

PCMD3180 Octal-Channel, PDM Input to TDM or I²S Output Converter

1 Features

- 8-channel PDM microphones simultaneous conversion
- PDM input to TDM or I²S output converter performance:
 - 127dB dynamic range (DR) with high performance 5th order PDM input
 - 117dB dynamic range (DR) with high performance 4th order PDM input
- Channel summing mode, DR performance with high performance 4th order PDM input:
 - 120dB, 2-channel summing
 - 123dB, 4-channel summing
- Programmable PDM clock output :
 - 768kHz to 6.144MHz
- Programmable output sample rate (f_S) :
 - 8kHz to 768kHz
- Programmable channel settings:
 - Digital volume control: –100dB to 27dB
 - Gain calibration with 0.1dB resolution
 - Phase calibration with 163ns resolution
- Microphone bias or supply voltage generation
- Low-latency signal processing filter selection
- Programmable HPF and biquad digital filters
- I²C or SPI controls
- Integrated high-performance audio PLL
- Automatic clock divider setting configurations
- Audio serial data interface:
 - Format: TDM, I²S, or left-justified (LJ)
 - Word length: 16 bits, 20 bits, 24 bits, or 32 bits
 - Master or slave interface
- Single-supply operation: 3.3V or 1.8V
- I/O-supply operation: 3.3V or 1.8V
- Power consumption for 1.8V supply:
 - 2.9mW/channel at 16kHz sample rate
 - 2.5mW/channel at 48kHz sample rate

2 Applications

- Video doorbell
- Smart speakers
- Building security gateway
- IP network cameras
- GPS personal navigation device
- Video conference systems

3 Description

The PCMD3180 is a high-performance, pulse-density-modulation (PDM) input to time-division multiplexing (TDM) or I²S output converter that supports simultaneous sampling of up to eight digital channels for the PDM microphone input. The device integrates programmable digital volume control, a microphone bias voltage, a phase-locked loop (PLL), a programmable high-pass filter (HPF), biquad filters, low-latency filter modes, and allows for output sample rates up to 768kHz. The device supports time-division multiplexing (TDM), I²S, or left-justified (LJ) audio formats, and can be controlled with either the I²C or SPI. Additionally, the PCMD3180 supports master and slave mode selection for the audio bus interface operation. These integrated high-performance features, along with the ability to be powered from a single-supply of 3.3V or 1.8V, make the device an excellent choice for space-constrained audio systems in far-field microphone recording applications.

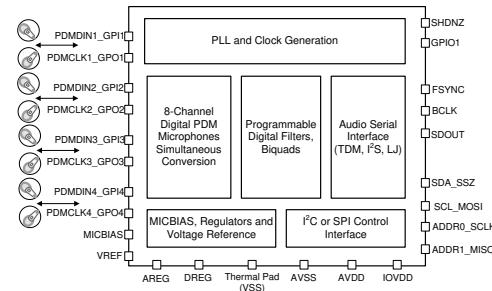
The PCMD3180 is specified from –40°C to +125°C, and is offered in a 24-pin WQFN package.

Package Information

PART NUMBER	PACKAGE ⁽¹⁾	PACKAGE SIZE ⁽²⁾
PCMD3180	RTW (WQFN, 24)	4.00mm × 4.00mm with 0.5mm pitch

(1) For all available packages, see the package option addendum at the end of the data sheet.

(2) The package size (length × width) is a nominal value and includes pins, where applicable.



Simplified Block Diagram



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4 Pin Configuration and Functions

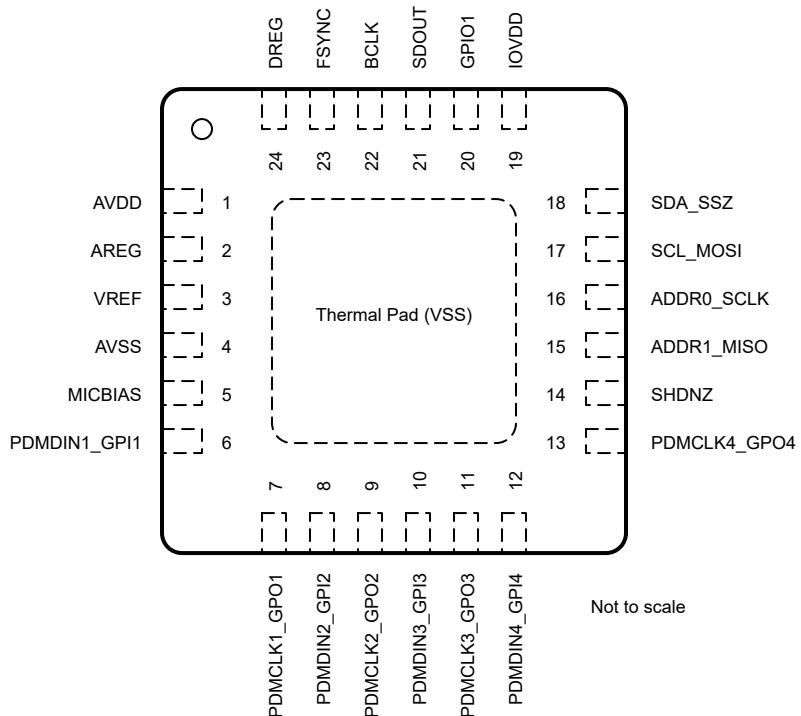


Figure 4-1. RTW Package 24-Pin WQFN With Exposed Thermal Pad Top View

Table 4-1. Pin Functions

PIN		TYPE	DESCRIPTION
NO.	NAME		
1	AVDD	Analog supply	Analog power (1.8V or 3.3V, nominal)
2	AREG	Analog supply	Analog on-chip regulator output voltage for analog supply (1.8V, nominal) or external analog power (1.8V, nominal). Connect a 10µF and 0.1µF low ESR capacitors in parallel to analog ground (AVSS)
3	VREF	Analog	Analog reference voltage filter output. Connect a 1µF to analog ground (AVSS)
4	AVSS	Analog supply	Analog ground. Short this pin directly to the board ground plane.
5	MICBIAS	Analog	MICBIAS output. Connect a 1µF to analog ground (AVSS)
6	PDMDIN1_GPI1	Digital input	PDM microphone data input 1 or general-purpose digital input 1 (multipurpose functions such as digital microphone data, PLL input clock source, and so forth)
7	PDMCLK1_GPO1	Digital output	PDM microphone clock output 1 or general-purpose digital output 1 (multipurpose functions such as digital microphone clock, interrupt, and so forth)
8	PDMDIN2_GPI2	Digital input	PDM microphone data input 2 or general-purpose digital input 2 (multipurpose functions such as digital microphone data, PLL input clock source, and so forth)
9	PDMCLK2_GPO2	Digital output	PDM microphone clock output 2 or general-purpose digital output 2 (multipurpose functions such as digital microphone clock, interrupt, and so forth)
10	PDMDIN3_GPI3	Digital input	PDM microphone data input 3 or general-purpose digital input 3 (multipurpose functions such as digital microphone data, PLL input clock source, and so forth)
11	PDMCLK3_GPO3	Digital output	PDM microphone clock output 3 or general-purpose digital output 3 (multipurpose functions such as digital microphone clock, interrupt, and so forth)
12	PDMDIN4_GPI4	Digital input	PDM microphone data input 4 or general-purpose digital input 4 (multipurpose functions such as digital microphone data, PLL input clock source, and so forth)
13	PDMCLK4_GPO4	Digital output	PDM microphone clock output 4 or general-purpose digital output 4 (multipurpose functions such as digital microphone clock, interrupt, and so forth)

Table 4-1. Pin Functions (continued)

PIN		TYPE	DESCRIPTION
NO.	NAME		
14	SHDNZ	Digital input	Device hardware shutdown and reset (active low)
15	ADDR1_MISO	Digital I/O	For I ² C operation: I ² C slave address A1 pin For SPI operation: SPI slave output pin
16	ADDR0_SCLK	Digital input	For I ² C operation: I ² C slave address A0 pin For SPI operation : SPI serial bit clock
17	SCL_MOSI	Digital input	For I ² C operation: clock pin for I ² C control bus For SPI operation: SPI slave input pin
18	SDA_SSZ	Digital I/O	For I ² C operation: data pin for I ² C control bus For SPI operation: SPI slave-select pin
19	IOVDD	Digital supply	Digital I/O power supply (1.8V or 3.3V, nominal)
20	GPIO1	Digital I/O	General-purpose digital input/output 1 (multipurpose functions such as digital microphones clock or data, PLL input clock source, interrupt, and so forth)
21	SDOUT	Digital output	Audio serial data interface bus output
22	BCLK	Digital I/O	Audio serial data interface bus bit clock
23	FSYNC	Digital I/O	Audio serial data interface bus frame synchronization signal
24	DREG	Digital supply	Digital regulator output voltage for digital core supply (1.5V, nominal). Connect a 10 μ F and 0.1 μ F low ESR capacitors in parallel to device ground (VSS)
Thermal Pad	Thermal Pad (VSS)	Ground supply	Thermal pad shorted to internal device ground. Short the thermal pad directly to the board ground plane.

5 Specifications

5.1 Absolute Maximum Ratings

over the operating ambient temperature range (unless otherwise noted)⁽¹⁾

			MIN	MAX	UNIT
Supply voltage	AVDD to AVSS		-0.3	3.9	V
	AREG to AVSS		-0.3	2.0	
	IOVDD to VSS (thermal pad)		-0.3	3.9	
Ground voltage differences	AVSS to VSS (thermal pad)		-0.3	0.3	V
Digital input voltage	Digital input except PDMDINx_GPIx pins voltage to VSS (thermal pad)		-0.3	IOVDD + 0.3	V
	Digital input PDMDINx_GPIx pins voltage to VSS (thermal pad)		-0.3	AVDD + 0.3	
Temperature	Operating ambient, T_A		-40	125	°C
	Junction, T_J		-40	150	
	Storage, T_{stg}		-65	150	

(1) Operation outside the *Absolute Maximum Ratings* may cause permanent device damage. *Absolute Maximum Ratings* do not imply functional operation of the device at these or any other conditions beyond those listed under *Recommended Operating Conditions*. If used outside the *Recommended Operating Conditions* but within the *Absolute Maximum Ratings*, the device may not be fully functional, and this may affect device reliability, functionality, performance, and shorten the device lifetime.

5.2 ESD Ratings

			VALUE	UNIT
V _(ESD)	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	V
		Charged-device model (CDM), per JEDEC specification JESD22C101 ⁽²⁾	±500	

(1) JEDEC document JEP155 states that 500V HBM allows safe manufacturing with a standard ESD control process.
 (2) JEDEC document JEP157 states that 250V CDM allows safe manufacturing with a standard ESD control process.

5.3 Recommended Operating Conditions

		MIN	NOM	MAX	UNIT
POWER					
AVDD, AREG ⁽¹⁾	Analog supply voltage AVDD to AVSS (AREG is generated using onchip regulator) - AVDD 3.3V operation	3.0	3.3	3.6	V
	Analog supply voltage AVDD and AREG to AVSS (AREG internal regulator is shutdown) - AVDD 1.8V operation	1.7	1.8	1.9	
IOVDD	IO supply voltage to VSS (thermal pad) - IOVDD 3.3V operation	3.0	3.3	3.6	V
	IO supply voltage to VSS (thermal pad) - IOVDD 1.8V operation	1.65	1.8	1.95	
INPUTS					
	Digital input except PDMDINx_GPIx pins voltage to VSS (thermal pad)	0	IOVDD	V	
	Digital input PDMDINx_GPIx pins voltage to VSS (thermal pad)	0	AVDD	V	
TEMPERATURE					
T_A	Operating ambient temperature	-40	125	°C	
OTHERS					
	GPIOx or GPIx (used as MCLK input) clock frequency		36.864	MHz	
C_b	SCL and SDA bus capacitance for I ² C interface supports standard-mode and fast-mode		400	pF	
	SCL and SDA bus capacitance for I ² C interface supports fast-mode plus		550		
C_L	Digital output load capacitance	20	50	pF	

(1) AVSS and VSS (thermal pad): all ground pins must be tied together and must not differ in voltage by more than 0.2V.

5.4 Thermal Information

THERMAL METRIC ⁽¹⁾		PCMD3180	UNIT
		RTW (WQFN)	
		24 PINS	
R _{θJA}	Junction-to-ambient thermal resistance	32.6	°C/W
R _{θJC(top)}	Junction-to-case (top) thermal resistance	25.0	°C/W
R _{θJB}	Junction-to-board thermal resistance	11.9	°C/W
Ψ _{JT}	Junction-to-top characterization parameter	0.2	°C/W
Ψ _{JB}	Junction-to-board characterization parameter	11.9	°C/W
R _{θJC(bot)}	Junction-to-case (bottom) thermal resistance	2.9	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application note.

5.5 Electrical Characteristics

at T_A = 25°C, AVDD = 3.3V, IOVDD = 3.3V, f_{IN} = 1kHz sinusoidal signal, f_S = 48kHz, PDMCLK_x = 64 × f_S, 32-bit audio data, BCLK = 256 × f_S, TDM slave mode, PLL on (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
PERFORMANCE FOR PDM INPUT CONVERSION						
SNR	Signal-to-noise ratio, A-weighted ^{(1) (2) (3)}	No signal, input generated using 5 th order PDM modulator		128		dB
		No signal, input generated using 4 th order PDM modulator		118		
DR	Dynamic range, A-weighted ^{(2) (3)}	–60dB full-scale signal input, input generated using 5 th order PDM modulator		127		dB
		–60dB full-scale signal input, input generated using 4 th order PDM modulator		117		
OTHER PARAMETERS						
	Digital volume control range	Programmable 0.5dB steps	–100	27		dB
	PDMCLK _x output rate	Programmable	0.7056	6.144		MHz
	Output data sample rate	Programmable	7.35	768		kHz
	Output data sample word length	Programmable	16	32		Bits
	Digital high-pass filter cutoff frequency	First-order IIR filter with programmable coefficients, –3dB point (default setting)		12		Hz
	MICBIAS voltage	MICBIAS programmed to VREF and VREF programmed to either 2.75V, 2.5V, or 1.375V		VREF		V
		Bypass to AVDD with 20mA load		AVDD – 0.2		
DIGITAL I/O						
V _{IL(SHDNZ)}	Low-level digital input logic voltage threshold	SHDNZ pin	–0.3	0.25 × IOVDD		V
V _{IH(SHDNZ)}	High-level digital input logic voltage threshold	SHDNZ pin	0.75 × IOVDD	IOVDD + 0.3		V
V _{IL}	Low-level digital input logic voltage threshold	All digital pins except PDMDIN _x _GPI _x , SDA and SCL, IOVDD 1.8V operation	–0.3	0.35 × IOVDD		V
		All digital pins except PDMDIN _x _GPI _x , SDA and SCL, IOVDD 3.3V operation	–0.3	0.8		
V _{IH}	High-level digital input logic voltage threshold	All digital pins except PDMDIN _x _GPI _x , SDA and SCL, IOVDD 1.8V operation	0.65 × IOVDD	IOVDD + 0.3		V
		All digital pins except PDMDIN _x _GPI _x , SDA and SCL, IOVDD 3.3V operation	2	IOVDD + 0.3		
V _{OL}	Low-level digital output voltage	All digital pins except PDMCLK _x _GPO _x , SDA and SCL, I _{OL} = –2mA, IOVDD 1.8V operation		0.45		V
		All digital pins except PDMCLK _x _GPO _x , SDA and SCL, I _{OL} = –2mA, IOVDD 3.3V operation		0.4		

5.5 Electrical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $\text{AVDD} = 3.3\text{V}$, $\text{IOVDD} = 3.3\text{V}$, $f_{IN} = 1\text{kHz}$ sinusoidal signal, $f_S = 48\text{kHz}$, $\text{PDMCLK}_x = 64 \times f_S$, 32-bit audio data, $\text{BCLK} = 256 \times f_S$, TDM slave mode, PLL on (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
V_{OH}	All digital pins except $\text{PDMCLK}_x\text{_GPO}_x$, SDA and SCL, $I_{OH} = 2\text{mA}$, $\text{IOVDD} = 1.8\text{V}$ operation	$\text{IOVDD} - 0.45$		2.4	V
	All digital pins except $\text{PDMCLK}_x\text{_GPO}_x$, SDA and SCL, $I_{OH} = 2\text{mA}$, $\text{IOVDD} = 3.3\text{V}$ operation	0.4			
$V_{IL(\text{I}2\text{C})}$	Low-level digital input logic voltage threshold	SDA and SCL	-0.5	$0.3 \times \text{IOVDD}$	V
$V_{IH(\text{I}2\text{C})}$	High-level digital input logic voltage threshold	SDA and SCL	$0.7 \times \text{IOVDD}$	$\text{IOVDD} + 0.5$	V
$V_{OL1(\text{I}2\text{C})}$	Low-level digital output voltage	SDA, $I_{OL(\text{I}2\text{C})} = -3\text{mA}$, $\text{IOVDD} > 2\text{V}$	0.4		V
$V_{OL2(\text{I}2\text{C})}$	Low-level digital output voltage	SDA, $I_{OL(\text{I}2\text{C})} = -2\text{mA}$, $\text{IOVDD} \leq 2\text{V}$	$0.2 \times \text{IOVDD}$		V
$I_{OL(\text{I}2\text{C})}$	Low-level digital output current	SDA, $V_{OL(\text{I}2\text{C})} = 0.4\text{V}$, standard-mode or fast-mode	3	20	mA
		SDA, $V_{OL(\text{I}2\text{C})} = 0.4\text{V}$, fast-mode plus	20		
I_{IL}	Input logic-low leakage for digital inputs	All digital pins except $\text{PDMDIN}_x\text{_GPIO}_x$ pins, input = 0V	-5	0.1	5
I_{IH}	Input logic-high leakage for digital inputs	All digital pins except $\text{PDMDIN}_x\text{_GPIO}_x$ pins, input = IOVDD	-5	0.1	5
$V_{IL(\text{GPIO}_x)}$	Low-level digital input logic voltage threshold	All $\text{PDMDIN}_x\text{_GPIO}_x$ digital pins, $\text{AVDD} = 1.8\text{V}$ operation	-0.3	$0.35 \times \text{AVDD}$	V
		All $\text{PDMDIN}_x\text{_GPIO}_x$ digital pins, $\text{AVDD} = 3.3\text{V}$ operation	-0.3	0.8	
$V_{IH(\text{GPIO}_x)}$	High-level digital input logic voltage threshold	All $\text{PDMDIN}_x\text{_GPIO}_x$ digital pins, $\text{AVDD} = 1.8\text{V}$ operation	$0.65 \times \text{AVDD}$	$\text{AVDD} + 0.3$	V
		All $\text{PDMDIN}_x\text{_GPIO}_x$ digital pins, $\text{AVDD} = 3.3\text{V}$ operation	2	$\text{AVDD} + 0.3$	
$V_{OL(\text{GPO}_x)}$	Low-level digital output voltage	All $\text{PDMCLK}_x\text{_GPO}_x$ digital pins, $I_{OL} = -2\text{mA}$, $\text{AVDD} = 1.8\text{V}$ operation	0.45		V
		All $\text{PDMCLK}_x\text{_GPO}_x$ digital pins, $I_{OL} = -2\text{mA}$, $\text{AVDD} = 3.3\text{V}$ operation	0.4		
$V_{OH(\text{GPO}_x)}$	High-level digital output voltage	All $\text{PDMCLK}_x\text{_GPO}_x$ digital pins, $I_{OH} = 2\text{mA}$, $\text{AVDD} = 1.8\text{V}$ operation	$\text{AVDD} - 0.45$		V
		All $\text{PDMCLK}_x\text{_GPO}_x$ digital pins, $I_{OH} = 2\text{mA}$, $\text{AVDD} = 3.3\text{V}$ operation	2.4		
$I_{IL(\text{GPIO}_x)}$	Input logic-high leakage for digital inputs	All $\text{PDMDIN}_x\text{_GPIO}_x$ digital pins, input = 0V	-5	0.1	5
$I_{IH(\text{GPIO}_x)}$	Input logic-high leakage for digital inputs	All $\text{PDMDIN}_x\text{_GPIO}_x$ digital pins, input = AVDD	-5	0.1	5
C_{IN}	Input capacitance for digital inputs	All digital pins	5		pF
R_{PD}	Pulldown resistance for digital I/O pins when asserted on	20		k Ω	

TYPICAL SUPPLY CURRENT CONSUMPTION

I_{AVDD}	Current consumption in hardware shutdown mode	SHDNZ = 0, $\text{AVDD} = 3.3\text{V}$	1	μA
I_{AVDD}		SHDNZ = 0, $\text{AVDD} = 1.8\text{V}$, external AREG supply (AREG shorted to AVDD)	1	
I_{IOVDD}		SHDNZ = 0, all external clocks stopped, $\text{IOVDD} = 3.3\text{V}$	0.1	
I_{IOVDD}		SHDNZ = 0, all external clocks stopped, $\text{IOVDD} = 1.8\text{V}$	0.1	
I_{AVDD}	Current consumption in sleep mode (software shutdown mode)	All external clocks stopped, $\text{AVDD} = 3.3\text{V}$	5	μA
I_{AVDD}		All external clocks stopped, $\text{AVDD} = 1.8\text{V}$, external AREG supply (AREG shorted to AVDD)	5	
I_{IOVDD}		All external clocks stopped, $\text{IOVDD} = 3.3\text{V}$	0.1	
I_{IOVDD}		All external clocks stopped, $\text{IOVDD} = 1.8\text{V}$	0.1	

5.5 Electrical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $\text{AVDD} = 3.3\text{V}$, $\text{IOVDD} = 3.3\text{V}$, $f_{IN} = 1\text{kHz}$ sinusoidal signal, $f_S = 48\text{kHz}$, $\text{PDMCLK}_x = 64 \times f_S$, 32-bit audio data, $\text{BCLK} = 256 \times f_S$, TDM slave mode, PLL on (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I_{AVDD}	Current consumption with 8-channel PDM input recording, $f_S = 48\text{kHz}$, $\text{PDMCLK}_x = 64 \times f_S$, PLL on and $\text{BCLK} = 256 \times f_S$	$\text{AVDD} = 3.3\text{V}$		11.9		mA
		$\text{AVDD} = 1.8\text{V}$, external AREG supply (AREG shorted to AVDD)		11.3		
		$\text{IOVDD} = 3.3\text{V}$		0.7		
		$\text{IOVDD} = 1.8\text{V}$		0.4		
I_{AVDD}	Current consumption with 4-channel PDM input recording, $f_S = 16\text{kHz}$, $\text{PDMCLK}_x = 96 \times f_S$, PLL on and $\text{BCLK} = 256 \times f_S$	$\text{AVDD} = 3.3\text{V}$		7.2		mA
		$\text{AVDD} = 1.8\text{V}$, external AREG supply (AREG shorted to AVDD)		6.5		
		$\text{IOVDD} = 3.3\text{V}$		0.2		
		$\text{IOVDD} = 1.8\text{V}$		0.1		

- (1) Ratio of output level with 1kHz full-scale sine-wave input, to the output level with no signal, measured A-weighted over a 20Hz to 20kHz bandwidth using an audio analyzer.
- (2) All performance measurements done with 20kHz low-pass filter and, where noted, A-weighted filter. Failure to use such a filter can result in higher THD and lower SNR and dynamic range readings than shown in the Electrical Characteristics. The low-pass filter removes out-of-band noise, which, although not audible, can affect dynamic specification values.
- (3) The device performance parameters, SNR, DR and THD+N, are mainly limited by single-bit PDM modulator generated data output. The THD+N performance for single-bit PDM modulator output is generally not so good for signal above -10dB full-scale. The data sheet measured numbers are with using audio equipments consist of high performance PDM modulator output generator.

5.6 Timing Requirements: I²C Interface

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V (unless otherwise noted); see [Figure 1](#) for timing diagram

		MIN	NOM	MAX	UNIT
STANDARD-MODE					
f_{SCL}	SCL clock frequency	0	100		kHz
$t_{\text{HD:STA}}$	Hold time (repeated) START condition. After this period, the first clock pulse is generated.	4			μs
t_{LOW}	Low period of the SCL clock	4.7			μs
t_{HIGH}	High period of the SCL clock	4			μs
$t_{\text{SU:STA}}$	Setup time for a repeated START condition	4.7			μs
$t_{\text{HD:DAT}}$	Data hold time	0	3.45		μs
$t_{\text{SU:DAT}}$	Data setup time	250			ns
t_r	SDA and SCL rise time		1000		ns
t_f	SDA and SCL fall time		300		ns
$t_{\text{SU:STO}}$	Setup time for STOP condition	4			μs
t_{BUF}	Bus free time between a STOP and START condition	4.7			μs
FAST-MODE					
f_{SCL}	SCL clock frequency	0	400		kHz
$t_{\text{HD:STA}}$	Hold time (repeated) START condition. After this period, the first clock pulse is generated.	0.6			μs
t_{LOW}	Low period of the SCL clock	1.3			μs
t_{HIGH}	High period of the SCL clock	0.6			μs
$t_{\text{SU:STA}}$	Setup time for a repeated START condition	0.6			μs
$t_{\text{HD:DAT}}$	Data hold time	0	0.9		μs
$t_{\text{SU:DAT}}$	Data setup time	100			ns
t_r	SDA and SCL rise time	20	300		ns
t_f	SDA and SCL fall time	$20 \times (\text{IOVDD} / 5.5\text{V})$	300		ns
$t_{\text{SU:STO}}$	Setup time for STOP condition	0.6			μs
t_{BUF}	Bus free time between a STOP and START condition	1.3			μs
FAST-MODE PLUS					
f_{SCL}	SCL clock frequency	0	1000		kHz
$t_{\text{HD:STA}}$	Hold time (repeated) START condition. After this period, the first clock pulse is generated.	0.26			μs
t_{LOW}	Low period of the SCL clock	0.5			μs
t_{HIGH}	High period of the SCL clock	0.26			μs
$t_{\text{SU:STA}}$	Setup time for a repeated START condition	0.26			μs
$t_{\text{HD:DAT}}$	Data hold time	0			μs
$t_{\text{SU:DAT}}$	Data setup time	50			ns
t_r	SDA and SCL rise time		120		ns
t_f	SDA and SCL fall time	$20 \times (\text{IOVDD} / 5.5\text{V})$	120		ns
$t_{\text{SU:STO}}$	Setup time for STOP condition	0.26			μs
t_{BUF}	Bus free time between a STOP and START condition	0.5			μs

5.7 Timing Requirements: SPI

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V and 20pF load on all outputs (unless otherwise noted); see [Figure 2](#) for timing diagram

		MIN	NOM	MAX	UNIT
t_{SCLK}	SCLK period	40			ns
$t_{\text{H}(\text{SCLK})}$	SCLK high pulse duration	18			ns
$t_{\text{L}(\text{SCLK})}$	SCLK low pulse duration	18			ns
t_{LEAD}	Enable lead time	16			ns
t_{TRAIL}	Enable trail time	16			ns
t_{DSEQ}	Sequential transfer delay	20			ns
$t_{\text{SU}(\text{MOSI})}$	MOSI data setup time	8			ns
$t_{\text{HLD}(\text{MOSI})}$	MOSI data hold time	8			ns
$t_{\text{r}(\text{SCLK})}$	SCLK rise time		10%	90%	rise time
$t_{\text{f}(\text{SCLK})}$	SCLK fall time		90%	10%	fall time
				6	ns
				6	ns

5.8 Timing Requirements: TDM, I²S or LJ Interface

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V and 20pF load on all outputs (unless otherwise noted); see [Figure 3](#) for timing diagram

		MIN	NOM	MAX	UNIT
t_{BCLK}	BCLK period	40			ns
$t_{\text{H}(\text{BCLK})}$	BCLK high pulse duration ⁽¹⁾	18			ns
$t_{\text{L}(\text{BCLK})}$	BCLK low pulse duration ⁽¹⁾	18			ns
$t_{\text{SU}(\text{FSYNC})}$	FSYNC setup time	8			ns
$t_{\text{HLD}(\text{FSYNC})}$	FSYNC hold time	8			ns
$t_{\text{r}(\text{BCLK})}$	BCLK rise time		10%	90%	rise time
$t_{\text{f}(\text{BCLK})}$	BCLK fall time		90%	10%	fall time
				10	ns
				10	ns

(1) The BCLK minimum high or low pulse duration must be higher than 25ns (to meet the timing specifications), if the SDOUT data line is latched on the opposite BCLK edge polarity than the edge used by the device to transmit SDOUT data.

5.9 Timing Requirements: PDM Digital Microphone Interface

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V and 20pF load on all outputs (unless otherwise noted); see [Figure 4](#) for timing diagram

		MIN	NOM	MAX	UNIT
$t_{\text{SU}(\text{PDMDINx})}$	PDMDINx setup time	30			ns
$t_{\text{HLD}(\text{PDMDINx})}$	PDMDINx hold time	0			ns

5.10 Switching Characteristics: I²C Interface

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V (unless otherwise noted); see [Figure 1](#) for timing diagram

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		Standard-mode	250	1250	
		Fast-mode	250	850	
$t_{\text{d}(\text{SDA})}$		Fast-mode plus		400	ns

5.11 Switching Characteristics: SPI

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V and 20pF load on all outputs (unless otherwise noted); see [Figure 2](#) for timing diagram

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_{\text{a}(\text{MISO})}$	MISO access time			16	ns
$t_{\text{d}(\text{MISO})}$	SCLK to MISO delay		50%	of SCLK to 50%	of MISO
$t_{\text{dis}(\text{MISO})}$	MISO disable time			20	ns

5.12 Switching Characteristics: TDM, I²S or LJ Interface

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V and 20pF load on all outputs (unless otherwise noted); see [Figure 3](#) for timing diagram

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$t_d(\text{SDOUT-BCLK})$	BCLK to SDOUT delay	50% of BCLK to 50% of SDOUT		18	ns
$t_d(\text{SDOUT-FSYNC})$	FSYNC to SDOUT delay in TDM or LJ mode (for MSB data with $\text{TX_OFFSET} = 0$)	50% of FSYNC to 50% of SDOUT		18	ns
$f(\text{BCLK})$	BCLK output clock frequency: master mode ⁽¹⁾			24.576	MHz
$t_H(\text{BCLK})$	BCLK high pulse duration: master mode		14		ns
$t_L(\text{BCLK})$	BCLK low pulse duration: master mode		14		ns
$t_d(\text{FSYNC})$	BCLK to FSYNC delay: master mode	50% of BCLK to 50% of FSYNC		18	ns
$t_r(\text{BCLK})$	BCLK rise time: master mode	10% - 90% rise time		8	ns
$t_f(\text{BCLK})$	BCLK fall time: master mode	90% - 10% fall time		8	ns

(1) The BCLK output clock frequency must be lower than 18.5MHz (to meet the timing specifications), if the SDOUT data line is latched on the opposite BCLK edge polarity than the edge used by the device to transmit SDOUT data.

5.13 Switching Characteristics: PDM Digital Microphone Interface

at $T_A = 25^\circ\text{C}$, $\text{IOVDD} = 3.3\text{V}$ or 1.8V and 20pF load on all outputs (unless otherwise noted); see [Figure 4](#) for timing diagram

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$f(\text{PDMCLK})$	PDMCLK clock frequency		0.768	6.144	MHz
$t_H(\text{PDMCLK})$	PDMCLK high pulse duration		72		ns
$t_L(\text{PDMCLK})$	PDMCLK low pulse duration		72		ns
$t_r(\text{PDMCLK})$	PDMCLK rise time	10% - 90% rise time		18	ns
$t_f(\text{PDMCLK})$	PDMCLK fall time	90% - 10% fall time		18	ns

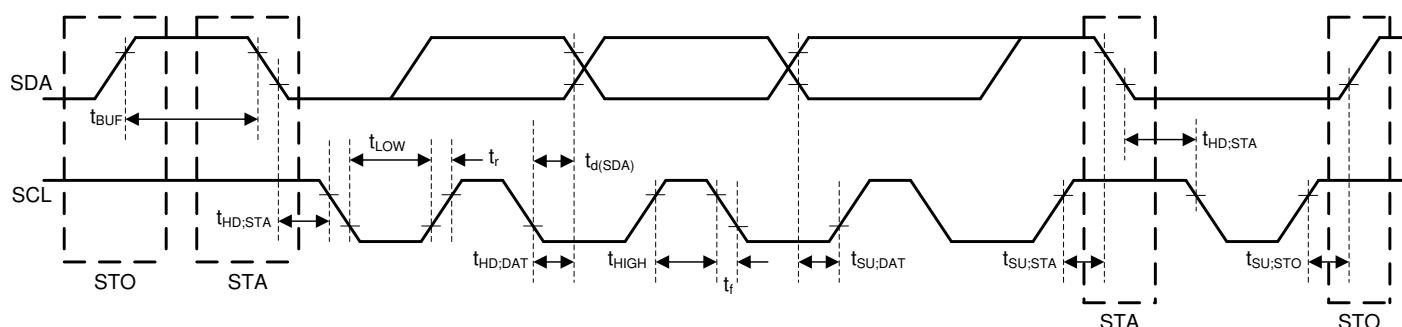


Figure 5-1. I²C Interface Timing Diagram

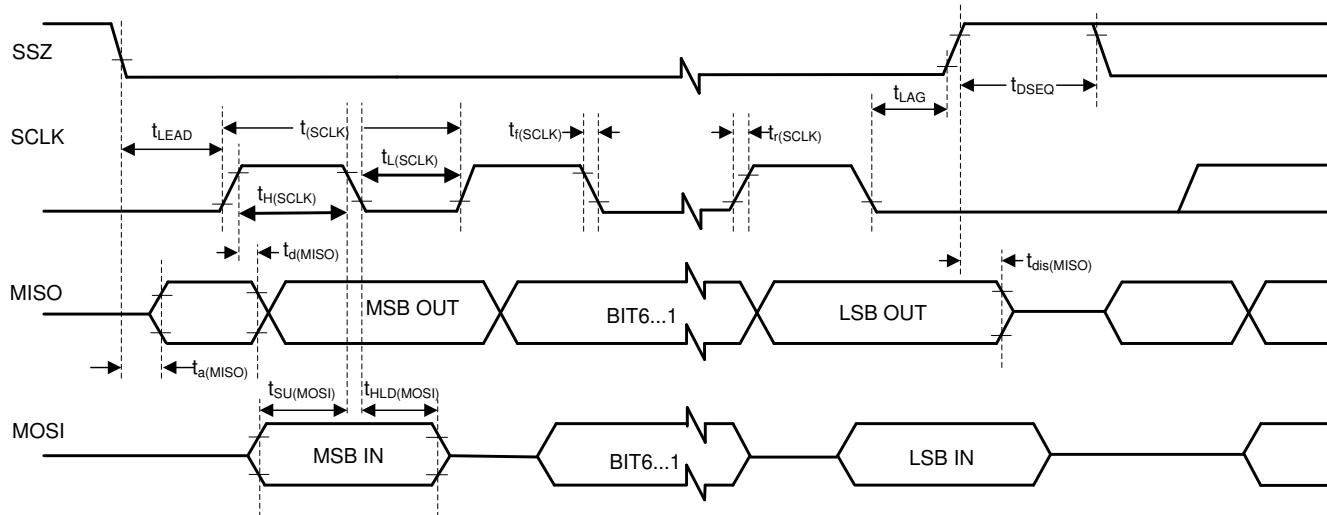


Figure 5-2. SPI Timing Diagram

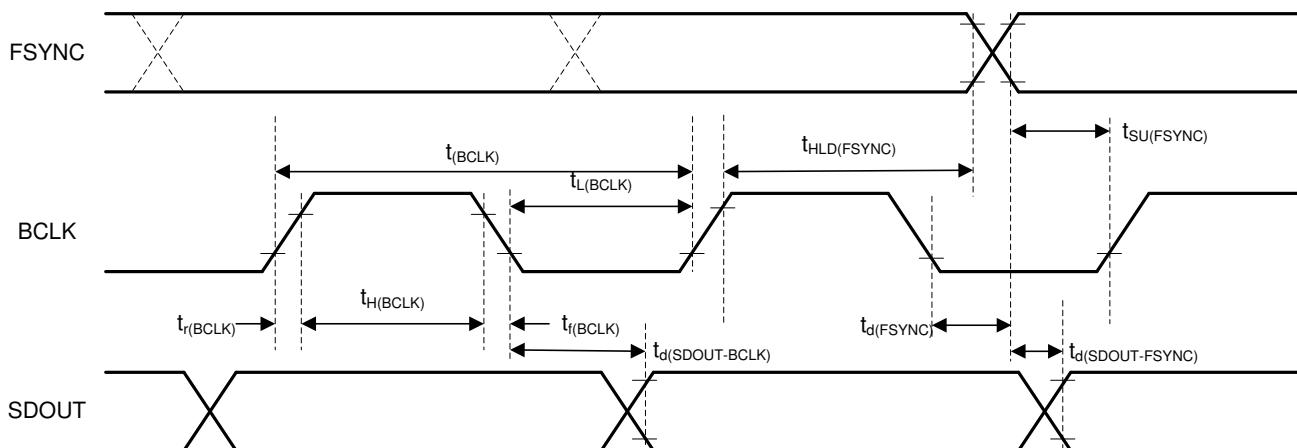


Figure 5-3. TDM (With BCLK_POL = 1), I²S, and LJ Interface Timing Diagram

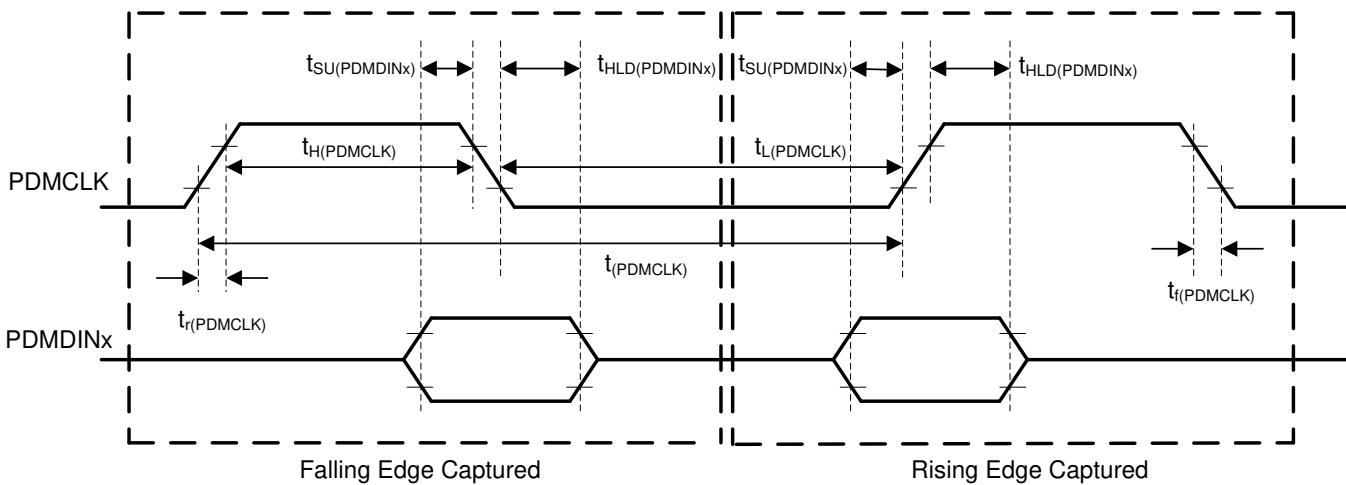
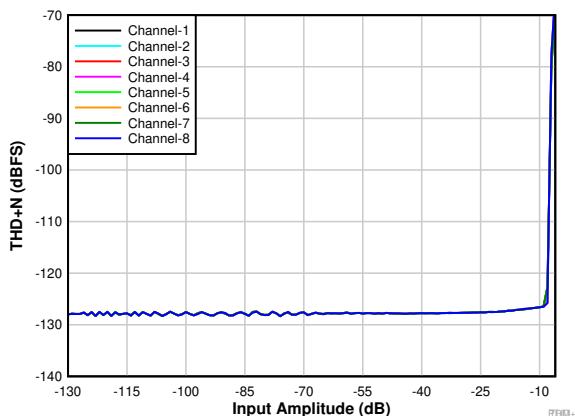


