage them.

8.3.8.1 Temperature Measurement with Multiple RTDs

The temperature measurement mode is selected by setting the MEAS_MODE bit in the CONFIG_2 register to 1. A temperature measurement is started by sending a trigger pulse. After the temperature measurement is complete, the TDC1011 returns to SLEEP mode. To return to TOF measurement mode, reset the *MEAS_MODE* bit to 0.

The temperature sensor measurement can be performed without the need of an external ADC. The temperature sensor block operates by converting the resistance of a reference, R_{REF}, and up to two RTDs into a series of START and STOP pulses. The interval between the pulses is proportional to the measured resistance, and therefore, the temperature. As shown in Figure 25, the TDC1011 performs three measurements per trigger event and generates the corresponding pulses on the START and STOP pins.

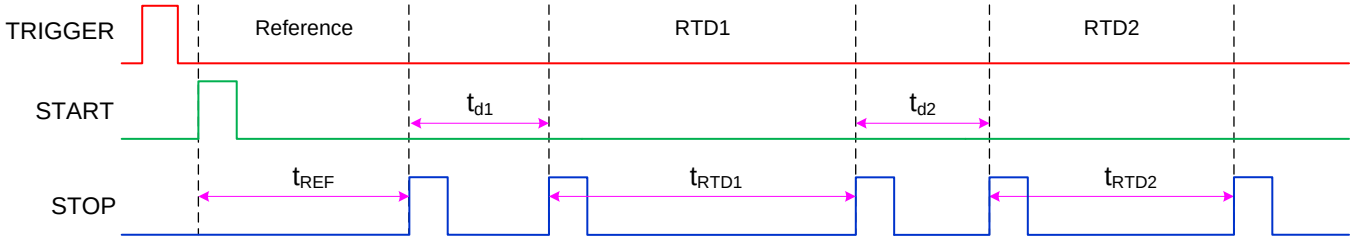

Figure 25. Temperature Measurement Output Timing

The resistance of RTD1 and RTD2 can be calculated from the time intervals in Figure 25 as follows:

$$R_{RTDx} = R_{REF} \times \frac{t_{REF}}{t_{RTDx}} \quad (3)$$

With a 1-kΩ reference resistor, the t_{REF} interval is approximately 200 µs. The following intervals, t_{RTD1} and t_{RTD2}, will depend on the resistance of the RTDs. The time delay between measurements, t_{d1} and t_{d2}, can be approximated as follows:

$$t_{d1} = (51 \times T_{TEMP}) + (t_{RTD1} \times 0.55) \quad (4)$$

$$t_{d2} = (51 \times T_{TEMP}) + (t_{RTD2} \times 0.55) \quad (5)$$

For example, two PT1000 sensors at 0°C will have an approximate resistance of 1 kΩ; the same as the reference resistor in this example. Given an external 8-MHz clock and the default temperature clock divide-by-8 from the *TEMP_CLK_DIV* bit, the overall measurement time between the START pulse and the last STOP pulse is approximately 922 µs.

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8.3.8.2 Temperature Measurement with a Single RTD

The temperature sensing block can be configured to measure a single RTD by setting the *TEMP_MODE* bit in register CONFIG_3 to 1. When the temperature measurement runs in PT1000 mode (*TEMP_RTD_SEL* = 0), the first interval corresponds to R_{REF}, the second interval is a redundant measurement on R_{REF} and should be neglected, and the third interval corresponds to RTD1. This operation is represented in Figure 26.

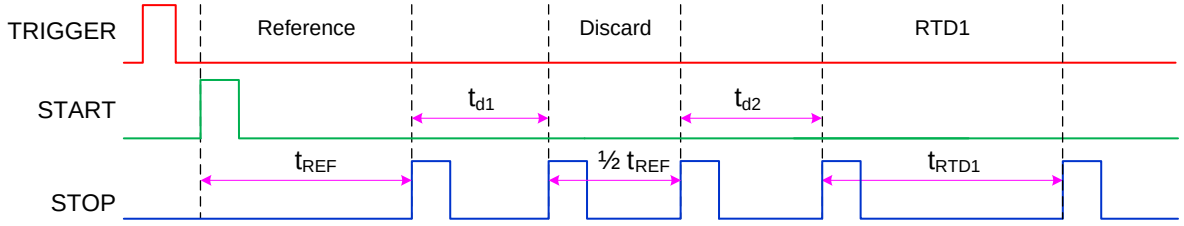

Figure 26. Temperature Measurement with a Single PT1000

The resistance of RTD1 can be calculated using Equation 3. The time delay between measurements can be approximated using Equation 4 and Equation 5, with the exception that in this case, t_{d1} is a function of ½ t_{REF} and t_{d2} is a function of t_{RTD1}.

If the temperature measurement runs in PT500 mode (*TEMP_RTD_SEL* = 1), the first interval is a redundant measurement on R_{REF} and should be neglected, the second interval corresponds to R_{REF}, and the third interval corresponds to RTD1. This operation is represented in Figure 27.

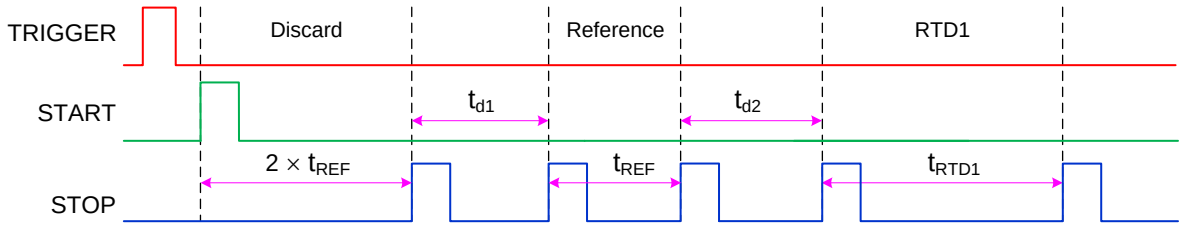

Figure 27. Temperature Measurement with a Single PT500

The resistance of RTD1 can be calculated using Equation 3. The time delay between measurements can be approximated using Equation 4 and Equation 5, with the exception that in this case, t_{d1} is a function of t_{REF} and t_{d2} is a function of t_{RTD1}.

8.4 Device Function Description

8.4.1 Time-of-Flight Measurement Mode

The TOF measurement mode is selected by setting the MEAS_MODE bit in the CONFIG_2 register to 0.

8.4.1.1 Liquid Level or Fluid Identification

The TDC1011 performs a single TOF measurement after receiving a trigger signal and returns to the SLEEP mode when the measurement is complete.

8.4.2 State Machine

A state machine in the TDC1011 manages the operation of the various measurement modes (see Figure 28). At power-on, the state machine is reset and most blocks are disabled. After the power-on sequence is complete, the device goes into SLEEP mode if the EN pin is low or into READY mode if the EN pin is high. In the SLEEP or READY state, the TDC1011 is able to receive SPI commands to set registers and configure the device for a measurement mode.

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Device Function Description (continued)

NOTE

Although the SPI block is always active, it is not recommended to perform configuration changes while the device is active. Configuration changes should be performed while the device is in the SLEEP state or in the READY state.

If the EN pin is high and a trigger signal is received, the state machine will begin the execution of the configured measurement. The state machine will return to the SLEEP state after the measurement is completed.

The device can be forced to exit a measurement by applying a logic high on the RESET pin high or a logic low on the EN pin.

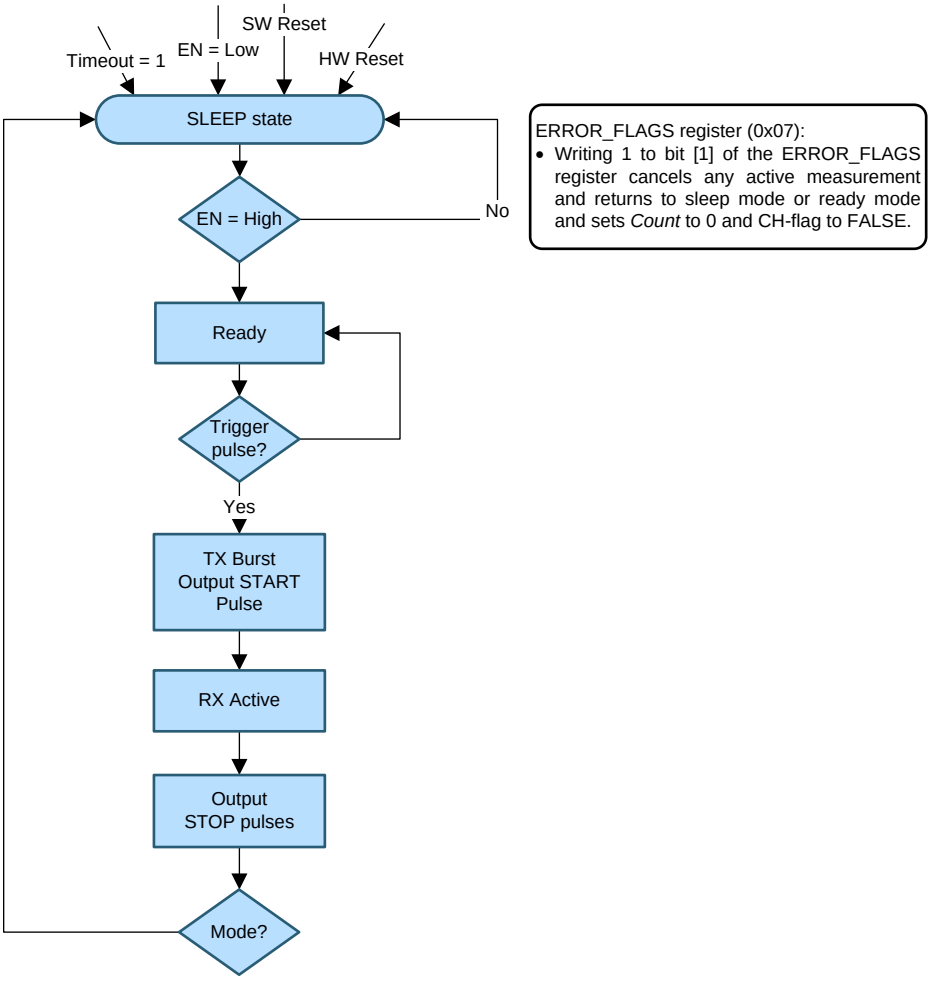

Figure 28. Simplified TDC1011 State Machine Diagram

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Device Function Description (continued)

8.4.3 TRANSMIT Operation

8.4.3.1 Transmission Pulse Count

The number of TX pulses generated by the TDC1011 to drive an ultrasonic transducer is programmable using the *NUM_TX* field located in the CONFIG_0 register.

8.4.3.2 TX 180° Pulse Shift

As shown in Figure 29, the transmitter block can add a 180° shift at a position in the TX signal. The position of the pulse shift is set by the *TX_PH_SHIFT_POS* field in the CONFIG_4 register and allows generating a specific signal pattern.

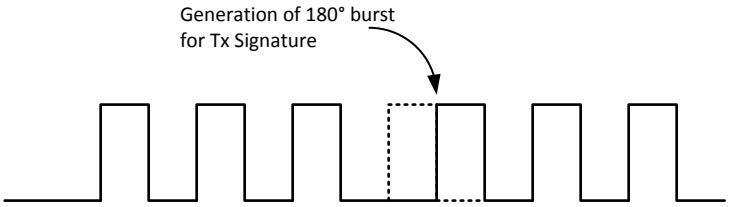

Figure 29. Transmitter Pulse Signature, 180° Burst

As shown in Figure 30, enabling the TX 180° pulse shift has the effect of decreasing the number of transmitted pulses by 1.

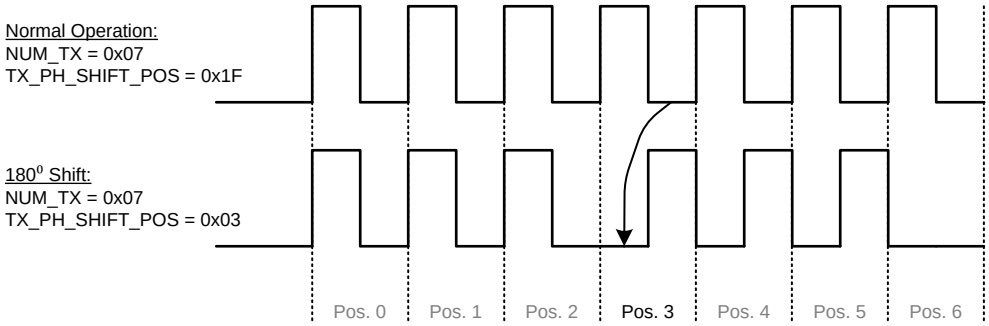

Figure 30. Transmitter Pulse Signature

In some cases, the 180° pulse shift may help improving the turn-off time of a transducer, and thus suppress the transmit ringing.

The 180° pulse shift is disabled by setting *TX_PH_SHIFT_POS* to position 31. Setting the 180° pulse shift to positions 0 or 1 is not recommended.

8.4.3.3 Transmitter Damping

The transmitter damping feature allows for improved control over the transducer signal generation. Damping extends the duration of the last TX pulse to help dissipate ringing and improve the transducer's turn-off time (see Figure 31 and Figure 32). The accuracy of measurements can be improved by having a faster transducer turn-off time. Damping is controlled with the *DAMPING* bit in the CONFIG_2 register.

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Device Function Description (continued)

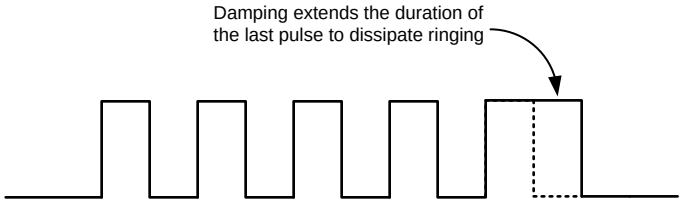

Figure 31. Transmitter Damping (5 Tx Pulses With a Damping Pulse)

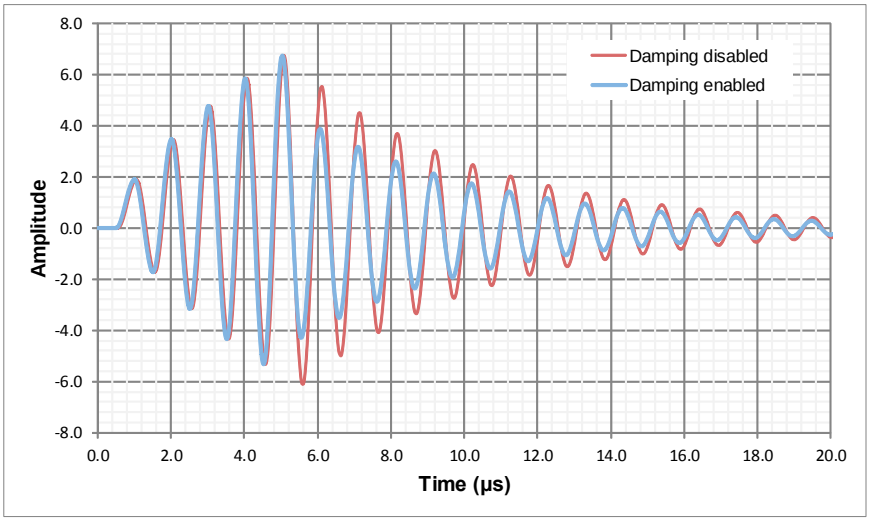

Figure 32. Transmitter Damped Echo

There are two invalid use combinations of the damping feature that may result in unexpected behavior. First, damping should not be combined with the 180° pulse shift described in the previous section. Second, damping should not be enabled if the number of TX pulses is set to 31.

8.4.4 RECEIVE Operation

8.4.4.1 Single Echo Receive Mode

Single Echo mode can be used for fluid identification measurements or level applications where transducer carrier frequency information is required. The device can be configured for Single Echo mode by setting the *RECEIVE_MODE* bit to 0 in the CONFIG_4 register. In Single Echo mode, the device will generate STOP pulses for every zero-cross qualified by the threshold comparator, up to the number of expected STOP events configured in the *NUM_RX* field in the CONFIG_1 register.

The threshold comparator qualifies the next zero-cross after an RX amplitude smaller than the programmed threshold voltage is detected. The zero-cross detector will provide output pulses corresponding to the rising edge of the received signal crossing the V_{COM} level, as shown in Figure 33. The threshold voltage can be set in the *ECHO_QUAL_THDL* field in the CONFIG_3 register.

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Device Function Description (continued)

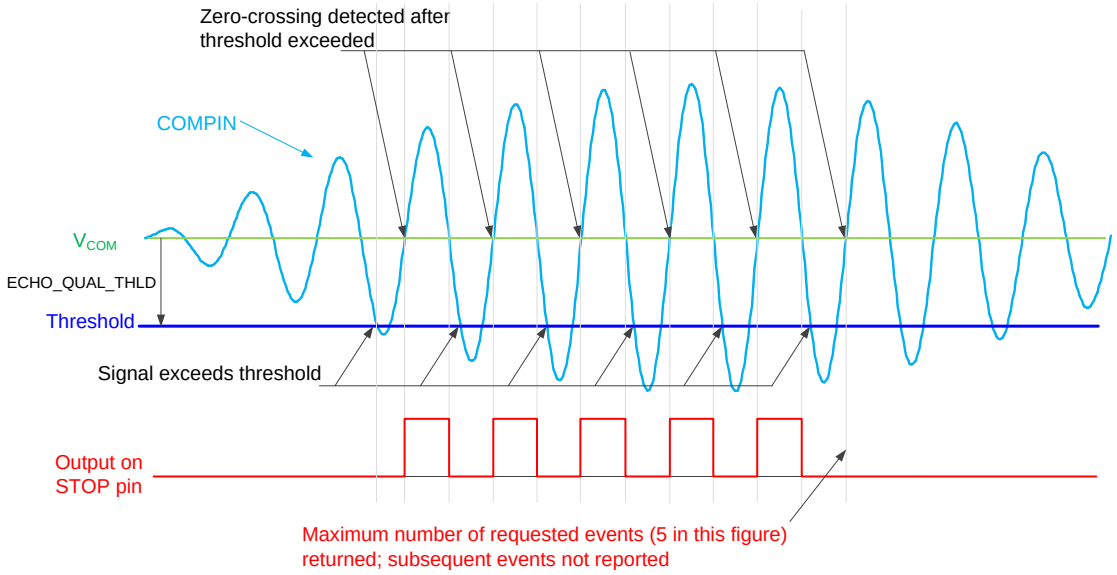

Figure 33. Single Echo Receive Mode (5 STOP Events)

If the number of expected pulses programmed in *NUM_RX* is not received or the time-of-flight operation times out, the TDC1011 will indicate an error condition in the ERROR_FLAGS register and will set the ERRB pin low.

8.4.4.2 Multiple Echo Receive Mode

The Multiple Echo mode is intended for use in level sensing applications and distance/displacement measurements in which multiple echoes (burst) are received. In this condition, each received echo group will be treated as a single pulse on the STOP pin. Up to 7 STOP pulses can be generated based on the value of the *NUM_RX* field in the CONFIG_1 register. Multi echo mode can be enabled by setting the *RECEIVE_MODE* bit to 1 in the CONFIG_4 register. A representation of multiple echo STOP pulse generation is shown in Figure 34.

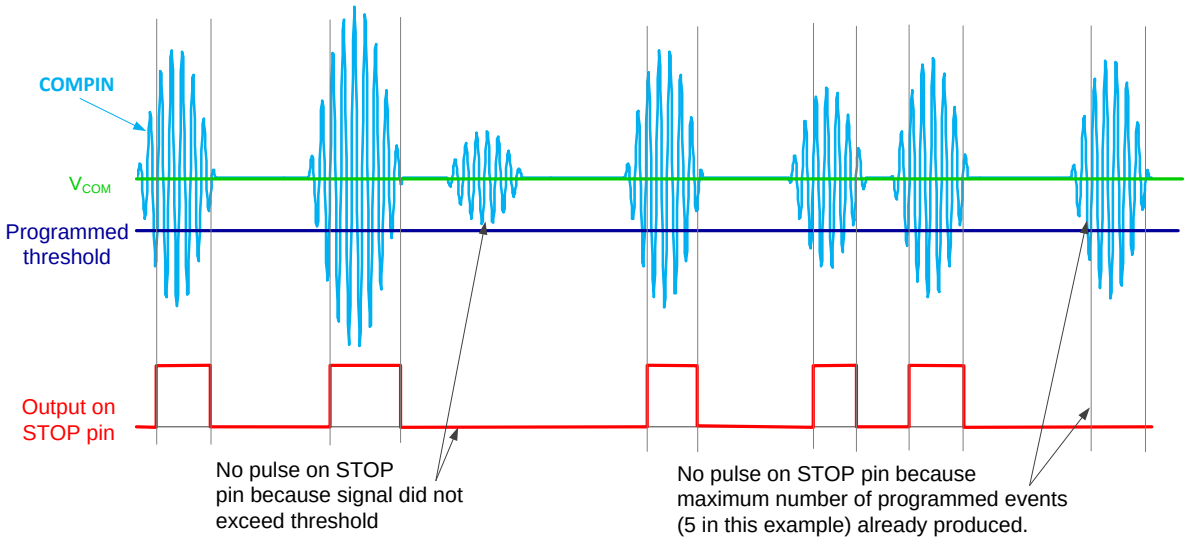

Figure 34. Multiple Echo Receive Mode (5 STOP Events)

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Device Function Description (continued)

The rising edge of a STOP pulse is generated by a zero-crossing event. As in the Single Echo Receive Mode, the threshold comparator qualifies the next zero-cross after an RX amplitude smaller than the programmed threshold voltage is detected. The STOP pulse will extend until a zero-cross after the RX amplitude is no longer smaller than the threshold voltage (see Figure 35).