Figure 5-4. PDM Digital Microphone Interface Timing Diagram

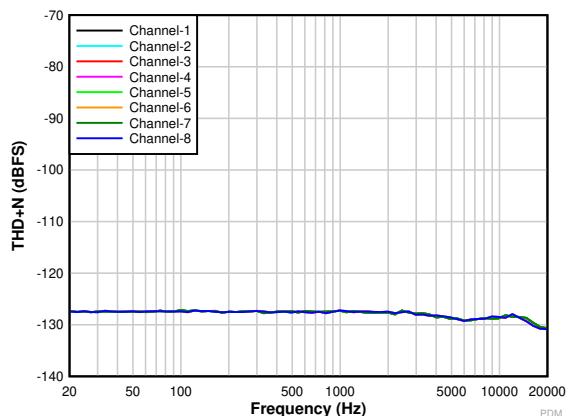
5.14 Typical Characteristics

at $T_A = 25^\circ\text{C}$, $\text{AVDD} = 3.3\text{V}$, $\text{IOVDD} = 3.3\text{V}$, $f_{IN} = 1\text{kHz}$ sinusoidal signal, $f_S = 48\text{kHz}$, $\text{PDMCLKx} = 64 \times f_S$, 32-bit audio data, $\text{BCLK} = 256 \times f_S$, TDM slave mode, PLL on, and linear phase decimation filter (unless otherwise noted); all performance measurements are done with a 20kHz, low-pass filter and an A-weighted filter (unless otherwise noted). All measurements are done by feeding the device PDM digital input signal using audio precision



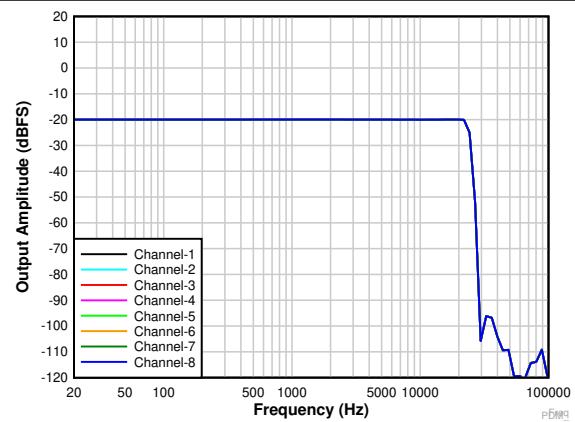
Fifth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-5. THD+N vs Input Amplitude



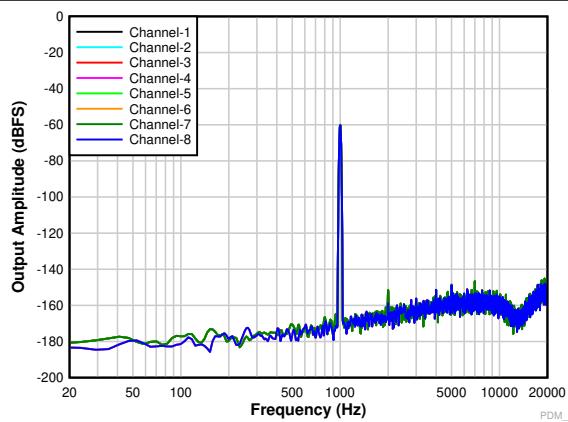
Fifth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-6. THD+N vs Input Frequency With a -20-dBr Input



Fifth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-7. Frequency Response With a -20-dBr Input

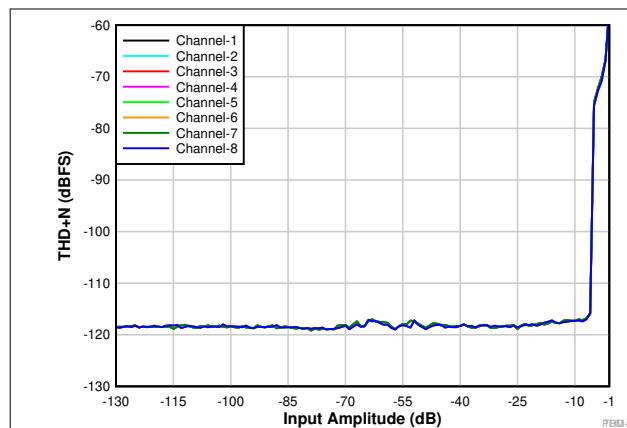


Fifth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-8. FFT With a -60-dBr Input

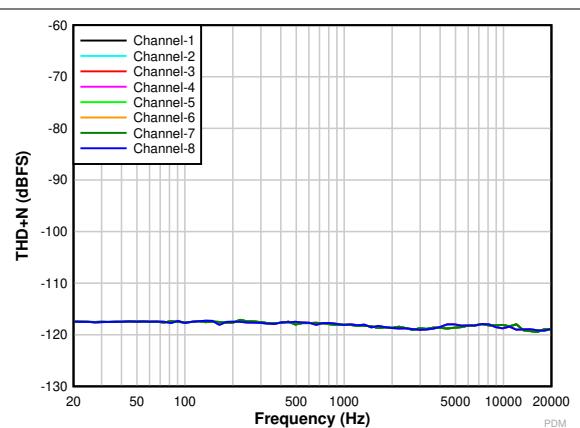
5.14 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $\text{AVDD} = 3.3\text{V}$, $\text{IOVDD} = 3.3\text{V}$, $f_{\text{IN}} = 1\text{kHz}$ sinusoidal signal, $f_S = 48\text{kHz}$, $\text{PDMCLKx} = 64 \times f_S$, 32-bit audio data, $\text{BCLK} = 256 \times f_S$, TDM slave mode, PLL on, and linear phase decimation filter (unless otherwise noted); all performance measurements are done with a 20kHz, low-pass filter and an A-weighted filter (unless otherwise noted). All measurements are done by feeding the device PDM digital input signal using audio precision



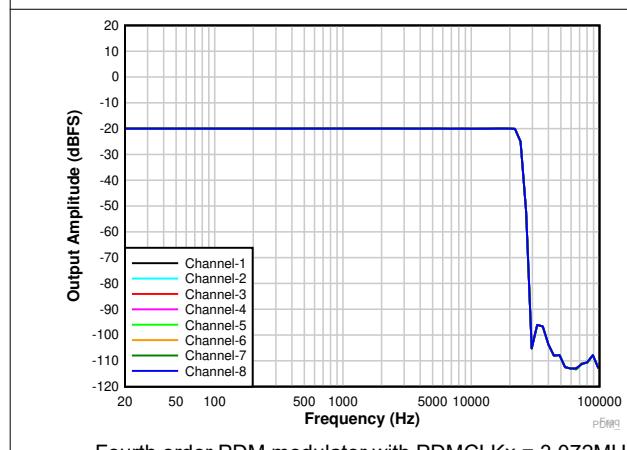
Fourth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-9. THD+N vs Input Amplitude



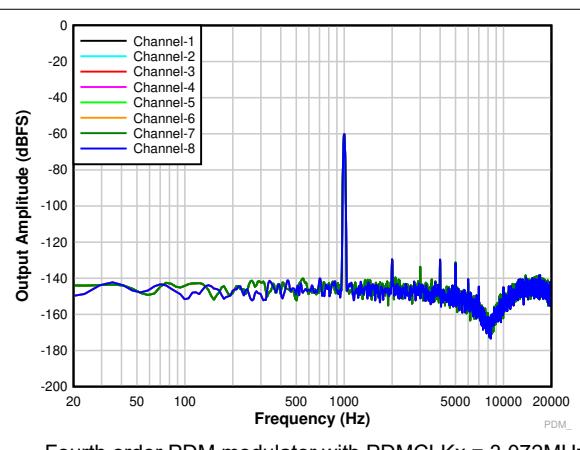
Fourth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-10. THD+N vs Input Frequency With a -20-dBr Input



Fourth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-11. Frequency Response With a -20-dBr Input

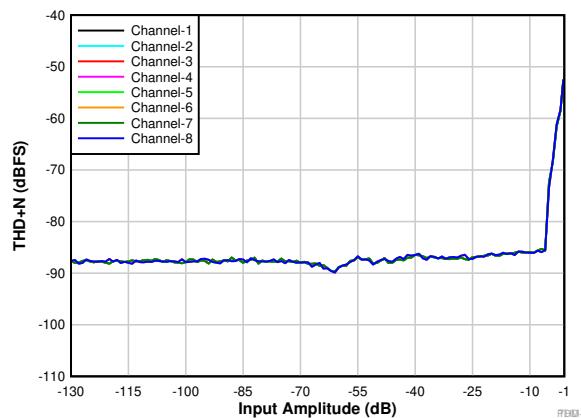


Fourth order PDM modulator with $\text{PDMCLKx} = 3.072\text{MHz}$

Figure 5-12. FFT With a -60-dBr Input

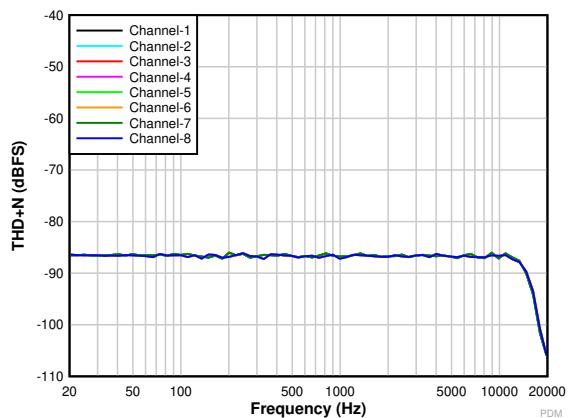
5.14 Typical Characteristics (continued)

at $T_A = 25^\circ\text{C}$, $\text{AVDD} = 3.3\text{V}$, $\text{IOVDD} = 3.3\text{V}$, $f_{IN} = 1\text{kHz}$ sinusoidal signal, $f_S = 48\text{kHz}$, $\text{PDMCLKx} = 64 \times f_S$, 32-bit audio data, $\text{BCLK} = 256 \times f_S$, TDM slave mode, PLL on, and linear phase decimation filter (unless otherwise noted); all performance measurements are done with a 20kHz, low-pass filter and an A-weighted filter (unless otherwise noted). All measurements are done by feeding the device PDM digital input signal using audio precision



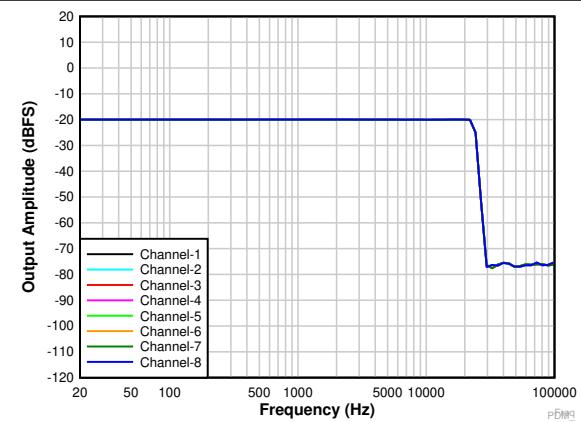
Fourth order PDM modulator with $\text{PDMCLKx} = 1.536\text{MHz}$

Figure 5-13. THD+N vs Input Amplitude



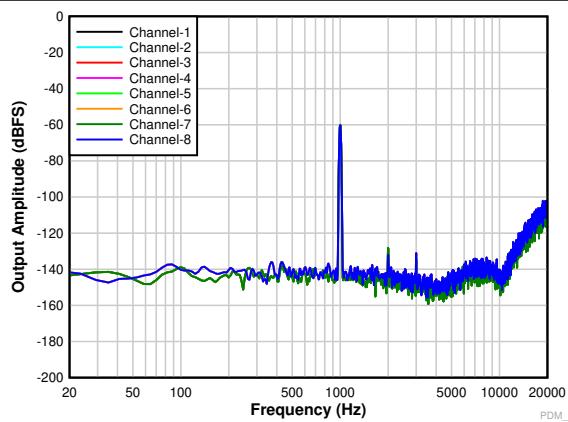
Fourth order PDM modulator with $\text{PDMCLKx} = 1.536\text{MHz}$

Figure 5-14. THD+N vs Input Frequency With a -20-dBr Input



Fourth order PDM modulator with $\text{PDMCLKx} = 1.536\text{MHz}$

Figure 5-15. Frequency Response With a -20-dBr Input



Fourth order PDM modulator with $\text{PDMCLKx} = 1.536\text{MHz}$

Figure 5-16. FFT With a -60-dBr Input

6 Detailed Description

6.1 Overview

The PCMD3180 is a high-performance, low-power, flexible, 8-channel, pulse-density-modulation (PDM) input to time-division multiplexing (TDM) or I²S audio output converter with extensive feature integration. This device is intended for applications in voice-activated systems, portable computing, communication, and entertainment applications. The low power consumption makes this device is designed for battery powered portable audio systems. This device integrates a host of features that reduces cost, board space, and power consumption in space-constrained, battery-powered, consumer, home, and industrial applications.

The PCMD3180 consists of the following blocks:

- Eight-channel, pulse density modulation (PDM) digital microphone interface with high-performance decimation filter
- Low-noise, microphone bias output to power the digital microphone
- Programmable decimation filters with linear-phase or low-latency filter
- Programmable digital volume control, biquad filters for each channel
- Programmable phase and gain calibration with fine resolution for each channel
- Programmable high-pass filter (HPF), and digital channel mixer
- Integrated low-jitter phase-locked loop (PLL) supporting a wide range of system clocks
- Integrated digital and analog voltage regulators to support single-supply operation

Communication to the PCMD3180 to configure the control registers is supported using an I²C or SPI. The device supports a highly flexible audio serial interface [time-division multiplexing (TDM), I²S, or left-justified (LJ)] to transmit audio data seamlessly in the system across devices.

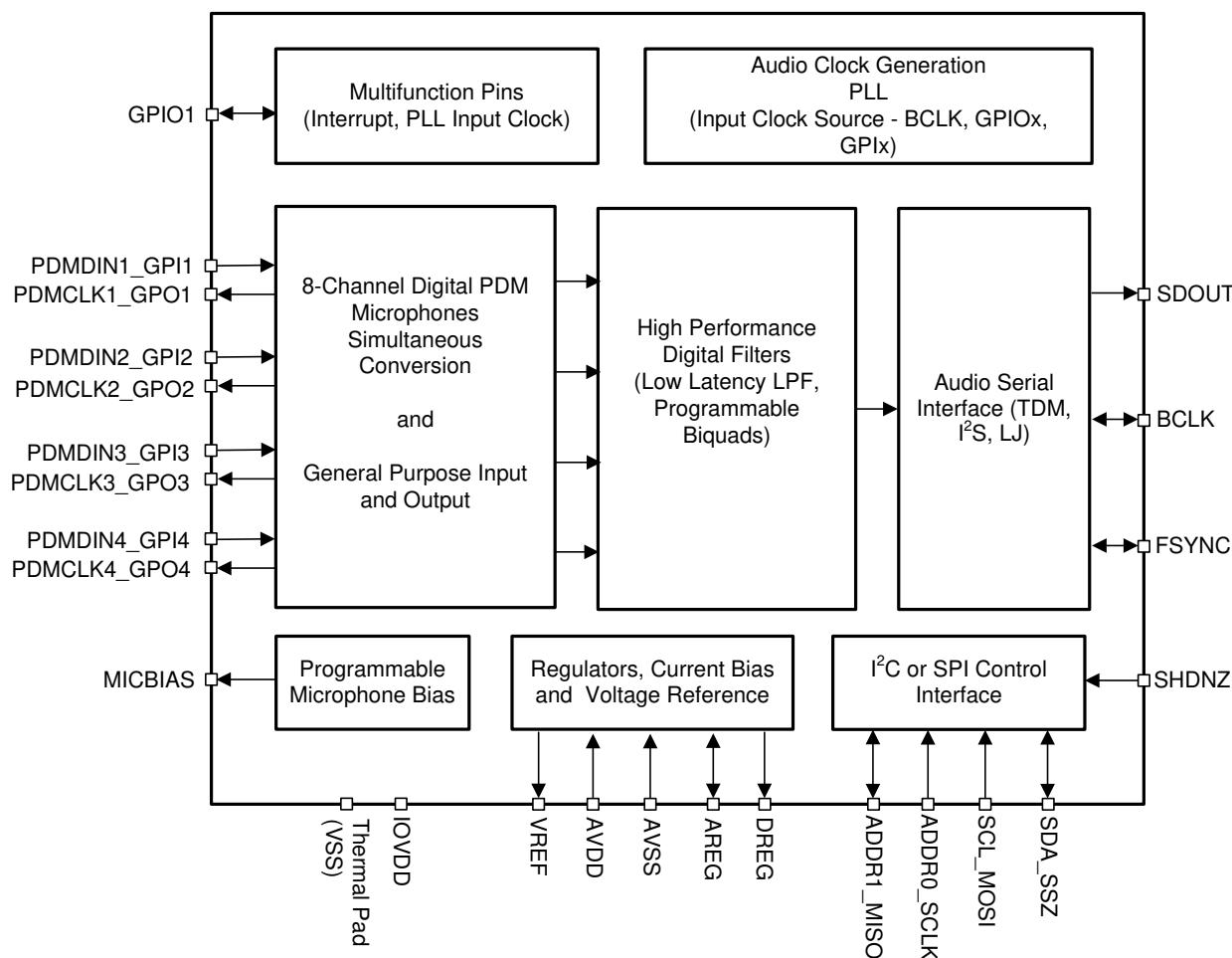
The device can support multiple devices by sharing the common I²C and TDM buses across devices. Moreover, the device includes a daisy-chain feature and a secondary audio serial output data pin. These features relax the shared TDM bus timing requirements and board design complexities when operating multiple devices for applications requiring high audio data bandwidth.

Table 6-1 lists the reference abbreviations used throughout this document to registers that control the device.

Table 6-1. Abbreviations for Register References

REFERENCE	ABBREVIATION	DESCRIPTION	EXAMPLE
Page y, register z, bit k	Py_Rz_Dk	Single data bit. The value of a single bit in a register.	Page 4, register 36, bit 0 = P4_R36_D0
Page y, register z, bits k:m	Py_Rz_D[k:m]	Range of data bits. A range of data bits (inclusive).	Page 4, register 36, bits 3:0 = P4_R36_D[3:0]
Page y, register z	Py_Rz	One entire register. All eight bits in the register as a unit.	Page 4, register 36 = P4_R36
Page y, registers z-n	Py_Rz-Rn	Range of registers. A range of registers in the same page.	Page 4, registers 36, 37, 38 = P4_R36R38

6.2 Functional Block Diagram



6.3 Feature Description

6.3.1 Serial Interfaces

This device has two serial interfaces: control and audio data. The control serial interface is used for device configuration. The audio data serial interface is used for transmitting audio data to the host device.

6.3.1.1 Control Serial Interfaces

The device contains configuration registers and programmable coefficients that can be set to the desired values for a specific system and application use. All these registers can be accessed using either I²C or SPI communication to the device. For more information, see the [Programming](#) section.

6.3.1.2 Audio Serial Interfaces

Digital audio data flows between the host processor and the PCMD3180 on the digital audio serial interface (ASI), or audio bus. This highly flexible ASI bus includes a TDM mode for multichannel operation, support for I²S or left-justified protocols format, programmable data length options, very flexible master-slave configurability for bus clock lines and the ability to communicate with multiple devices within a system directly.

The bus protocol TDM, I²S, or left-justified (LJ) format can be selected by using the ASI_FORMAT[1:0], P0_R7_D[7:6] register bits. As shown in [Table 6-2](#) and [Table 6-3](#), these modes are all most significant byte (MSB)-first, pulse code modulation (PCM) data format, with the output channel data word-length programmable as 16, 20, 24, or 32 bits by configuring the ASI_WLEN[1:0], P0_R7_D[5:4] register bits.

Table 6-2. Audio Serial Interface Format

P0_R7_D[7:6] : ASI_FORMAT[1:0]		AUDIO SERIAL INTERFACE FORMAT
00 (default)		Time division multiplexing (TDM) mode
01		Inter IC sound (I ² S) mode
10		Left-justified (LJ) mode
11		Reserved (do not use this setting)

Table 6-3. Audio Output Channel Data Word-Length

P0_R7_D[5:4] : ASI_WLEN[1:0]		AUDIO OUTPUT CHANNEL DATA WORD-LENGTH
00		Output channel data word-length set to 16 bits
01		Output channel data word-length set to 20 bits
10		Output channel data word-length set to 24 bits
11 (default)		Output channel data word-length set to 32 bits

The frame sync pin, FSYNC, is used in this audio bus protocol to define the beginning of a frame and has the same frequency as the output data sample rates. The bit clock pin, BCLK, is used to clock out the digital audio data across the serial bus. The number of bit-clock cycles in a frame must accommodate multiple device active output channels with the programmed data word length.

A frame consists of multiple time-division channel slots (up to 64) to allow all output channel audio data transmissions to complete on the audio bus by a device or multiple PCMD3180 devices sharing the same audio bus. The device supports up to eight output channels that can be configured to place the audio data on bus slot 0 to slot 63. [Table 6-4](#) lists the output channel slot configuration settings. In I²S and LJ mode, the slots are divided into two sets, left-channel slots and right-channel slots, as described in the [Inter IC Sound \(I²S\) Interface](#) and [Left-Justified \(LJ\) Interface](#) sections.

Table 6-4. Output Channel Slot Assignment Settings

P0_R11_D[5:0] : CH1_SLOT[5:0]		OUTPUT CHANNEL 1 SLOT ASSIGNMENT
00 0000 = 0d (default)		Slot 0 for TDM or left slot 0 for I ² S, LJ.
00 0001 = 1d		Slot 1 for TDM or left slot 1 for I ² S, LJ.
...		...
01 1111 = 31d		Slot 31 for TDM or left slot 31 for I ² S, LJ.
10 0000 = 32d		Slot 32 for TDM or right slot 0 for I ² S, LJ.
...		...
11 1110 = 62d		Slot 62 for TDM or right slot 30 for I ² S, LJ.
11 1111 = 63d		Slot 63 for TDM or right slot 31 for I ² S, LJ.

Similarly, the slot assignment setting for output channel 2 to channel 8 can be done using the CH2_SLOT (P0_R12) to CH8_SLOT (P0_R18) registers, respectively.

The slot word length is the same as the output channel data word length set for the device. The output channel data word length must be set to the same value for all PCMD3180 devices if all devices share the same ASI bus in a system. The maximum number of slots possible for the ASI bus in a system is limited by the available bus bandwidth, which depends upon the BCLK frequency, output data sample rate used, and the channel data word length configured.

The device also includes a feature that offsets the start of the slot data transfer with respect to the frame sync by up to 31 cycles of the bit clock. [Table 6-5](#) lists the programmable offset configuration settings.

Table 6-5. Programmable Offset Settings for the ASI Slot Start

P0_R8_D[4:0] : TX_OFFSET[4:0]	PROGRAMMABLE OFFSET SETTING FOR SLOT DATA TRANSMISSION START
0 0000 = 0d (default)	The device follows the standard protocol timing without any offset.
0 0001 = 1d	Slot start is offset by one BCLK cycle, as compared to standard protocol timing. For I ² S or LJ, the left and right slot start is offset by one BCLK cycle, as compared to standard protocol timing.
.....
1 1110 = 30d	Slot start is offset by 30 BCLK cycles, as compared to standard protocol timing. For I ² S or LJ, the left and right slot start is offset by 30 BCLK cycles, as compared to standard protocol timing.
1 1111 = 31d	Slot start is offset by 31 BCLK cycles, as compared to standard protocol timing. For I ² S or LJ, the left and right slot start is offset by 31 BCLK cycles, as compared to standard protocol timing.

The device also features the ability to invert the polarity of the frame sync pin, FSYNC, used to transfer the audio data as compared to the default FSYNC polarity used in standard protocol timing. This feature can be set using the FSYNC_POL, P0_R7_D3 register bit. Similarly, the device can invert the polarity of the bit clock pin, BCLK, which can be set using the BCLK_POL, P0_R7_D2 register bit.

6.3.1.2.1 Time Division Multiplexed Audio (TDM) Interface

In TDM mode, also known as DSP mode, the rising edge of FSYNC starts the data transfer with the slot 0 data first. Immediately after the slot 0 data transmission, the remaining slot data are transmitted in order. FSYNC and each data bit (except the MSB of slot 0 when TX_OFFSET equals 0) is transmitted on the rising edge of BCLK. [Figure 6-1](#) to [Figure 6-4](#) illustrate the protocol timing for TDM operation with various configurations.

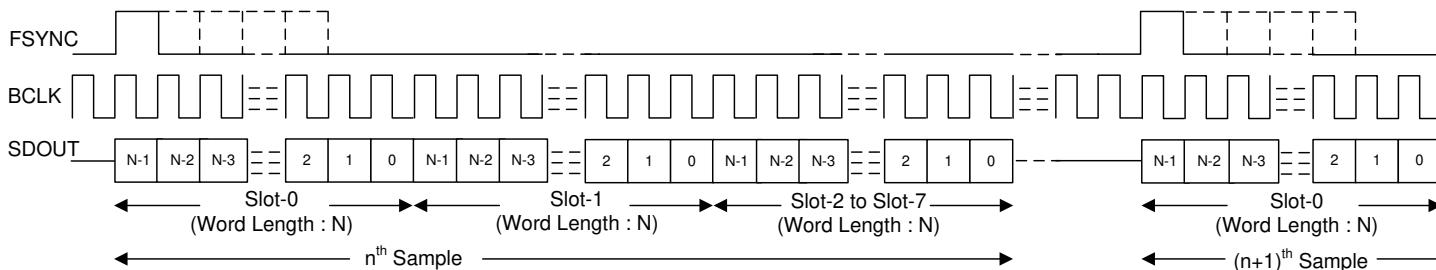


Figure 6-1. TDM Mode Standard Protocol Timing (TX_OFFSET = 0)

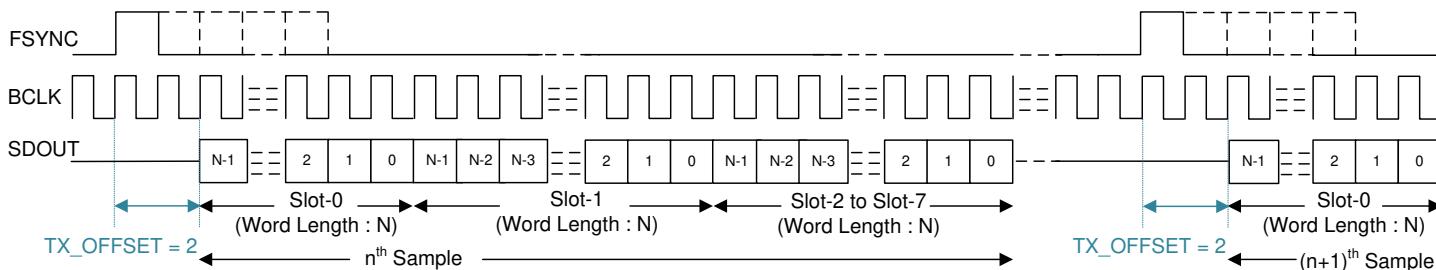


Figure 6-2. TDM Mode Protocol Timing (TX_OFFSET = 2)

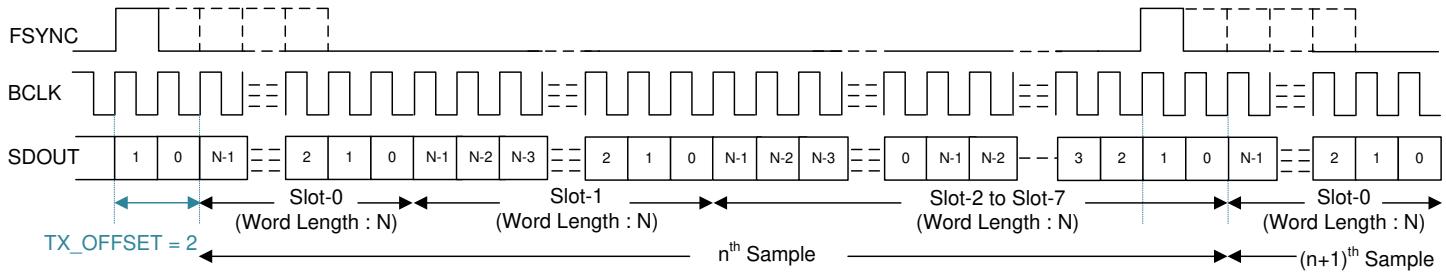


Figure 6-3. TDM Mode Protocol Timing (No Idle BCLK Cycles, TX_OFFSET = 2)

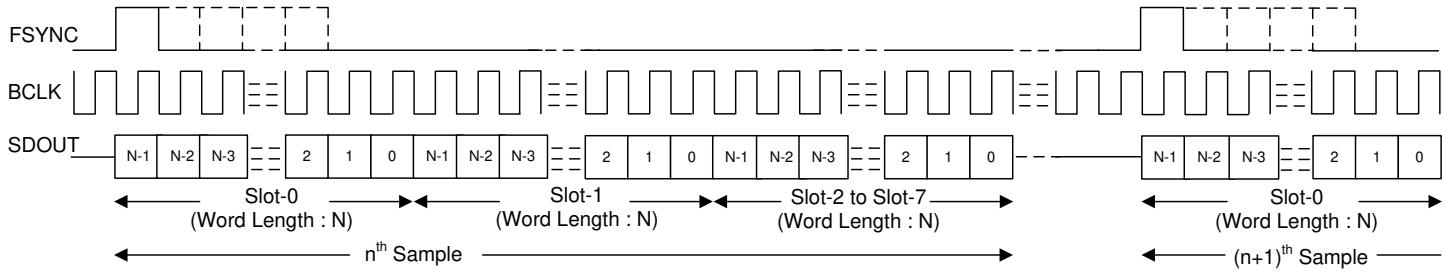


Figure 6-4. TDM Mode Protocol Timing (TX_OFFSET = 0 and BCLK_POL = 1)

For proper operation of the audio bus in TDM mode, the number of bit clocks per frame must be greater than or equal to the number of active output channels times the programmed word length of the output channel data. The device supports FSYNC as a pulse with a 1-cycle-wide bit clock, but also supports multiples as well. For a higher BCLK frequency operation, using TDM mode with a TX_OFFSET value higher than 0 is recommended.

6.3.1.2.2 Inter IC Sound (I²S) Interface

The standard I²S protocol is defined for only two channels: left and right. The device extends the same protocol timing for multichannel operation. In I²S mode, the MSB of the left slot 0 is transmitted on the falling edge of BCLK in the second cycle after the *falling* edge of FSYNC. Immediately after the left slot 0 data transmission, the remaining left slot data are transmitted in order. The MSB of the right slot 0 is transmitted on the falling edge of BCLK in the second cycle after the *rising* edge of FSYNC. Immediately after the right slot 0 data transmission, the remaining right slot data are transmitted in order. FSYNC and each data bit is transmitted on the falling edge of BCLK. Figure 6-5 to Figure 6-8 illustrate the protocol timing for I²S operation with various configurations.