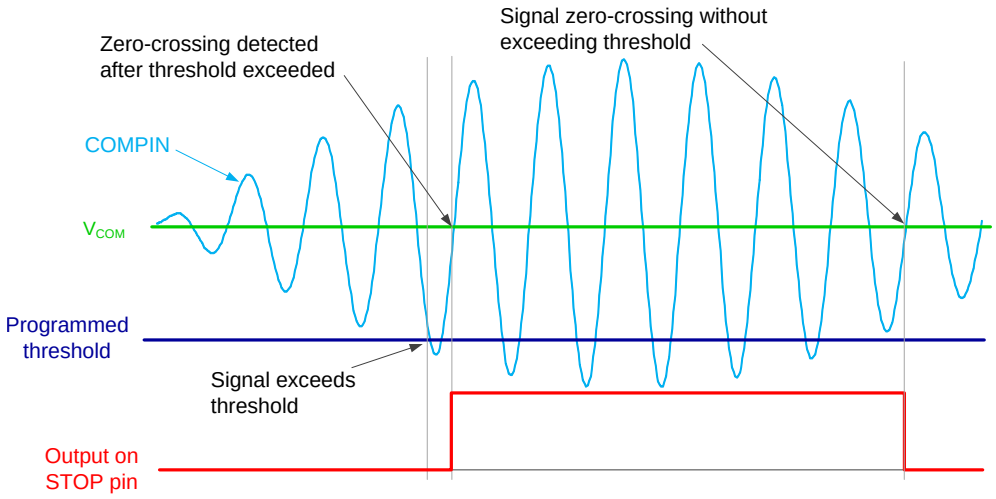

Figure 35. Multiple Echo Receive Mode (Zoom-in)

If the number of expected pulses programmed in *NUM_RX* is not received or the time-of-flight operation times out, the TDC1011 will indicate an error condition in the ERROR_FLAGS register and will set the ERRB pin low.

8.4.5 Timing

8.4.5.1 Timing Control and Frequency Scaling (CLKIN)

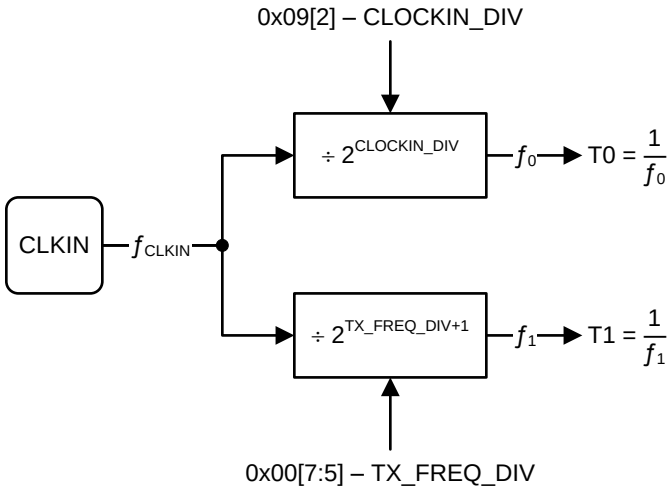

Figure 36. External Clock Division Tree

All transmit and receive function sequencing is synchronous to the external clock applied to the CLKIN pin. The external clock is divided to generate two internal clocks with corresponding time periods denoted as T0 and T1 in Figure 36. The division factor used to generate T0 is controlled with the *CLOCKIN_DIV* bit in the CLOCK_RATE register. The division factor used to generate T1 is controlled with the *TX_FREQ_DIV* field located in the CONFIG_0 register.

The SPI block is synchronous with the clock applied to the SCLK pin, and it is independent of the clock applied to CLKIN. See the *Serial Peripheral Interface (SPI)* section for a complete description of the SPI block.

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Device Function Description (continued)

8.4.5.2 TX/RX Measurement Sequencing and Timing

The TDC1011 automatically sequences the TX and RX functionality. After receiving a pulse edge on the TRIGGER pin, the TDC1011 resynchronizes to the CLKIN signal, and sends a TX burst.

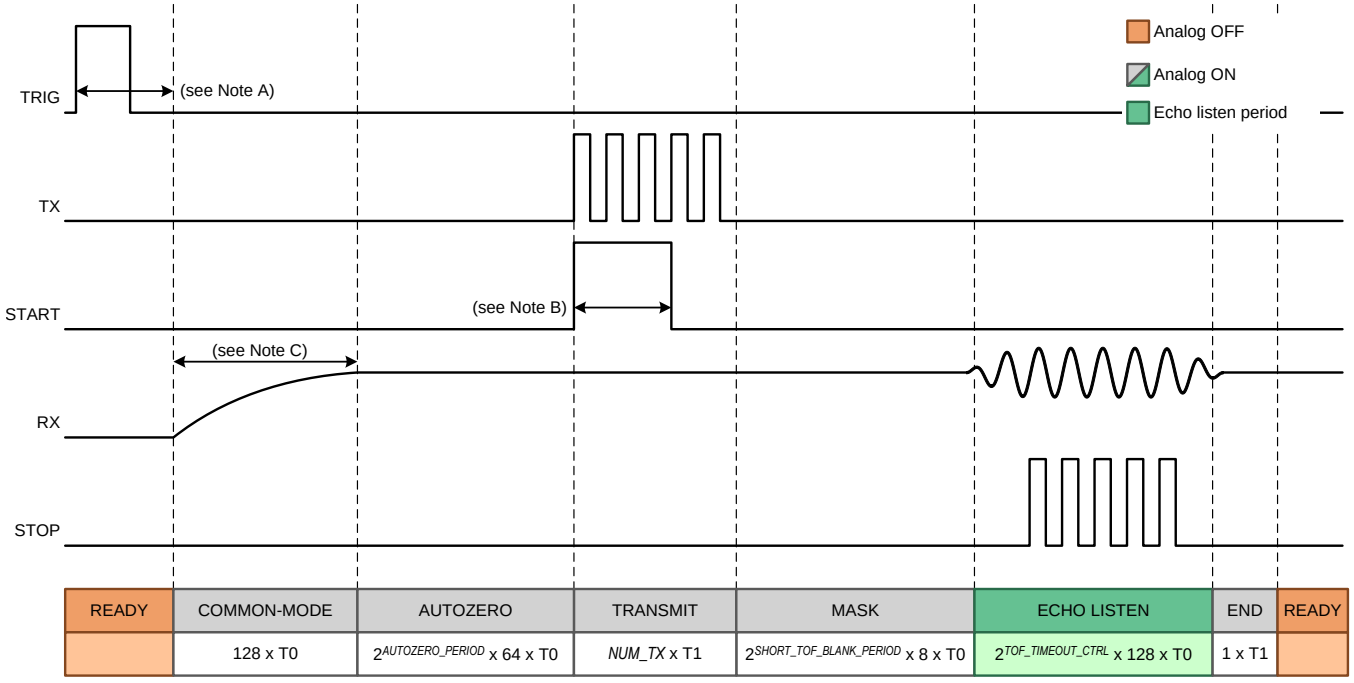
The trigger edge polarity is configured to rising edge by default, but it can be changed to falling edge by setting the *TRIG_EDGE_POLARITY* bit in the CONFIG_4 register to 1.

After a device reset, the system must wait a determined time before sending the next trigger signal. The typical reset to trigger wait time is 3 × T1 + 50 ns.

8.4.6 Time-of-Flight (TOF) Control

The possible configurations of the TX/RX sequencing during a time-of-flight measurement can be divided into three cases: *Short TOF Measurement*, *Standard TOF Measurement* and *Standard TOF Measurement with Power Blanking*. Overall, the cases differ in the order of sequencing, power saving and echo listening windows. The behavior of each case is described in the sections to follow.

8.4.6.1 Short TOF Measurement



A. Common-mode settling time.

B. If *NUM_TX* < 3, the width of the START pulse is equal to *NUM_TX* × T1. If *NUM_TX* ≥ 3, the width of the START pulse is equal to 3 × T1.

Figure 37. Short TOF Measurement

In a short time of flight measurement, the RX path is activated before the TX burst, as shown in *Figure 37*.

The short TOF is the default measurement sequence selected at power-on. The short TOF measurement is selected if the value of the *TIMING_REG[9:0]* field is less than 30, or if the *FORCE_SHORT_TOF* bit is set to 1. The *TIMING_REG[9:0]* is a 10-bit wide field, with its 2 most significant bits located in the TOF_1 register, and the 8 least significant bits located in the TOF_0 register. The FORCE_SHORT_TOF bit is located in the TIMEOUT register.

The comparator's input offset is stored in an internal capacitor during the auto-zero period. The length of the auto-zero period is controlled by the AUTOZERO_PERIOD field in the CLOCK_RATE register.