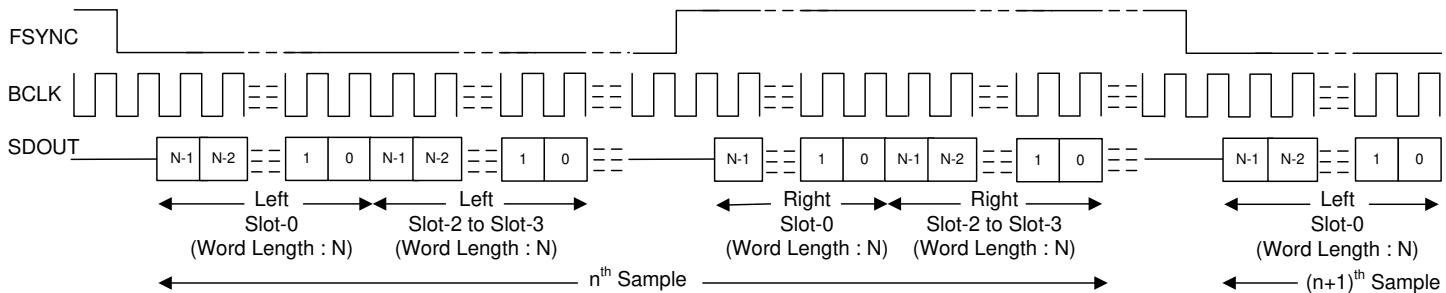


Figure 6-5. I²S Mode Standard Protocol Timing (TX_OFFSET = 0)

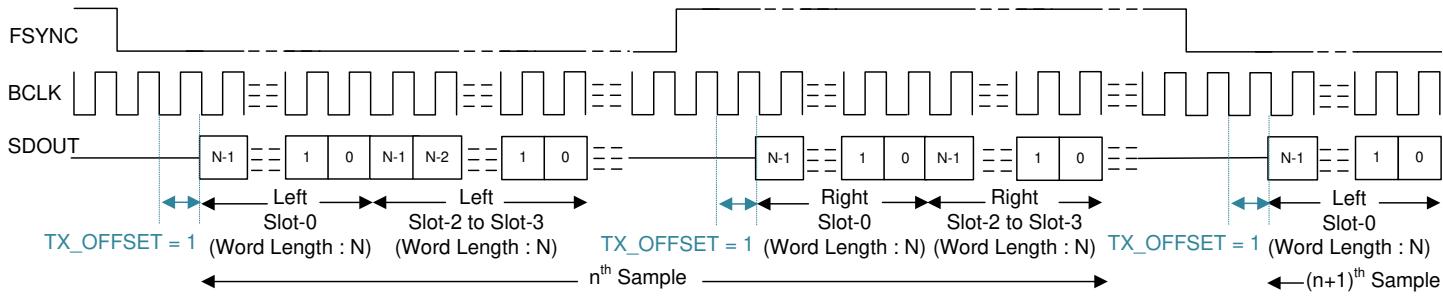


Figure 6-6. I²S Protocol Timing (TX_OFFSET = 1)

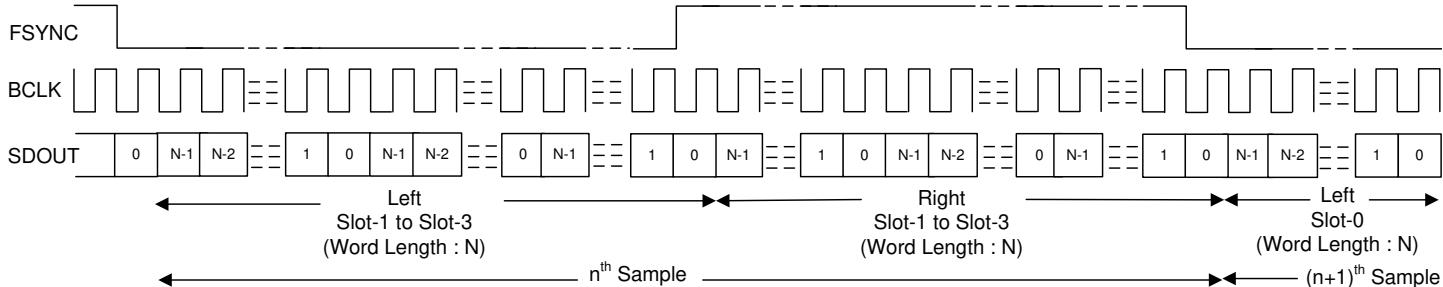


Figure 6-7. I²S Protocol Timing (No Idle BCLK Cycles, TX_OFFSET = 0)

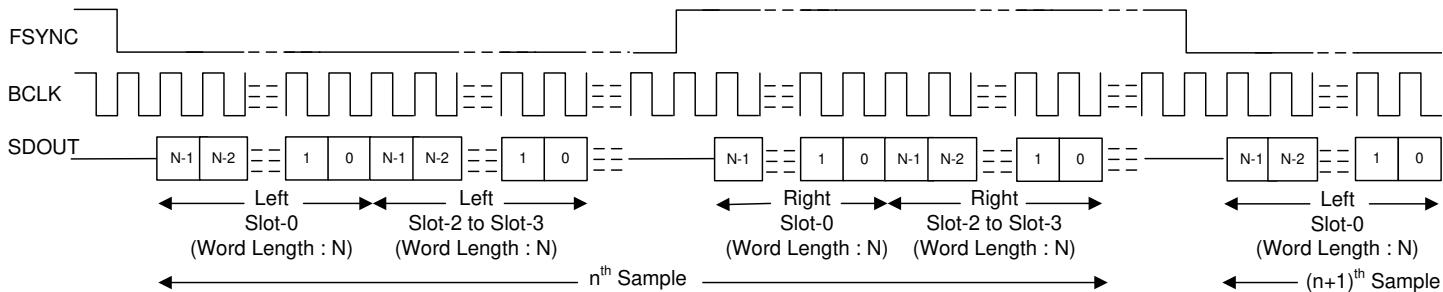


Figure 6-8. I²S Protocol Timing (TX_OFFSET = 0 and BCLK_POL = 1)

For proper operation of the audio bus in I²S mode, the number of bit clocks per frame must be greater than or equal to the number of active output channels (including left and right slots) times the programmed word length of the output channel data. The device FSYNC low pulse must be a number of BCLK cycles wide that is greater than or equal to the number of active left slots times the data word length configured. Similarly, the FSYNC high pulse must be a number of BCLK cycles wide that is greater than or equal to the number of active right slots times the data word length configured.

6.3.1.2.3 Left-Justified (LJ) Interface

The standard LJ protocol is defined for only two channels: left and right. The device extends the same protocol timing for multichannel operation. In LJ mode, the MSB of the left slot 0 is transmitted in the same BCLK cycle after the *rising* edge of FSYNC. Each subsequent data bit is transmitted on the falling edge of BCLK. Immediately after the left slot 0 data transmission, the remaining left slot data are transmitted in order. The MSB of the right slot 0 is transmitted in the same BCLK cycle after the *falling* edge of FSYNC. Each subsequent data bit is transmitted on the falling edge of BCLK. Immediately after the right slot 0 data transmission, the remaining right slot data are transmitted in order. FSYNC is transmitted on the falling edge of BCLK. [Figure 6-9 to Figure 6-12](#) illustrate the protocol timing for LJ operation with various configurations.

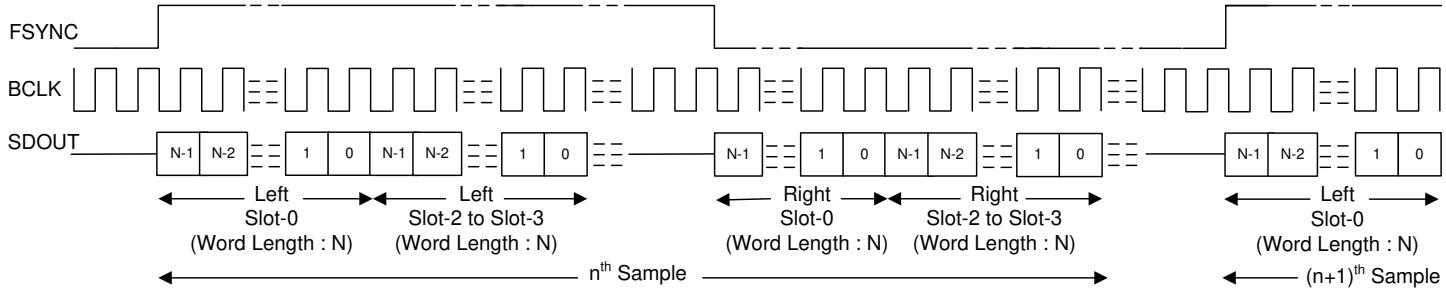


Figure 6-9. LJ Mode Standard Protocol Timing (TX_OFFSET = 0)

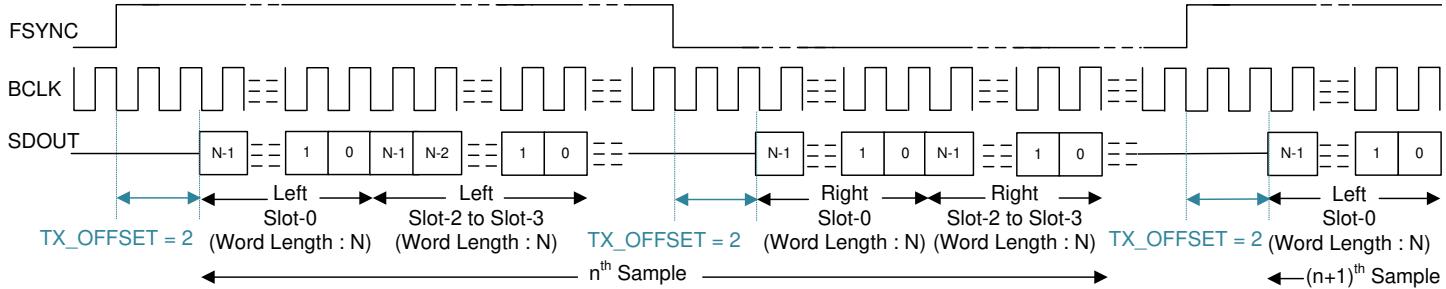


Figure 6-10. LJ Protocol Timing (TX_OFFSET = 2)

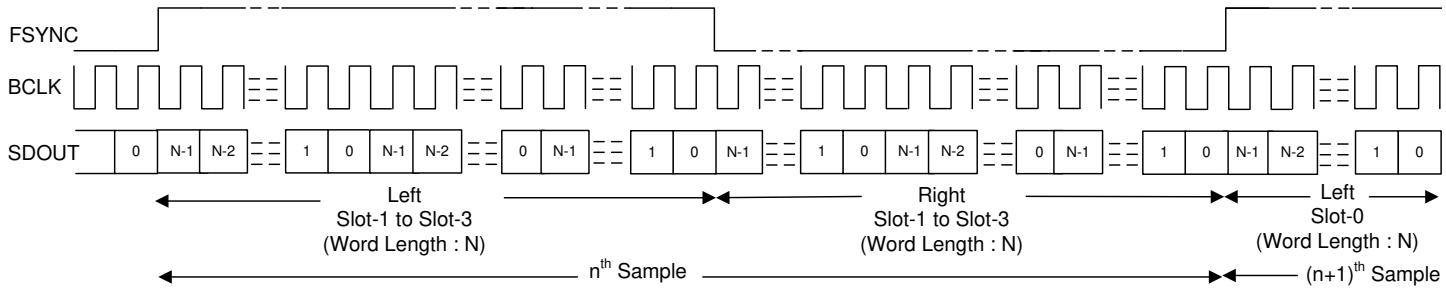


Figure 6-11. LJ Protocol Timing (No Idle BCLK Cycles, TX_OFFSET = 0)

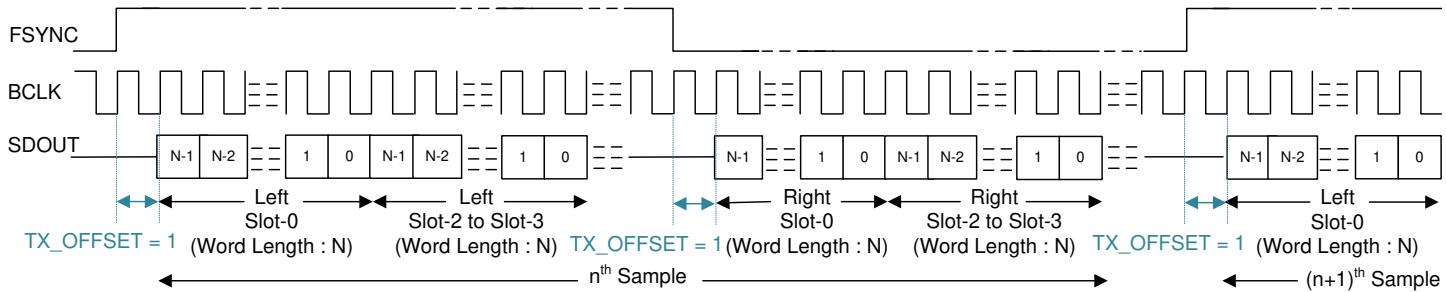


Figure 6-12. LJ Protocol Timing (TX_OFFSET = 1 and BCLK_POL = 1)

For proper operation of the audio bus in LJ mode, the number of bit clocks per frame must be greater than or equal to the number of active output channels (including left and right slots) times the programmed word length of the output channel data. The device FSYNC high pulse must be a number of BCLK cycles wide that is greater than or equal to the number of active left slots times the data word length configured. Similarly, the FSYNC low pulse must be number of BCLK cycles wide that is greater than or equal to the number of active right slots times the data word length configured. For a higher BCLK frequency operation, using LJ mode with a TX_OFFSET value higher than 0 is recommended.

6.3.1.3 Using Multiple Devices With Shared Buses

The device has many supported features and flexible options that can be used in the system to seamlessly connect multiple PCMD3180 devices by sharing a single common I²C control bus and an audio serial interface bus. This architecture enables multiple applications to be applied to a system that require a microphone array for beam-forming operation, audio conferencing, noise cancellation, and so forth. [Figure 6-13](#) shows a diagram of multiple PCMD3180 devices in a configuration where the control and audio data buses are shared.

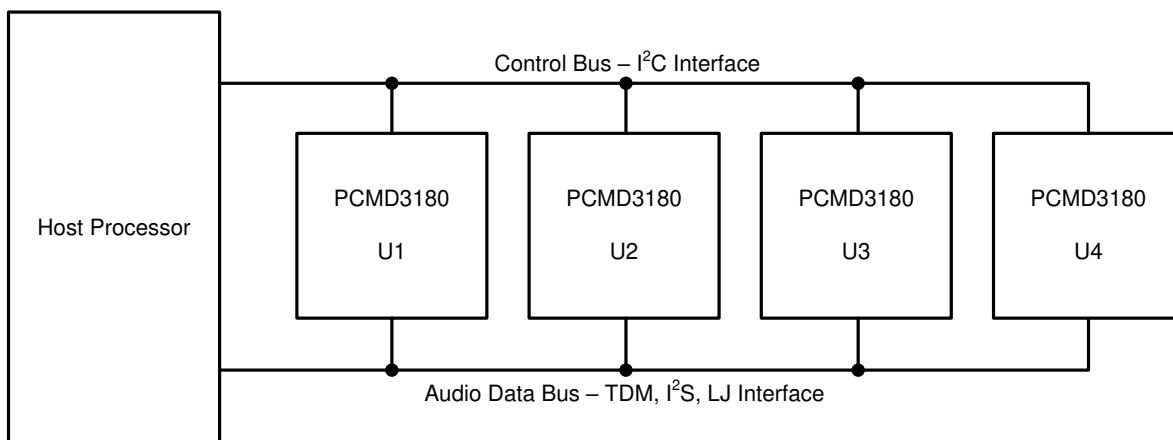


Figure 6-13. Multiple PCMD3180 Devices With Shared Control and Audio Data Buses

The PCMD3180 consists of the following features to enable seamless connection and interaction of multiple devices using a shared bus:

- Supports up to four pin-programmable I²C slave addresses
- I²C broadcast simultaneously writes to (or triggers) all PCMD3180 devices
- Supports up to 64 configuration output channel slots for the audio serial interface
- Tri-state feature (with enable and disable) for the unused audio data slots of the device
- Supports a bus-holder feature (with enable and disable) to keep the last driven value on the audio bus
- The GPIO1 or GPO_x pin can be configured as a secondary output data lane for the audio serial interface
- The GPIO1 or GPI_x pin can be used in a daisy-chain configuration of multiple PCMD3180 devices
- Supports one BCLK cycle data latching timing to relax the timing requirement for the high-speed interface
- Programmable master and slave options for the audio serial interface
- Ability to synchronize the multiple devices for the simultaneous sampling requirement across devices

The system can also connect multiple PCMD3180 devices in combination with TLV320ADCx140 devices by sharing a single common I²C control bus and an audio serial interface bus. See the [Multiple TLV320ADCx140 Devices With Shared TDM and I²C Bus application note](#) for further details.

6.3.2 Phase-Locked Loop (PLL) and Clock Generation

The device has a smart auto-configuration block to generate all necessary internal clocks required for the PDM clock generation and the digital filter engine used for signal processing. This configuration is done by monitoring the frequency of the FSYNC and BCLK signal on the audio bus.

The device supports the various output data sample rates (of the FSYNC signal frequency) and the BCLK to FSYNC ratio to configure all clock dividers, including the PLL configuration, internally without host programming. **Table 6-6** and **Table 6-7** list the supported FSYNC and BCLK frequencies.

Table 6-6. Supported FSYNC (Multiples or Submultiples of 48kHz) and BCLK Frequencies

BCLK TO FSYNC RATIO	BCLK (MHz)								
	FSYNC (8kHz)	FSYNC (16kHz)	FSYNC (24kHz)	FSYNC (32kHz)	FSYNC (48kHz)	FSYNC (96kHz)	FSYNC (192kHz)	FSYNC (384kHz)	FSYNC (768kHz)
16	Reserved	0.256	0.384	0.512	0.768	1.536	3.072	6.144	12.288
24	Reserved	0.384	0.576	0.768	1.152	2.304	4.608	9.216	18.432
32	0.256	0.512	0.768	1.024	1.536	3.072	6.144	12.288	24.576
48	0.384	0.768	1.152	1.536	2.304	4.608	9.216	18.432	Reserved
64	0.512	1.024	1.536	2.048	3.072	6.144	12.288	24.576	Reserved
96	0.768	1.536	2.304	3.072	4.608	9.216	18.432	Reserved	Reserved
128	1.024	2.048	3.072	4.096	6.144	12.288	24.576	Reserved	Reserved
192	1.536	3.072	4.608	6.144	9.216	18.432	Reserved	Reserved	Reserved
256	2.048	4.096	6.144	8.192	12.288	24.576	Reserved	Reserved	Reserved
384	3.072	6.144	9.216	12.288	18.432	Reserved	Reserved	Reserved	Reserved
512	4.096	8.192	12.288	16.384	24.576	Reserved	Reserved	Reserved	Reserved
1024	8.192	16.384	24.576	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
2048	16.384	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved

Table 6-7. Supported FSYNC (Multiples or Submultiples of 44.1kHz) and BCLK Frequencies

BCLK TO FSYNC RATIO	BCLK (MHz)								
	FSYNC (7.35kHz)	FSYNC (14.7kHz)	FSYNC (22.05kHz)	FSYNC (29.4kHz)	FSYNC (44.1kHz)	FSYNC (88.2kHz)	FSYNC (176.4kHz)	FSYNC (352.8kHz)	FSYNC (705.6kHz)
16	Reserved	Reserved	0.3528	0.4704	0.7056	1.4112	2.8224	5.6448	11.2896
24	Reserved	0.3528	0.5292	0.7056	1.0584	2.1168	4.2336	8.4672	16.9344
32	Reserved	0.4704	0.7056	0.9408	1.4112	2.8224	5.6448	11.2896	22.5792
48	0.3528	0.7056	1.0584	1.4112	2.1168	4.2336	8.4672	16.9344	Reserved
64	0.4704	0.9408	1.4112	1.8816	2.8224	5.6448	11.2896	22.5792	Reserved
96	0.7056	1.4112	2.1168	2.8224	4.2336	8.4672	16.9344	Reserved	Reserved
128	0.9408	1.8816	2.8224	3.7632	5.6448	11.2896	22.5792	Reserved	Reserved
192	1.4112	2.8224	4.2336	5.6448	8.4672	16.9344	Reserved	Reserved	Reserved
256	1.8816	3.7632	5.6448	7.5264	11.2896	22.5792	Reserved	Reserved	Reserved
384	2.8224	5.6448	8.4672	11.2896	16.9344	Reserved	Reserved	Reserved	Reserved
512	3.7632	7.5264	11.2896	15.0528	22.5792	Reserved	Reserved	Reserved	Reserved
1024	7.5264	15.0528	22.5792	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
2048	15.0528	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved

The status register ASI_STS, P0_R21, captures the device auto detect result for the FSYNC frequency and the BCLK to FSYNC ratio. If the device finds any unsupported combinations of FSYNC frequency and BCLK to FSYNC ratios, the device generates an ASI clock-error interrupt and mutes the record channels accordingly.

The device uses an integrated, low-jitter, phase-locked loop (PLL) to generate internal clocks required for the PDM clock generation and digital filter engine, as well as other control blocks. The device also supports an option to use BCLK, GPIO1, or the GPIx pin (as MCLK) as the audio clock source without using the PLL to reduce power consumption. However, the PDM microphone performance can degrade based on jitter from the external clock source, and some processing features are not always supported if the external audio clock source frequency is not high enough. Therefore, TI recommends using the PLL for high-performance applications. More details and information on how to configure and use the device in low-power mode without using the PLL are discussed in the [TLV320ADCx140 Power Consumption Matrix Across Various Usage Scenario application note](#).

The device also supports an audio bus master mode operation using the GPIO1 or GPIx pin (as MCLK) as the reference input clock source and supports various flexible options and a wide variety of system clocks. More details and information on master mode configuration and operation are discussed in the [Configuring and Operating the TLV320ADCx140 as Audio Bus Master application note](#).

The audio bus clock error detection and auto-detect feature automatically generates all internal clocks, but can be disabled using the ASI_ERR, P0_R9_D5 and AUTO_CLK_CFG, P0_R19_D6, register bits, respectively. In the system, this disable feature can be used to support custom clock frequencies that are not covered by the auto detect scheme. For such application use cases, care must be taken to verify that the multiple clock dividers are all configured appropriately. Therefore, TI recommends using the PPC3 GUI for device configuration settings; for more details see the [ADCx140EVM-PDK user's guide](#) and the [PurePath™ Console Graphical Development Suite for Audio System Design and Development](#).

6.3.3 Reference Voltage

The PCMD3180 achieves low-noise performance by internally generating a low-noise reference voltage. This reference voltage is generated using a band-gap circuit with high PSRR performance and must be filtered externally using a 1 μ F capacitor connected from the VREF pin to analog ground (AVSS).

The value of this reference voltage can be configured using the P0_R59_D[1:0] register bits and must be set to an appropriate value based on the AVDD supply voltage available in the system. The default VREF value is set to 2.75V, which require minimum AVDD voltage for this mode is 3V. [Table 6-8](#) lists the various VREF settings supported along with required AVDD range for that configuration.

Table 6-8. VREF Programmable Settings

P0_R59_D[1:0] : VREF_SEL[1:0]	VREF OUTPUT VOLTAGE	AVDD RANGE REQUIREMENT
00 (default)	2.75V	3V to 3.6V
01	2.5V	2.8V to 3.6V
10	1.375V	1.7V to 1.9V
11	Reserved	Reserved

To achieve low-power consumption, this audio reference block is powered down as described in the [Sleep Mode or Software Shutdown](#) section. When exiting sleep mode, the audio reference block is powered up using the internal fast-charge scheme and the VREF pin settles to the steady-state voltage after the settling time (a function of the decoupling capacitor on the VREF pin). This time is approximately equal to 3.5ms when using a 1 μ F decoupling capacitor. If a higher-value decoupling capacitor is used on the VREF pin, the fast-charge setting must be reconfigured using the VREF_QCHG, P0_R2_D[4:3] register bits, which support options of 3.5ms (default), 10ms, 50ms, or 100ms.

6.3.4 Microphone Bias

The device integrates a built-in, low-noise programmable microphone bias pin that can be used in the system for providing the supply to the MEMS digital microphone. The MICBIAS pin must be connected to external 1 μ F to analog ground (AVSS). The MICBIAS pin supports up to 20mA of load current that can be used for multiple microphones. When using this MICBIAS pin for biasing or supplying to multiple microphones, avoid any common impedance on the board layout for the MICBIAS connection to minimize coupling across microphones. In system, if MICBIAS pin is used as supply for digital microphones then TI recommends to use MICBIAS configuration as AVDD, so that the digital microphone's PDMCLK x and PDMDIN x signals can be directly interface to PCMD3180 without using any external level shifters. [Table 6-9](#) shows the available microphone bias programmable options.

Table 6-9. MICBIAS Programmable Settings

P0_R59_D[6:4] : MBIAS_VAL[2:0]	P0_R59_D[1:0] : VREF_SEL[1:0]	MICBIAS OUTPUT VOLTAGE
000 (default)	00 (default)	2.75V (same as the VREF output)
	01	2.5V (same as the VREF output)
	10	1.375V (same as the VREF output)
001 to 101	XX	Reserved (do not use these settings)
110	XX	Same as AVDD
111	XX	Reserved (do not use this setting)

The microphone bias output can be powered on or powered off (default) by configuring the MICBIAS_PDZ, P0_R117_D7 register bit. Additionally, the device provides an option to configure the GPIO1 or GPIx pin to directly control the microphone bias output powering on or off. This feature is useful to control the microphone directly without engaging the host for I²C or SPI communication. The MICBIAS_PDZ, P0_R117_D7 register bit value is ignored if the GPIO1 or GPIx pin is configured to set the microphone bias on or off.

6.3.5 Digital PDM Microphone Record Channel

The device interfaces up to eight digital pulse-density-modulation (PDM) microphones for simultaneous conversion and uses high-order and high-performance decimation filters to generate pulse code modulation (PCM) output data that can be transmitted on the audio serial interface to the host using either time-division multiplexing (TDM), I²S, or left-justified (LJ) audio formats.

The device internally generates PCMCLK with a programmable frequency of either 6.144MHz, 3.072MHz, 1.536MHz, or 768kHz (for output data sample rates in multiples or submultiples of 48kHz) or 5.6448MHz, 2.8224MHz, 1.4112MHz, or 705.6kHz (for output data sample rates in multiples or submultiples of 44.1kHz) using the PDMCLK_DIV[1:0], P0_R31_D[1:0] register bits. PDMCLK can be routed on the PDMCLK_x_GPO_x pin. This clock can be connected to the external digital microphone device. The device also support control register to independently configure each channel PDMDIN_x data to be latched using either rising edge or falling edge. Figure 6-14 shows a connection diagram of the digital PDM microphones.

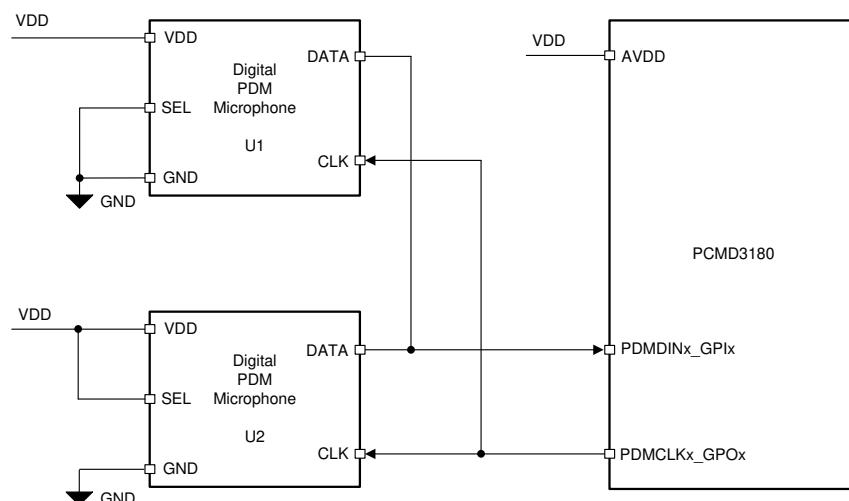


Figure 6-14. Digital PDM Microphones Connection Diagram to the PCMD3180

The single-bit output of the external digital microphone device can be connected to the GPIO_x pin. This single data line can be shared by two digital microphones to place the data on the opposite edge of PDMCLK. Internally, the device latches the steady value of the data on the rising edge of PDMCLK or the falling edge of PDMCLK based on the configuration register bits set in P0_R32_D[7:4]. Figure 6-15 shows the digital PDM microphone interface timing diagram.

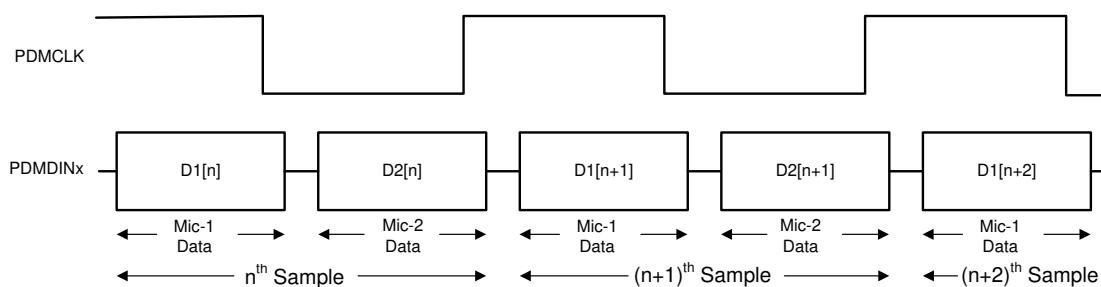


Figure 6-15. Digital PDM Microphone Protocol Timing Diagram

6.3.6 Signal-Chain Processing

The PCMD3180 signal chain is comprised of high-performance, low-power and highly flexible and programmable digital processing blocks. The high performance and flexibility combined with a compact package makes the PCMD3180 optimized for a wide variety of end-equipments and applications that require multichannel audio capture. [Figure 6-16](#) shows a conceptual block diagram that highlights the various building blocks used in the signal chain, and how the blocks interact in the signal chain.

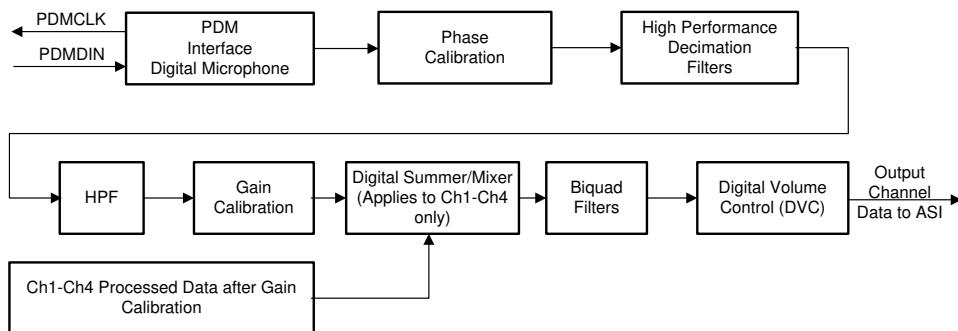


Figure 6-16. Signal-Chain Processing Flowchart

The device supports up to eight digital PDM microphone recording channels for simultaneous operation. The signal chain consists of various highly programmable digital processing blocks such as phase calibration, gain calibration, high-pass filter, digital summer or mixer, biquad filters, and volume control. The details on these processing blocks are discussed further in this section. Channels 1 to 4 in the signal chain block diagram of [Figure 6-16](#) are as described in this section, however, channels 5 to 8 do not support the digital summer or mixer option.

The desired input channels for recording can be enabled or disabled by using the IN_CH_EN (P0_R115) register, and the output channels for the audio serial interface can be enabled or disabled by using the ASI_OUT_EN (P0_R116) register. In general, the device supports simultaneous power-up and power-down of all active channels for simultaneous recording. However, based on the application needs, if some channels must be powered-up or powered-down dynamically when the other channel recording is on, then that use case is supported by setting the DYN_CH_PUPD_EN, P0_R117_D4 register bit to 1'b1 but do not power-down channel 1 in this mode of operation.

The device supports an input signal bandwidth up to 80kHz, which allows the high-frequency non-audio signal to be recorded by using a 176.4kHz (or higher) sample rate.

For output sample rates of 48kHz or lower, the device supports all features for 8-channel recording and various programmable processing blocks. However, for output sample rates higher than 48kHz, there are limitations in the number of simultaneous channel recordings supported and the number of biquad filters and such. See the [TLV320ADCx140 Sampling Rates and Programmable Processing Blocks Supported](#) application note for further details.

6.3.6.1 Programmable Digital Volume Control

The device has a programmable digital volume control with a range from -100dB to 27dB in steps of 0.5 dB with the option to mute the channel recording. The digital volume control value can be changed dynamically while the channel is powered-up and recording. During volume control changes, the soft ramp-up or ramp-down volume feature is used internally to avoid any audible artifacts. Soft-stepping can be entirely disabled using the DISABLE_SOFT_STEP (P0_R108_D4) register bit.

The digital volume control setting is independently available for each output channel, including the digital microphone record channel. However, the device also supports an option to gang-up the volume control setting for all channels together using the channel 1 digital volume control setting, regardless if channel 1 is powered up or powered down. This gang-up can be enabled using the DVOL_GANG (P0_R108_D7) register bit.

[Table 6-10](#) shows the programmable options available for the digital volume control.

Table 6-10. Digital Volume Control (DVC) Programmable Settings

P0_R62_D[7:0] : CH1_DVOL[7:0]	DVC SETTING FOR OUTPUT CHANNEL 1
0000 0000 = 0d	Output channel 1 DVC is set to mute
0000 0001 = 1d	Output channel 1 DVC is set to -100dB
0000 0010 = 2d	Output channel 1 DVC is set to -99.5dB
0000 0011 = 3d	Output channel 1 DVC is set to -99dB
...	...
1100 1000 = 200d	Output channel 1 DVC is set to -0.5dB
1100 1001 = 201d (default)	Output channel 1 DVC is set to 0dB
1100 1010 = 202d	Output channel 1 DVC is set to 0.5dB
...	...
1111 1101 = 253d	Output channel 1 DVC is set to 26dB
1111 1110 = 254d	Output channel 1 DVC is set to 26.5dB
1111 1111 = 255d	Output channel 1 DVC is set to 27dB

Similarly, the digital volume control setting for output channel 2 to channel 8 can be configured using the CH2_DVOL (P0_R67) to CH8_DVOL (P0_R97) register bits, respectively.

The internal digital processing engine soft ramps up the volume from a muted level to the programmed volume level when the channel is powered up, and the internal digital processing engine soft ramps down the volume from a programmed volume to mute when the channel is powered down. This soft-stepping of volume is done to prevent abruptly powering up and powering down the record channel. This feature can also be entirely disabled using the DISABLE_SOFT_STEP (P0_R108_D4) register bit.

6.3.6.2 Programmable Channel Gain Calibration

Along with the programmable channel gain and digital volume, this device also provides programmable channel gain calibration. The gain of each channel can be finely calibrated or adjusted in steps of 0.1dB for a range of -0.8dB to 0.7dB gain error. This adjustment is useful when trying to match the gain across channels resulting from external components and microphone sensitivity. This feature, in combination with the regular digital volume control, allows the gains across all channels to be matched for a wide gain error range with a resolution of 0.1 dB. [Table 6-11](#) shows the programmable options available for the channel gain calibration.

Table 6-11. Channel Gain Calibration Programmable Settings

P0_R63_D[7:4] : CH1_GCAL[3:0]	CHANNEL GAIN CALIBRATION SETTING FOR INPUT CHANNEL 1
0000 = 0d	Input channel 1 gain calibration is set to -0.8dB
0001 = 1d	Input channel 1 gain calibration is set to -0.7dB
...	...
1000 = 8d (default)	Input channel 1 gain calibration is set to 0dB
...	...
1110 = 14d	Input channel 1 gain calibration is set to 0.6dB
1111 = 15d	Input channel 1 gain calibration is set to 0.7dB

Similarly, the channel gain calibration setting for input channel 2 to channel 8 can be configured using the CH2_GCAL (P0_R68) to CH8_GCAL (P0_R98) register bits, respectively.

6.3.6.3 Programmable Channel Phase Calibration

In addition to the gain calibration, the phase delay in each channel can be finely calibrated or adjusted in steps of one modulator clock cycle for a cycle range of 0 to 255 for the phase error. The modulator clock is 6.144MHz (the output data sample rate is multiples or submultiples of 48kHz) or 5.6448MHz (the output data sample rate is multiples or submultiples of 44.1kHz) irrespective of the PDMCLK frequency used for digital microphone. This feature is very useful for many applications that must match the phase with fine resolution between each channel, including any phase mismatch across channels resulting from external components or microphones. [Table 6-12](#) shows the available programmable options for channel phase calibration.

Table 6-12. Channel Phase Calibration Programmable Settings

P0_R64_D[7:0] : CH1_PCAL[7:0]	CHANNEL PHASE CALIBRATION SETTING FOR INPUT CHANNEL 1
0000 0000 = 0d (default)	Input channel 1 phase calibration with no delay
0000 0001 = 1d	Input channel 1 phase calibration delay is set to one cycle of the modulator clock
0000 0010 = 2d	Input channel 1 phase calibration delay is set to two cycles of the modulator clock
...	...
1111 1110 = 254d	Input channel 1 phase calibration delay is set to 254 cycles of the modulator clock
1111 1111 = 255d	Input channel 1 phase calibration delay is set to 255 cycles of the modulator clock

Similarly, the channel phase calibration setting for input channel 2 to channel 8 can be configured using the CH2_PCAL (P0_R69) to CH8_PCAL (P0_R99) register bits, respectively.

6.3.6.4 Programmable Digital High-Pass Filter

To remove the DC offset component and attenuate the undesired low-frequency noise content in the record data, the device supports a programmable high-pass filter (HPF). The HPF is not a channel-independent filter setting but is globally applicable for all the channels. This HPF is constructed using the first-order infinite impulse response (IIR) filter, and is efficient enough to filter out possible DC components of the signal. [Table 6-13](#) shows the predefined -3dB cutoff frequencies available that can be set by using the `HPF_SEL[1:0]` register bits of `P0_R107`. Additionally, to achieve a custom -3dB cutoff frequency for a specific application, the device also allows the first-order IIR filter coefficients to be programmed when the `HPF_SEL[1:0]` register bits are set to 2'b00. [Figure 6-17](#) illustrates a frequency response plot for the HPF filter.

Table 6-13. HPF Programmable Settings

<code>P0_R107_D[1:0] : HPF_SEL[1:0]</code>	-3dB CUTOFF FREQUENCY SETTING	-3dB CUTOFF FREQUENCY AT 16kHz SAMPLE RATE	-3dB CUTOFF FREQUENCY AT 48kHz SAMPLE RATE
00	Programmable 1st-order IIR filter	Programmable 1st-order IIR filter	Programmable 1st-order IIR filter
01 (default)	$0.00025 \times f_s$	4Hz	12Hz
10	$0.002 \times f_s$	32Hz	96Hz
11	$0.008 \times f_s$	128Hz	384Hz

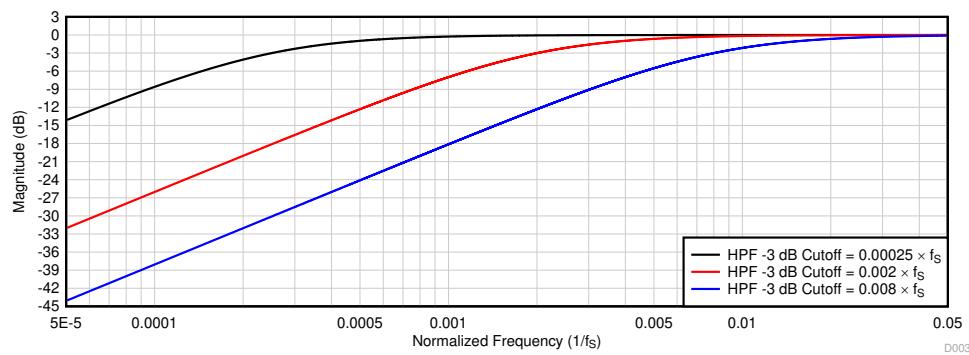


Figure 6-17. HPF Filter Frequency Response Plot

[Equation 1](#) gives the transfer function for the first-order programmable IIR filter:

$$H(z) = \frac{N_0 + N_1 z^{-1}}{2^{31} - D_1 z^{-1}} \quad (1)$$

The frequency response for this first-order programmable IIR filter with default coefficients is flat at a gain of 0dB (all-pass filter). The host device can override the frequency response by programming the IIR coefficients in [Table 6-14](#) to achieve the desired frequency response for high-pass filtering or any other desired filtering. If `HPF_SEL[1:0]` are set to 2'b00, the host device must write these coefficients values for the desired frequency response before powering-up any PDM channel for recording. These programmable coefficients are 32-bit, two's complement numbers. [Table 6-14](#) shows the filter coefficients for the first-order IIR filter.