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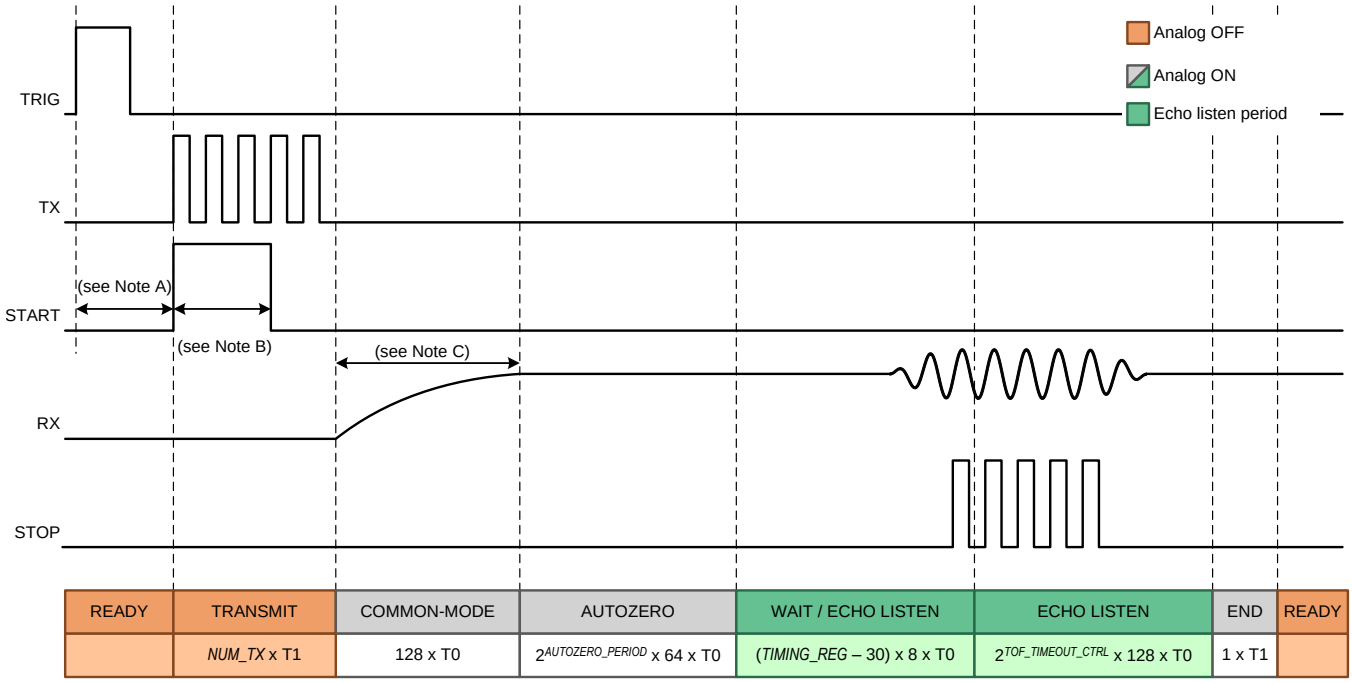
Device Function Description (continued)

The length of the window when the comparators are able to qualify and generate STOP pulses is configured by the *TOF_TIMEOUT_CTRL* field. A timeout will occur if the number of expected pulses is not received during the allocated time and an error condition is reported to the ERROR_FLAGS register and the ERRB pin. It is possible to disable the echo timeout (see TOF Measurement Interval). The *TOF_TIMEOUT_CTRL* field is located in the TIMEOUT register.

See the *Timing Control and Frequency Scaling (CLKIN)* section for the definition of the time periods T0 and T1.

8.4.6.2 Standard TOF Measurement



A. Clock alignment.

B. If *NUM_TX* < 3, the width of the START pulse is equal to *NUM_TX* × T1. If *NUM_TX* ≥ 3, the width of the START pulse is equal to 3 × T1.

C. Common-mode settling time.

Figure 38. Standard TOF Measurement

In a standard time of flight measurement, the RX path is activated after the TX burst is completed, as shown in *Figure 38*.

The standard TOF measurement sequence is enabled if the value of the *TIMING_REG* field is greater than or equal to 30, and only if the *FORCE_SHORT_TOF* bit is set to 0. The *TIMING_REG* is a 10-bit wide field, with its 2 most significant bits located in the TOF_1 register, and the 8 least significant bits located in the TOF_0 register. The FORCE_SHORT_TOF bit is located in the TIMEOUT register.

The comparator's input offset is stored in an internal capacitor during the auto-zero period. The length of the auto-zero period is controlled by the AUTOZERO_PERIOD field in the CLOCK_RATE register.

The length of the window when the comparators are able to qualify and generate STOP pulses is configured by a combination of the *TIMING_REG* field and the *TOF_TIMEOUT_CTRL* field. With the addition of the TIMING_REG in the calculation, the standard TOF measurement allows for a longer wait time and listening window. A timeout will occur if the number of expected pulses is not received during the allocated time and an error condition is reported to the ERROR_FLAGS register and the ERRB pin. It is possible to disable the echo timeout (see *TOF Measurement Interval*). The *TOF_TIMEOUT_CTRL* field is located in the TIMEOUT register.

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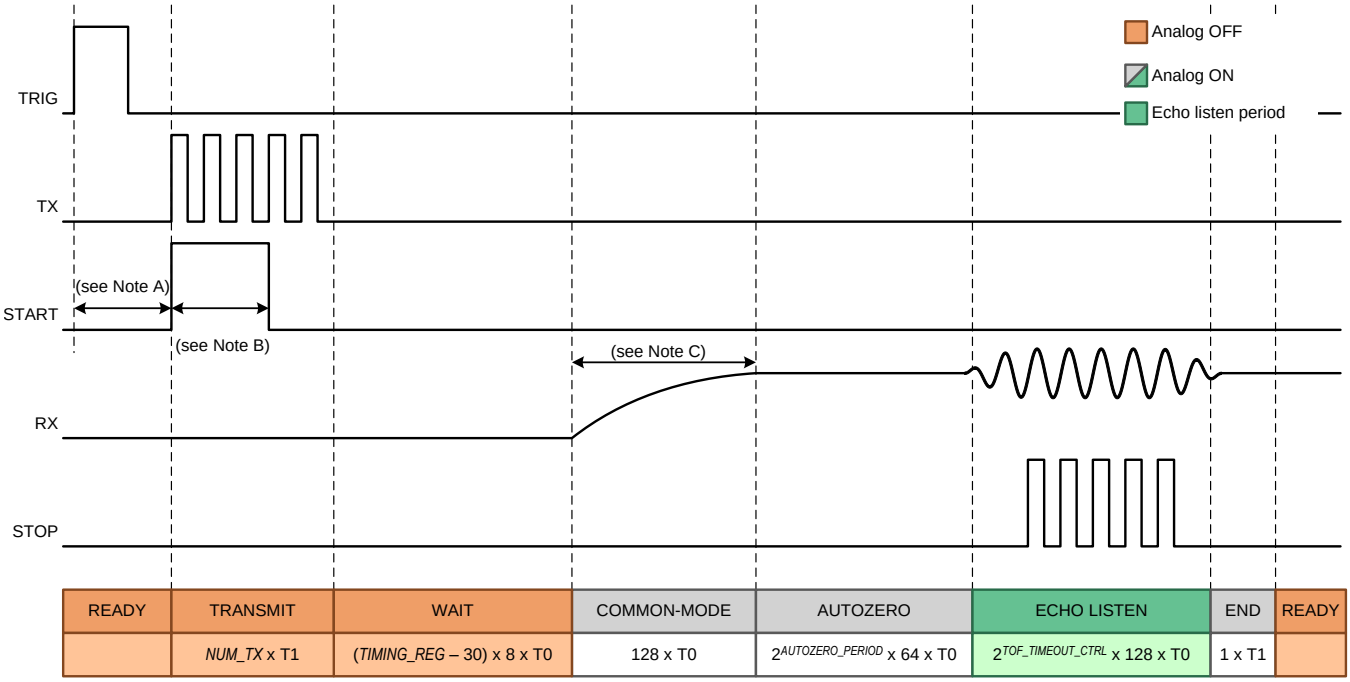
Device Function Description (continued)

NOTE

If the *FORCE_SHORT_TOF* bit = 1, the measurement sequencing will behave as a **Short TOF Measurement**, thus overriding the setting of the *TIMING_REG* field.

8.4.6.3 Standard TOF Measurement with Power Blanking



A. Clock alignment.

B. If *NUM_TX* < 3, the width of the START pulse is equal to *NUM_TX* × T1. If *NUM_TX* ≥ 3, the width of the START pulse is equal to 3 × T1.

C. Common-mode settling time.

Figure 39. Standard TOF Measurement with Blanking Enabled

The power blanking sequence is a variation to the standard TOF measurement sequence, and can be enabled by setting the *BLANKING* bit to 1. In addition, all other conditions described in the *Standard TOF Measurement* should be met. The *BLANKING* bit can be found in the CONFIG_3 register.

Power blanking allows the device to remain in a low-power state while the TX signals propagate to the RX transducer in situations when the expected time-of-flight is long. Power blanking uses the *TIMING_REG* to control a wait time between the transmit sequence and the receive sequence, during which, the complete RX chain is disabled, as shown in Figure 39. The *TIMING_REG* is a 10-bit wide field, with its 2 most significant bits located in the TOF_1 register, and the 7 least significant bits located in the TOF_0 register.

8.4.6.4 Common-mode Reference Settling Time

The duration of the common-mode settling *time* is defined by the VCOM capacitor. With a 10-nF VCOM capacitor, the common-mode reference requires 16 µs to settle. On the other hand, the duration of the common-mode settling *window* is defined as 128 × T0, where the time unit T0 is determined by the external clock frequency and the value of the CLOCKIN_DIV bit, as explained in the *Timing Control and Frequency Scaling (CLKIN)* section.

A frequency of 8 MHz will result in a settling *window* of 128 × 1 / 8 MHz, which equals to 16 µs. Increasing the value of the VCOM capacitor will increase the common-mode settling *time*, but for the same 8-MHz frequency, the duration of the common-mode settling *window* will remain at 16 µs. In such situation, the common-mode reference will take multiple TOF cycles to reach its final value when starting from zero initial conditions.

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Device Function Description (continued)

8.4.6.5 TOF Measurement Interval

The comparators in the TDC1011's RX path can qualify and generate STOP pulses from a received echo within an interval set by the *TOF_TIMEOUT_CTRL* field in the TIMEOUT register. The listening interval can be extended in the standard TOF measurement (without blanking) by a period controlled with the *TIMING_REG* field (see *Standard TOF Measurement*).

If the number of STOP events programmed in the *NUM_RX* field is not received within the listening interval, a timeout event will occur and the device will return to the READY state. In addition, an error will be reported to the ERROR_FLAGS register and the ERRB pin will be driven low.

The echo timeout can be disabled by setting the *ECHO_TIMEOUT* bit to 1 in the TIMEOUT register. If the echo timeout is disabled, the device will not exit from the receive state until the expected number of STOP events set in *NUM_RX* occur. If the number of events does not occur, the device can be forced out of the receive state by writing a value of 0x03 to the ERROR_FLAGS register, or by de-asserting the EN pin, or asserting the RESET pin.

NOTE

Writing a logic 1 to bit [1] of the ERROR_FLAGS register clears the state machine. Writing a logic 1 to bit[0] clears the error flags.

NOTE

It is not recommended to hold the RX in an active state for intervals longer than 100ms, as the comparator auto-zero may no longer be accurate.

8.4.7 Error Reporting

The TDC1011 will report an error when the receive signals do not match the expected configuration. The ERRB pin will go low to indicate the presence of an error condition. Reading the ERROR_FLAGS register provides information about the condition(s) that caused the error.

The *ERR_SIG_WEAK* bit indicates that the number of received and qualified zero-crossings was less than the expected number set in the *NUM_RX* register field and a timeout occurred. This error is cleared when bit [0] is written to 1.

The *ERR_NO_SIG* bit indicates that no signals were received and a timeout occurred. Writing a 1 to this bit resets the state machine, halts active measurements and returns the device to SLEEP or READY mode. This error is cleared when bit [0] is written to 1.

The *ERR_SIG_HIGH* bit indicates that the received echo amplitude exceeds the largest echo qualification threshold at the input of the comparators. The *ERR_SIG_HIGH* error is only reported when the *ECHO_QUAL_THDL* register field is set to 7h. Writing a 1 to this bit will reset all the error flags and reset the ERRB pin to high.

NOTE

It is recommended to reset the state machine when the error flags are cleared. This can be done simultaneously by writing a value of 0x03 to the ERROR_FLAGS register.

8.5 Programming

8.5.1 Serial Peripheral Interface (SPI)

The serial interface consists of serial data input (SDI), serial data output (SDO), serial interface clock (SCLK) and chip select bar (CSB). The serial interface is used to configure the TDC1011 parameters available in various configuration registers. All the registers are organized into individually addressable byte-long registers with a unique address.

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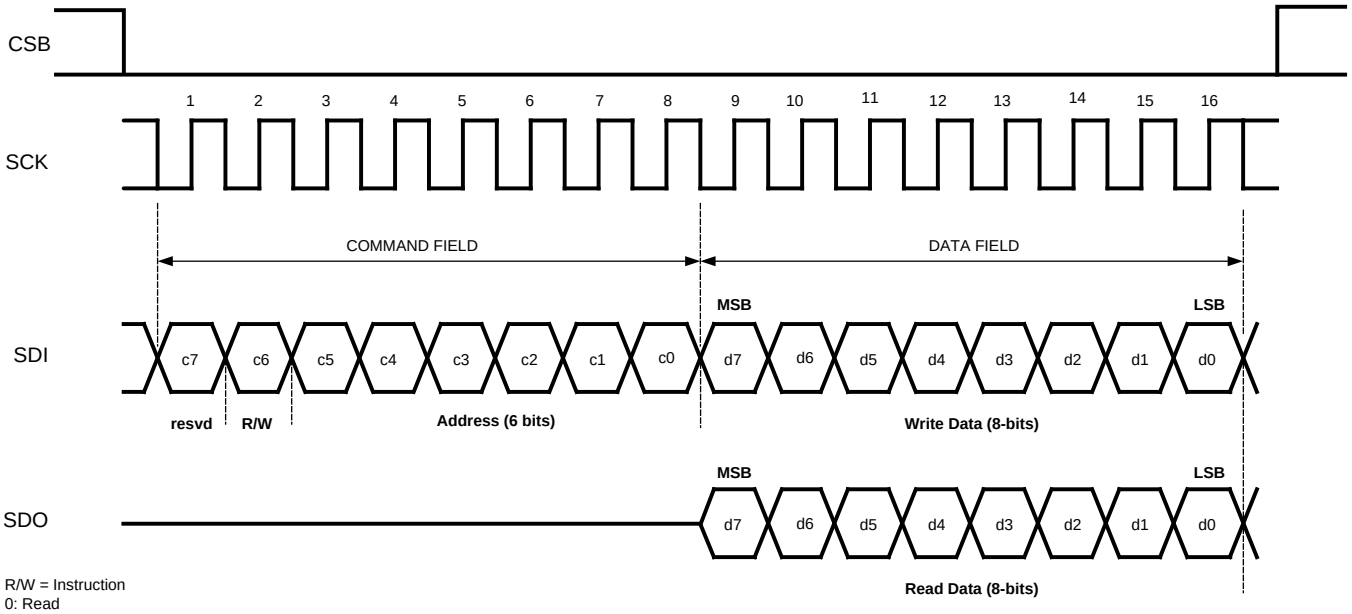
Programming (continued)

The communication on the SPI bus normally supports write and read transactions. A write transaction consists of a single write command byte, followed by single data byte. A read transaction consists of a single read command byte followed by 8 SCLK cycles. The write and read command bytes consist of 1 reserved bit, a 1-bit instruction, and a 6-bit register address. Figure 40 shows the SPI protocol for a transaction involving one byte of data (read or write).



R/W = Instruction
0: Read
1: Write
Note: Specifying any value other than zero in bit[7] of the command byte is prohibited.

Figure 40. SPI Protocol

8.5.1.1 Chip Select Bar (CSB)

CSB is an active-low signal and needs to be low throughout a transaction. That is, CSB should not pulse between the command byte and the data byte of a single transaction.

De-asserting CSB always terminates an ongoing transaction, even if it is not yet complete. Re-asserting CSB will always bring the device into a state ready for the next transaction, regardless of the termination status of a previous transaction.

8.5.1.2 Serial Clock (SCLK)

SCLK can idle high or low. It is recommended to keep SCLK as clean as possible to prevent glitches from corrupting the SPI frame.

8.5.1.3 Serial Data Input (SDI)

SDI is driven by the SPI master by sending the command and the data byte to configure the AFE.

8.5.1.4 Serial Data Output (SDO)

SDO is driven by the AFE when the SPI master initiates a read transaction.

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8.6 Register Maps

NOTE

- Reserved bits must be written to 0 unless otherwise indicated.
- Read-back value of reserved bits and registers is unspecified and should be discarded.
- Recommended values must be programmed and forbidden values must not be programmed where they are indicated to avoid unexpected results.

8.6.1 TDC1011 Registers

Table 3 list the memory-mapped registers for the TDC1011. All register addresses not listed in Table 3 should be considered as reserved locations and the register contents should not be modified.

Table 3. TDC1011 REGISTERS

Address (Hex)	Acronym	Register Name	Reset Value	Section
0h	CONFIG_0	Config-0	45h	See here
1h	CONFIG_1	Config-1	40h	See here
2h	CONFIG_2	Config-2	0h	See here
3h	CONFIG_3	Config-3	3h	See here
4h	CONFIG_4	Config-4	1Fh	See here
5h	TOF_1	TOF-1	0h	See here
6h	TOF_0	TOF-0	0h	See here
7h	ERROR_FLAGS	Error Flags	0h	See here
8h	TIMEOUT	Timeout	19h	See here
9h	CLOCK_RATE	Clock Rate	0h	See here

8.6.1.1 CONFIG_0 Register (address = 0h) [reset = 45h] (map)

Figure 41. CONFIG_0 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
TX_FREQ_DIV			**NUM_TX**				
R/W-2h			R/W-5h				

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 4. CONFIG_0 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7:5]	TX_FREQ_DIV(1)	R/W	2h	Frequency divider for TX clock and T1 0h: Divide by 2 1h: Divide by 4 2h: Divide by 8 (default) 3h: Divide by 16 4h: Divide by 32 5h: Divide by 64 6h: Divide by 128 7h: Divide by 256
[4:0]	NUM_TX	R/W	5h	Number of TX pulses in a burst, ranging from 0 to 31. 5h: 5 pulses (default)

(1) See Timing Control and Frequency Scaling (CLKIN) for the definition of the time period T1.

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8.6.1.2 CONFIG_1 Register (address = 1h) [reset = 40h] (map)

Figure 42. CONFIG_1 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
RESERVED			**RESERVED**			**NUM_RX**	
R/W-1h			R/W-0h			R/W-0h	

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 5. CONFIG_1 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7:6]	RESERVED	R/W	1h	1h: Reserved (default)
[5:3]	RESERVED	R/W	0h	Must always be written 0h (default)
[2:0]	NUM_RX	R/W	0h	Number of expected receive events 0h: Do not count events (32 STOP pulses output) (default) 1h: 1 event (1 STOP pulse output) 2h: 2 events (2 STOP pulses output) 3h: 3 events (3 STOP pulses output) 4h: 4 events (4 STOP pulses output) 5h: 5 events (5 STOP pulses output) 6h: 6 events (6 STOP pulses output) 7h: 7 events (7 STOP pulses output)

8.6.1.3 CONFIG_2 Register (address = 2h) [reset = 0h] (map)

Figure 43. CONFIG_2 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
VCOM_SEL	**MEAS_MODE**	**DAMPING**	**RESERVED**				
R/W-0h	R/W-0h	R/W-0h	R/W-0h				

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 6. CONFIG_2 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7]	VCOM_SEL	R/W	0h	Common-mode voltage reference control 0h: Internal (default) 1h: External
[6]	MEAS_MODE	R/W	0h	AFE measurement type 0h: Time-of-flight measurement (default) 1h: Temperature measurement
[5]	DAMPING	R/W	0h	TX burst damping 0h: Disable damping (default) 1h: Enable damping
[4:0]	RESERVED	R/W	0h	Must always be written 0h (default)

8.6.1.4 CONFIG_3 Register (address 3h) [reset = 3h] (map)

Figure 44. CONFIG_3 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
RESERVED	**TEMP_MODE**	**TEMP_RTD_SEL**	**TEMP_CLK_DIV**	**BLANKING**	**ECHO_QUAL_THLD**		
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-3h		

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

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Table 7. CONFIG_3 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7]	RESERVED	R/W	0h	0h: Reserved (default)
[6]	TEMP_MODE	R/W	0h	Temperature measurement channels 0h: Measure REF, RTD1 and RTD2 (default) 1h: Measure REF and RTD1
[5]	TEMP_RTD_SEL	R/W	0h	RTD type 0h: PT1000 (default) 1h: PT500
[4]	TEMP_CLK_DIV	R/W	0h	Clock divider for temperature mode 0h: Divide by 8 (default) 1h: Use TX_FREQ_DIV
[3]	BLANKING	R/W	0h	Power blanking in standard TOF measurements. The blanking length is controlled with the TIMING_REG field (see Standard TOF Measurement with Power Blanking). 0h: Disable power blanking (default) 1h: Enable power blanking
[2:0]	ECHO_QUAL_THLD	R/W	3h	Echo qualification DAC threshold level with respect to V_{COM} 0h: –35 mV 1h: –50 mV 2h: –75 mV 3h: –125 mV (default) 4h: –220 mV 5h: –410 mV 6h: –775 mV 7h: –1500 mV

8.6.1.5 CONFIG_4 Register (address = 4h) [reset = 1Fh] (map)

Figure 45. CONFIG_4 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
RESERVED	**RECEIVE_ MODE**	**TRIG_EDGE_ POLARITY**	**TX_PH_SHIFT_POS**				
R/W-0h	R/W-0h	R/W-0h	R/W-1Fh				

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 8. CONFIG_4 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7]	RESERVED	R/W	0h	0h: Reserved (default)
[6]	RECEIVE_MODE	R/W	0h	Receive echo mode 0h: Single echo (default) 1h: Multi echo
[5]	TRIG_EDGE_POLARITY	R/W	0h	Trigger edge polarity 0h: Rising edge (default) 1h: Falling edge
[4:0]	TX_PH_SHIFT_POS	R/W	1Fh	TX 180° pulse shift position, ranging from 0 to 31. 1Fh: Position 31 (default) It is not recommended to set TX_PH_SHIFT_POS to 0 or 1.