Table 6-14. 1st-Order IIR Filter Coefficients

FILTER	FILTER COEFFICIENT	DEFAULT COEFFICIENT VALUE	COEFFICIENT REGISTER MAPPING
Programmable 1st-order IIR filter (can be allocated to HPF or any other desired filter)	N_0	0x7FFFFFFF	P4_R72R75
	N_1	0x00000000	P4_R76R79
	D_1	0x00000000	P4_R80R83

6.3.6.5 Programmable Digital Biquad Filters

The device supports up to 12 programmable digital biquad filters. These highly efficient filters achieve the desired frequency response. In digital signal processing, a digital biquad filter is a second-order, recursive linear filter with two poles and two zeros. [Equation 2](#) gives the transfer function of each biquad filter:

$$H(z) = \frac{N_0 + 2N_1 z^{-1} + N_2 z^{-2}}{2^{31} - 2D_1 z^{-1} - D_2 z^{-2}} \quad (2)$$

The frequency response for the biquad filter section with default coefficients is flat at a gain of 0dB (all-pass filter). The host device can override the frequency response by programming the biquad coefficients to achieve the desired frequency response for a low-pass, high-pass, or any other desired frequency shaping. The programmable coefficients for the mixer operation are located in the [Programmable Coefficient Registers: Page = 0x02](#) and [Programmable Coefficient Registers: Page = 0x03](#) sections. If biquad filtering is required, then the host device must write these coefficients values before powering up any PDM channels for recording. These programmable coefficients are 32-bit, two's complement numbers. As described in [Table 6-15](#), these biquad filters can be allocated for each output channel based on the BIQUEAD_CFG[1:0] register setting of P0_R108. By setting BIQUEAD_CFG[1:0] to 2'b00, the biquad filtering for all record channels is disabled and the host device can choose this setting if no additional filtering is required for the system application. See the [TLV320ADCx140 Programmable Biquad Filter Configuration and Applications](#) application note for further details.

Table 6-15. Biquad Filter Allocation to the Record Output Channel

PROGRAMMABLE BIQUAD FILTER	RECORD OUTPUT CHANNEL ALLOCATION USING P0_R108_D[6:5] REGISTER SETTING		
	BIQUEAD_CFG[1:0] = 2'b01 (1 Biquad per Channel)	BIQUEAD_CFG[1:0] = 2'b10 (Default) (2 Biquads per Channel)	BIQUEAD_CFG[1:0] = 2'b11 (3 Biquads per Channel)
	SUPPORTS ALL 8 CHANNELS	SUPPORTS UP TO 6 CHANNELS	SUPPORTS UP TO 4 CHANNELS
Biquad filter 1	Allocated to output channel 1	Allocated to output channel 1	Allocated to output channel 1
Biquad filter 2	Allocated to output channel 2	Allocated to output channel 2	Allocated to output channel 2
Biquad filter 3	Allocated to output channel 3	Allocated to output channel 3	Allocated to output channel 3
Biquad filter 4	Allocated to output channel 4	Allocated to output channel 4	Allocated to output channel 4
Biquad filter 5	Not used	Allocated to output channel 1	Allocated to output channel 1
Biquad filter 6	Not used	Allocated to output channel 2	Allocated to output channel 2
Biquad filter 7	Not used	Allocated to output channel 3	Allocated to output channel 3
Biquad filter 8	Not used	Allocated to output channel 4	Allocated to output channel 4
Biquad filter 9	Allocated to output channel 5	Allocated to output channel 5	Allocated to output channel 1
Biquad filter 10	Allocated to output channel 6	Allocated to output channel 6	Allocated to output channel 2
Biquad filter 11	Allocated to output channel 7	Allocated to output channel 5	Allocated to output channel 3
Biquad filter 12	Allocated to output channel 8	Allocated to output channel 6	Allocated to output channel 4

[Table 6-16](#) shows the biquad filter coefficients mapping to the register space.

Table 6-16. Biquad Filter Coefficients Register Mapping

PROGRAMMABLE BIQUAD FILTER	BIQUAD FILTER COEFFICIENTS REGISTER MAPPING	PROGRAMMABLE BIQUAD FILTER	BIQUAD FILTER COEFFICIENTS REGISTER MAPPING
Biquad filter 1	P2_R8R27	Biquad filter 7	P3_R8R27
Biquad filter 2	P2_R28R47	Biquad filter 8	P3_R28R47
Biquad filter 3	P2_R48R67	Biquad filter 9	P3_R48R67
Biquad filter 4	P2_R68R87	Biquad filter 10	P3_R68R87
Biquad filter 5	P2_R88R107	Biquad filter 11	P3_R88R107
Biquad filter 6	P2_R108R127	Biquad filter 12	P3_R108R127

6.3.6.6 Programmable Channel Summer and Digital Mixer

For applications that require an even higher SNR than that supported for each channel, the device digital summing mode can be used. In this mode, the digital record data are summed up across the channel with an equal weightage factor, which helps in reducing the effective record noise. [Table 6-17](#) lists the configuration settings available for channel summing mode.

Table 6-17. Channel Summing Mode Programmable Settings

P0_R107_D[3:2] : CH_SUM[2:0]	CHANNEL SUMMING MODE FOR INPUT CHANNELS	SNR AND DYNAMIC RANGE BOOST
00 (Default)	Channel summing mode is disabled	Not applicable
01	Output channel 1 = (input channel 1 + input channel 2) / 2	3dB boost in SNR and dynamic range
	Output channel 2 = (input channel 1 + input channel 2) / 2	
	Output channel 3 = (input channel 3 + input channel 4) / 2	3dB boost in SNR and dynamic range
	Output channel 4 = (input channel 3 + input channel 4) / 2	
10	Output channel 1 = (input channel 1 + input channel 2 + input channel 3 + input channel 4) / 4	6dB boost in SNR and dynamic range
	Output channel 2 = (input channel 1 + input channel 2 + input channel 3 + input channel 4) / 4	
	Output channel 3 = (input channel 1 + input channel 2 + input channel 3 + input channel 4) / 4	
	Output channel 4 = (input channel 1 + input channel 2 + input channel 3 + input channel 4) / 4	
11	Reserved (do not use this setting)	Not applicable

The device additionally supports a fully programmable mixer feature that can mix the various input channels with the custom programmable scale factor to generate the final output channels. The programmable mixer feature is available only if CH_SUM[2:0] is set to 2'b00. The mixer function is only supported for input channel 1 to channel 4. [Figure 6-18](#) shows a block diagram that describes the mixer 1 operation to generate output channel 1. The programmable coefficients for the mixer operation are located in the [Programmable Coefficient Registers: Page = 0x04](#) section. All mixer coefficients are 32-bit, two's complement numbers using a 1.31 number format. The value of 0xFFFFFFFF is equivalent to +1 (0dB gain), the value 0x00000000 is equivalent to mute (zero data), and any values in between set the mixer attenuation computed using [Equation 3](#). If the MSB is set to '1' then the attenuation remains the same but the signal phase is inverted.

$$\text{hex2dec}(\text{value}) / 2^{31} \quad (3)$$

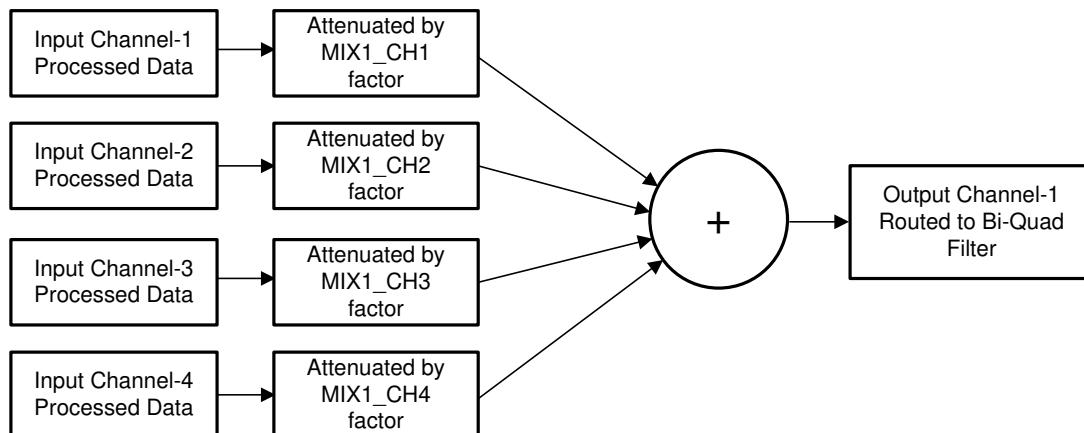


Figure 6-18. Programmable Digital Mixer Block Diagram

A similar mixer operation is performed by mixer 2, mixer 3, and mixer 4 to generate output channel 2, channel 3, and channel 4, respectively.

6.3.6.7 Configurable Digital Decimation Filters

The device record channel includes a high dynamic range, built-in digital decimation filter to process the oversampled PDM data stream from the digital microphone to generate digital data at the same Nyquist sampling rate as the FSYNC rate. The decimation filter can be chosen from three different types, depending on the required frequency response, group delay, and phase linearity requirements for the target application. The selection of the decimation filter option can be done by configuring the DECI_FILT, P0_R107_D[5:4] register bits. [Table 6-18](#) shows the configuration register setting for the decimation filter mode selection for the record channel.

Table 6-18. Decimation Filter Mode Selection for the Record Channel

P0_R107_D[5:4] : DECI_FILT[1:0]	DECIMATION FILTER MODE SELECTION
00 (default)	Linear phase filters are used for the decimation
01	Low latency filters are used for the decimation
10	Ultra-low latency filters are used for the decimation
11	Reserved (do not use this setting)

6.3.6.7.1 Linear Phase Filters

The linear phase decimation filters are the default filters set by the device and can be used for all applications that require a perfect linear phase with zero-phase deviation within the pass-band specification of the filter. The filter performance specifications and various plots for all supported output sampling rates are listed in this section.

6.3.6.7.1.1 Sampling Rate: 8kHz or 7.35kHz

[Figure 6-19](#) and [Figure 6-20](#) respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 8kHz or 7.35kHz. [Table 6-19](#) lists the specifications for a decimation filter with an 8kHz or 7.35kHz sampling rate.

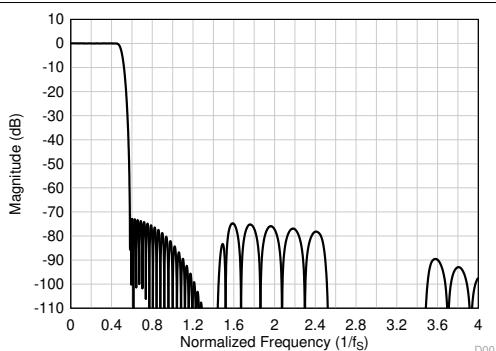


Figure 6-19. Linear Phase Decimation Filter Magnitude Response

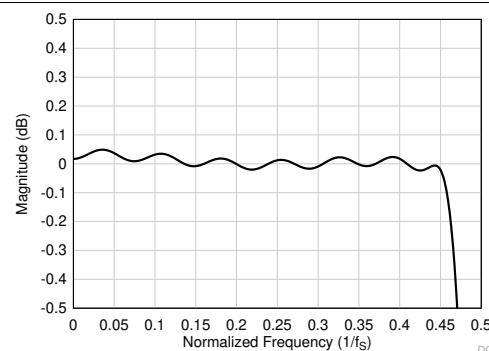


Figure 6-20. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-19. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.454 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	72.7			dB
	Frequency range is $4 \times f_S$ onwards	81.2			
Group delay or latency	Frequency range is 0 to $0.454 \times f_S$		17.1		$1/f_S$

6.3.6.7.1.2 Sampling Rate: 16kHz or 14.7kHz

Figure 6-21 and Figure 6-22 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 16kHz or 14.7kHz. Table 6-20 lists the specifications for a decimation filter with an 16kHz or 14.7kHz sampling rate.

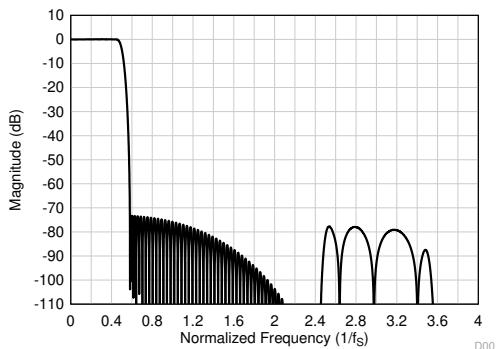


Figure 6-21. Linear Phase Decimation Filter Magnitude Response

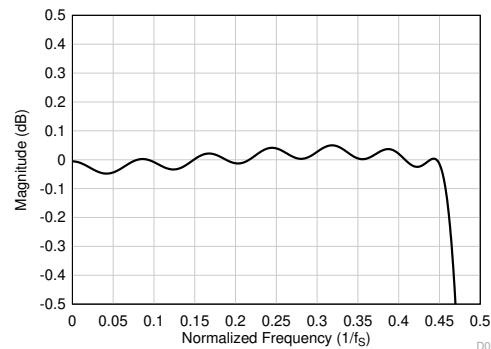


Figure 6-22. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-20. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.454 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	73.3			dB
	Frequency range is $4 \times f_S$ onwards	95.0			
Group delay or latency	Frequency range is 0 to $0.454 \times f_S$		15.7		$1/f_S$

6.3.6.7.1.3 Sampling Rate: 24kHz or 22.05kHz

Figure 6-23 and Figure 6-24 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 24kHz or 22.05kHz. Table 6-21 lists the specifications for a decimation filter with an 24kHz or 22.05kHz sampling rate.

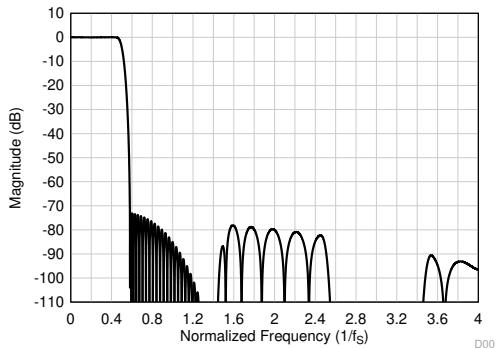


Figure 6-23. Linear Phase Decimation Filter Magnitude Response

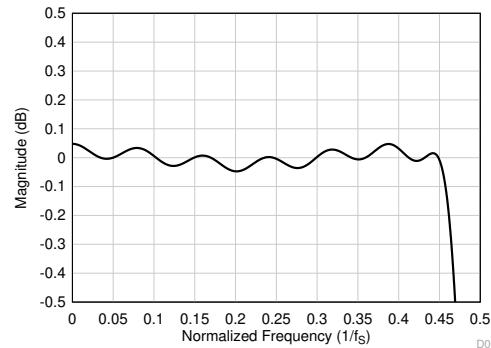


Figure 6-24. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-21. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.454 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	73.0			dB
	Frequency range is $4 \times f_S$ onwards	96.4			
Group delay or latency	Frequency range is 0 to $0.454 \times f_S$		16.6		$1/f_S$

6.3.6.7.1.4 Sampling Rate: 32kHz or 29.4kHz

Figure 6-25 and Figure 6-26 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 32kHz or 29.4kHz. Table 6-22 lists the specifications for a decimation filter with an 32kHz or 29.4kHz sampling rate.

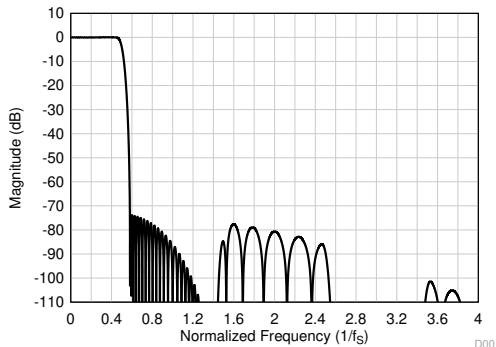


Figure 6-25. Linear Phase Decimation Filter Magnitude Response

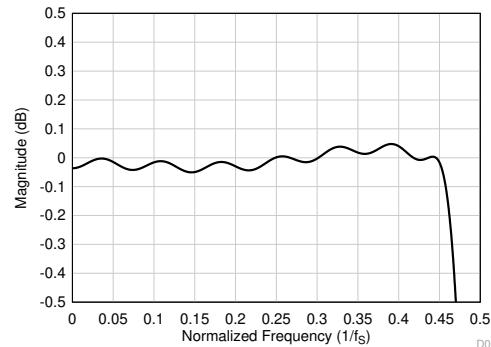


Figure 6-26. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-22. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.454 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	73.7			dB
	Frequency range is $4 \times f_S$ onwards	107.2			
Group delay or latency	Frequency range is 0 to $0.454 \times f_S$		16.9		$1/f_S$

6.3.6.7.1.5 Sampling Rate: 48kHz or 44.1kHz

Figure 6-27 and Figure 6-28 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 48kHz or 44.1kHz. Table 6-23 lists the specifications for a decimation filter with an 48kHz or 44.1kHz sampling rate.

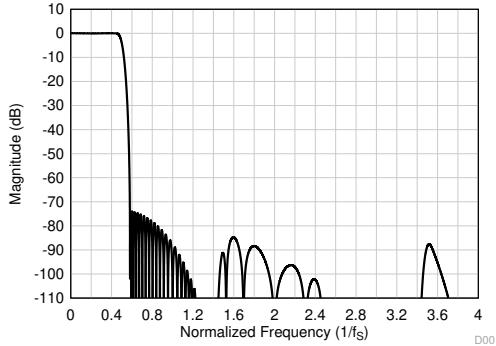


Figure 6-27. Linear Phase Decimation Filter Magnitude Response

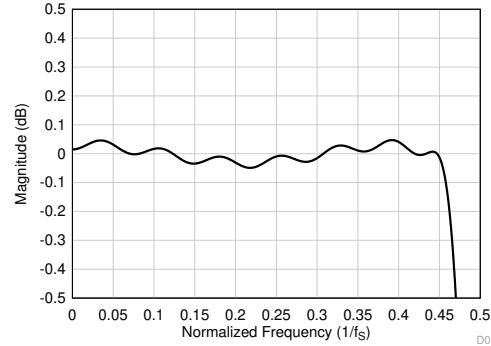


Figure 6-28. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-23. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.454 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	73.8			dB
	Frequency range is $4 \times f_S$ onwards	98.1			
Group delay or latency	Frequency range is 0 to $0.454 \times f_S$		17.1		$1/f_S$

6.3.6.7.1.6 Sampling Rate: 96kHz or 88.2kHz

Figure 6-29 and Figure 6-30 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 96kHz or 88.2kHz. Table 6-24 lists the specifications for a decimation filter with an 96kHz or 88.2kHz sampling rate.

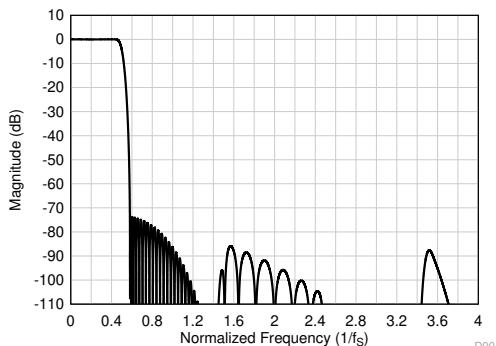


Figure 6-29. Linear Phase Decimation Filter Magnitude Response

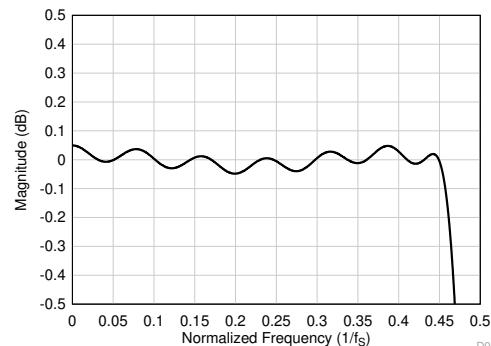


Figure 6-30. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-24. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.454 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	73.6			dB
	Frequency range is $4 \times f_S$ onwards	97.9			
Group delay or latency	Frequency range is 0 to $0.454 \times f_S$		17.1		$1/f_S$

6.3.6.7.1.7 Sampling Rate: 192kHz or 176.4kHz

Figure 6-31 and Figure 6-32 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 192kHz or 176.4kHz. Table 6-25 lists the specifications for a decimation filter with an 192kHz or 176.4kHz sampling rate.

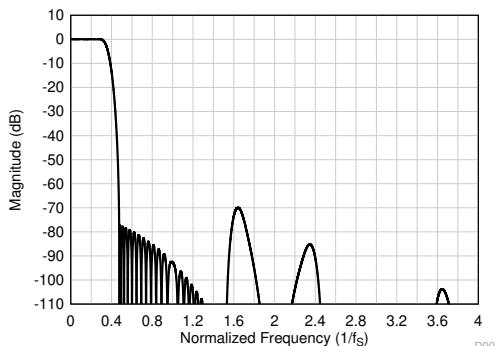


Figure 6-31. Linear Phase Decimation Filter Magnitude Response

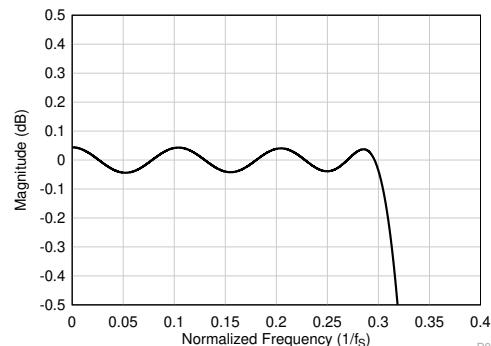


Figure 6-32. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-25. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.3 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.473 \times f_S$ to $4 \times f_S$	70.0			dB
	Frequency range is $4 \times f_S$ onwards	111.0			
Group delay or latency	Frequency range is 0 to $0.3 \times f_S$		11.9		$1/f_S$

6.3.6.7.1.8 Sampling Rate: 384kHz or 352.8kHz

Figure 6-33 and Figure 6-34 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 384kHz or 352.8kHz. Table 6-26 lists the specifications for a decimation filter with an 384kHz or 352.8kHz sampling rate.

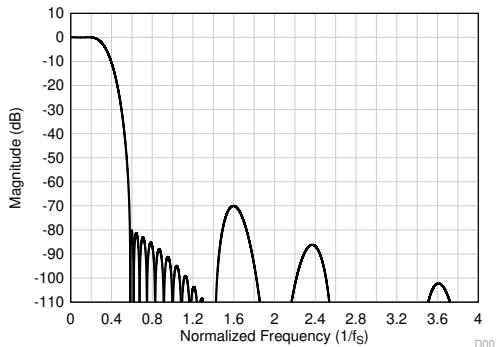


Figure 6-33. Linear Phase Decimation Filter Magnitude Response

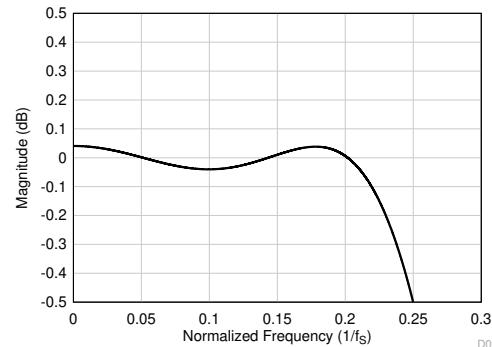


Figure 6-34. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-26. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.212 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $4 \times f_S$	70.0			dB
	Frequency range is $4 \times f_S$ onwards	108.8			
Group delay or latency	Frequency range is 0 to $0.212 \times f_S$		7.2		$1/f_S$

6.3.6.7.1.9 Sampling Rate 768kHz or 705.6kHz

Figure 6-35 and Figure 6-36 respectively show the magnitude response and the pass-band ripple for a decimation filter with a sampling rate of 768kHz or 705.6kHz. Table 6-27 lists the specifications for a decimation filter with an 768kHz or 705.6kHz sampling rate.

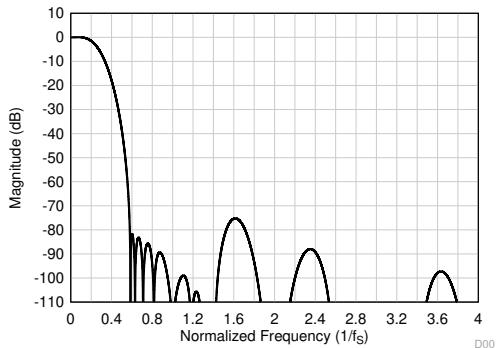


Figure 6-35. Linear Phase Decimation Filter Magnitude Response

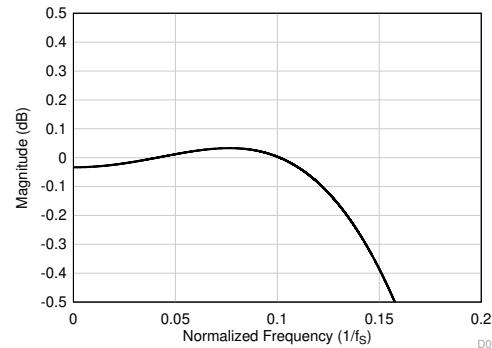


Figure 6-36. Linear Phase Decimation Filter Pass-Band Ripple

Table 6-27. Linear Phase Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.113 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.58 \times f_S$ to $2 \times f_S$	75.0			dB
	Frequency range is $2 \times f_S$ onwards	88.0			
Group Delay or Latency	Frequency range is 0 to $0.113 \times f_S$		5.9		$1/f_S$

6.3.6.7.2 Low-Latency Filters

For applications where low latency with minimal phase deviation (within the audio band) is critical, the low-latency decimation filters on the PCMD3180 can be used. The device supports these filters with a group delay of approximately seven samples with an almost linear phase response within the $0.365 \times f_S$ frequency band. This section provides the filter performance specifications and various plots for all supported output sampling rates for the low-latency filters.

6.3.6.7.2.1 Sampling Rate: 16kHz or 14.7kHz

Figure 6-37 shows the magnitude response and Figure 6-38 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 16kHz or 14.7kHz. Table 6-28 lists the specifications for a decimation filter with a 16kHz or 14.7kHz sampling rate.

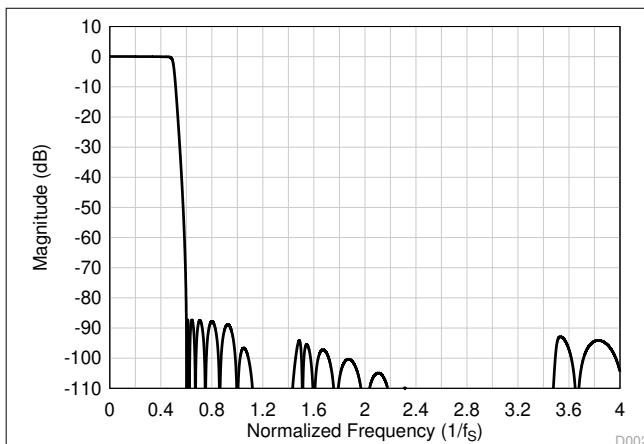


Figure 6-37. Low-Latency Decimation Filter Magnitude Response

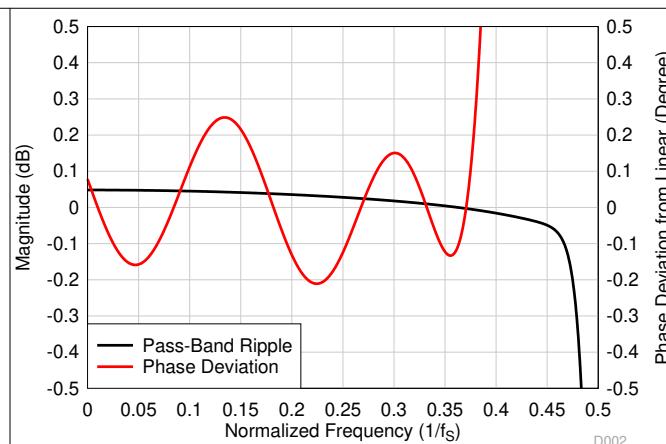


Figure 6-38. Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-28. Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.451 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.61 \times f_S$ onwards	87.3			dB
Group delay or latency	Frequency range is 0 to $0.363 \times f_S$		7.6		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.363 \times f_S$	-0.022		0.022	$1/f_S$
Phase deviation	Frequency range is 0 to $0.363 \times f_S$	-0.21		0.25	Degrees

6.3.6.7.2.2 Sampling Rate: 24kHz or 22.05kHz

Figure 6-39 shows the magnitude response and Figure 6-40 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 24kHz or 22.05kHz. Table 6-29 lists the specifications for a decimation filter with a 24kHz or 22.05kHz sampling rate.

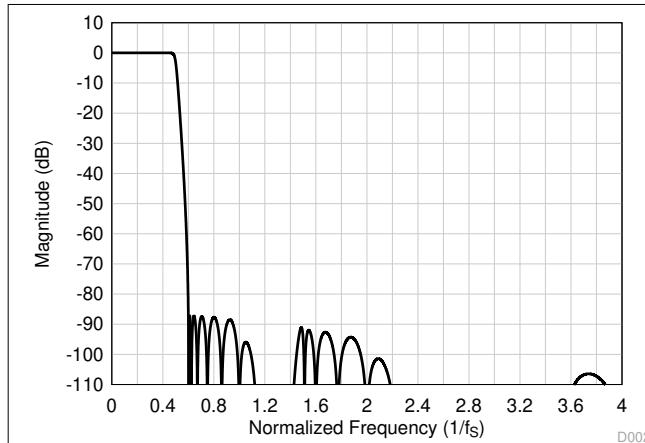


Figure 6-39. Low-Latency Decimation Filter Magnitude Response

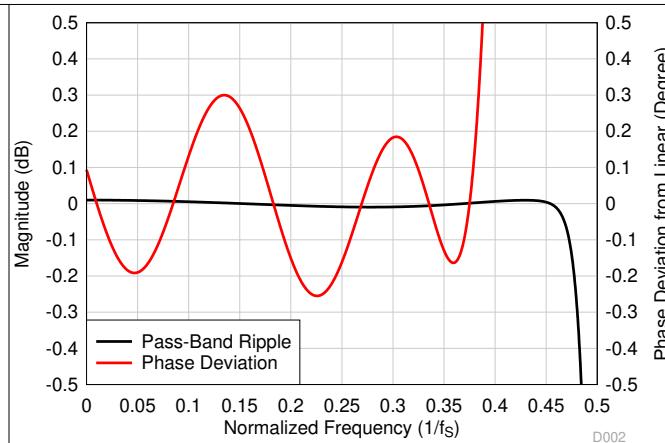


Figure 6-40. Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-29. Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.459 \times f_s$	-0.01		0.01	dB
Stop-band attenuation	Frequency range is $0.6 \times f_s$ onwards	87.2			dB
Group delay or latency	Frequency range is 0 to $0.365 \times f_s$		7.5		$1/f_s$
Group delay deviation	Frequency range is 0 to $0.365 \times f_s$	-0.026		0.026	$1/f_s$
Phase deviation	Frequency range is 0 to $0.365 \times f_s$	-0.26		0.30	Degrees

6.3.6.7.2.3 Sampling Rate: 32kHz or 29.4kHz

Figure 6-41 shows the magnitude response and Figure 6-42 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 32kHz or 29.4kHz. Table 6-30 lists the specifications for a decimation filter with a 32kHz or 29.4kHz sampling rate.

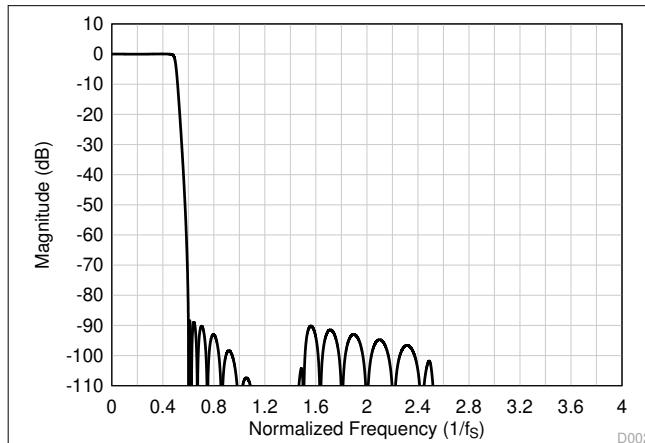


Figure 6-41. Low-Latency Decimation Filter Magnitude Response

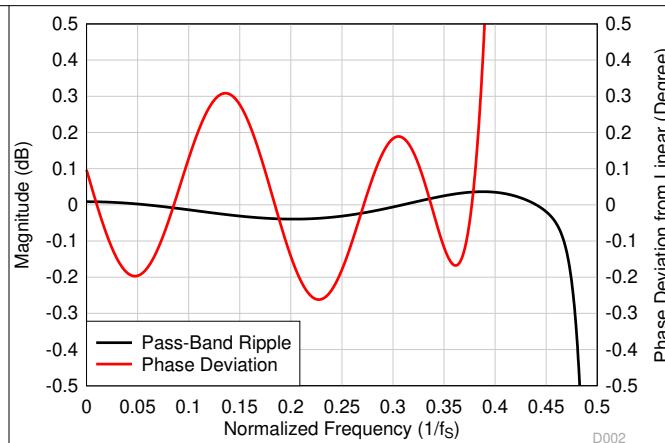


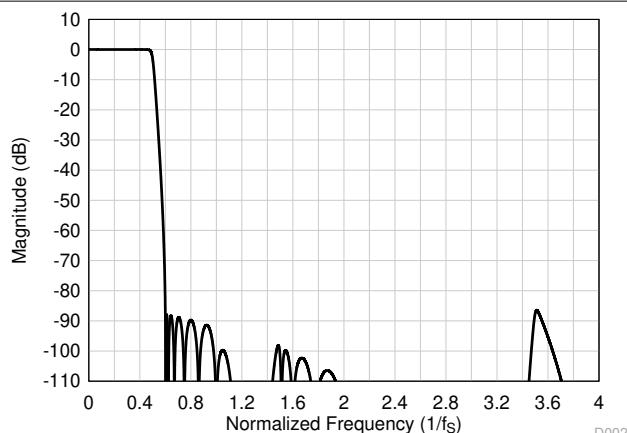
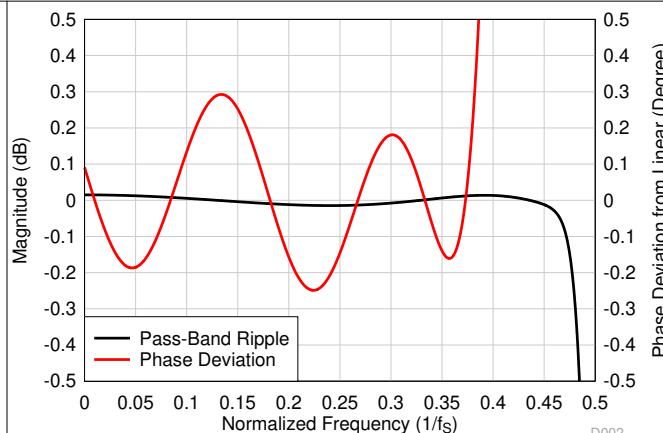
Figure 6-42. Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-30. Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.457 \times f_S$	-0.04		0.04	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	88.3			dB
Group delay or latency	Frequency range is 0 to $0.368 \times f_S$		8.7		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.368 \times f_S$	-0.026		0.026	$1/f_S$
Phase deviation	Frequency range is 0 to $0.368 \times f_S$	-0.26		0.31	Degrees

6.3.6.7.2.4 Sampling Rate: 48kHz or 44.1kHz

Figure 6-43 shows the magnitude response and Figure 6-44 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 48kHz or 44.1kHz. Table 6-31 lists the specifications for a decimation filter with a 48kHz or 44.1kHz sampling rate.


Figure 6-43. Low-Latency Decimation Filter Magnitude Response

Figure 6-44. Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation
Table 6-31. Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.452 \times f_S$	-0.015		0.015	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	86.4			dB
Group delay or latency	Frequency range is 0 to $0.365 \times f_S$		7.7		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.365 \times f_S$	-0.027		0.027	$1/f_S$
Phase deviation	Frequency range is 0 to $0.365 \times f_S$	-0.25		0.30	Degrees

6.3.6.7.2.5 Sampling Rate: 96kHz or 88.2kHz

Figure 6-45 shows the magnitude response and Figure 6-46 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 96kHz or 88.2kHz. Table 6-32 lists the specifications for a decimation filter with a 96kHz or 88.2kHz sampling rate.