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8.6.1.6 TOF_1 Register (address = 5h) [reset = 0h] (map)

Figure 46. TOF_1 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
PGA_GAIN			**PGA_CTRL**	**LNA_CTRL**	**LNA_FB**	**TIMING_REG[9:8]**	
R/W-0h			R/W-0h	R/W-0h	R/W-0h	R/W-0h	

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 9. TOF_1 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7:5]	PGA_GAIN	R/W	0h	PGA gain 0h: 0 dB (default) 1h: 3 dB 2h: 6 dB 3h: 9 dB 4h: 12 dB 5h: 15 dB 6h: 18 dB 7h: 21 dB
[4]	PGA_CTRL	R/W	0h	PGA control 0h: Active (default) 1h: Bypassed and powered off
[3]	LNA_CTRL	R/W	0h	LNA control 0h: Active (default) 1h: Bypassed and powered off
[2]	LNA_FB	R/W	0h	LNA feedback mode 0h: Capacitive feedback (default) 1h: Resistive feedback
[1:0]	TIMING_REG[9:8]	R/W	0h	TIMING_REG field's 2 most-significant bits (see Standard TOF Measurement and Standard TOF Measurement with Power Blanking) 0h: 0 (default)

8.6.1.7 TOF_0 Register (address = 6h) [reset = 0h] (map)

Figure 47. TOF_0 Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
TIMING_REG[7:0]							
R/W-0h							

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 10. TOF_0 Register Field Descriptions

Bit	Field	Type	Reset	Description
[7:0]	TIMING_REG[7:0]	R/W	0h	TIMING_REG field's 8 least-significant bits (see Standard TOF Measurement and Standard TOF Measurement with Power Blanking) 0h: 0 (default)

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8.6.1.8 ERROR_FLAGS Register (address = 7h) [reset = 0h] (map)

Figure 48. ERROR_FLAGS Register

7 (MSB)	6	5	4	3	2	1	0 (LSB)
RESERVED					ERR_SIG_WEAK	ERR_NO_SIG	ERR_SIG_HIGH
R-0h					R-0h	R/W1C-0	R/W1C-0

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 11. ERROR_FLAGS Register Field Descriptions(1)(2)

Bit	Field	Type	Reset	Description
[7:3]	RESERVED	R	0h	0h: Reserved (default)
[2]	ERR_SIG_WEAK	R	0h	1h: The number of received and qualified zero-crossings was less than the expected number set in NUM_RX field and a timeout occurred.
[1]	ERR_NO_SIG	R/W1C	0h	1h: No signals were received and timeout occurred. Writing a 1 to this field resets the state machine, halts active measurements and returns the device to the SLEEP or READY mode.
[0]	ERR_SIG_HIGH	R/W1C	0h	1h: The received echo amplitude exceeds the largest echo qualification threshold at the input of the comparators. The error is only reported when ECHO_QUAL_THLD = 0x07. Writing a 1 to this field will reset all the error flags and reset the ERRB pin to high.

(1) It is recommended to read the error status register or the ERRB pin before initiating a new measurement.
(2) All error flags should be cleared before initiating a new measurement.

8.6.1.9 TIMEOUT Register (address = 8h) [reset = 19h] (map)

Figure 49. TIMEOUT Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
RESERVED	FORCE_SHORT_TOF	SHORT_TOF_BLANK_PERIOD			ECHO_TIMEOUT	TOF_TIMEOUT_CTRL	
R/W-0h	R/W-0h	R/W-3h			R/W-0h	R/W-1h	

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 12. TIMEOUT Register Field Descriptions

Bit	Field	Type	Reset	Description
[7]	RESERVED	R/W	0h	0h: Reserved (default)
[6]	FORCE_SHORT_TOF	R/W	0h	Short time-of-flight control 0h: Disabled (default) 1h: Force a short time-of-flight measurement
[5:3]	SHORT_TOF_BLANK_PERIOD(1)	R/W	3h	Short time-of-flight blanking period (see Short TOF Measurement) 0h: 8 × T0 1h: 16 × T0 2h: 32 × T0 3h: 64 × T0 (default) 4h: 128 × T0 5h: 256 × T0 6h: 512 × T0 7h: 1024 × T0

(1) See Timing Control and Frequency Scaling (CLKIN) for the definition of the time period T0.

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Table 12. TIMEOUT Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
[2]	ECHO_TIMEOUT	R/W	0h	Echo receive timeout control (see TOF Measurement Interval) 0h: Enable echo timeout (default) 1h: Disable timeout
[1:0]	TOF_TIMEOUT_CTRL(1)	R/W	1h	Echo listening window timeout (see TOF Measurement Interval) 0h: 128 × T0 1h: 256 × T0 (default) 2h: 512 × T0 3h: 1024 × T0

8.6.1.10 CLOCK_RATE Register (address = 9h) [reset = 0h] (map)

Figure 50. CLOCK_RATE Register

(MSB) 7	6	5	4	3	2	1	0 (LSB)
RESERVED					**CLOCKIN_DIV**	**AUTOZERO_PERIOD**	
R/W-0h					R/W-0h	R/W-0h	

LEGEND: R/W = Read or write; R = Read only; R/W1C = Read or write 1 to clear

Table 13. CLOCK_RATE Register Field Descriptions(1)

Bit	Field	Type	Reset	Description
[7:3]	RESERVED	R/W	0h	0h: Reserved (default)
[2]	CLOCKIN_DIV(1)	R/W	0h	CLKIN divider to generate T0 0h: Divide by 1 (default) 1h: Divide by 2
[1:0]	AUTOZERO_PERIOD(1)	R/W	0h	Receiver auto-zero period 0h: 64 × T0 (default) 1h: 128 × T0 2h: 256 × T0 3h: 512 × T0

(1) See Timing Control and Frequency Scaling (CLKIN) for the definition of the time period T0.

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9 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

9.1 Application Information

The TDC1011 is an analog front-end for ultrasonic sensing applications. The device is typically used for the driving and sensing of ultrasonic transducers to perform accurate time-of-flight measurements. Ultrasonic time-of-flight sensing allows for fluid level, fluid identification or concentration measurements.

9.2 Typical Applications

9.2.1 Level and Fluid Identification Measurements

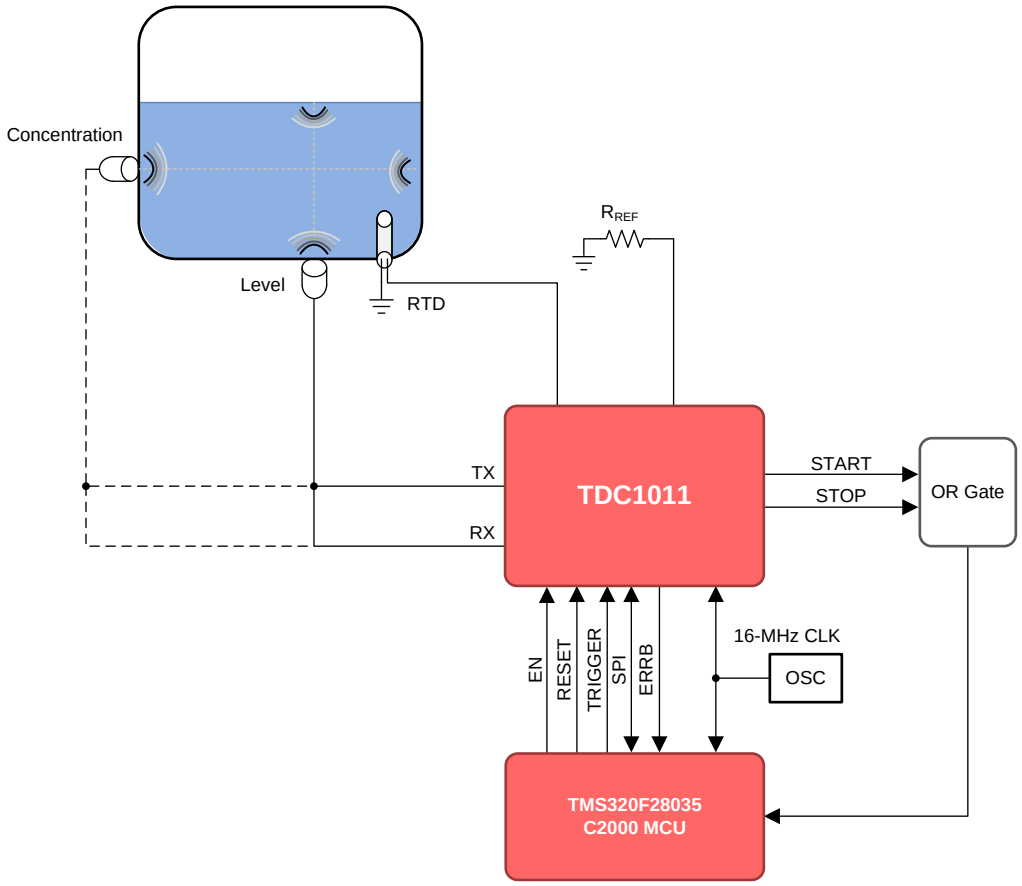

Figure 51. Level or Concentration Measurement Application Diagram

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Typical Applications (continued)

9.2.1.1 Design Requirements

Table 14. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Fluid Level	
Range	2 – 10 cm
Fluid Identification	
Accuracy	0.5% concentration variation
Distance	5.08 cm

9.2.1.2 Detailed Design Procedure

9.2.1.2.1 Level Measurements

For level sensing applications, the total time-of-flight (TOF) of the sound wave in the fluid is measured. The pulses transmitted by a bottom mounted transducer travel through the fluid, to the surface of the fluid. The discontinuity between the fluid and air generates a reflected wave which returns back to the bottom mounted transducer.

At the beginning of a measurement cycle, the transducer is connected to a transmit channel of the AFE, and the transmit burst excites the transducer to generate an ultrasound wave. Synchronous to the TX burst, a START pulse is generated by the TDC1011 to indicate the start of a measurement. After the transmission is completed, and depending on the device configuration, the transducer is connected to a receive channel of the AFE.

When a valid echo is received, the TDC1011 generates a STOP pulse. Generation of multiple STOP pulses is possible through register configuration of the device. The START and STOP signal times are compared to determine the TOF.

The level of the fluid can be determined using the following equation:

$$d = \frac{TOF \times c}{2} \quad (6)$$

where

- *d* is the fluid level in meters (m)
- *TOF* is the time-of-flight in seconds (s)
- *c* is the speed of sound in the fluid in meters per second (m/s)

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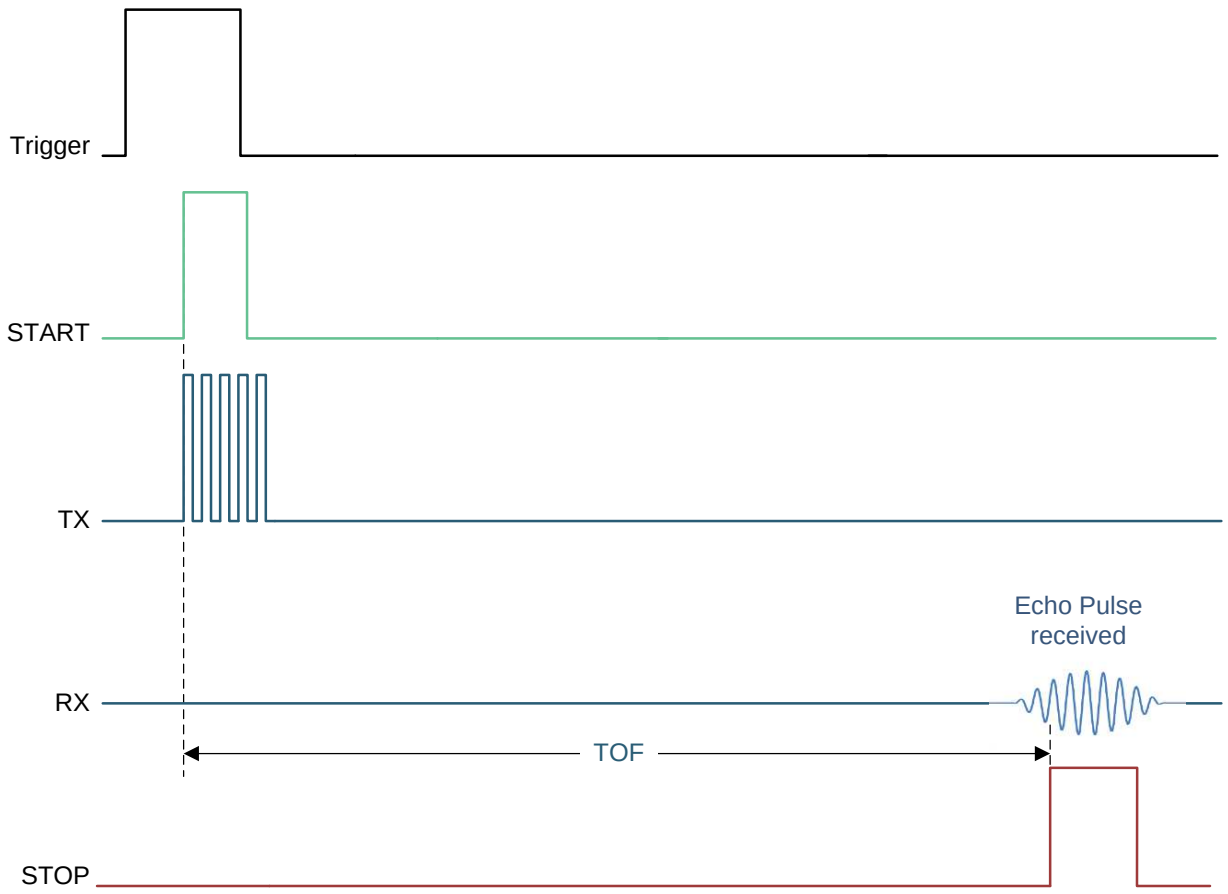

Figure 52. Relation Between Transmit and Receive Pulses in Level Measurements

Level measurements have 2 main criteria: resolution and range (maximum height). Resolution accuracies of 1-2 mm are achievable but are impractical due to any environmental disturbances, such as tank vibrations, creating millimeter level surface waves. Ranges of up to 1 m are measurable utilizing VDD level excitation pulses, but surface disturbances and signal loss over longer distances make the reliable echo reception an issue. Greater level measurement reception can be achieved by mechanical means (level guide tube) and/or electronic means (level shifting the TX pulses to greater voltages; see TIDA-00322).

9.2.1.2.2 Fluid Identification

The TDC1011 can be used to measure the time-of-flight for a known distance to calculate the speed of sound (c_{medium}) in the fluid. This application utilizes the same circuitry as the level example but a transducer in a side mounted configuration transmitting across the container or to a target at a known distance from the transducer.

The temperature can also be measured to compensate for the temperature variation of sound. With the known distance, TOF and temperature, the speed of sound in the fluid can be determined and the identity of the medium verified.

After measuring the time-of-flight for the fixed distance, the speed of sound can be calculated as follows:

$$c_{medium} = \frac{2 \times d}{TOF} \quad (7)$$

where

- c_{medium} is the speed of sound in the fluid in meters per second (m/s)
- d is the level in meters (m)
- TOF is the time of flight in seconds (s)

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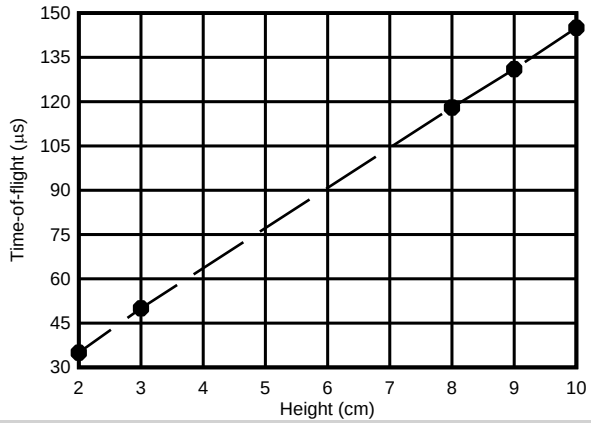
The measurement process is identical to the level example above. The speed of sound can be used to uniquely identify a variety of fluids. In this example, the concentration of diesel exhaust fluid (DEF) is measured with a desired accuracy resolution of 0.5% of concentration variation. For most fluids, the speed of sound varies over temperature, so every application will be different. In this example, all samples were all at ambient temperature of 23°C.

9.2.1.3 Application Curves

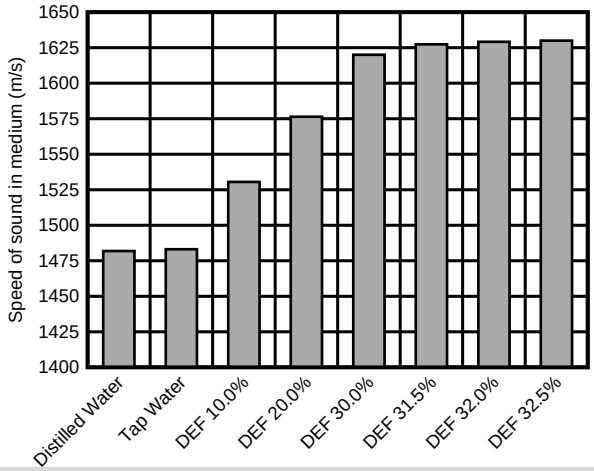
The data used in the following level and fluid identification graphs was collected using an ultrasonic test cell. The test cell is an acrylic plastic container with width of 2.54 cm and ultrasonic transducers attached to the outside using cyanoacrylate glue. The transducers in this experiment were STEMiNC 1MHz piezo electric ceramic discs (SMD10T2R111). Equivalent transducers with the following characteristics could be used:

- Piezo material: SM111
- Dimensions: 10mm diameter x 2mm thickness
- Resonant frequency: 1050 kHz (thickness mode)



Fluid Height in Tank	Time-of-Flight (µs)
Full (10 cm)	145
Full – 1 (9 cm)	131
Full – 2 (8 cm)	118
3 cm	50
2 cm	35

Figure 53. Time-of-Flight for Fluid Height in Tank



Fluid	Speed of sound (m/s)
Distilled water	1481.87
Tap water	1483.13
DEF 10.0%	1530.49
DEF 20.0%	1576.42
DEF 30.0%	1620.00
DEF 31.5%	1627.37
DEF 32.0%	1629.15
DEF 32.5%	1630.00

Figure 54. Speed of Sound for Various Fluids and Diesel Exhaust Fluid (DEF) Concentration

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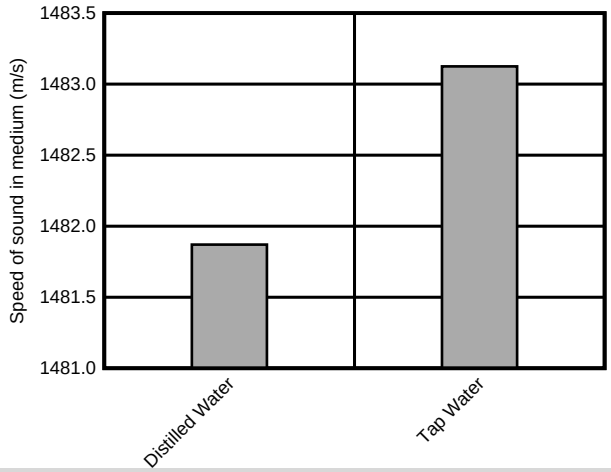
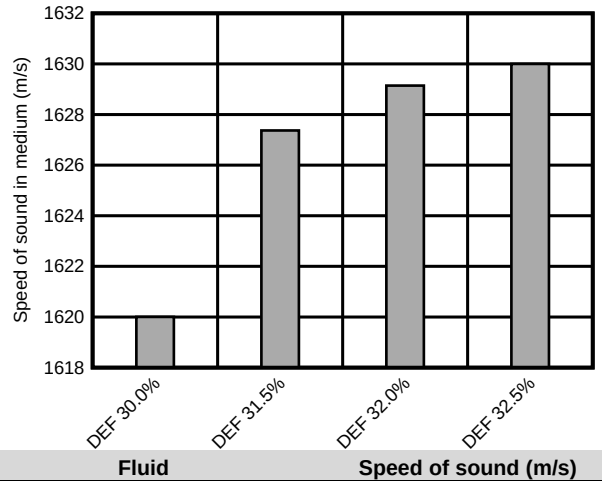


Fluid	Speed of sound (m/s)
Distilled water	1481.87
Tap water	1483.13

Figure 55. Speed of Sound in Distilled Water and Tap Water



Fluid	Speed of sound (m/s)
DEF 30.0%	1620.00
DEF 31.5%	1627.37
DEF 32.0%	1629.15
DEF 32.5%	1630.00

Figure 56. Speed of Sound of Various Diesel Exhaust Fluid (DEF) Concentrations

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10 Power Supply Recommendations

The analog circuitry of the TDC1011 is designed to operate from an input voltage supply range between 2.7V and 5.5V. It is recommended to place a 100nF ceramic bypass capacitor to ground as close as possible to the VDD pins. In addition, an electrolytic or tantalum capacitor with value greater than 1µF is recommended. The bulk capacitor does not need to be in close vicinity with the TDC1011 and could be close to the voltage source terminals or at the output of the voltage regulators powering the TDC1011.