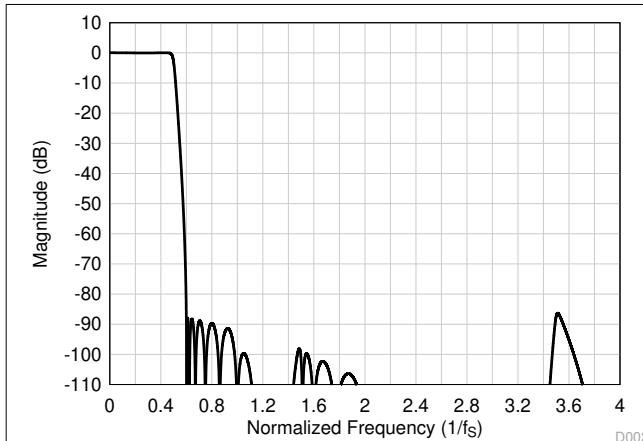


Figure 6-45. Low-Latency Decimation Filter Magnitude Response

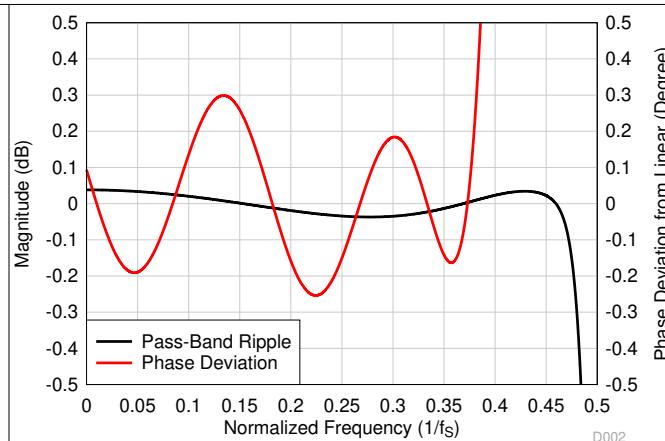


Figure 6-46. Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-32. Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.466 \times f_s$	-0.04	0.04	0.04	dB
Stop-band attenuation	Frequency range is $0.6 \times f_s$ onwards	86.3			dB
Group delay or latency	Frequency range is 0 to $0.365 \times f_s$		7.7		$1/f_s$
Group delay deviation	Frequency range is 0 to $0.365 \times f_s$	-0.027		0.027	$1/f_s$
Phase deviation	Frequency range is 0 to $0.365 \times f_s$	-0.26		0.30	Degrees

6.3.6.7.2.6 Sampling Rate 192kHz or 176.4kHz

Figure 6-47 shows the magnitude response and Figure 6-48 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 192kHz or 176.4kHz. Table 6-33 lists the specifications for a decimation filter with a 192kHz or 176.4kHz sampling rate.

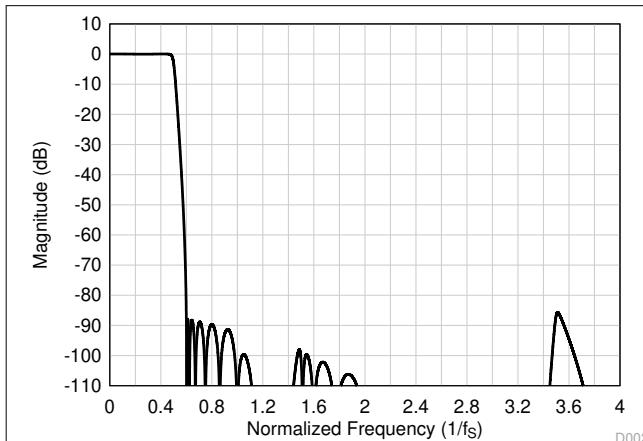


Figure 6-47. Low-Latency Decimation Filter Magnitude Response

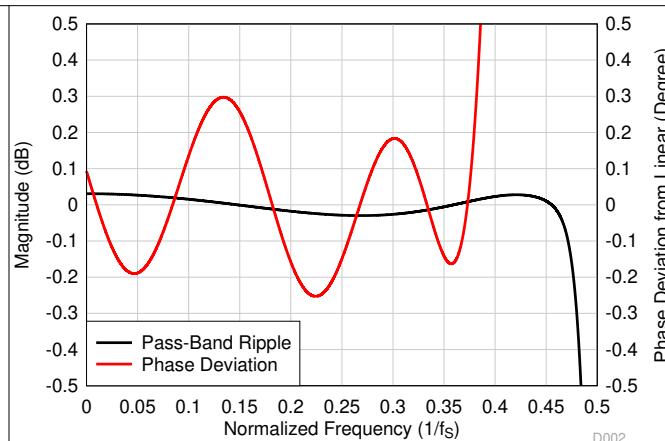


Figure 6-48. Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-33. Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $463 \times f_S$	-0.03		0.03	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	85.6			dB
Group delay or latency	Frequency range is 0 to $0.365 \times f_S$		7.7		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.365 \times f_S$	-0.027		0.027	$1/f_S$
Phase deviation	Frequency range is 0 to $0.365 \times f_S$	-0.26		0.30	Degrees

6.3.6.7.3 Ultra-Low-Latency Filters

For applications where ultra-low latency (within the audio band) is critical, the ultra-low-latency decimation filters on the PCMD3180 can be used. The device supports these filters with a group delay of approximately four samples with an almost linear phase response within the $0.325 \times f_S$ frequency band. This section provides the filter performance specifications and various plots for all supported output sampling rates for the ultra-low-latency filters.

6.3.6.7.3.1 Sampling Rate: 16kHz or 14.7kHz

Figure 6-49 shows the magnitude response and Figure 6-50 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 16kHz or 14.7kHz. Table 6-34 lists the specifications for a decimation filter with a 16kHz or 14.7kHz sampling rate.

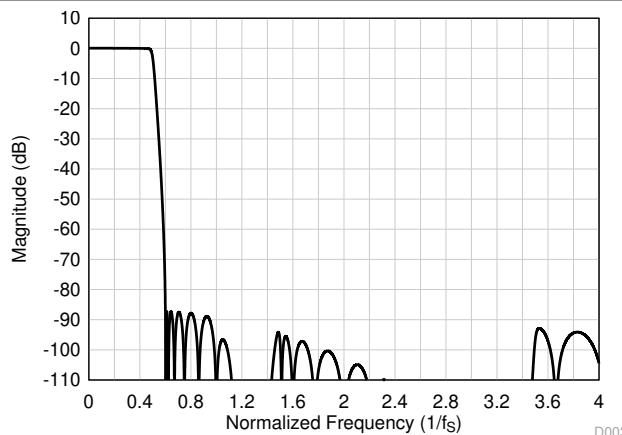


Figure 6-49. Ultra-Low-Latency Decimation Filter Magnitude Response

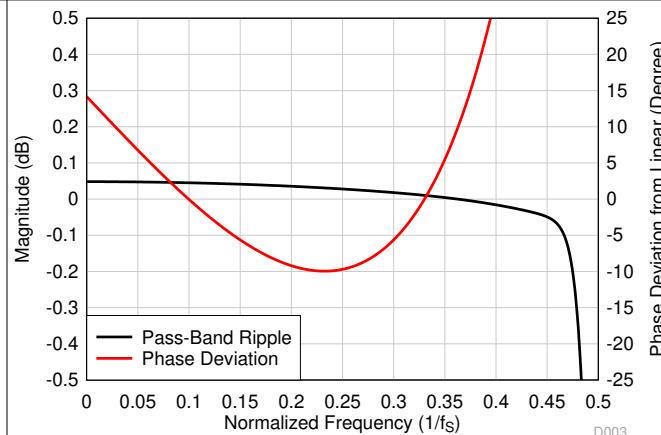


Figure 6-50. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-34. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.45 \times f_S$	-0.05		0.05	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	87.2			dB
Group delay or latency	Frequency range is 0 to $0.325 \times f_S$		4.3		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.325 \times f_S$	-0.512		0.512	$1/f_S$
Phase deviation	Frequency range is 0 to $0.325 \times f_S$	-10.0		14.2	Degrees

6.3.6.7.3.2 Sampling Rate: 24kHz or 22.05kHz

Figure 6-51 shows the magnitude response and Figure 6-52 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 24kHz or 22.05kHz. Table 6-35 lists the specifications for a decimation filter with a 24kHz or 22.05kHz sampling rate.

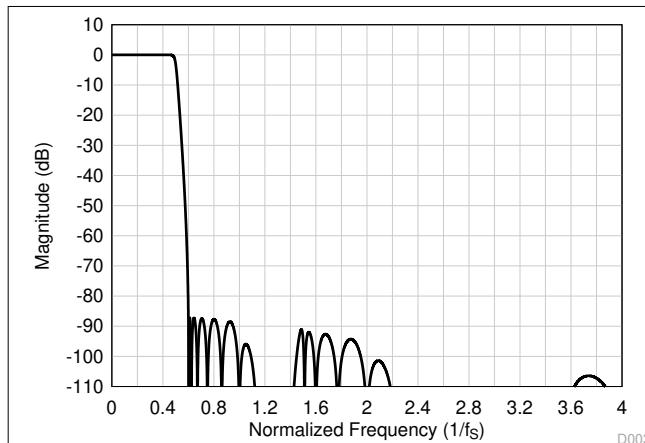


Figure 6-51. Ultra-Low-Latency Decimation Filter Magnitude Response

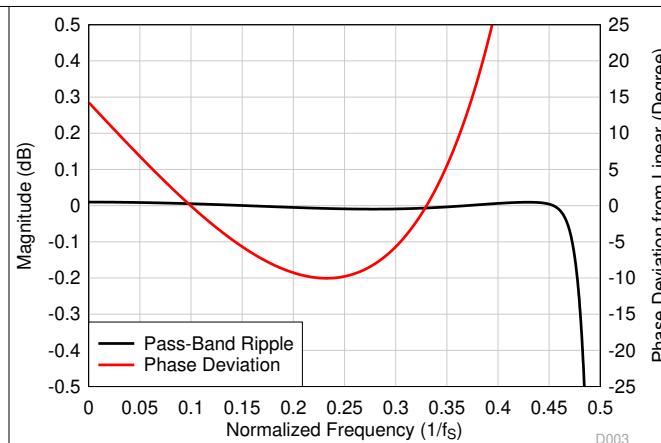


Figure 6-52. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-35. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.46 \times f_s$	-0.01		0.01	dB
Stop-band attenuation	Frequency range is $0.6 \times f_s$ onwards	87.1			dB
Group delay or latency	Frequency range is 0 to $0.325 \times f_s$		4.1		$1/f_s$
Group delay deviation	Frequency range is 0 to $0.325 \times f_s$	-0.514		0.514	$1/f_s$
Phase deviation	Frequency range is 0 to $0.325 \times f_s$	-10.0		14.3	Degrees

6.3.6.7.3.3 Sampling Rate: 32kHz or 29.4kHz

Figure 6-53 shows the magnitude response and Figure 6-54 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 32kHz or 29.4kHz. Table 6-36 lists the specifications for a decimation filter with an 32kHz or 29.4kHz sampling rate.

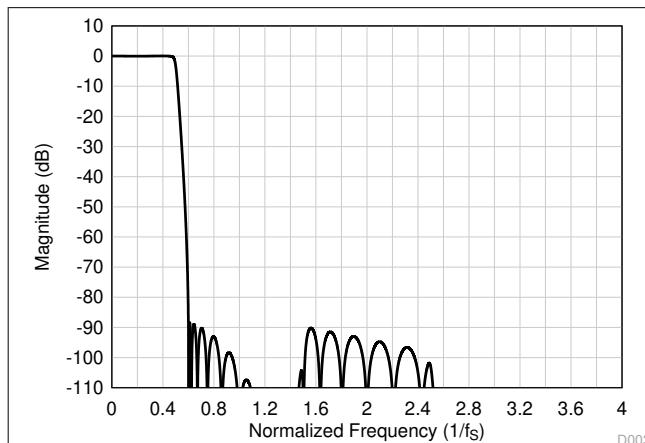


Figure 6-53. Ultra-Low-Latency Decimation Filter Magnitude Response

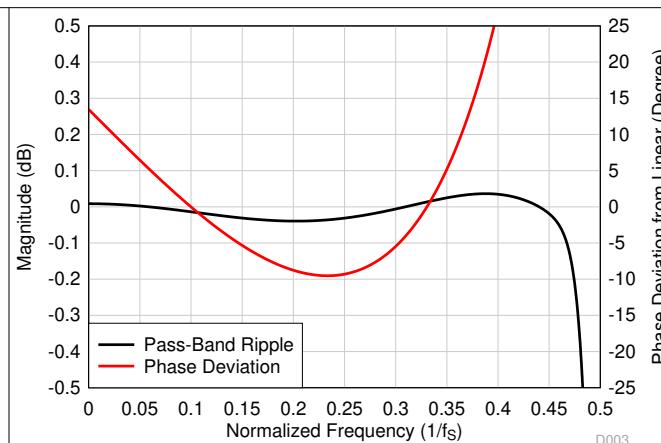


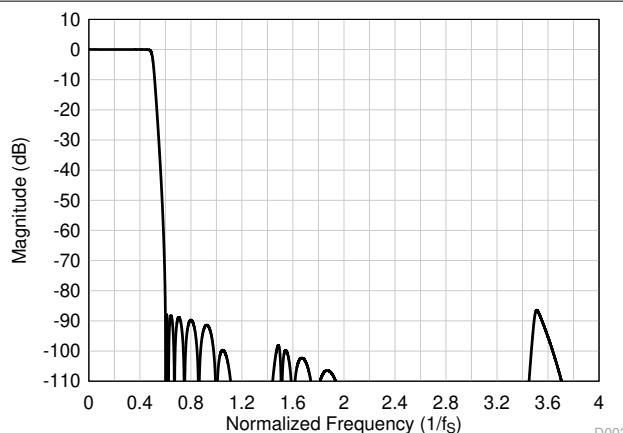
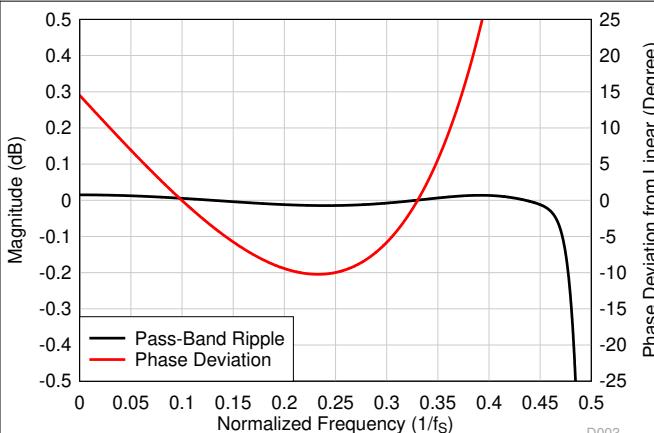
Figure 6-54. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-36. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.457 \times f_S$	-0.04		0.04	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	88.3			dB
Group delay or latency	Frequency range is 0 to $0.325 \times f_S$		5.2		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.325 \times f_S$	-0.492		0.492	$1/f_S$
Phase deviation	Frequency range is 0 to $0.325 \times f_S$	-9.5		13.5	Degrees

6.3.6.7.3.4 Sampling Rate: 48kHz or 44.1kHz

Figure 6-55 shows the magnitude response and Figure 6-56 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 48kHz or 44.1kHz. Table 6-37 lists the specifications for a decimation filter with a 48kHz or 44.1kHz sampling rate.


Figure 6-55. Ultra-Low-Latency Decimation Filter Magnitude Response

Figure 6-56. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation
Table 6-37. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.452 \times f_S$	-0.015		0.015	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	86.4			dB
Group delay or latency	Frequency range is 0 to $0.325 \times f_S$		4.1		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.325 \times f_S$	-0.525		0.525	$1/f_S$
Phase deviation	Frequency range is 0 to $0.325 \times f_S$	-10.3		14.5	Degrees

6.3.6.7.3.5 Sampling Rate: 96kHz or 88.2kHz

Figure 6-57 shows the magnitude response and Figure 6-58 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 96kHz or 88.2kHz. Table 6-38 lists the specifications for a decimation filter with a 96kHz or 88.2kHz sampling rate.

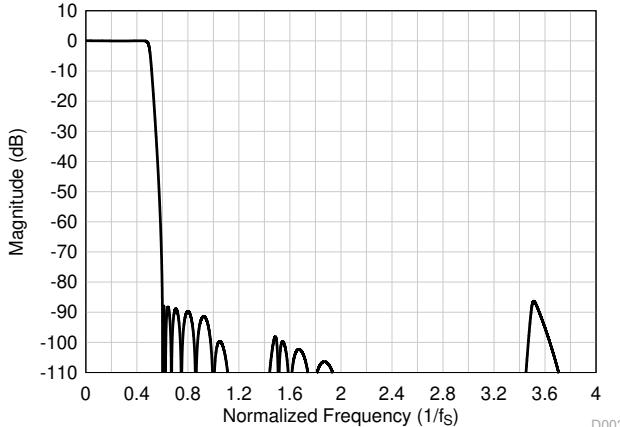


Figure 6-57. Ultra-Low-Latency Decimation Filter Magnitude Response

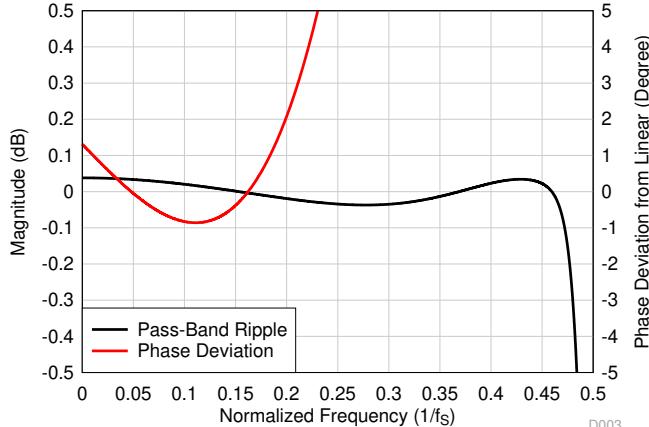


Figure 6-58. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-38. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.466 \times f_S$	-0.04	0.04	0.04	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	86.3			dB
Group delay or latency	Frequency range is 0 to $0.1625 \times f_S$		3.7		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.1625 \times f_S$	-0.091		0.091	$1/f_S$
Phase deviation	Frequency range is 0 to $0.1625 \times f_S$	-0.86		1.30	Degrees

6.3.6.7.3.6 Sampling Rate 192kHz or 176.4kHz

Figure 6-59 shows the magnitude response and Figure 6-60 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 192kHz or 176.4kHz. Table 6-39 lists the specifications for a decimation filter with a 192kHz or 176.4kHz sampling rate.

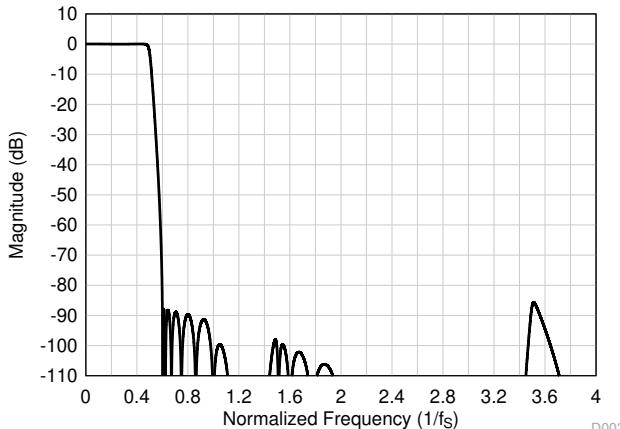


Figure 6-59. Ultra-Low-Latency Decimation Filter Magnitude Response

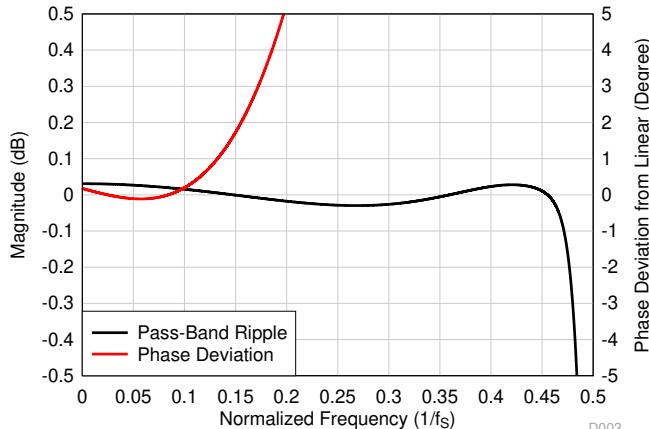


Figure 6-60. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-39. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.463 \times f_S$	-0.03		0.03	dB
Stop-band attenuation	Frequency range is $0.6 \times f_S$ onwards	85.6			dB
Group delay or latency	Frequency range is 0 to $0.085 \times f_S$		3.7		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.085 \times f_S$	-0.024		0.024	$1/f_S$
Phase deviation	Frequency range is 0 to $0.085 \times f_S$	-0.12		0.18	Degrees

6.3.6.7.3.7 Sampling Rate 384kHz or 352.8kHz

Figure 6-61 shows the magnitude response and Figure 6-62 shows the pass-band ripple and phase deviation for a decimation filter with a sampling rate of 384kHz or 352.8kHz. Table 6-40 lists the specifications for a decimation filter with a 384kHz or 352.8kHz sampling rate.

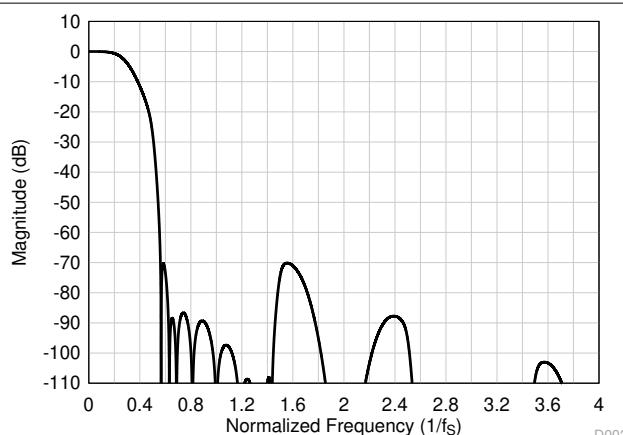


Figure 6-61. Ultra-Low-Latency Decimation Filter Magnitude Response

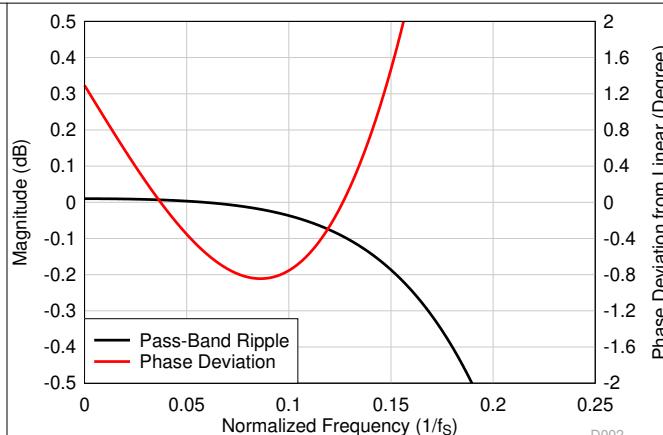


Figure 6-62. Ultra-Low-Latency Decimation Filter Pass-Band Ripple and Phase Deviation

Table 6-40. Ultra-Low-Latency Decimation Filter Specifications

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Pass-band ripple	Frequency range is 0 to $0.1 \times f_S$	-0.04		0.01	dB
Stop-band attenuation	Frequency range is $0.56 \times f_S$ onwards	70.1			dB
Group delay or latency	Frequency range is 0 to $0.157 \times f_S$		4.1		$1/f_S$
Group delay deviation	Frequency range is 0 to $0.157 \times f_S$	-0.18		0.18	$1/f_S$
Phase deviation	Frequency range is 0 to $0.157 \times f_S$	-0.85		2.07	Degrees

6.3.7 Interrupts, Status, and Digital I/O Pin Multiplexing

Certain events in the device can require host processor intervention and can be used to trigger interrupts to the host processor. One such event is an audio serial interface (ASI) bus error. The device powers down the record channels if any faults are detected with the ASI bus error clocks, such as:

- Invalid FSYNC frequency
- Invalid SBCLK to FSYNC ratio
- Long pauses of the SBCLK or FSYNC clocks

When an ASI bus clock error is detected, the device shuts down the record channel as quickly as possible. After all ASI bus clock errors are resolved, the device volume ramps back to the previous state to recover the record channel. During an ASI bus clock error, the internal interrupt request (IRQ) interrupt signal asserts low if the clock error interrupt mask register bit INT_MASK0[7], P0_R51_D7 is set low. The clock fault is also available for readback in the latched fault status register bit INT_LTCH0, P0_R54, which is a read-only register. Reading the latched fault status register, INT_LTCH0, clears all latched fault status. The device can be additionally configured to route the internal IRQ interrupt signal on the GPIO1 or GPOx pins and also can be configured as open-drain outputs so that these pins can be wire-ANDed to the open-drain interrupt outputs of other devices.

The IRQ interrupt signal can either be configured as active low or active high polarity by setting the INT_POL, P0_R50_D7 register bit. This signal can also be configured as a single pulse or a series of pulses by programming the INT_EVENT[1:0], P0_R50_D[6:5] register bits. If the interrupts are configured as a series of pulses, the events trigger the start of pulses that stop when the latched fault status register is read to determine the cause of the interrupt.

The device also supports read-only live-status registers to determine if the channels are powered up or down and if the device is in sleep mode or not. These status registers are located in P0_R118, DEV_STS0 and P0_R119, DEV_STS1.

The device has a multifunctional GPIO1 pin that can be configured for a desired specific function. Additionally, PDMINx_GPIx and PDMCLKx_GPOx can be repurposed as multifunction pins GPIx and GPOx respectively, as required for system application. The maximum number of GPO pins supported by the device is four and the maximum number of GPI pins are four. [Table 6-41](#) shows all possible allocations of these multifunctional pins for the various features.

Table 6-41. Multifunction Pin Assignments

ROW	Pin Function ⁽⁴⁾	GPIO1	GPO1	GPO2	GPO3	GPO4	GPI1	GPI2	GPI3	GPI4
—	—	GPIO1_CFG	GPO1_CFG	GPO2_CFG	GPO3_CFG	GPO4_CFG	GPI1_CFG	GPI2_CFG	GPI3_CFG	GPI4_CFG
—	—	P0_R33[7:4]	P0_R34[7:4]	P0_R35[7:4]	P0_R36[7:4]	P0_R37[7:4]	P0_R43[6:4]	P0_R43[2:0]	P0_R44[6:4]	P0_R44[2:0]
A	Pin disabled	S ⁽¹⁾	S (default)	S (default)	S (default)	S (default)				
B	General-purpose output (GPO)	S	S	S	S	S	NS ⁽²⁾	NS	NS	NS
C	Interrupt output (IRQ)	S (default)	S	S	S	S	NS	NS	NS	NS
D	Secondary ASI output (SDOUT2) ⁽³⁾	S	S	S	S	S	NS	NS	NS	NS
E	PDM clock output (PDMCLK)	S	S	S	S	S	NS	NS	NS	NS
F	MICBIAS on/off input (BIASEN)	S	NS	NS	NS	NS	NS	NS	NS	NS
G	General-purpose input (GPI)	S	NS	NS	NS	NS	S	S	S	S
H	Master clock input (MCLK)	S	NS	NS	NS	NS	S	S	S	S
I	ASI daisy-chain input (SDIN)	S	NS	NS	NS	NS	S	S	S	S
J	PDM data input 1 (PDMIN1)	S	NS	NS	NS	NS	S	S	S	S
K	PDM data input 2 (PDMIN2)	S	NS	NS	NS	NS	S	S	S	S
L	PDM data input 3 (PDMIN3)	S	NS	NS	NS	NS	S	S	S	S
M	PDM data input 4 (PDMIN4)	S	NS	NS	NS	NS	S	S	S	S

(1) S means the feature mentioned in this row is *supported* for the respective GPIO1, GPOx, or GPIx pin mentioned in this column.

(2) NS means the feature mentioned in this row is *not supported* for the respective GPIO1, GPOx, or GPIx pin mentioned in this column.

- (3) For the high-speed ASI output, GPIO1 must be used instead of GPOx for the secondary ASI output. GPOx can be used only if the bus speed requirement is less than 6.144MHz.
- (4) Only the GPIO1 pin is with reference to the IOVDD supply, the other GPOx and GPIx pins are with reference to the AVDD supply and the primary pin functions are for the PDMCLK or PDMDIN function.

Each GPOx or GPIOx pin can be independently set for the desired drive configurations setting using the GPO_x_DRV[3:0] or GPIO1_DRV[3:0] register bits. [Table 6-42](#) lists the drive configuration settings.

Table 6-42. GPIO or GPOx Pins Drive Configuration Settings

P0_R33_D[3:0] : GPIO1_DRV[3:0]	GPIO OUTPUT DRIVE CONFIGURATION SETTINGS FOR GPIO1
000	The GPIO1 pin is set to high impedance (floated)
001	The GPIO1 pin is set to be driven active low or active high
010 (default)	The GPIO1 pin is set to be driven active low or weak high (on-chip pullup)
011	The GPIO1 pin is set to be driven active low or Hi-Z (floated)
100	The GPIO1 pin is set to be driven weak low (on-chip pulldown) or active high
101	The GPIO1 pin is set to be driven Hi-Z (floated) or active high
110 and 111	Reserved (do not use these settings)

Similarly, the GPO1 to GPO4 pins can be configured using the GPO1_DRV(P0_R34) to GPO4_DRV(P0_R37) register bits, respectively.

When configured as a general-purpose output (GPO), the GPIO1 or GPOx pin values can be driven by writing the GPIO_VAL or GPO_x_VAL, P0_R41 registers. The GPIO_MON, P0_R42 register can be used to readback the status of the GPIO1 pin when configured as a general-purpose input (GPI). Similarly, the GPI_MON, P0_R47 register can be used to readback the status of the GPIx pins when configured as a general-purpose input (GPI).

6.4 Device Functional Modes

6.4.1 Hardware Shutdown

The device enters hardware shutdown mode when the SHDNZ pin is asserted low or the AVDD supply voltage is not applied to the device. In hardware shutdown mode, the device consumes the minimum quiescent current from the AVDD supply. All configuration registers and programmable coefficients lose the value in this mode, and I²C or SPI communication to the device is not supported.

If the SHDNZ pin is asserted low when the device is in active mode, the device ramps down volume on the record data, powers down the analog and digital blocks, and puts the device into hardware shutdown mode in 25ms (typical). The device can also be immediately put into hardware shutdown mode from active mode if the SHDNZ_CFG[1:0], P0_R5_D[3:2], register bits are set to 2'b00. After the SHDNZ pin is asserted low, and after the device enters hardware shutdown mode, keep the SHDNZ pin low for at least 1ms before releasing SHDNZ for further device operation.

Assert the SHDNZ pin high only when the IOVDD supply settles to a steady voltage level. When the SHDNZ pin goes high, the device sets all configuration registers and programmable coefficients to the default values, and then enters sleep mode.

6.4.2 Sleep Mode or Software Shutdown

In sleep mode or software shutdown mode, the device consumes very low quiescent current from the AVDD supply and, at the same time, allows the I²C or SPI communication to wake the device for active operation.

The device can also enter sleep mode when the host device sets the SLEEP_ENZ, P0_R2_D0 bit to 1'b0. If the SLEEP_ENZ bit is asserted low when the device is in active mode, the device ramps down the volume on the record data, powers down the analog and digital blocks, and enters sleep mode. However, the device still continues to retain the last programmed value of the device configuration registers and programmable coefficients.

In sleep mode, do not perform any I²C or SPI transactions, except for exiting sleep mode to enter active mode. After entering sleep mode, wait at least 10ms before starting I²C or SPI transactions to exit sleep mode.

While exiting sleep mode, the host device must configure the PCMD3180 to use either an external 1.8V AREG supply (default setting) or an on-chip-regulator-generated AREG supply. To configure the AREG supply, write to AREG_SELECT, bit D7 in the same P0_R2 register.

6.4.3 Active Mode

If the host device exits sleep mode by setting the SLEEP_ENZ bit to 1'b1, the device enters active mode. In active mode, I²C or SPI transactions can be done to configure and power-up the device for active operation. After entering active mode, wait at least 1ms before starting any I²C or SPI transactions to allow the device to complete the internal wake-up sequence.

After configuring all other registers for the target application and system settings, configure the input and output channel enable registers, P0_R115 (IN_CH_EN) and P0_R116 (ASI_OUT_CH_EN), respectively. Lastly, configure the device power-up register, P0_R117 (PWR_CFG). All the programmable coefficient values must be written before powering up the respective channel.

In active mode, the power-up and power-down status of various blocks is monitored by reading the read-only device status bits located in the P0_R117 (DEV_STS0) and P0_R118 (DEV_STS1) registers.

6.4.4 Software Reset

A software reset can be done any time by asserting the SW_RESET bit, P0_R1_D0, which is a self-clearing bit. This software reset immediately shuts down the device, and restores all device configuration registers and programmable coefficients to the default values.

6.5 Programming

The device contains configuration registers and programmable coefficients that can be set to the desired values for a specific system and application use. These registers are called *device control registers* and are each eight bits in width, mapped using a page scheme.

Each page contains 128 configuration registers. All device configuration registers are stored in page 0, which is the default page setting at power up and after a software reset. All programmable coefficient registers are located in page 2, page 3, and page 4. The current page of the device can be switched to a new desired page by using the PAGE[7:0] bits located in register 0 of every page.

6.5.1 Control Serial Interfaces

The device control registers can be accessed using either I²C or SPI communication to the device.

By monitoring the SDA_SSZ, SCL_MOSI, ADDR0_SCLK, and ADDR1_MISO device pins, which are the multiplexed pins for the I²C or SPI, the device automatically detects whether the host device is using I²C or SPI communication to configure the device. For a given end application, the host device must always use either the I²C or SPI, but not both, to configure the device.

6.5.1.1 I²C Control Interface

The device supports the I²C control protocol as a slave device, and is capable of operating in standard mode, fast mode, and fast mode plus. The I²C control protocol requires a 7-bit slave address. The five most significant bits (MSBs) of the slave address are fixed at 10011 and cannot be changed. The two least significant bits (LSBs) are programmable and are controlled by the ADDR0_SCLK and ADDR1_MISO pins. These two pins must always be either pulled to VSS or IOVDD. If the I2C_BRDCAST_EN (P0_R2_D2) bit is set to 1'b1, then the I²C slave address is fixed to 1001100 to allow simultaneous I²C broadcast communication to all PCMD3180 devices in the system. [Table 6-43](#) lists the four possible device addresses resulting from this configuration.

Table 6-43. I²C Slave Address Settings

ADDR1_MISO	ADDR0_SCLK	I2C_BRDCAST_EN (P0_R2_D2)	I ² C SLAVE ADDRESS
0	0	0 (default)	1001 100
0	1	0 (default)	1001 101
1	0	0 (default)	1001 110
1	1	0 (default)	1001 111
X	X	1	1001 100

6.5.1.1.1 General I²C Operation

The I²C bus employs two signals, SDA (data) and SCL (clock), to communicate between integrated circuits in a system using serial data transmission. The address and data 8-bit bytes are transferred MSB first. In addition, each byte transferred on the bus is acknowledged by the receiving device with an acknowledge bit. Each transfer operation begins with the master device driving a start condition on the bus and ends with the master device driving a stop condition on the bus. The bus uses transitions on the data pin (SDA) while the clock is at logic high to indicate start and stop conditions. A high-to-low transition on SDA indicates a start, and a low-to-high transition indicates a stop. Normal data-bit transitions must occur within the low time of the clock period.

The master device drives a start condition followed by the 7-bit slave address and the read/write (R/W) bit to open communication with another device and then waits for an acknowledgment condition. The slave device holds SDA low during the acknowledge clock period to indicate acknowledgment. When this occurs, the master device transmits the next byte of the sequence. Each slave device is addressed by a unique 7-bit slave address plus the R/W bit (1 byte). All compatible devices share the same signals via a bidirectional bus using a wired-AND connection.

There is no limit on the number of bytes that can be transmitted between start and stop conditions. When the last word transfers, the master device generates a stop condition to release the bus. [Figure 6-63](#) shows a generic data transfer sequence.

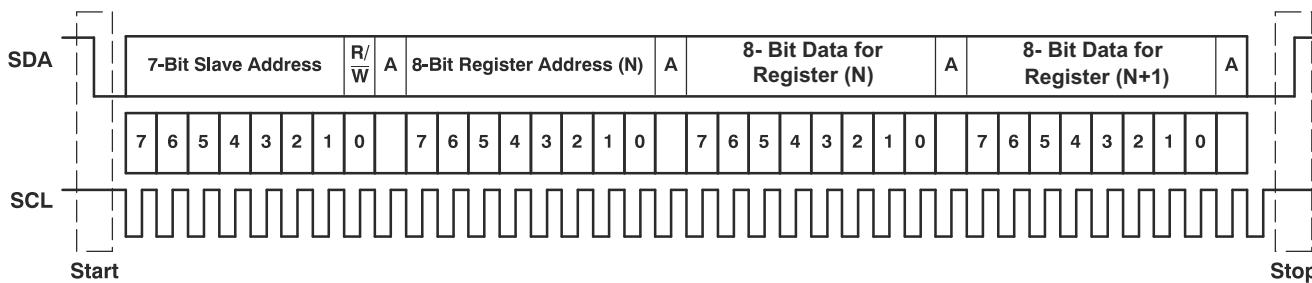


Figure 6-63. Typical I²C Sequence

In the system, use external pullup resistors for the SDA and SCL signals to set the logic high level for the bus. The SDA and SCL voltages must not exceed the device supply voltage, IOVDD.

6.5.1.1.2 I²C Single-Byte and Multiple-Byte Transfers

The device I²C interface supports both single-byte and multiple-byte read/write operations for all registers. During multiple-byte read operations, the device responds with data, a byte at a time, starting at the register assigned, as long as the master device continues to respond with acknowledges.

The device supports sequential I²C addressing. For write transactions, if a register is issued followed by data for that register and all the remaining registers that follow, a sequential I²C write transaction takes place. For I²C sequential write transactions, the register issued then serves as the starting point, and the amount of data subsequently transmitted, before a stop or start is transmitted, determines how many registers are written.

6.5.1.1.2.1 I²C Single-Byte Write

As shown in [Figure 6-64](#), a single-byte data write transfer begins with the master device transmitting a start condition followed by the I²C device address and the read/write bit. The read/write bit determines the direction of the data transfer. For a write-data transfer, the read/write bit must be set to 0. After receiving the correct I²C slave address and the read/write bit, the device responds with an acknowledge bit (ACK). Next, the master device transmits the register byte corresponding to the device internal register address being accessed. After receiving the register byte, the device again responds with an acknowledge bit (ACK). Then, the master transmits the byte of data to be written to the specified register. When finished, the slave device responds with an acknowledge bit (ACK). Finally, the master device transmits a stop condition to complete the single-byte data write transfer.

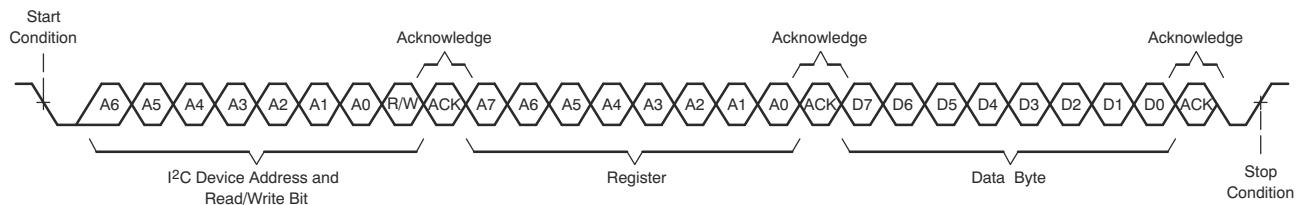


Figure 6-64. I²C Single-Byte Write Transfer

6.5.1.1.2.2 I²C Multiple-Byte Write

As shown in [Figure 6-65](#), a multiple-byte data write transfer is identical to a single-byte data write transfer except that multiple data bytes are transmitted by the master device to the slave device. After receiving each data byte, the device responds with an acknowledge bit (ACK). Finally, the master device transmits a stop condition after the last data-byte write transfer.

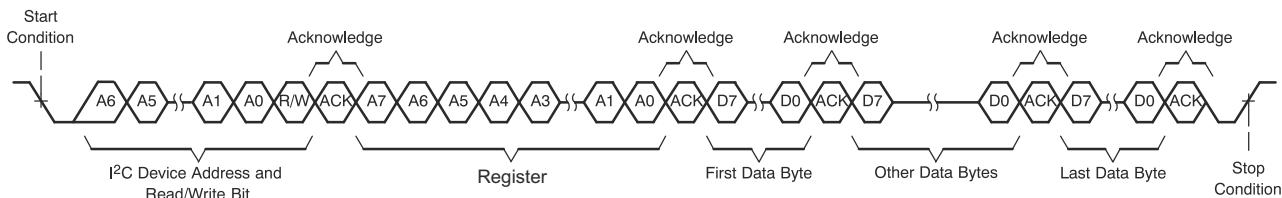


Figure 6-65. I²C Multiple-Byte Write Transfer

6.5.1.1.2.3 I²C Single-Byte Read

As shown in [Figure 6-66](#), a single-byte data read transfer begins with the master device transmitting a start condition followed by the I²C slave address and the read/write bit. For the data read transfer, both a write followed by a read are done. Initially, a write is done to transfer the address byte of the internal register address to be read. As a result, the read/write bit is set to 0.

After receiving the slave address and the read/write bit, the device responds with an acknowledge bit (ACK). The master device then sends the internal register address byte, after which the device issues an acknowledge bit (ACK). The master device transmits another start condition followed by the slave address and the read/write bit again. This time, the read/write bit is set to 1, indicating a read transfer. Next, the device transmits the data byte from the register address being read. After receiving the data byte, the master device transmits a not-acknowledge (NACK) followed by a stop condition to complete the single-byte data read transfer.

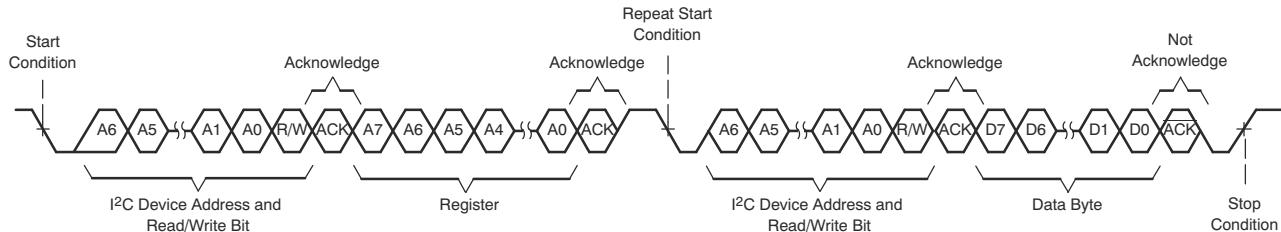


Figure 6-66. I²C Single-Byte Read Transfer

6.5.1.1.2.4 I²C Multiple-Byte Read

As shown in [Figure 6-67](#), a multiple-byte data read transfer is identical to a single-byte data read transfer except that multiple data bytes are transmitted by the device to the master device. With the exception of the last data byte, the master device responds with an acknowledge bit after receiving each data byte. After receiving the last data byte, the master device transmits a not-acknowledge (NACK) followed by a stop condition to complete the data read transfer.

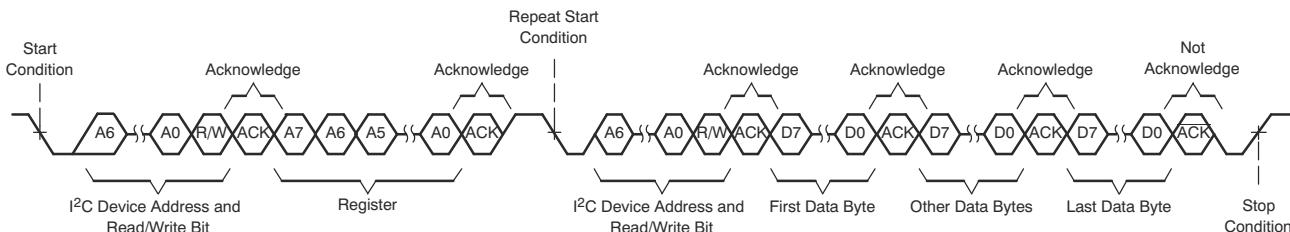


Figure 6-67. I²C Multiple-Byte Read Transfer

6.5.1.2 SPI Control Interface

The general SPI protocol allows full-duplex, synchronous, serial communication between a host processor (the master) and peripheral devices (slaves). The SPI master (in this case, the host processor) generates the synchronizing clock (driven onto SCLK) and initiates transmissions by taking the slave-select pin SSZ from high to low. The SPI slave devices (such as the PCMD3180) depend on a master to start and synchronize transmissions. A transmission begins when initiated by an SPI master. The byte from the SPI master begins shifting in on the slave MOSI pin under the control of the master serial clock (driven onto SCLK). When the byte shifts in on the MOSI pin, a byte shifts out on the MISO pin to the master shift register.

The PCMD3180 supports a standard SPI control protocol with a clock polarity setting of 0 (typical microprocessor SPI control bit CPOL = 0) and a clock phase setting of 1 (typical microprocessor SPI control bit CPHA = 1). The SSZ pin can remain low between transmissions; however, the device only interprets the first eight bits transmitted after the falling edge of SSZ as a command byte, and the next eight bits as a data byte only if writing to a register. The device is entirely controlled by registers. Reading and writing these registers is accomplished by an 8-bit command sent to the MOSI pin prior to the data for that register. [Table 6-44](#) shows the command structure. The first seven bits specify the address of the register that is being written or read, from 0 to 127 (decimal). The command word ends with an R/W bit, which specifies the direction of data flow on the serial bus.

In the case of a register write, set the R/W bit to 0. A second byte of data is sent to the MOSI pin and contains the data to be written to the register. A register read is accomplished in a similar fashion. The 8-bit command word sends the 7-bit register address, followed by the R/W bit equal to 1 to signify a register read. The 8-bit register data is then clocked out of the device on the MISO pin during the second eight SCLK clocks in the frame. The device supports sequential SPI addressing for a multiple-byte data write/read transfer until the SSZ pin is pulled high. A multiple-byte data write or read transfer is identical to a single-byte data write or read transfer, respectively, until all data byte transfers complete. The host device must keep the SSZ pin low during all data byte transfers. [Figure 6-68](#) shows the single-byte write transfer and [Figure 6-69](#) illustrates the single-byte read transfer.

Table 6-44. SPI Command Word

BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
ADDR(6)	ADDR(5)	ADDR(4)	ADDR(3)	ADDR(2)	ADDR(1)	ADDR(0)	R/WZ

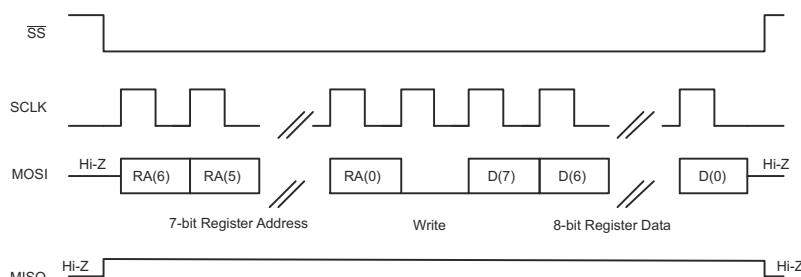


Figure 6-68. SPI Single-Byte Write Transfer

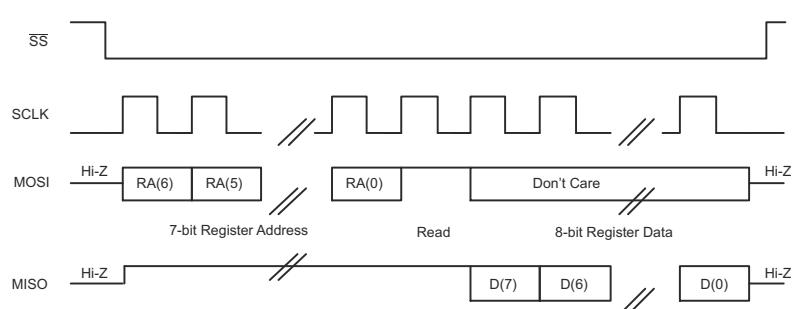


Figure 6-69. SPI Single-Byte Read Transfer

7 Register Maps

This section describes the control registers for the device in detail. All these registers are eight bits in width and allocated to device configuration and programmable coefficients settings. These registers are mapped internally using a page scheme that can be controlled using either I²C or SPI communication to the device. Each page contains 128 bytes of registers. All device configuration registers are stored in page 0, which is the default page setting at power up (and after a software reset). All programmable coefficient registers are located in page 2, page 3, and page 4. The device current page can be switch to a new desired page by using the PAGE[7:0] bits located in register 0 of every page.

Do not read from or write to reserved pages or reserved registers. Write only default values for the reserved bits in the valid registers.

The procedure for register access across pages is:

- Select page N (write data *N* to register 0 regardless of the current page number)
- Read or write data from or to valid registers in page N
- Select the new page M (write data *M* to register 0 regardless of the current page number)
- Read or write data from or to valid registers in page M
- Repeat as needed

7.1 Device Configuration Registers

This section describes the device configuration registers for page 0.

7.1.1 Register Summary Table Page=0x00

ADDRESS	REGISTER	DESCRIPTION	SECTION
0x00	PAGE_CFG	Device page register	PAGE_CFG Register (P0_R0)
0x01	SW_RESET	Software reset register	SW_RESET Register (P0_R1)
0x02	SLEEP_CFG	Sleep mode register	SLEEP_CFG Register (P0_R2)
0x05	SHDN_CFG	Shutdown configuration register	SHDN_CFG Register (P0_R5)
0x07	ASI_CFG0	ASI configuration register 0	ASI_CFG0 Register (P0_R7)
0x08	ASI_CFG1	ASI configuration register 1	ASI_CFG1 Register (P0_R8)
0x09	ASI_CFG2	ASI configuration register 2	ASI_CFG2 Register (P0_R9)
0x0B	ASI_CH1	Channel 1 ASI slot configuration register	ASI_CH1 Register (P0_R11)
0x0C	ASI_CH2	Channel 2 ASI slot configuration register	ASI_CH2 Register (P0_R12)
0x0D	ASI_CH3	Channel 3 ASI slot configuration register	ASI_CH3 Register (P0_R13)
0x0E	ASI_CH4	Channel 4 ASI slot configuration register	ASI_CH4 Register (P0_R14)
0x0F	ASI_CH5	Channel 5 ASI slot configuration register	ASI_CH5 Register (P0_R15)
0x10	ASI_CH6	Channel 6 ASI slot configuration register	ASI_CH6 Register (P0_R16)
0x11	ASI_CH7	Channel 7 ASI slot configuration register	ASI_CH7 Register (P0_R17)
0x12	ASI_CH8	Channel 8 ASI slot configuration register	ASI_CH8 Register (P0_R18)
0x13	MST_CFG0	ASI master mode configuration register 0	MST_CFG0 Register (P0_R19)
0x14	MST_CFG1	ASI master mode configuration register 1	MST_CFG1 Register (P0_R20)
0x15	ASI_STS	ASI bus clock monitor status register	ASI_STS Register (P0_R21)
0x16	CLK_SRC	Clock source configuration register 0	CLK_SRC Register (P0_R22)
0x1F	PDMCLK_CFG	PDM clock generation configuration register	PDMCLK_CFG Register (P0_R31)
0x20	PDMIN_CFG	PDM DINx sampling edge register	PDMIN_CFG Register (P0_R32)
0x21	GPIO_CFG0	GPIO configuration register 0	GPIO_CFG0 Register (P0_R33)
0x22	GPO_CFG0	GPO configuration register 0	GPO_CFG0 Register (P0_R34)
0x23	GPO_CFG1	GPO configuration register 1	GPO_CFG1 Register (P0_R35)
0x24	GPO_CFG2	GPO configuration register 2	GPO_CFG2 Register (P0_R36)
0x25	GPO_CFG3	GPO configuration register 3	GPO_CFG3 Register (P0_R37)
0x29	GPO_VAL	GPIO, GPO output value register	GPO_VAL Register (P0_R41)
0x2A	GPIO_MON	GPIO monitor value register	GPIO_MON Register (P0_R42)
0x2B	GPI_CFG0	GPI configuration register 0	GPI_CFG0 Register (P0_R43)

0x2C	GPI_CFG1	GPI configuration register 1	GPI_CFG1 Register (P0_R44)
0x2F	GPI_MON	GPI monitor value register	GPI_MON Register (P0_R47)
0x32	INT_CFG	Interrupt configuration register	INT_CFG Register (P0_R50)
0x33	INT_MASK0	Interrupt mask register 0	INT_MASK0 Register (P0_R51)
0x36	INT_LTCH0	Latched interrupt readback register 0	INT_LTCH0 Register (P0_R54)
0x3B	BIAS_CFG	MICBIAS and VREF configuration register	BIAS_CFG Register (P0_R59)
0x3C	CH1_CFG0	Channel 1 configuration register 0	CH1_CFG0 Register (P0_R60)
0x3E	CH1_CFG2	Channel 1 configuration register 2	CH1_CFG2 Register (P0_R62)
0x3F	CH1_CFG3	Channel 1 configuration register 3	CH1_CFG3 Register (P0_R63)
0x40	CH1_CFG4	Channel 1 configuration register 4	CH1_CFG4 Register (P0_R64)
0x41	CH2_CFG0	Channel 2 configuration register 0	CH2_CFG0 Register (P0_R65)
0x43	CH2_CFG2	Channel 2 configuration register 2	CH2_CFG2 Register (P0_R67)
0x44	CH2_CFG3	Channel 2 configuration register 3	CH2_CFG3 Register (P0_R68)
0x45	CH2_CFG4	Channel 2 configuration register 4	CH2_CFG4 Register (P0_R69)
0x46	CH3_CFG0	Channel 3 configuration register 0	CH3_CFG0 Register (P0_R70)
0x48	CH3_CFG2	Channel 3 configuration register 2	CH3_CFG2 Register (P0_R72)
0x49	CH3_CFG3	Channel 3 configuration register 3	CH3_CFG3 Register (P0_R73)
0x4A	CH3_CFG4	Channel 3 configuration register 4	CH3_CFG4 Register (P0_R74)
0x4B	CH4_CFG0	Channel 4 configuration register 0	CH4_CFG0 Register (P0_R75)
0x4D	CH4_CFG2	Channel 4 configuration register 2	CH4_CFG2 Register (P0_R77)
0x4E	CH4_CFG3	Channel 4 configuration register 3	CH4_CFG3 Register (P0_R78)
0x4F	CH4_CFG4	Channel 4 configuration register 4	CH4_CFG4 Register (P0_R79)
0x50	CH5_CFG0	Channel 5 configuration register 0	CH5_CFG0 Register (P0_R80)
0x52	CH5_CFG2	Channel 5 configuration register 2	CH5_CFG2 Register (P0_R82)
0x53	CH5_CFG3	Channel 5 configuration register 3	CH5_CFG3 Register (P0_R83)
0x54	CH5_CFG4	Channel 5 configuration register 4	CH5_CFG4 Register (P0_R84)
0x55	CH6_CFG0	Channel 6 configuration register 0	CH6_CFG0 Register (P0_R85)
0x57	CH6_CFG2	Channel 6 configuration register 2	CH6_CFG2 Register (P0_R87)
0x58	CH6_CFG3	Channel 6 configuration register 3	CH6_CFG3 Register (P0_R88)
0x59	CH6_CFG4	Channel 6 configuration register 4	CH6_CFG4 Register (P0_R89)
0x5A	CH7_CFG0	Channel 7 configuration register 0	CH7_CFG0 Register (P0_R90)
0x5C	CH7_CFG2	Channel 7 configuration register 2	CH7_CFG2 Register (P0_R92)
0x5D	CH7_CFG3	Channel 7 configuration register 3	CH7_CFG3 Register (P0_R93)
0x5E	CH7_CFG4	Channel 7 configuration register 4	CH7_CFG4 Register (P0_R94)
0x5F	CH8_CFG0	Channel 8 configuration register 0	CH8_CFG0 Register (P0_R95)
0x61	CH8_CFG2	Channel 8 configuration register 2	CH8_CFG2 Register (P0_R97)
0x62	CH8_CFG3	Channel 8 configuration register 3	CH8_CFG3 Register (P0_R98)
0x63	CH8_CFG4	Channel 8 configuration register 4	CH8_CFG4 Register (P0_R99)
0x6B	DSP_CFG0	DSP configuration register 0	DSP_CFG0 Register (P0_R107)
0x6C	DSP_CFG1	DSP configuration register 1	DSP_CFG1 Register (P0_R108)
0x73	IN_CH_EN	Input channel enable configuration register	IN_CH_EN Register (P0_R115)
0x74	ASI_OUT_CH_EN	ASI output channel enable configuration register	ASI_OUT_CH_EN Register (P0_R116)
0x75	PWR_CFG	Power up configuration register	PWR_CFG Register (P0_R117)
0x76	DEV_STS0	Device status value register 0	DEV_STS0 Register (P0_R118)
0x77	DEV_STS1	Device status value register 1	DEV_STS1 Register (P0_R119)
0x7E	I2C_CKSUM	I2C checksum register	I2C_CKSUM Register (P0_R126)

Table 45 lists the access codes used for the PCMD3180 registers.

Table 7-1. PCMD3180 Access Type Codes

Access Type	Code	Description
Read Type		
R	R	Read
R-W	R/W	Read or write
Write Type		
W	W	Write
Reset or Default Value		
-n		Value after reset or the default value

7.1.2 Register Descriptions

7.1.2.1 PAGE_CFG Register (page = 0x00, address = 0x00) [reset = 0h]

The device memory map is divided into pages. This register sets the page.

Figure 7-1. PAGE_CFG Register

7	6	5	4	3	2	1	0
PAGE[7:0]							
R/W-0h							

Table 7-2. PAGE_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	PAGE[7:0]	R/W	0h	These bits set the device page. 0d = Page 0 1d = Page 1 ... 255d = Page 255

7.1.2.2 SW_RESET Register (page = 0x00, address = 0x01) [reset = 0h]

This register is the software reset register. Asserting a software reset places all register values in the default power-on-reset (POR) state.

Figure 7-2. SW_RESET Register

7	6	5	4	3	2	1	0
Reserved							
R-0h							

Table 7-3. SW_RESET Register Field Descriptions

Bit	Field	Type	Reset	Description
7-1	Reserved	R	0h	Reserved
0	SW_RESET	R/W	0h	Software reset. This bit is self clearing. 0d = Do not reset 1d = Reset

7.1.2.3 SLEEP_CFG Register (page = 0x00, address = 0x02) [reset = 0h]

This register configures the regulator, VREF quick charge, I²C broadcast and sleep mode.

Figure 7-3. SLEEP_CFG Register

7	6	5	4	3	2	1	0
AREG_SELEC_T	Reserved		VREF_QCHG[1:0]		I2C_BRDCAST_EN	Reserved	SLEEP_ENZ
R/W-0h	R/W-0h		R/W-0h		R/W-0h	R-0h	R/W-0h

Table 7-4. SLEEP_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7	AREG_SELECT	R/W	0h	The analog supply selection from either the internal regulator supply or the external AREG supply. 0d = External 1.8V AREG supply (use this setting when AVDD is 1.8V and short AREG with AVDD) 1d = Internally generated 1.8V AREG supply using an on-chip regulator (use this setting when AVDD is 3.3V)
6-5	Reserved	R/W	0h	Reserved
4-3	VREF_QCHG[1:0]	R/W	0h	The duration of the quick-charge for the VREF external capacitor is set using an internal series impedance of 200ohm. 0d = VREF quick-charge duration of 3.5ms (typical) 1d = VREF quick-charge duration of 10ms (typical) 2d = VREF quick-charge duration of 50ms (typical) 3d = VREF quick-charge duration of 100ms (typical)
2	I2C_BRDCAST_EN	R/W	0h	I ² C broadcast addressing setting. 0d = I ² C broadcast mode disabled; the I ² C slave address is determined based on the ADDR pins 1d = I ² C broadcast mode enabled; the I ² C slave address is fixed at 1001 100
1	Reserved	R	0h	Reserved
0	SLEEP_ENZ	R/W	0h	Sleep mode setting. 0d = Device is in sleep mode 1d = Device is not in sleep mode

7.1.2.4 SHDN_CFG Register (page = 0x00, address = 0x05) [reset = 5h]

This register configures the device shutdown

Figure 7-4. SHDN_CFG Register

7	6	5	4	3	2	1	0
Reserved		Reserved			SHDNZ_CFG[1:0]	DREG_KA_TIME[1:0]	
R-0h		R/W-0h			R/W-1h	R/W-1h	

Table 7-5. SHDN_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7-6	Reserved	R	0h	Reserved
5-4	Reserved	R/W	0h	Reserved
3-2	SHDNZ_CFG[1:0]	R/W	1h	Shutdown configuration. 0d = DREG is powered down immediately after SHDNZ asserts 1d = DREG remains active to enable a clean shut down until a time-out is reached; after the time-out period, DREG is forced to power off 2d = DREG remains active until the device cleanly shuts down 3d = Reserved
1-0	DREG_KA_TIME[1:0]	R/W	1h	These bits set how long DREG remains active after SHDNZ asserts. 0d = DREG remains active for 30ms (typical) 1d = DREG remains active for 25ms (typical) 2d = DREG remains active for 10ms (typical) 3d = DREG remains active for 5ms (typical)

7.1.2.5 ASI_CFG0 Register (page = 0x00, address = 0x07) [reset = 30h]

This register is the ASI configuration register 0.

Figure 7-5. ASI_CFG0 Register

7	6	5	4	3	2	1	0
ASI_FORMAT[1:0]		ASI_WLEN[1:0]			FSYNC_POL	BCLK_POL	TX_EDGE
R/W-0h		R/W-3h			R/W-0h	R/W-0h	R/W-0h

Table 7-6. ASI_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-6	ASI_FORMAT[1:0]	R/W	0h	ASI protocol format. 0d = TDM mode 1d = I ² S mode 2d = LJ (left-justified) mode 3d = Reserved
5-4	ASI_WLEN[1:0]	R/W	3h	ASI word or slot length. 0d = 16 bits 1d = 20 bits 2d = 24 bits 3d = 32 bits
3	FSYNC_POL	R/W	0h	ASI FSYNC polarity. 0d = Default polarity as per standard protocol 1d = Inverted polarity with respect to standard protocol
2	BCLK_POL	R/W	0h	ASI BCLK polarity. 0d = Default polarity as per standard protocol 1d = Inverted polarity with respect to standard protocol
1	TX_EDGE	R/W	0h	ASI data output (on the primary and secondary data pin) transmit edge. 0d = Default edge as per the protocol configuration setting in bit 2 (BCLK_POL) 1d = Inverted following edge (half cycle delay) with respect to the default edge setting
0	TX_FILL	R/W	0h	ASI data output (on the primary and secondary data pin) for any unused cycles 0d = Always transmit 0 for unused cycles 1d = Always use Hi-Z for unused cycles

7.1.2.6 ASI_CFG1 Register (page = 0x00, address = 0x08) [reset = 0h]

This register is the ASI configuration register 1.

Figure 7-6. ASI_CFG1 Register

7	6	5	4	3	2	1	0
TX_LSB	TX_KEEPER[1:0]		TX_OFFSET[4:0]				
R/W-0h	R/W-0h		R/W-0h				

Table 7-7. ASI_CFG1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	TX_LSB	R/W	0h	ASI data output (on the primary and secondary data pin) for LSB transmissions. 0d = Transmit the LSB for a full cycle 1d = Transmit the LSB for the first half cycle and Hi-Z for the second half cycle
6-5	TX_KEEPER[1:0]	R/W	0h	ASI data output (on the primary and secondary data pin) bus keeper. 0d = Bus keeper is always disabled 1d = Bus keeper is always enabled 2d = Bus keeper is enabled during LSB transmissions only for one cycle 3d = Bus keeper is enabled during LSB transmissions only for one and half cycles
4-0	TX_OFFSET[4:0]	R/W	0h	ASI data MSB slot 0 offset (on the primary and secondary data pin). 0d = ASI data MSB location has no offset and is as per standard protocol 1d = ASI data MSB location (TDM mode is slot 0 or I ² S, LJ mode is the left and right slot 0) offset of one BCLK cycle with respect to standard protocol 2d = ASI data MSB location (TDM mode is slot 0 or I ² S, LJ mode is the left and right slot 0) offset of two BCLK cycles with respect to standard protocol 3d to 30d = ASI data MSB location (TDM mode is slot 0 or I ² S, LJ mode is the left and right slot 0) offset assigned as per configuration 31d = ASI data MSB location (TDM mode is slot 0 or I ² S, LJ mode is the left and right slot 0) offset of 31 BCLK cycles with respect to standard protocol

7.1.2.7 ASI_CFG2 Register (page = 0x00, address = 0x09) [reset = 0h]

This register is the ASI configuration register 2.

Figure 7-7. ASI_CFG2 Register

7	6	5	4	3	2	1	0
ASI_DAISY	Reserved	ASI_ERR	ASI_ERR_RCOV			Reserved	
R/W-0h	R-0h	R/W-0h	R/W-0h			R-0h	

Table 7-8. ASI_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	ASI_DAISY	R/W	0h	ASI daisy chain connection. 0d = All devices are connected in the common ASI bus 1d = All devices are daisy-chained for the ASI bus
6	Reserved	R	0h	Reserved
5	ASI_ERR	R/W	0h	ASI bus error detection. 0d = Enable bus error detection 1d = Disable bus error detection
4	ASI_ERR_RCOV	R/W	0h	ASI bus error auto resume. 0d = Enable auto resume after bus error recovery 1d = Disable auto resume after bus error recovery and remain powered down until the host configures the device
3-0	Reserved	R	0h	Reserved

7.1.2.8 ASI_CH1 Register (page = 0x00, address = 0x0B) [reset = 0h]

This register is the ASI slot configuration register for channel 1.

Figure 7-8. ASI_CH1 Register

7	6	5	4	3	2	1	0
Reserved	CH1_OUTPUT			CH1_SLOT[5:0]			
R-0h	R/W-0h			R/W-0h			

Table 7-9. ASI_CH1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH1_OUTPUT	R/W	0h	Channel 1 output line. 0d = Channel 1 output is on the ASI primary output pin (SDOUT) 1d = Channel 1 output is on the ASI secondary output pin (GPIO1 or GPOx)
5-0	CH1_SLOT[5:0]	R/W	0h	Channel 1 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.9 ASI_CH2 Register (page = 0x00, address = 0x0C) [reset = 1h]

This register is the ASI slot configuration register for channel 2.

Figure 7-9. ASI_CH2 Register

7	6	5	4	3	2	1	0
Reserved	CH2_OUTPUT			CH2_SLOT[5:0]			
R-0h	R/W-0h			R/W-1h			

Table 7-10. ASI_CH2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH2_OUTPUT	R/W	0h	Channel 2 output line. 0d = Channel 2 output is on the ASI primary output pin (SDOUT) 1d = Channel 2 output is on the ASI secondary output pin (GPIO1 or GPOx)
5-0	CH2_SLOT[5:0]	R/W	1h	Channel 2 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.10 ASI_CH3 Register (page = 0x00, address = 0x0D) [reset = 2h]

This register is the ASI slot configuration register for channel 3.

Figure 7-10. ASI_CH3 Register

7	6	5	4	3	2	1	0
Reserved	CH3_OUTPUT			CH3_SLOT[5:0]			
R-0h	R/W-0h			R/W-2h			

Table 7-11. ASI_CH3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH3_OUTPUT	R/W	0h	Channel 3 output line. 0d = Channel 3 output is on the ASI primary output pin (SDOUT) 1d = Channel 3 output is on the ASI secondary output pin (GPIO1 or GPOx)
5-0	CH3_SLOT[5:0]	R/W	2h	Channel 3 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.11 ASI_CH4 Register (page = 0x00, address = 0x0E) [reset = 3h]

This register is the ASI slot configuration register for channel 4.

Figure 7-11. ASI_CH4 Register

7	6	5	4	3	2	1	0
Reserved	CH4_OUTPUT			CH4_SLOT[5:0]			
R-0h	R/W-0h			R/W-3h			

Table 7-12. ASI_CH4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH4_OUTPUT	R/W	0h	Channel 4 output line. 0d = Channel 4 output is on the ASI primary output pin (SDOUT) 1d = Channel 4 output is on the ASI secondary output pin (GPIO1 or GPOx)

Table 7-12. ASI_CH4 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
5-0	CH4_SLOT[5:0]	R/W	3h	Channel 4 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.12 ASI_CH5 Register (page = 0x00, address = 0x0F) [reset = 4h]

This register is the ASI slot configuration register for channel 5.

Figure 7-12. ASI_CH5 Register

7	6	5	4	3	2	1	0
Reserved	CH5_OUTPUT			CH5_SLOT[5:0]			
R-0h	R/W-0h			R/W-4h			

Table 7-13. ASI_CH5 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH5_OUTPUT	R/W	0h	Channel 5 output line. 0d = Channel 5 output is on the ASI primary output pin (SDOUT) 1d = Channel 5 output is on the ASI secondary output pin (GPIO1 or GPOx)
5-0	CH5_SLOT[5:0]	R/W	4h	Channel 5 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.13 ASI_CH6 Register (page = 0x00, address = 0x10) [reset = 5h]

This register is the ASI slot configuration register for channel 6.

Figure 7-13. ASI_CH6 Register

7	6	5	4	3	2	1	0
Reserved	CH6_OUTPUT			CH6_SLOT[5:0]			
R-0h	R/W-0h			R/W-5h			

Table 7-14. ASI_CH6 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH6_OUTPUT	R/W	0h	Channel 6 output line. 0d = Channel 6 output is on the ASI primary output pin (SDOUT) 1d = Channel 6 output is on the ASI secondary output pin (GPIO1 or GPOx)

Table 7-14. ASI_CH6 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
5-0	CH6_SLOT[5:0]	R/W	5h	Channel 6 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.14 ASI_CH7 Register (page = 0x00, address = 0x11) [reset = 6h]

This register is the ASI slot configuration register for channel 7.

Figure 7-14. ASI_CH7 Register

7	6	5	4	3	2	1	0
Reserved	CH7_OUTPUT			CH7_SLOT[5:0]			
R-0h	R/W-0h			R/W-6h			

Table 7-15. ASI_CH7 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH7_OUTPUT	R/W	0h	Channel 7 output line. 0d = Channel 7 output is on the ASI primary output pin (SDOUT) 1d = Channel 7 output is on the ASI secondary output pin (GPIO1 or GPOx)
5-0	CH7_SLOT[5:0]	R/W	6h	Channel 7 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.15 ASI_CH8 Register (page = 0x00, address = 0x12) [reset = 7h]

This register is the ASI slot configuration register for channel 8.

Figure 7-15. ASI_CH8 Register

7	6	5	4	3	2	1	0
Reserved	CH8_OUTPUT			CH8_SLOT[5:0]			
R-0h	R/W-0h			R/W-7h			

Table 7-16. ASI_CH8 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6	CH8_OUTPUT	R/W	0h	Channel 8 output line. 0d = Channel 8 output is on the ASI primary output pin (SDOUT) 1d = Channel 8 output is on the ASI secondary output pin (GPIO1 or GPOx)

Table 7-16. ASI_CH8 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
5-0	CH8_SLOT[5:0]	R/W	7h	Channel 8 slot assignment. 0d = TDM is slot 0 or I ² S, LJ is left slot 0 1d = TDM is slot 1 or I ² S, LJ is left slot 1 2d to 30d = Slot assigned as per configuration 31d = TDM is slot 31 or I ² S, LJ is left slot 31 32d = TDM is slot 32 or I ² S, LJ is right slot 0 33d = TDM is slot 33 or I ² S, LJ is right slot 1 34d to 62d = Slot assigned as per configuration 63d = TDM is slot 63 or I ² S, LJ is right slot 31

7.1.2.16 MST_CFG0 Register (page = 0x00, address = 0x13) [reset = 2h]

This register is the ASI master mode configuration register 0.

Figure 7-16. MST_CFG0 Register

7	6	5	4	3	2	1	0
MST_SLV_CFG	AUTO_CLK_CFG	AUTO_MODE_PLL_DIS	BCLK_FSYNC_GATE	FS_MODE			MCLK_FREQ_SEL[2:0]
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h			R/W-2h

Table 7-17. MST_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	MST_SLV_CFG	R/W	0h	ASI master or slave configuration register setting. 0d = Device is in slave mode (both BCLK and FSYNC are inputs to the device) 1d = Device is in master mode (both BCLK and FSYNC are generated from the device)
6	AUTO_CLK_CFG	R/W	0h	Automatic clock configuration setting. 0d = Auto clock configuration is enabled (all internal clock divider and PLL configurations are auto derived) 1d = Auto clock configuration is disabled (custom mode and device GUI must be used for the device configuration settings)
5	AUTO_MODE_PLL_DIS	R/W	0h	Automatic mode PLL setting. 0d = PLL is enabled in auto clock configuration 1d = PLL is disabled in auto clock configuration
4	BCLK_FSYNC_GATE	R/W	0h	BCLK and FSYNC clock gate (valid when the device is in master mode). 0d = Do not gate BCLK and FSYNC 1d = Force gate BCLK and FSYNC when being transmitted from the device in master mode
3	FS_MODE	R/W	0h	Sample rate setting (valid when the device is in master mode). 0d = fS is a multiple (or submultiple) of 48kHz 1d = fS is a multiple (or submultiple) of 44.1kHz
2-0	MCLK_FREQ_SEL[2:0]	R/W	2h	These bits select the MCLK (GPIO or GPIOx) frequency for the PLL source clock input (valid when the device is in master mode and MCLK_FREQ_SEL_MODE = 0). 0d = 12MHz 1d = 12.288MHz 2d = 13MHz 3d = 16MHz 4d = 19.2MHz 5d = 19.68MHz 6d = 24MHz 7d = 24.576MHz

7.1.2.17 MST_CFG1 Register (page = 0x00, address = 0x14) [reset = 48h]

This register is the ASI master mode configuration register 1.