The IO circuitry of the TDC1011 is designed to operate from an input voltage supply range between 1.8V and 5.5V. The IO voltage supply (V_{IO}) can be lower than the analog voltage supply (V_{DD}), but it should not exceed it. It is also recommended to place a 100nF ceramic bypass capacitor to ground as close as possible to the VIO pin. If a separate source or regulator is used for VIO, an additional electrolytic or tantalum capacitor with value greater than 1µF is recommended.

In some cases an additional 10µF bypass capacitor may further reduce the supply noise.

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11 Layout

11.1 Layout Guidelines

- In a 4-layer board design, the recommended layer stack order from top to bottom is: signal, ground, power and signal.
- Bypass capacitors should be placed in close proximity to the VDD and VIO pins.
- The length of the START and STOP traces from the DUT to the stopwatch/MCU should be matched to prevent uneven signal delays. Also, avoid unnecessary via-holes on these traces and keep the routing as short/direct as possible to minimize parasitic capacitance on the PCB.
- Match the length (or resistance) of the traces leading to the RTD sensors. PCB series resistance will be added in series to the RTD sensors.
- Route the SPI signal traces close together. Place a series resistor at the source of SDO (close to the DUT) and series resistors at the sources of SDI, SCLK and CSB (close to the master MCU).

11.2 Layout Example

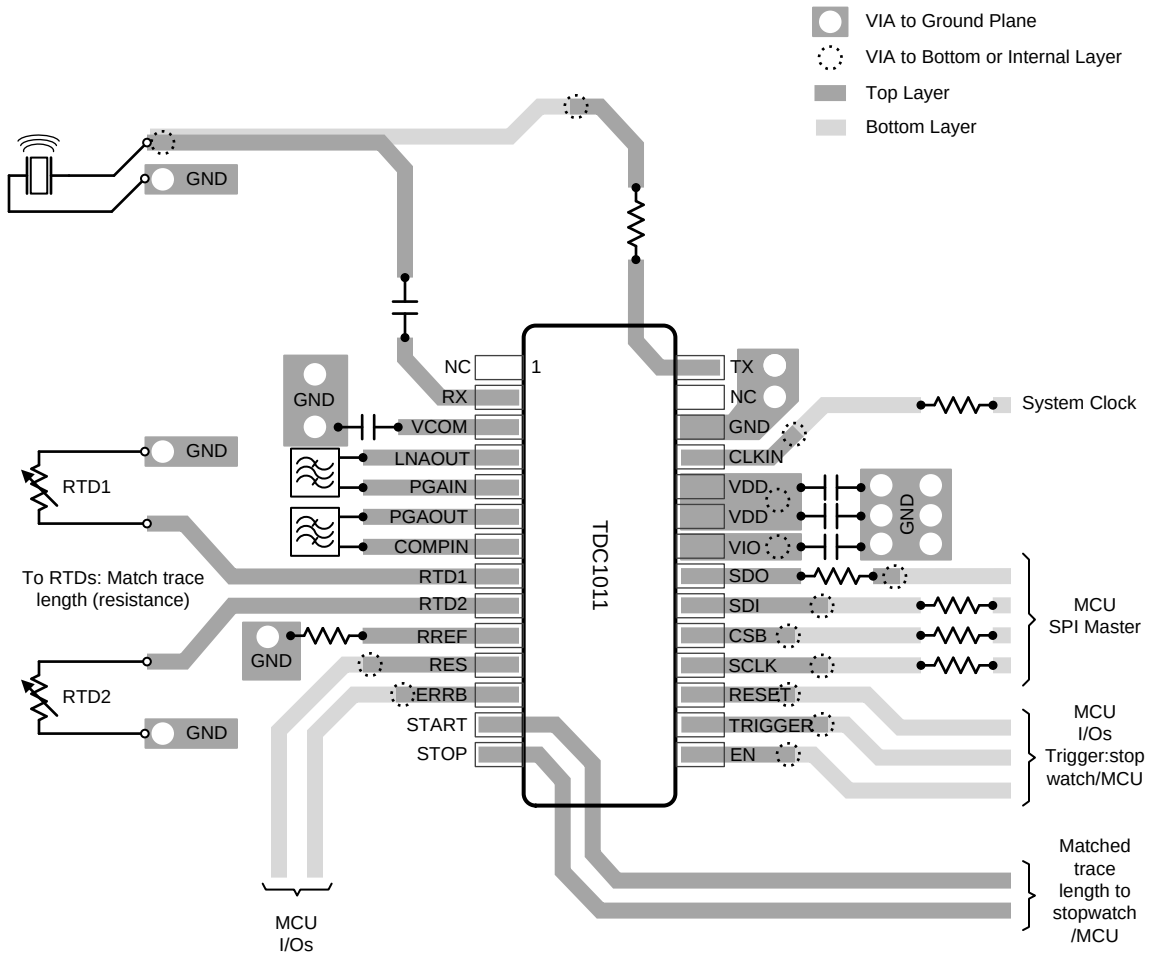

Figure 57. TDC1011 Board Layout (Capacitive Feedback Mode)

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12 器件和文档支持

12.1 器件支持

12.1.1 Third-Party Products Disclaimer

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12.1.2 开发支持

相关开发支持请参阅以下文档：

- *《汽车类超声波液位/质量测量参考设计》*，TIDA-00322

12.2 社区资源

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12.5 Glossary

SLYZ022 — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

13 机械、封装和可订购信息

以下页中包括机械、封装和可订购信息。 这些信息是针对指定器件可提供的最新数据。 这些数据会在无通知且不对本文档进行修订的情况下发生改变。 欲获得该数据表的浏览器版本，请查阅左侧的导航栏。

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package | Pins	Package qty | Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
TDC1011PW	Active	Production	TSSOP (PW) | 28	48 | TUBE	Yes	SN	Level-1-260C-UNLIM	-40 to 125	TDC1011PW
TDC1011PW.A	Active	Production	TSSOP (PW) | 28	48 | TUBE	Yes	SN	Level-1-260C-UNLIM	-40 to 125	TDC1011PW
TDC1011PWR	Active	Production	TSSOP (PW) | 28	2500 | LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	TDC1011PW
TDC1011PWR.A	Active	Production	TSSOP (PW) | 28	2500 | LARGE T&R	Yes	SN	Level-1-260C-UNLIM	-40 to 125	TDC1011PW

(1) **Status:** For more details on status, see our product life cycle.

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the TI RoHS Statement for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

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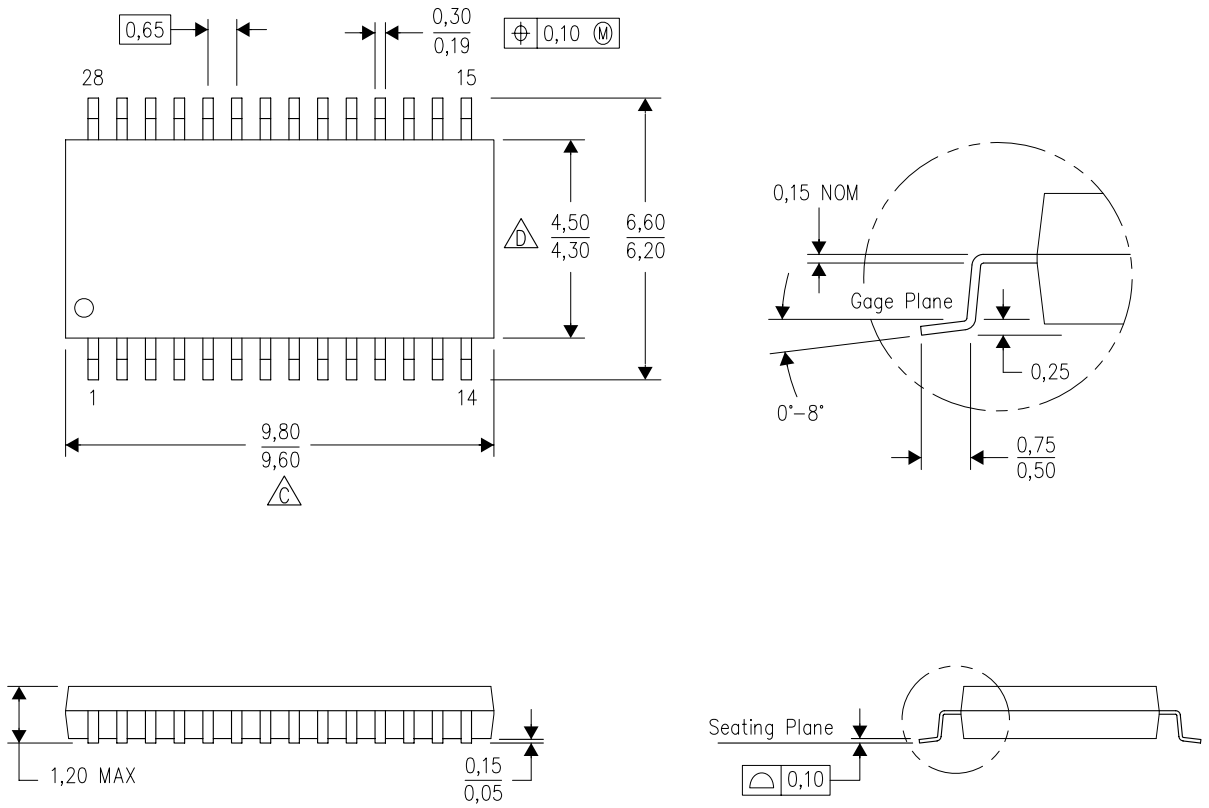
OTHER QUALIFIED VERSIONS OF TDC1011 :

- Automotive : TDC1011-Q1

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

PW (R–PDSO–G28) PLASTIC SMALL OUTLINE



4040064–7/G 02/11

NOTES:
A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M–1994.
B. This drawing is subject to change without notice.
C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
E. Falls within JEDEC MO–153

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