Figure 7-17. MST_CFG1 Register

7	6	5	4	3	2	1	0
FS_RATE[3:0]				FS_BCLK_RATIO[3:0]			
R/W-4h				R/W-8h			

Table 7-18. MST_CFG1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	FS_RATE[3:0]	R/W	4h	Programmed sample rate of the ASI bus (not used when the device is configured in slave mode auto clock configuration). 0d = 7.35kHz or 8kHz 1d = 14.7kHz or 16kHz 2d = 22.05kHz or 24kHz 3d = 29.4kHz or 32kHz 4d = 44.1kHz or 48kHz 5d = 88.2kHz or 96kHz 6d = 176.4kHz or 192kHz 7d = 352.8kHz or 384kHz 8d = 705.6kHz or 768kHz 9d to 15d = Reserved
3-0	FS_BCLK_RATIO[3:0]	R/W	8h	Programmed BCLK to FSYNC frequency ratio of the ASI bus (not used when the device is configured in slave mode auto clock configuration). 0d = Ratio of 16 1d = Ratio of 24 2d = Ratio of 32 3d = Ratio of 48 4d = Ratio of 64 5d = Ratio of 96 6d = Ratio of 128 7d = Ratio of 192 8d = Ratio of 256 9d = Ratio of 384 10d = Ratio of 512 11d = Ratio of 1024 12d = Ratio of 2048 13d to 15d = Reserved

7.1.2.18 ASI_STS Register (page = 0x00, address = 0x15) [reset = FFh]

This register is the ASI bus clock monitor status register

Figure 7-18. ASI_STS Register

7	6	5	4	3	2	1	0
FS_RATE_STS[3:0]				FS_RATIO_STS[3:0]			
R-Fh				R-Fh			

Table 7-19. ASI_STS Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	FS_RATE_STS[3:0]	R	Fh	Detected sample rate of the ASI bus. 0d = 7.35kHz or 8kHz 1d = 14.7kHz or 16kHz 2d = 22.05kHz or 24kHz 3d = 29.4kHz or 32kHz 4d = 44.1kHz or 48kHz 5d = 88.2kHz or 96kHz 6d = 176.4kHz or 192kHz 7d = 352.8kHz or 384kHz 8d = 705.6kHz or 768kHz 9d to 14d = Reserved 15d = Invalid sample rate

Table 7-19. ASI_STS Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
3-0	FS_RATIO_STS[3:0]	R	Fh	Detected BCLK to FSYNC frequency ratio of the ASI bus. 0d = Ratio of 16 1d = Ratio of 24 2d = Ratio of 32 3d = Ratio of 48 4d = Ratio of 64 5d = Ratio of 96 6d = Ratio of 128 7d = Ratio of 192 8d = Ratio of 256 9d = Ratio of 384 10d = Ratio of 512 11d = Ratio of 1024 12d = Ratio of 2048 13d to 14d = Reserved 15d = Invalid ratio

7.1.2.19 CLK_SRC Register (page = 0x00, address = 0x16) [reset = 10h]

This register is the clock source configuration register.

Figure 7-19. CLK_SRC Register

7	6	5	4	3	2	1	0
DIS_PLL_SLV_CLK_SRC	MCLK_FREQ_SEL_MODE	MCLK_RATIO_SEL[2:0]				Reserved	
R/W-0h	R/W-0h	R/W-2h				R-0h	

Table 7-20. CLK_SRC Register Field Descriptions

Bit	Field	Type	Reset	Description
7	DIS_PLL_SLV_CLK_SRC	R/W	0h	Audio root clock source setting when the device is configured with the PLL disabled in the auto clock configuration for slave mode (AUTO_MODE_PLL_DIS = 1). 0d = BCLK is used as the audio root clock source 1d = MCLK (GPIO or GPIx) is used as the audio root clock source (the MCLK to FSYNC ratio is as per MCLK_RATIO_SEL setting)
6	MCLK_FREQ_SEL_MODE	R/W	0h	Master mode MCLK (GPIO or GPIx) frequency selection mode (valid when the device is in auto clock configuration). 0d = MCLK frequency is based on the MCLK_FREQ_SEL (P0_R19) configuration 1d = MCLK frequency is specified as a multiple of FSYNC in the MCLK_RATIO_SEL (P0_R22) configuration
5-3	MCLK_RATIO_SEL[2:0]	R/W	2h	These bits select the MCLK (GPIO or GPIx) to FSYNC ratio for master mode or when MCLK is used as the audio root clock source in slave mode. 0d = Ratio of 64 1d = Ratio of 256 2d = Ratio of 384 3d = Ratio of 512 4d = Ratio of 768 5d = Ratio of 1024 6d = Ratio of 1536 7d = Ratio of 2304
2-0	Reserved	R	0h	Reserved

7.1.2.20 PDMCLK_CFG Register (page = 0x00, address = 0x1F) [reset = 40h]

This register is the PDM clock generation configuration register.

Figure 7-20. PDMCLK_CFG Register

7	6	5	4	3	2	1	0
Reserved	Reserved				PDMCLK_DIV[1:0]		
R/W-0h	R/W-10h				R/W-0h		

Table 7-21. PDMCLK_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R/W	0h	Reserved
6-2	Reserved	R/W	10h	Reserved
1-0	PDMCLK_DIV[1:0]	R/W	0h	PDMCLK divider value. 0d = PDMCLK is 2.8224MHz or 3.072MHz 1d = PDMCLK is 1.4112MHz or 1.536MHz 2d = PDMCLK is 705.6kHz or 768kHz 3d = PDMCLK is 5.6448MHz or 6.144MHz

7.1.2.21 PDMIN_CFG Register (page = 0x00, address = 0x20) [reset = 0h]

This register is the PDM DINx sampling edge configuration register.

Figure 7-21. PDMIN_CFG Register

7	6	5	4	3	2	1	0
PDMDIN1_EDGE	PDMDIN2_EDGE	PDMDIN3_EDGE	PDMDIN4_EDGE			Reserved	
R/W-0h	R/W-0h	R/W-0h	R/W-0h			R-0h	

Table 7-22. PDMIN_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7	PDMDIN1_EDGE	R/W	0h	PDMCLK latching edge used for channel 1 and channel 2 data. 0d = Channel 1 data are latched on the negative edge, channel 2 data are latched on the positive edge 1d = Channel 1 data are latched on the positive edge, channel 2 data are latched on the negative edge
6	PDMDIN2_EDGE	R/W	0h	PDMCLK latching edge used for channel 3 and channel 4 data. 0d = Channel 3 data are latched on the negative edge, channel 4 data are latched on the positive edge 1d = Channel 3 data are latched on the positive edge, channel 4 data are latched on the negative edge
5	PDMDIN3_EDGE	R/W	0h	PDMCLK latching edge used for channel 5 and channel 6 data. 0d = Channel 5 data are latched on the negative edge, channel 6 data are latched on the positive edge 1d = Channel 5 data are latched on the positive edge, channel 6 data are latched on the negative edge
4	PDMDIN4_EDGE	R/W	0h	PDMCLK latching edge used for channel 7 and channel 8 data. 0d = Channel 7 data are latched on the negative edge, channel 8 data are latched on the positive edge 1d = Channel 7 data are latched on the positive edge, channel 8 data are latched on the negative edge
3-0	Reserved	R	0h	Reserved

7.1.2.22 GPIO_CFG0 Register (page = 0x00, address = 0x21) [reset = 22h]

This register is the GPIO configuration register 0.

Figure 7-22. GPIO_CFG0 Register

7	6	5	4	3	2	1	0	
GPIO1_CFG[3:0]				Reserved	GPIO1_DRV[2:0]			
R/W-2h				R-0h	R/W-2h			

Table 7-23. GPIO_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	GPIO1_CFG[3:0]	R/W	2h	GPIO1 configuration. 0d = GPIO1 is disabled 1d = GPIO1 is configured as a general-purpose output (GPO) 2d = GPIO1 is configured as a device interrupt output (IRQ) 3d = GPIO1 is configured as a secondary ASI output (SDOUT2) 4d = GPIO1 is configured as a PDM clock output (PDMCLK) 5d to 7d = Reserved 8d = GPIO1 is configured as an input to control when MICBIAS turns on or off (MICBIAS_EN) 9d = GPIO1 is configured as a general-purpose input (GPI) 10d = GPIO1 is configured as a master clock input (MCLK) 11d = GPIO1 is configured as an ASI input for daisy-chain (SDIN) 12d = GPIO1 is configured as a PDM data input for channel 1 and channel 2 (PDMDIN1) 13d = GPIO1 is configured as a PDM data input for channel 3 and channel 4 (PDMDIN2) 14d = GPIO1 is configured as a PDM data input for channel 5 and channel 6 (PDMDIN3) 15d = GPIO1 is configured as a PDM data input for channel 7 and channel 8 (PDMDIN4)
3	Reserved	R	0h	Reserved
2-0	GPIO1_DRV[2:0]	R/W	2h	GPIO1 output drive configuration (not used when GPIO1 is configured as SDOUT2). 0d = Hi-Z output 1d = Drive active low and active high 2d = Drive active low and weak high 3d = Drive active low and Hi-Z 4d = Drive weak low and active high 5d = Drive Hi-Z and active high 6d to 7d = Reserved

7.1.2.23 GPO_CFG0 Register (page = 0x00, address = 0x22) [reset = 0h]

This register is the GPO configuration register 0.

Figure 7-23. GPO_CFG0 Register

7	6	5	4	3	2	1	0
GPO1_CFG[3:0]				Reserved			
R/W-0h				R-0h			

Table 7-24. GPO_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	GPO1_CFG[3:0]	R/W	0h	PDMCLK1_GPO1 (GPO1) configuration. 0d = GPO1 is disabled 1d = GPO1 is configured as a general-purpose output (GPO) 2d = GPO1 is configured as a device interrupt output (IRQ) 3d = GPO1 is configured as a secondary ASI output (SDOUT2) 4d = GPO1 is configured as a PDM clock output (PDMCLK) 5d to 15d = Reserved
3-1	Reserved	R	0h	Reserved
0	GPO1_DRV	R/W	0h	0d = GPO1 is in a Hi-Z state 1d = GPO1 is driven as Active High/Active Low

7.1.2.24 GPO_CFG1 Register (page = 0x00, address = 0x23) [reset = 0h]

This register is the GPO configuration register 1.

Figure 7-24. GPO_CFG1 Register

7	6	5	4	3	2	1	0
GPO2_CFG[3:0]				Reserved			

Figure 7-24. GPO_CFG1 Register (continued)

R/W-0h	R-0h
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Table 7-25. GPO_CFG1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	GPO2_CFG[3:0]	R/W	0h	PDMCLK2_GPO2 (GPO2) configuration. 0d = GPO2 is disabled 1d = GPO2 is configured as a general-purpose output (GPO) 2d = GPO2 is configured as a device interrupt output (IRQ) 3d = GPO2 is configured as a secondary ASI output (SDOUT2) 4d = GPO2 is configured as a PDM clock output (PDMCLK) 5d to 15d = Reserved
3-1	Reserved	R	0h	Reserved
0	GPO2_DRV	R/W	0h	0d = GPO2 is in a Hi-Z state 1d = GPO2 is driven as Active High/Active Low

7.1.2.25 GPO_CFG2 Register (page = 0x00, address = 0x24) [reset = 0h]

This register is the GPO configuration register 2.

Figure 7-25. GPO_CFG2 Register

7	6	5	4	3	2	1	0
GPO3_CFG[3:0]				Reserved			
R/W-0h				R-0h			

Table 7-26. GPO_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	GPO3_CFG[3:0]	R/W	0h	PDMCLK3_GPO3 (GPO3) configuration. 0d = GPO3 is disabled 1d = GPO3 is configured as a general-purpose output (GPO) 2d = GPO3 is configured as a device interrupt output (IRQ) 3d = GPO3 is configured as a secondary ASI output (SDOUT2) 4d = GPO3 is configured as a PDM clock output (PDMCLK) 5d to 15d = Reserved
3-1	Reserved	R	0h	Reserved
0	GPO3_DRV	R/W	0h	0d = GPO3 is in a Hi-Z state 1d = GPO3 is driven as Active High/Active Low

7.1.2.26 GPO_CFG3 Register (page = 0x00, address = 0x25) [reset = 0h]

This register is the GPO configuration register 3.

Figure 7-26. GPO_CFG3 Register

7	6	5	4	3	2	1	0
GPO4_CFG[3:0]				Reserved			
R/W-0h				R-0h			

Table 7-27. GPO_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	GPO4_CFG[3:0]	R/W	0h	PDMCLK4_GPO4 (GPO4) configuration. 0d = GPO4 is disabled 1d = GPO4 is configured as a general-purpose output (GPO) 2d = GPO4 is configured as a device interrupt output (IRQ) 3d = GPO4 is configured as a secondary ASI output (SDOUT2) 4d = GPO4 is configured as a PDM clock output (PDMCLK) 5d to 15d = Reserved
3-1	Reserved	R	0h	Reserved

Table 7-27. GPO_CFG3 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
0	GPO4_DRV	R/W	0h	0d = GPO4 is in a Hi-Z state 1d = GPO4 is driven as Active High/Active Low

7.1.2.27 GPO_VAL Register (page = 0x00, address = 0x29) [reset = 0h]

This register is the GPIO and GPO output value register.

Figure 7-27. GPO_VAL Register

7	6	5	4	3	2	1	0
GPIO1_VAL	GPO1_VAL	GPO2_VAL	GPO3_VAL	GPO4_VAL		Reserved	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h		R-0h	

Table 7-28. GPO_VAL Register Field Descriptions

Bit	Field	Type	Reset	Description
7	GPIO1_VAL	R/W	0h	GPIO1 output value when configured as a GPO. 0d = Drive the output with a value of 0 1d = Drive the output with a value of 1
6	GPO1_VAL	R/W	0h	GPO1 output value when configured as a GPO. 0d = Drive the output with a value of 0 1d = Drive the output with a value of 1
5	GPO2_VAL	R/W	0h	GPO2 output value when configured as a GPO. 0d = Drive the output with a value of 0 1d = Drive the output with a value of 1
4	GPO3_VAL	R/W	0h	GPO3 output value when configured as a GPO. 0d = Drive the output with a value of 0 1d = Drive the output with a value of 1
3	GPO4_VAL	R/W	0h	GPO4 output value when configured as a GPO. 0d = Drive the output with a value of 0 1d = Drive the output with a value of 1
2-0	Reserved	R	0h	Reserved

7.1.2.28 GPIO_MON Register (page = 0x00, address = 0x2A) [reset = 0h]

This register is the GPIO monitor value register.

Figure 7-28. GPIO_MON Register

7	6	5	4	3	2	1	0
GPIO1_MON					Reserved		
R-0h					R-0h		

Table 7-29. GPIO_MON Register Field Descriptions

Bit	Field	Type	Reset	Description
7	GPIO1_MON	R	0h	GPIO1 monitor value when configured as a GPI. 0d = Input monitor value 0 1d = Input monitor value 1
6-0	Reserved	R	0h	Reserved

7.1.2.29 GPI_CFG0 Register (page = 0x00, address = 0x2B) [reset = 0h]

This register is the GPI configuration register 0.

Figure 7-29. GPI_CFG0 Register

7	6	5	4	3	2	1	0
Reserved			GPI1_CFG[2:0]	Reserved		GPI2_CFG[2:0]	
R-0h			R/W-0h	R-0h		R/W-0h	

Table 7-30. GPI_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-4	GPI1_CFG[2:0]	R/W	0h	PDMDIN1_GPI1 (GPI1) configuration. 0d = GPI1 is disabled 1d = GPI1 is configured as a general-purpose input (GPI) 2d = GPI1 is configured as a master clock input (MCLK) 3d = GPI1 is configured as an ASI input for daisy-chain (SDIN) 4d = GPI1 is configured as a PDM data input for channel 1 and channel 2 (PDMDIN1) 5d = GPI1 is configured as a PDM data input for channel 3 and channel 4 (PDMDIN2) 6d = GPI1 is configured as a PDM data input for channel 5 and channel 6 (PDMDIN3) 7d = GPI1 is configured as a PDM data input for channel 7 and channel 8 (PDMDIN4)
3	Reserved	R	0h	Reserved
2-0	GPI2_CFG[2:0]	R/W	0h	PDMDIN2_GPI2 (GPI2) configuration. 0d = GPI2 is disabled 1d = GPI2 is configured as a general-purpose input (GPI) 2d = GPI2 is configured as a master clock input (MCLK) 3d = GPI2 is configured as an ASI input for daisy-chain (SDIN) 4d = GPI2 is configured as a PDM data input for channel 1 and channel 2 (PDMDIN1) 5d = GPI2 is configured as a PDM data input for channel 3 and channel 4 (PDMDIN2) 6d = GPI2 is configured as a PDM data input for channel 5 and channel 6 (PDMDIN3) 7d = GPI2 is configured as a PDM data input for channel 7 and channel 8 (PDMDIN4)

7.1.2.30 GPI_CFG1 Register (page = 0x00, address = 0x2C) [reset = 0h]

This register is the GPI configuration register 1.

Figure 7-30. GPI_CFG1 Register

7	6	5	4	3	2	1	0
Reserved	GPI3_CFG[2:0]			Reserved	GPI4_CFG[2:0]		
R-0h	R/W-0h			R-0h	R/W-0h		

Table 7-31. GPI_CFG1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-4	GPI3_CFG[2:0]	R/W	0h	PDMDIN3_GPI3 (GPI3) configuration. 0d = GPI3 is disabled 1d = GPI3 is configured as a general-purpose input (GPI) 2d = GPI3 is configured as a master clock input (MCLK) 3d = GPI3 is configured as an ASI input for daisy-chain (SDIN) 4d = GPI3 is configured as a PDM data input for channel 1 and channel 2 (PDMDIN1) 5d = GPI3 is configured as a PDM data input for channel 3 and channel 4 (PDMDIN2) 6d = GPI3 is configured as a PDM data input for channel 5 and channel 6 (PDMDIN3) 7d = GPI3 is configured as a PDM data input for channel 7 and channel 8 (PDMDIN4)
3	Reserved	R	0h	Reserved

Table 7-31. GPI_CFG1 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
2-0	GPI4_CFG[2:0]	R/W	0h	PDMDIN4_GPI4 (GPI4) configuration. 0d = GPI4 is disabled 1d = GPI4 is configured as a general-purpose input (GPI) 2d = GPI4 is configured as a master clock input (MCLK) 3d = GPI4 is configured as an ASI input for daisy-chain (SDIN) 4d = GPI4 is configured as a PDM data input for channel 1 and channel 2 (PDMDIN1) 5d = GPI4 is configured as a PDM data input for channel 3 and channel 4 (PDMDIN2) 6d = GPI4 is configured as a PDM data input for channel 5 and channel 6 (PDMDIN3) 7d = GPI4 is configured as a PDM data input for channel 7 and channel 8 (PDMDIN4)

7.1.2.31 GPI_MON Register (page = 0x00, address = 0x2F) [reset = 0h]

This register is the GPI monitor value register.

Figure 7-31. GPI_MON Register

7	6	5	4	3	2	1	0
GPI1_MON	GPI2_MON	GPI3_MON	GPI4_MON	Reserved			
R-0h	R-0h	R-0h	R-0h	R-0h			

Table 7-32. GPI_MON Register Field Descriptions

Bit	Field	Type	Reset	Description
7	GPI1_MON	R	0h	GPI1 monitor value when configured as a GPIO. 0d = Input monitor value 0 1d = Input monitor value 1
6	GPI2_MON	R	0h	GPI2 monitor value when configured as a GPIO. 0d = Input monitor value 0 1d = Input monitor value 1
5	GPI3_MON	R	0h	GPI3 monitor value when configured as a GPIO. 0d = Input monitor value 0 1d = Input monitor value 1
4	GPI4_MON	R	0h	GPI4 monitor value when configured as a GPIO. 0d = Input monitor value 0 1d = Input monitor value 1
3-0	Reserved	R	0h	Reserved

7.1.2.32 INT_CFG Register (page = 0x00, address = 0x32) [reset = 0h]

This register is the interrupt configuration register.

Figure 7-32. INT_CFG Register

7	6	5	4	3	2	1	0
INT_POL	INT_EVENT[1:0]		Reserved		LTCH_READ_CFG	Reserved	
R/W-0h	R/W-0h		R-0h		R/W-0h	R-0h	

Table 7-33. INT_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7	INT_POL	R/W	0h	Interrupt polarity. 0d = Active low (IRQZ) 1d = Active high (IRQ)

Table 7-33. INT_CFG Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
6-5	INT_EVENT[1:0]	R/W	0h	Interrupt event configuration. 0d = INT asserts on any unmasked latched interrupts event 1d = Reserved 2d = INT asserts for 2ms (typical) for every 4ms (typical) duration on any unmasked latched interrupts event 3d = INT asserts for 2ms (typical) one time on each pulse for any unmasked interrupts event
4-3	Reserved	R	0h	Reserved
2	LTCH_READ_CFG	R/W	0h	Interrupt latch registers readback configuration. 0d = All interrupts can be read through the LTCH registers 1d = Only unmasked interrupts can be read through the LTCH registers
1-0	Reserved	R	0h	Reserved

7.1.2.33 INT_MASK0 Register (page = 0x00, address = 0x33) [reset = FFh]

This register is the interrupt masks register 0.

Figure 7-33. INT_MASK0 Register

7	6	5	4	3	2	1	0
INT_MASK0[7]	INT_MASK0[6]	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
R/W-1h	R/W-1h	R/W-1h	R/W-1h	R/W-1h	R/W-1h	R/W-1h	R/W-1h

Table 7-34. INT_MASK0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	INT_MASK0[7]	R/W	1h	ASI clock error mask. 0d = Do not mask 1d = Mask
6	INT_MASK0[6]	R/W	1h	PLL Lock interrupt mask. 0d = Do not mask 1d = Mask
5-0	Reserved	R/W	3Fh	Reserved

7.1.2.34 INT_LTCH0 Register (page = 0x00, address = 0x36) [reset = 0h]

This register is the latched Interrupt readback register 0.

Figure 7-34. INT_LTCH0 Register

7	6	5	4	3	2	1	0
INT_LTCH0[7]	INT_LTCH0[6]	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h	R-0h

Table 7-35. INT_LTCH0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	INT_LTCH0[7]	R	0h	Interrupt caused by an ASI bus clock error (self-clearing bit). 0d = No interrupt 1d = Interrupt
6	INT_LTCH0[6]	R	0h	Interrupt caused by PLL LOCK (self-clearing bit). 0d = No interrupt 1d = Interrupt
5-0	Reserved	R	0h	Reserved

7.1.2.35 BIAS_CFG Register (page = 0x00, address = 0x3B) [reset = 0h]

This register is the MICBIAS and VREF configuration register

Figure 7-35. BIAS_CFG Register

7	6	5	4	3	2	1	0
Reserved	MBIAS_VAL[2:0]				Reserved	VREF_SEL[1:0]	
R-0h	R/W-0h				R-0h	R/W-0h	

Table 7-36. BIAS_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-4	MBIAS_VAL[2:0]	R/W	0h	MICBIAS value. 0d = Microphone bias is set to VREF (2.750V, 2.500V, or 1.375V) 1d to 5d = Reserved 2d = Reserved 6d = Microphone bias is set to AVDD 7d = Reserved
3-2	Reserved	R	0h	Reserved
1-0	VREF_SEL[1:0]	R/W	0h	VREF voltage setting (configure this setting based on the AVDD supply minimum voltage used). 0d = VREF is set to 2.75V 1d = VREF is set to 2.5V 2d = VREF is set to 1.375V (this option must be used for 1.8V AVDD) 3d = Reserved

7.1.2.36 CH1_CFG0 Register (page = 0x00, address = 0x3C) [reset = 0h]

This register is configuration register 0 for channel 1.

Figure 7-36. CH1_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH1_INSRC[1:0]		Reserved	Reserved		Reserved	Reserved
R/W-0h	R/W-0h		R/W-0h	R/W-0h		R-0h	R/W-0h

Table 7-37. CH1_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R/W	0h	Reserved
6-5	CH1_INSRC[1:0]	R/W	0h	Channel 1 input configuration. 0d = Input source is not enabled 1d = Reserved 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN1 and PDMCLK) 3d = Reserved
4-0	Reserved	R/W	0h	Reserved

7.1.2.37 CH1_CFG2 Register (page = 0x00, address = 0x3E) [reset = C9h]

This register is configuration register 2 for channel 1.

Figure 7-37. CH1_CFG2 Register

7	6	5	4	3	2	1	0
CH1_DVOL[7:0]							
R/W-C9h							

Table 7-38. CH1_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH1_DVOL[7:0]	R/W	C9h	Channel 1 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.38 CH1_CFG3 Register (page = 0x00, address = 0x3F) [reset = 80h]

This register is configuration register 3 for channel 1.

Figure 7-38. CH1_CFG3 Register

7	6	5	4	3	2	1	0
CH1_GCAL[3:0]				Reserved			
R/W-8h				R-0h			

Table 7-39. CH1_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH1_GCAL[3:0]	R/W	8h	Channel 1 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.39 CH1_CFG4 Register (page = 0x00, address = 0x40) [reset = 0h]

This register is configuration register 4 for channel 1.

Figure 7-39. CH1_CFG4 Register

7	6	5	4	3	2	1	0
CH1_PCAL[7:0]				Reserved			
R/W-0h				R-0h			

Table 7-40. CH1_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH1_PCAL[7:0]	R/W	0h	Channel 1 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.40 CH2_CFG0 Register (page = 0x00, address = 0x41) [reset = 0h]

This register is configuration register 0 for channel 2.

Figure 7-40. CH2_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH2_INSRC[1:0]	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R-0h	R/W-0h	

Table 7-41. CH2_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R/W	0h	Reserved
6-5	CH2_INSRC[1:0]	R/W	0h	Channel 2 input configuration. 0d = Input source is not enabled 1d = Reserved 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN1 and PDMLCK) 3d = Reserved
4-0	Reserved	R/W	0h	Reserved

7.1.2.41 CH2_CFG2 Register (page = 0x00, address = 0x43) [reset = C9h]

This register is configuration register 2 for channel 2.

Figure 7-41. CH2_CFG2 Register

7	6	5	4	3	2	1	0
CH2_DVOL[7:0]							
R/W-C9h							

Table 7-42. CH2_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH2_DVOL[7:0]	R/W	C9h	Channel 2 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.42 CH2_CFG3 Register (page = 0x00, address = 0x44) [reset = 80h]

This register is configuration register 3 for channel 2.

Figure 7-42. CH2_CFG3 Register

7	6	5	4	3	2	1	0
CH2_GCAL[3:0]							
R/W-8h							

Table 7-43. CH2_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH2_GCAL[3:0]	R/W	8h	Channel 2 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.43 CH2_CFG4 Register (page = 0x00, address = 0x45) [reset = 0h]

This register is configuration register 4 for channel 2.

Figure 7-43. CH2_CFG4 Register

7	6	5	4	3	2	1	0
CH2_PCAL[7:0]							
R/W-0h							

Table 7-44. CH2_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH2_PCAL[7:0]	R/W	0h	Channel 2 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.44 CH3_CFG0 Register (page = 0x00, address = 0x46) [reset = 0h]

This register is configuration register 0 for channel 3.

Figure 7-44. CH3_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH3_INSRC[1:0]		Reserved	Reserved		Reserved	Reserved
R/W-0h	R/W-0h		R/W-0h	R/W-0h		R-0h	R/W-0h

Table 7-45. CH3_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R/W	0h	Reserved
6-5	CH3_INSRC[1:0]	R/W	0h	Channel 3 input configuration. 0d = Input source is not enabled 1d = Reserved 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN2 and PDMCLK) 3d = Reserved
4-0	Reserved	R/W	0h	Reserved

7.1.2.45 CH3_CFG2 Register (page = 0x00, address = 0x48) [reset = C9h]

This register is configuration register 2 for channel 3.

Figure 7-45. CH3_CFG2 Register

7	6	5	4	3	2	1	0
CH3_DVOL[7:0]							
R/W-C9h							

Table 7-46. CH3_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH3_DVOL[7:0]	R/W	C9h	Channel 3 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.46 CH3_CFG3 Register (page = 0x00, address = 0x49) [reset = 80h]

This register is configuration register 3 for channel 3.

Figure 7-46. CH3_CFG3 Register

7	6	5	4	3	2	1	0
CH3_GCAL[3:0]				Reserved			
R/W-8h				R-0h			

Table 7-47. CH3_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH3_GCAL[3:0]	R/W	8h	Channel 3 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.47 CH3_CFG4 Register (page = 0x00, address = 0x4A) [reset = 0h]

This register is configuration register 4 for channel 3.

Figure 7-47. CH3_CFG4 Register

7	6	5	4	3	2	1	0
CH3_PCAL[7:0]							
R/W-0h							

Table 7-48. CH3_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH3_PCAL[7:0]	R/W	0h	Channel 3 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.48 CH4_CFG0 Register (page = 0x00, address = 0x4B) [reset = 0h]

This register is configuration register 0 for channel 4.

Figure 7-48. CH4_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH4_INSRC[1:0]	Reserved	Reserved	Reserved	Reserved	Reserved	Reserved
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R-0h	R-0h	R/W-0h

Table 7-49. CH4_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R/W	0h	Reserved
6-5	CH4_INSRC[1:0]	R/W	0h	Channel 4 input configuration. 0d = Input source is not enabled 1d = Reserved 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN2 and PDMLCK) 3d = Reserved
4-0	Reserved	R/W	0h	Reserved

7.1.2.49 CH4_CFG2 Register (page = 0x00, address = 0x4D) [reset = C9h]

This register is configuration register 2 for channel 4.

Figure 7-49. CH4_CFG2 Register

7	6	5	4	3	2	1	0
CH4_DVOL[7:0]							
R/W-C9h							

Table 7-50. CH4_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH4_DVOL[7:0]	R/W	C9h	Channel 4 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.50 CH4_CFG3 Register (page = 0x00, address = 0x4E) [reset = 80h]

This register is configuration register 3 for channel 4.

Figure 7-50. CH4_CFG3 Register

7	6	5	4	3	2	1	0
CH4_GCAL[3:0]				Reserved			
R/W-8h				R-0h			

Table 7-51. CH4_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH4_GCAL[3:0]	R/W	8h	Channel 4 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.51 CH4_CFG4 Register (page = 0x00, address = 0x4F) [reset = 0h]

This register is configuration register 4 for channel 4.

Figure 7-51. CH4_CFG4 Register

7	6	5	4	3	2	1	0
CH4_PCAL[7:0]							
R/W-0h							

Table 7-52. CH4_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH4_PCAL[7:0]	R/W	0h	Channel 4 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.52 CH5_CFG0 Register (page = 0x00, address = 0x50) [reset = 0h]

This register is configuration register 0 for Channel 5.

Figure 7-52. CH5_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH5_INSRC[1:0]						Reserved
R-0h	R/W-0h						R-0h

Table 7-53. CH5_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-5	CH5_INSRC[1:0]	R/W	0h	Channel 5 Input Configuration 0d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN3 and PDMCLK) 1d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN3 and PDMCLK) 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN3 and PDMCLK) 3d = Reserved
4-0	Reserved	R	0h	Reserved

7.1.2.53 CH5_CFG2 Register (page = 0x00, address = 0x52) [reset = C9h]

This register is configuration register 2 for Channel 5.

Figure 7-53. CH5_CFG2 Register

7	6	5	4	3	2	1	0
CH5_DVOL[7:0]							
R/W-C9h							

Table 7-54. CH5_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH5_DVOL[7:0]	R/W	C9h	Channel 5 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.54 CH5_CFG3 Register (page = 0x00, address = 0x53) [reset = 80h]

This register is configuration register 3 for Channel 5.

Figure 7-54. CH5_CFG3 Register

7	6	5	4	3	2	1	0
CH5_GCAL[3:0]				Reserved			
R/W-8h				R-0h			

Table 7-55. CH5_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH5_GCAL[3:0]	R/W	8h	Channel 5 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.55 CH5_CFG4 Register (page = 0x00, address = 0x54) [reset = 0h]

This register is configuration register 4 for Channel 5.

Figure 7-55. CH5_CFG4 Register

7	6	5	4	3	2	1	0
CH5_PCAL[7:0]				Reserved			
R/W-0h				R-0h			

Table 7-56. CH5_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH5_PCAL[7:0]	R/W	0h	Channel 5 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.56 CH6_CFG0 Register (page = 0x00, address = 0x55) [reset = 0h]

This register is configuration register 0 for Channel 6.

Figure 7-56. CH6_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH6_INSRC[1:0]				Reserved		
R-0h	R/W-0h				R-0h		

Table 7-57. CH6_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-5	CH6_INSRC[1:0]	R/W	0h	Channel 6 Input Configuration 0d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN3 and PDMCLK) 1d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN3 and PDMCLK) 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN3 and PDMCLK) 3d = Reserved
4-0	Reserved	R	0h	Reserved

7.1.2.57 CH6_CFG2 Register (page = 0x00, address = 0x57) [reset = C9h]

This register is configuration register 2 for Channel 6.

Figure 7-57. CH6_CFG2 Register

7	6	5	4	3	2	1	0
CH6_DVOL[7:0]							
R/W-C9h							

Table 7-58. CH6_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH6_DVOL[7:0]	R/W	C9h	Channel 6 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.58 CH6_CFG3 Register (page = 0x00, address = 0x58) [reset = 80h]

This register is configuration register 3 for Channel 6.

Figure 7-58. CH6_CFG3 Register

7	6	5	4	3	2	1	0
CH6_GCAL[3:0]				Reserved			
R/W-8h				R-0h			

Table 7-59. CH6_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH6_GCAL[3:0]	R/W	8h	Channel 6 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.59 CH6_CFG4 Register (page = 0x00, address = 0x59) [reset = 0h]

This register is configuration register 4 for Channel 6.

Figure 7-59. CH6_CFG4 Register

7	6	5	4	3	2	1	0
CH6_PCAL[7:0]							
R/W-0h							

Table 7-60. CH6_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH6_PCAL[7:0]	R/W	0h	Channel 6 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.60 CH7_CFG0 Register (page = 0x00, address = 0x5A) [reset = 0h]

This register is configuration register 0 for Channel 7.

Figure 7-60. CH7_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH7_INSRC[1:0]						
R-0h	R/W-0h						

Table 7-61. CH7_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-5	CH7_INSRC[1:0]	R/W	0h	Channel 7 Input Configuration 0d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN4 and PDMCLK) 1d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN4 and PDMCLK) 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN4 and PDMCLK) 3d = Reserved
4-0	Reserved	R	0h	Reserved

7.1.2.61 CH7_CFG2 Register (page = 0x00, address = 0x5C) [reset = C9h]

This register is configuration register 2 for Channel 7.

Figure 7-61. CH7_CFG2 Register

7	6	5	4	3	2	1	0
CH7_DVOL[7:0]							
R/W-C9h							

Table 7-62. CH7_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH7_DVOL[7:0]	R/W	C9h	Channel 7 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.62 CH7_CFG3 Register (page = 0x00, address = 0x5D) [reset = 80h]

This register is configuration register 3 for Channel 7.

Figure 7-62. CH7_CFG3 Register

7	6	5	4	3	2	1	0
CH7_GCAL[3:0]							
R/W-8h							

Table 7-63. CH7_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH7_GCAL[3:0]	R/W	8h	Channel 7 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.63 CH7_CFG4 Register (page = 0x00, address = 0x5E) [reset = 0h]

This register is configuration register 4 for Channel 7.

Figure 7-63. CH7_CFG4 Register

7	6	5	4	3	2	1	0
CH7_PCAL[7:0]							
R/W-0h							

Table 7-64. CH7_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH7_PCAL[7:0]	R/W	0h	Channel 7 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.64 CH8_CFG0 Register (page = 0x00, address = 0x5F) [reset = 0h]

This register is configuration register 0 for Channel 8.

Figure 7-64. CH8_CFG0 Register

7	6	5	4	3	2	1	0
Reserved	CH8_INSRC[1:0]			Reserved			
R-0h	R/W-0h			R-0h			

Table 7-65. CH8_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	Reserved	R	0h	Reserved
6-5	CH8_INSRC[1:0]	R/W	0h	Channel 8 Input Configuration 0d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN4 and PDMCLK) 1d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN4 and PDMCLK) 2d = Digital microphone PDM input (configure the GPO and GPI pins accordingly for PDMDIN4 and PDMCLK) 3d = Reserved
4-0	Reserved	R	0h	Reserved

7.1.2.65 CH8_CFG2 Register (page = 0x00, address = 0x61) [reset = C9h]

This register is configuration register 2 for Channel 8.

Figure 7-65. CH8_CFG2 Register

7	6	5	4	3	2	1	0
CH8_DVOL[7:0]							
R/W-C9h							

Table 7-66. CH8_CFG2 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH8_DVOL[7:0]	R/W	C9h	Channel 8 digital volume control. 0d = Digital volume is muted 1d = Digital volume control is set to -100dB 2d = Digital volume control is set to -99.5dB 3d to 200d = Digital volume control is set as per configuration 201d = Digital volume control is set to 0dB 202d = Digital volume control is set to 0.5dB 203d to 253d = Digital volume control is set as per configuration 254d = Digital volume control is set to 26.5dB 255d = Digital volume control is set to 27dB

7.1.2.66 CH8_CFG3 Register (page = 0x00, address = 0x62) [reset = 80h]

This register is configuration register 3 for Channel 8.

Figure 7-66. CH8_CFG3 Register

7	6	5	4	3	2	1	0
CH8_GCAL[3:0]				Reserved			
R/W-8h				R-0h			

Table 7-67. CH8_CFG3 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-4	CH8_GCAL[3:0]	R/W	8h	Channel 8 gain calibration. 0d = Gain calibration is set to -0.8dB 1d = Gain calibration is set to -0.7dB 2d = Gain calibration is set to -0.6dB 3d to 7d = Gain calibration is set as per configuration 8d = Gain calibration is set to 0dB 9d = Gain calibration is set to 0.1dB 10d to 13d = Gain calibration is set as per configuration 14d = Gain calibration is set to 0.6dB 15d = Gain calibration is set to 0.7dB
3-0	Reserved	R	0h	Reserved

7.1.2.67 CH8_CFG4 Register (page = 0x00, address = 0x63) [reset = 0h]

This register is configuration register 4 for Channel 8.

Figure 7-67. CH8_CFG4 Register

7	6	5	4	3	2	1	0
CH8_PCAL[7:0]							
R/W-0h							

Table 7-68. CH8_CFG4 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	CH8_PCAL[7:0]	R/W	0h	Channel 8 phase calibration with modulator clock resolution. 0d = No phase calibration 1d = Phase calibration delay is set to one cycle of the modulator clock 2d = Phase calibration delay is set to two cycles of the modulator clock 3d to 254d = Phase calibration delay as per configuration 255d = Phase calibration delay is set to 255 cycles of the modulator clock

7.1.2.68 DSP_CFG0 Register (page = 0x00, address = 0x6B) [reset = 1h]

This register is the digital signal processor (DSP) configuration register 0.

Figure 7-68. DSP_CFG0 Register

7	6	5	4	3	2	1	0
Reserved		DECI_FILT[1:0]			CH_SUM[1:0]		HPF_SEL[1:0]
R-0h		R/W-0h			R/W-0h		R/W-1h

Table 7-69. DSP_CFG0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-6	Reserved	R	0h	Reserved
5-4	DECI_FILT[1:0]	R/W	0h	Decimation filter response. 0d = Linear phase 1d = Low latency 2d = Ultra-low latency 3d = Reserved
3-2	CH_SUM[1:0]	R/W	0h	Channel summation mode for higher SNR 0d = Channel summation mode is disabled 1d = 2-channel summation mode is enabled to generate a (CH1 + CH2) / 2 and a (CH3 + CH4) / 2 output 2d = 4-channel summation mode is enabled to generate a (CH1 + CH2 + CH3 + CH4) / 4 output 3d = Reserved

Table 7-69. DSP_CFG0 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
1-0	HPF_SEL[1:0]	R/W	1h	High-pass filter (HPF) selection. 0d = Programmable first-order IIR filter for a custom HPF with default coefficient values in P4_R72 to P4_R83 set as the all-pass filter 1d = HPF with a cutoff of $0.00025x f_S$ (12Hz at $f_S = 48\text{kHz}$) is selected 2d = HPF with a cutoff of $0.002x f_S$ (96Hz at $f_S = 48\text{kHz}$) is selected 3d = HPF with a cutoff of $0.008x f_S$ (384Hz at $f_S = 48\text{kHz}$) is selected

7.1.2.69 DSP_CFG1 Register (page = 0x00, address = 0x6C) [reset = 40h]

This register is the digital signal processor (DSP) configuration register 1.

Figure 7-69. DSP_CFG1 Register

7	6	5	4	3	2	1	0
DVOL_GANG		BIQUAD_CFG[1:0]	DISABLE_SOFT_STEP	Reserved	Reserved	Reserved	
R/W-0h		R/W-2h	R/W-0h	R/W-0h	R/W-0h	R-0h	

Table 7-70. DSP_CFG1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	DVOL_GANG	R/W	0h	DVOL control ganged across channels. 0d = Each channel has a dedicated DVOL CTRL settings as programmed in the CHx_DVOL bits 1d = All active channels must use the channel 1 DVOL setting (CH1_DVOL) irrespective of whether channel 1 is turned on or not
6-5	BIQUAD_CFG[1:0]	R/W	2h	Number of biquads per channel configuration. 0d = No biquads per channel; biquads are all disabled 1d = 1 biquad per channel 2d = 2 biquads per channel 3d = 3 biquads per channel
4	DISABLE_SOFT_STEP	R/W	0h	Soft-stepping disable during DVOL change, mute, and unmute. 0d = Soft-stepping enabled 1d = Soft-stepping disabled
3-0	Reserved	R/W	0h	Reserved

7.1.2.70 IN_CH_EN Register (page = 0x00, address = 0x73) [reset = F0h]

This register is the input channel enable configuration register.

Figure 7-70. IN_CH_EN Register

7	6	5	4	3	2	1	0
IN_CH1_EN	IN_CH2_EN	IN_CH3_EN	IN_CH4_EN	IN_CH5_EN	IN_CH6_EN	IN_CH7_EN	IN_CH8_EN
R/W-1h	R/W-1h	R/W-1h	R/W-1h	R/W-0h	R/W-0h	R/W-0h	R/W-0h

Table 7-71. IN_CH_EN Register Field Descriptions

Bit	Field	Type	Reset	Description
7	IN_CH1_EN	R/W	1h	Input channel 1 enable setting. 0d = Channel 1 is disabled 1d = Channel 1 is enabled
6	IN_CH2_EN	R/W	1h	Input channel 2 enable setting. 0d = Channel 2 is disabled 1d = Channel 2 is enabled
5	IN_CH3_EN	R/W	1h	Input channel 3 enable setting. 0d = Channel 3 is disabled 1d = Channel 3 is enabled
4	IN_CH4_EN	R/W	1h	Input channel 4 enable setting. 0d = Channel 4 is disabled 1d = Channel 4 is enabled

Table 7-71. IN_CH_EN Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
3	IN_CH5_EN	R/W	0h	Input channel 5 enable setting. 0d = Channel 5 is disabled 1d = Channel 5 is enabled
2	IN_CH6_EN	R/W	0h	Input channel 6 enable setting. 0d = Channel 6 is disabled 1d = Channel 6 is enabled
1	IN_CH7_EN	R/W	0h	Input channel 7 enable setting. 0d = Channel 7 is disabled 1d = Channel 7 is enabled
0	IN_CH8_EN	R/W	0h	Input channel 8 enable setting. 0d = Channel 8 is disabled 1d = Channel 8 is enabled

7.1.2.71 ASI_OUT_CH_EN Register (page = 0x00, address = 0x74) [reset = 0h]

This register is the ASI output channel enable configuration register.

Figure 7-71. ASI_OUT_CH_EN Register

7	6	5	4	3	2	1	0
ASI_OUT_CH1_EN	ASI_OUT_CH2_EN	ASI_OUT_CH3_EN	ASI_OUT_CH4_EN	ASI_OUT_CH5_EN	ASI_OUT_CH6_EN	ASI_OUT_CH7_EN	ASI_OUT_CH8_EN
R/W-0h							

Table 7-72. ASI_OUT_CH_EN Register Field Descriptions

Bit	Field	Type	Reset	Description
7	ASI_OUT_CH1_EN	R/W	0h	ASI output channel 1 enable setting. 0d = Channel 1 output slot is in a tri-state condition 1d = Channel 1 output slot is enabled
6	ASI_OUT_CH2_EN	R/W	0h	ASI output channel 2 enable setting. 0d = Channel 2 output slot is in a tri-state condition 1d = Channel 2 output slot is enabled
5	ASI_OUT_CH3_EN	R/W	0h	ASI output channel 3 enable setting. 0d = Channel 3 output slot is in a tri-state condition 1d = Channel 3 output slot is enabled
4	ASI_OUT_CH4_EN	R/W	0h	ASI output channel 4 enable setting. 0d = Channel 4 output slot is in a tri-state condition 1d = Channel 4 output slot is enabled
3	ASI_OUT_CH5_EN	R/W	0h	ASI output channel 5 enable setting. 0d = Channel 5 output slot is in a tri-state condition 1d = Channel 5 output slot is enabled
2	ASI_OUT_CH6_EN	R/W	0h	ASI output channel 6 enable setting. 0d = Channel 6 output slot is in a tri-state condition 1d = Channel 6 output slot is enabled
1	ASI_OUT_CH7_EN	R/W	0h	ASI output channel 7 enable setting. 0d = Channel 7 output slot is in a tri-state condition 1d = Channel 7 output slot is enabled
0	ASI_OUT_CH8_EN	R/W	0h	ASI output channel 8 enable setting. 0d = Channel 8 output slot is in a tri-state condition 1d = Channel 8 output slot is enabled

7.1.2.72 PWR_CFG Register (page = 0x00, address = 0x75) [reset = 0h]

This register is the power-up configuration register.

Figure 7-72. PWR_CFG Register

7	6	5	4	3	2	1	0
MICBIAS_PDZ	PDM_PDZ	PLL_PDZ	DYN_CH_PUPD_EN	DYN_MAXCH_SEL[1:0]	Reserved	Reserved	
R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R/W-0h	R-0h

Table 7-73. PWR_CFG Register Field Descriptions

Bit	Field	Type	Reset	Description
7	MICBIAS_PDZ	R/W	0h	Power control for MICBIAS. 0d = Power down MICBIAS 1d = Power up MICBIAS
6	PDM_PDZ	R/W	0h	Power control for PDM channels. 0d = Power down all PDM channels 1d = Power up all enabled PDM channels
5	PLL_PDZ	R/W	0h	Power control for the PLL. 0d = Power down the PLL 1d = Power up the PLL
4	DYN_CH_PUPD_EN	R/W	0h	Dynamic channel power-up, power-down enable. 0d = Channel power-up, power-down is not supported if any channel recording is on 1d = Channel can be powered up or down individually, even if channel recording is on
3-2	DYN_MAXCH_SEL[1:0]	R/W	0h	Dynamic mode maximum channel select configuration. 0d = Channel 1 and channel 2 are used with dynamic channel power-up, power-down feature enabled 1d = Channel 1 to channel 4 are used with dynamic channel power-up, power-down feature enabled 2d = Channel 1 to channel 6 are used with dynamic channel power-up, power-down feature enabled 3d = Channel 1 to channel 8 are used with dynamic channel power-up, power-down feature enabled
1	Reserved	R/W	0h	Reserved
0	Reserved	R	0h	Reserved

7.1.2.73 DEV_STS0 Register (page = 0x00, address = 0x76) [reset = 0h]

This register is the device status value register 0.

Figure 7-73. DEV_STS0 Register

7	6	5	4	3	2	1	0
CH1_STATUS	CH2_STATUS	CH3_STATUS	CH4_STATUS	CH5_STATUS	CH6_STATUS	CH7_STATUS	CH8_STATUS
R-0h							

Table 7-74. DEV_STS0 Register Field Descriptions

Bit	Field	Type	Reset	Description
7	CH1_STATUS	R	0h	PDM channel 1 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up
6	CH2_STATUS	R	0h	PDM channel 2 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up
5	CH3_STATUS	R	0h	PDM channel 3 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up
4	CH4_STATUS	R	0h	PDM channel 4 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up

Table 7-74. DEV_STS0 Register Field Descriptions (continued)

Bit	Field	Type	Reset	Description
3	CH5_STATUS	R	0h	PDM channel 5 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up
2	CH6_STATUS	R	0h	PDM channel 6 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up
1	CH7_STATUS	R	0h	PDM channel 7 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up
0	CH8_STATUS	R	0h	PDM channel 8 power status. 0d = PDM channel is powered down 1d = PDM channel is powered up

7.1.2.74 DEV_STS1 Register (page = 0x00, address = 0x77) [reset = 80h]

This register is the device status value register 1.

Figure 7-74. DEV_STS1 Register

7	6	5	4	3	2	1	0	
MODE_STS[2:0]			Reserved					
R-4h			R-0h					

Table 7-75. DEV_STS1 Register Field Descriptions

Bit	Field	Type	Reset	Description
7-5	MODE_STS[2:0]	R	4h	Device mode status. 4d = Device is in sleep mode or software shutdown mode 6d = Device is in active mode with all PDM channels turned off 7d = Device is in active mode with at least one PDM channel turned on
4-0	Reserved	R	0h	Reserved

7.1.2.75 I2C_CKSUM Register (page = 0x00, address = 0x7E) [reset = 0h]

This register returns the I²C transactions checksum value.

Figure 7-75. I2C_CKSUM Register

7	6	5	4	3	2	1	0
I2C_CKSUM[7:0]							
R/W-0h							

Table 7-76. I2C_CKSUM Register Field Descriptions

Bit	Field	Type	Reset	Description
7-0	I2C_CKSUM[7:0]	R/W	0h	These bits return the I ² C transactions checksum value. Writing to this register resets the checksum to the written value. This register is updated on writes to other registers on all pages.

7.2 Programmable Coefficient Registers

7.2.1 Programmable Coefficient Registers: Page = 0x02

This register page (shown in [Table 7-77](#)) consists of the programmable coefficients for the biquad 1 to biquad 6 filters. To optimize the coefficients register transaction time for page 2, page 3, and page 4, the device also supports (by default) auto-incremented pages for the I²C and SPI burst writes and reads. After a transaction of register address 0x7F, the device auto increments to the next page at register 0x08 to transact the next coefficient value. These programmable coefficients are 32-bit, two's complement numbers. For a successful coefficient register transaction, the host device must write and read all the four bytes starting with the most significant byte (BYT1) for a target coefficient register transaction. While using SPI for a coefficient register read transaction, the device gives out first byte as dummy read byte therefore the host must read five bytes which includes first byte as dummy read byte and the last four bytes corresponds to the coefficient register value starting with the most significant byte (BYT1).

Table 7-77. Page 0x02 Programmable Coefficient Registers

ADDRESS	REGISTER	RESET	DESCRIPTION
0x00	PAGE[7:0]	0x00	Device page register
0x08	BQ1_N0_BYT1[7:0]	0x7F	Programmable biquad 1, N0 coefficient byte[31:24]
0x09	BQ1_N0_BYT2[7:0]	0xFF	Programmable biquad 1, N0 coefficient byte[23:16]
0x0A	BQ1_N0_BYT3[7:0]	0xFF	Programmable biquad 1, N0 coefficient byte[15:8]
0x0B	BQ1_N0_BYT4[7:0]	0xFF	Programmable biquad 1, N0 coefficient byte[7:0]
0x0C	BQ1_N1_BYT1[7:0]	0x00	Programmable biquad 1, N1 coefficient byte[31:24]
0x0D	BQ1_N1_BYT2[7:0]	0x00	Programmable biquad 1, N1 coefficient byte[23:16]
0x0E	BQ1_N1_BYT3[7:0]	0x00	Programmable biquad 1, N1 coefficient byte[15:8]
0x0F	BQ1_N1_BYT4[7:0]	0x00	Programmable biquad 1, N1 coefficient byte[7:0]
0x10	BQ1_N2_BYT1[7:0]	0x00	Programmable biquad 1, N2 coefficient byte[31:24]
0x11	BQ1_N2_BYT2[7:0]	0x00	Programmable biquad 1, N2 coefficient byte[23:16]
0x12	BQ1_N2_BYT3[7:0]	0x00	Programmable biquad 1, N2 coefficient byte[15:8]
0x13	BQ1_N2_BYT4[7:0]	0x00	Programmable biquad 1, N2 coefficient byte[7:0]
0x14	BQ1_D1_BYT1[7:0]	0x00	Programmable biquad 1, D1 coefficient byte[31:24]
0x15	BQ1_D1_BYT2[7:0]	0x00	Programmable biquad 1, D1 coefficient byte[23:16]
0x16	BQ1_D1_BYT3[7:0]	0x00	Programmable biquad 1, D1 coefficient byte[15:8]
0x17	BQ1_D1_BYT4[7:0]	0x00	Programmable biquad 1, D1 coefficient byte[7:0]
0x18	BQ1_D2_BYT1[7:0]	0x00	Programmable biquad 1, D2 coefficient byte[31:24]
0x19	BQ1_D2_BYT2[7:0]	0x00	Programmable biquad 1, D2 coefficient byte[23:16]
0x1A	BQ1_D2_BYT3[7:0]	0x00	Programmable biquad 1, D2 coefficient byte[15:8]
0x1B	BQ1_D2_BYT4[7:0]	0x00	Programmable biquad 1, D2 coefficient byte[7:0]
0x1C	BQ2_N0_BYT1[7:0]	0x7F	Programmable biquad 2, N0 coefficient byte[31:24]
0x1D	BQ2_N0_BYT2[7:0]	0xFF	Programmable biquad 2, N0 coefficient byte[23:16]
0x1E	BQ2_N0_BYT3[7:0]	0xFF	Programmable biquad 2, N0 coefficient byte[15:8]
0x1F	BQ2_N0_BYT4[7:0]	0xFF	Programmable biquad 2, N0 coefficient byte[7:0]
0x20	BQ2_N1_BYT1[7:0]	0x00	Programmable biquad 2, N1 coefficient byte[31:24]
0x21	BQ2_N1_BYT2[7:0]	0x00	Programmable biquad 2, N1 coefficient byte[23:16]
0x22	BQ2_N1_BYT3[7:0]	0x00	Programmable biquad 2, N1 coefficient byte[15:8]
0x23	BQ2_N1_BYT4[7:0]	0x00	Programmable biquad 2, N1 coefficient byte[7:0]
0x24	BQ2_N2_BYT1[7:0]	0x00	Programmable biquad 2, N2 coefficient byte[31:24]
0x25	BQ2_N2_BYT2[7:0]	0x00	Programmable biquad 2, N2 coefficient byte[23:16]
0x26	BQ2_N2_BYT3[7:0]	0x00	Programmable biquad 2, N2 coefficient byte[15:8]
0x27	BQ2_N2_BYT4[7:0]	0x00	Programmable biquad 2, N2 coefficient byte[7:0]
0x28	BQ2_D1_BYT1[7:0]	0x00	Programmable biquad 2, D1 coefficient byte[31:24]
0x29	BQ2_D1_BYT2[7:0]	0x00	Programmable biquad 2, D1 coefficient byte[23:16]
0x2A	BQ2_D1_BYT3[7:0]	0x00	Programmable biquad 2, D1 coefficient byte[15:8]
0x2B	BQ2_D1_BYT4[7:0]	0x00	Programmable biquad 2, D1 coefficient byte[7:0]

Table 7-77. Page 0x02 Programmable Coefficient Registers (continued)

0x2C	BQ2_D2_BYT1[7:0]	0x00	Programmable biquad 2, D2 coefficient byte[31:24]
0x2D	BQ2_D2_BYT2[7:0]	0x00	Programmable biquad 2, D2 coefficient byte[23:16]
0x2E	BQ2_D2_BYT3[7:0]	0x00	Programmable biquad 2, D2 coefficient byte[15:8]
0x2F	BQ2_D2_BYT4[7:0]	0x00	Programmable biquad 2, D2 coefficient byte[7:0]
0x30	BQ3_N0_BYT1[7:0]	0x7F	Programmable biquad 3, N0 coefficient byte[31:24]
0x31	BQ3_N0_BYT2[7:0]	0xFF	Programmable biquad 3, N0 coefficient byte[23:16]
0x32	BQ3_N0_BYT3[7:0]	0xFF	Programmable biquad 3, N0 coefficient byte[15:8]
0x33	BQ3_N0_BYT4[7:0]	0xFF	Programmable biquad 3, N0 coefficient byte[7:0]
0x34	BQ3_N1_BYT1[7:0]	0x00	Programmable biquad 3, N1 coefficient byte[31:24]
0x35	BQ3_N1_BYT2[7:0]	0x00	Programmable biquad 3, N1 coefficient byte[23:16]
0x36	BQ3_N1_BYT3[7:0]	0x00	Programmable biquad 3, N1 coefficient byte[15:8]
0x37	BQ3_N1_BYT4[7:0]	0x00	Programmable biquad 3, N1 coefficient byte[7:0]
0x38	BQ3_N2_BYT1[7:0]	0x00	Programmable biquad 3, N2 coefficient byte[31:24]
0x39	BQ3_N2_BYT2[7:0]	0x00	Programmable biquad 3, N2 coefficient byte[23:16]
0x3A	BQ3_N2_BYT3[7:0]	0x00	Programmable biquad 3, N2 coefficient byte[15:8]
0x3B	BQ3_N2_BYT4[7:0]	0x00	Programmable biquad 3, N2 coefficient byte[7:0]
0x3C	BQ3_D1_BYT1[7:0]	0x00	Programmable biquad 3, D1 coefficient byte[31:24]
0x3D	BQ3_D1_BYT2[7:0]	0x00	Programmable biquad 3, D1 coefficient byte[23:16]
0x3E	BQ3_D1_BYT3[7:0]	0x00	Programmable biquad 3, D1 coefficient byte[15:8]
0x3F	BQ3_D1_BYT4[7:0]	0x00	Programmable biquad 3, D1 coefficient byte[7:0]
0x40	BQ3_D2_BYT1[7:0]	0x00	Programmable biquad 3, D2 coefficient byte[31:24]
0x41	BQ3_D2_BYT2[7:0]	0x00	Programmable biquad 3, D2 coefficient byte[23:16]
0x42	BQ3_D2_BYT3[7:0]	0x00	Programmable biquad 3, D2 coefficient byte[15:8]
0x43	BQ3_D2_BYT4[7:0]	0x00	Programmable biquad 3, D2 coefficient byte[7:0]
0x44	BQ4_N0_BYT1[7:0]	0x7F	Programmable biquad 4, N0 coefficient byte[31:24]
0x45	BQ4_N0_BYT2[7:0]	0xFF	Programmable biquad 4, N0 coefficient byte[23:16]
0x46	BQ4_N0_BYT3[7:0]	0xFF	Programmable biquad 4, N0 coefficient byte[15:8]
0x47	BQ4_N0_BYT4[7:0]	0xFF	Programmable biquad 4, N0 coefficient byte[7:0]
0x48	BQ4_N1_BYT1[7:0]	0x00	Programmable biquad 4, N1 coefficient byte[31:24]
0x49	BQ4_N1_BYT2[7:0]	0x00	Programmable biquad 4, N1 coefficient byte[23:16]
0x4A	BQ4_N1_BYT3[7:0]	0x00	Programmable biquad 4, N1 coefficient byte[15:8]
0x4B	BQ4_N1_BYT4[7:0]	0x00	Programmable biquad 4, N1 coefficient byte[7:0]
0x4C	BQ4_N2_BYT1[7:0]	0x00	Programmable biquad 4, N2 coefficient byte[31:24]
0x4D	BQ4_N2_BYT2[7:0]	0x00	Programmable biquad 4, N2 coefficient byte[23:16]
0x4E	BQ4_N2_BYT3[7:0]	0x00	Programmable biquad 4, N2 coefficient byte[15:8]
0x4F	BQ4_N2_BYT4[7:0]	0x00	Programmable biquad 4, N2 coefficient byte[7:0]
0x50	BQ4_D1_BYT1[7:0]	0x00	Programmable biquad 4, D1 coefficient byte[31:24]
0x51	BQ4_D1_BYT2[7:0]	0x00	Programmable biquad 4, D1 coefficient byte[23:16]
0x52	BQ4_D1_BYT3[7:0]	0x00	Programmable biquad 4, D1 coefficient byte[15:8]
0x53	BQ4_D1_BYT4[7:0]	0x00	Programmable biquad 4, D1 coefficient byte[7:0]
0x54	BQ4_D2_BYT1[7:0]	0x00	Programmable biquad 4, D2 coefficient byte[31:24]
0x55	BQ4_D2_BYT2[7:0]	0x00	Programmable biquad 4, D2 coefficient byte[23:16]
0x56	BQ4_D2_BYT3[7:0]	0x00	Programmable biquad 4, D2 coefficient byte[15:8]
0x57	BQ4_D2_BYT4[7:0]	0x00	Programmable biquad 4, D2 coefficient byte[7:0]
0x58	BQ5_N0_BYT1[7:0]	0x7F	Programmable biquad 5, N0 coefficient byte[31:24]
0x59	BQ5_N0_BYT2[7:0]	0xFF	Programmable biquad 5, N0 coefficient byte[23:16]
0x5A	BQ5_N0_BYT3[7:0]	0xFF	Programmable biquad 5, N0 coefficient byte[15:8]
0x5B	BQ5_N0_BYT4[7:0]	0xFF	Programmable biquad 5, N0 coefficient byte[7:0]
0x5C	BQ5_N1_BYT1[7:0]	0x00	Programmable biquad 5, N1 coefficient byte[31:24]
0x5D	BQ5_N1_BYT2[7:0]	0x00	Programmable biquad 5, N1 coefficient byte[23:16]
0x5E	BQ5_N1_BYT3[7:0]	0x00	Programmable biquad 5, N1 coefficient byte[15:8]

Table 7-77. Page 0x02 Programmable Coefficient Registers (continued)

0x5F	BQ5_N1_BYT4[7:0]	0x00	Programmable biquad 5, N1 coefficient byte[7:0]
0x60	BQ5_N2_BYT1[7:0]	0x00	Programmable biquad 5, N2 coefficient byte[31:24]
0x61	BQ5_N2_BYT2[7:0]	0x00	Programmable biquad 5, N2 coefficient byte[23:16]
0x62	BQ5_N2_BYT3[7:0]	0x00	Programmable biquad 5, N2 coefficient byte[15:8]
0x63	BQ5_N2_BYT4[7:0]	0x00	Programmable biquad 5, N2 coefficient byte[7:0]
0x64	BQ5_D1_BYT1[7:0]	0x00	Programmable biquad 5, D1 coefficient byte[31:24]
0x65	BQ5_D1_BYT2[7:0]	0x00	Programmable biquad 5, D1 coefficient byte[23:16]
0x66	BQ5_D1_BYT3[7:0]	0x00	Programmable biquad 5, D1 coefficient byte[15:8]
0x67	BQ5_D1_BYT4[7:0]	0x00	Programmable biquad 5, D1 coefficient byte[7:0]
0x68	BQ5_D2_BYT1[7:0]	0x00	Programmable biquad 5, D2 coefficient byte[31:24]
0x69	BQ5_D2_BYT2[7:0]	0x00	Programmable biquad 5, D2 coefficient byte[23:16]
0x6A	BQ5_D2_BYT3[7:0]	0x00	Programmable biquad 5, D2 coefficient byte[15:8]
0x6B	BQ5_D2_BYT4[7:0]	0x00	Programmable biquad 5, D2 coefficient byte[7:0]
0x6C	BQ6_N0_BYT1[7:0]	0x7F	Programmable biquad 6, N0 coefficient byte[31:24]
0x6D	BQ6_N0_BYT2[7:0]	0xFF	Programmable biquad 6, N0 coefficient byte[23:16]
0x6E	BQ6_N0_BYT3[7:0]	0xFF	Programmable biquad 6, N0 coefficient byte[15:8]
0x6F	BQ6_N0_BYT4[7:0]	0xFF	Programmable biquad 6, N0 coefficient byte[7:0]
0x70	BQ6_N1_BYT1[7:0]	0x00	Programmable biquad 6, N1 coefficient byte[31:24]
0x71	BQ6_N1_BYT2[7:0]	0x00	Programmable biquad 6, N1 coefficient byte[23:16]
0x72	BQ6_N1_BYT3[7:0]	0x00	Programmable biquad 6, N1 coefficient byte[15:8]
0x73	BQ6_N1_BYT4[7:0]	0x00	Programmable biquad 6, N1 coefficient byte[7:0]
0x74	BQ6_N2_BYT1[7:0]	0x00	Programmable biquad 6, N2 coefficient byte[31:24]
0x75	BQ6_N2_BYT2[7:0]	0x00	Programmable biquad 6, N2 coefficient byte[23:16]
0x76	BQ6_N2_BYT3[7:0]	0x00	Programmable biquad 6, N2 coefficient byte[15:8]
0x77	BQ6_N2_BYT4[7:0]	0x00	Programmable biquad 6, N2 coefficient byte[7:0]
0x78	BQ6_D1_BYT1[7:0]	0x00	Programmable biquad 6, D1 coefficient byte[31:24]
0x79	BQ6_D1_BYT2[7:0]	0x00	Programmable biquad 6, D1 coefficient byte[23:16]
0x7A	BQ6_D1_BYT3[7:0]	0x00	Programmable biquad 6, D1 coefficient byte[15:8]
0x7B	BQ6_D1_BYT4[7:0]	0x00	Programmable biquad 6, D1 coefficient byte[7:0]
0x7C	BQ6_D2_BYT1[7:0]	0x00	Programmable biquad 6, D2 coefficient byte[31:24]
0x7D	BQ6_D2_BYT2[7:0]	0x00	Programmable biquad 6, D2 coefficient byte[23:16]
0x7E	BQ6_D2_BYT3[7:0]	0x00	Programmable biquad 6, D2 coefficient byte[15:8]
0x7F	BQ6_D2_BYT4[7:0]	0x00	Programmable biquad 6, D2 coefficient byte[7:0]

7.2.2 Programmable Coefficient Registers: Page = 0x03

This register page (shown in [Table 7-78](#)) consists of the programmable coefficients for the biquad 7 to biquad 12 filters. To optimize the coefficients register transaction time for page 2, page 3, and page 4, the device also supports (by default) auto-incremented pages for the I²C and SPI burst writes and reads. After a transaction of register address 0x7F, the device auto increments to the next page at register 0x08 to transact the next coefficient value. These programmable coefficients are 32-bit, two's complement numbers. For a successful coefficient register transaction, the host device must write and read all the four bytes starting with the most significant byte (BYT1) for a target coefficient register transaction. While using SPI for a coefficient register read transaction, the device gives out first byte as dummy read byte therefore the host must read five bytes which includes first byte as dummy read byte and the last four bytes corresponds to the coefficient register value starting with the most significant byte (BYT1).

Table 7-78. Page 0x03 Programmable Coefficient Registers

ADDR	REGISTER	RESET	DESCRIPTION
0x00	PAGE[7:0]	0x00	Device page register
0x08	BQ7_N0_BYT1[7:0]	0x7F	Programmable biquad 7, N0 coefficient byte[31:24]
0x09	BQ7_N0_BYT2[7:0]	0xFF	Programmable biquad 7, N0 coefficient byte[23:16]
0x0A	BQ7_N0_BYT3[7:0]	0xFF	Programmable biquad 7, N0 coefficient byte[15:8]
0x0B	BQ7_N0_BYT4[7:0]	0xFF	Programmable biquad 7, N0 coefficient byte[7:0]
0x0C	BQ7_N1_BYT1[7:0]	0x00	Programmable biquad 7, N1 coefficient byte[31:24]
0x0D	BQ7_N1_BYT2[7:0]	0x00	Programmable biquad 7, N1 coefficient byte[23:16]
0x0E	BQ7_N1_BYT3[7:0]	0x00	Programmable biquad 7, N1 coefficient byte[15:8]
0x0F	BQ7_N1_BYT4[7:0]	0x00	Programmable biquad 7, N1 coefficient byte[7:0]
0x10	BQ7_N2_BYT1[7:0]	0x00	Programmable biquad 7, N2 coefficient byte[31:24]
0x11	BQ7_N2_BYT2[7:0]	0x00	Programmable biquad 7, N2 coefficient byte[23:16]
0x12	BQ7_N2_BYT3[7:0]	0x00	Programmable biquad 7, N2 coefficient byte[15:8]
0x13	BQ7_N2_BYT4[7:0]	0x00	Programmable biquad 7, N2 coefficient byte[7:0]
0x14	BQ7_D1_BYT1[7:0]	0x00	Programmable biquad 7, D1 coefficient byte[31:24]
0x15	BQ7_D1_BYT2[7:0]	0x00	Programmable biquad 7, D1 coefficient byte[23:16]
0x16	BQ7_D1_BYT3[7:0]	0x00	Programmable biquad 7, D1 coefficient byte[15:8]
0x17	BQ7_D1_BYT4[7:0]	0x00	Programmable biquad 7, D1 coefficient byte[7:0]
0x18	BQ7_D2_BYT1[7:0]	0x00	Programmable biquad 7, D2 coefficient byte[31:24]
0x19	BQ7_D2_BYT2[7:0]	0x00	Programmable biquad 7, D2 coefficient byte[23:16]
0x1A	BQ7_D2_BYT3[7:0]	0x00	Programmable biquad 7, D2 coefficient byte[15:8]
0x1B	BQ7_D2_BYT4[7:0]	0x00	Programmable biquad 7, D2 coefficient byte[7:0]
0x1C	BQ8_N0_BYT1[7:0]	0x7F	Programmable biquad 8, N0 coefficient byte[31:24]
0x1D	BQ8_N0_BYT2[7:0]	0xFF	Programmable biquad 8, N0 coefficient byte[23:16]
0x1E	BQ8_N0_BYT3[7:0]	0xFF	Programmable biquad 8, N0 coefficient byte[15:8]
0x1F	BQ8_N0_BYT4[7:0]	0xFF	Programmable biquad 8, N0 coefficient byte[7:0]
0x20	BQ8_N1_BYT1[7:0]	0x00	Programmable biquad 8, N1 coefficient byte[31:24]
0x21	BQ8_N1_BYT2[7:0]	0x00	Programmable biquad 8, N1 coefficient byte[23:16]
0x22	BQ8_N1_BYT3[7:0]	0x00	Programmable biquad 8, N1 coefficient byte[15:8]
0x23	BQ8_N1_BYT4[7:0]	0x00	Programmable biquad 8, N1 coefficient byte[7:0]
0x24	BQ8_N2_BYT1[7:0]	0x00	Programmable biquad 8, N2 coefficient byte[31:24]
0x25	BQ8_N2_BYT2[7:0]	0x00	Programmable biquad 8, N2 coefficient byte[23:16]
0x26	BQ8_N2_BYT3[7:0]	0x00	Programmable biquad 8, N2 coefficient byte[15:8]
0x27	BQ8_N2_BYT4[7:0]	0x00	Programmable biquad 8, N2 coefficient byte[7:0]
0x28	BQ8_D1_BYT1[7:0]	0x00	Programmable biquad 8, D1 coefficient byte[31:24]
0x29	BQ8_D1_BYT2[7:0]	0x00	Programmable biquad 8, D1 coefficient byte[23:16]
0x2A	BQ8_D1_BYT3[7:0]	0x00	Programmable biquad 8, D1 coefficient byte[15:8]
0x2B	BQ8_D1_BYT4[7:0]	0x00	Programmable biquad 8, D1 coefficient byte[7:0]
0x2C	BQ8_D2_BYT1[7:0]	0x00	Programmable biquad 8, D2 coefficient byte[31:24]

Table 7-78. Page 0x03 Programmable Coefficient Registers (continued)

0x2D	BQ8_D2_BYT2[7:0]	0x00	Programmable biquad 8, D2 coefficient byte[23:16]
0x2E	BQ8_D2_BYT3[7:0]	0x00	Programmable biquad 8, D2 coefficient byte[15:8]
0x2F	BQ8_D2_BYT4[7:0]	0x00	Programmable biquad 8, D2 coefficient byte[7:0]
0x30	BQ9_N0_BYT1[7:0]	0x7F	Programmable biquad 9, N0 coefficient byte[31:24]
0x31	BQ9_N0_BYT2[7:0]	0xFF	Programmable biquad 9, N0 coefficient byte[23:16]
0x32	BQ9_N0_BYT3[7:0]	0xFF	Programmable biquad 9, N0 coefficient byte[15:8]
0x33	BQ9_N0_BYT4[7:0]	0xFF	Programmable biquad 9, N0 coefficient byte[7:0]
0x34	BQ9_N1_BYT1[7:0]	0x00	Programmable biquad 9, N1 coefficient byte[31:24]
0x35	BQ9_N1_BYT2[7:0]	0x00	Programmable biquad 9, N1 coefficient byte[23:16]
0x36	BQ9_N1_BYT3[7:0]	0x00	Programmable biquad 9, N1 coefficient byte[15:8]
0x37	BQ9_N1_BYT4[7:0]	0x00	Programmable biquad 9, N1 coefficient byte[7:0]
0x38	BQ9_N2_BYT1[7:0]	0x00	Programmable biquad 9, N2 coefficient byte[31:24]
0x39	BQ9_N2_BYT2[7:0]	0x00	Programmable biquad 9, N2 coefficient byte[23:16]
0x3A	BQ9_N2_BYT3[7:0]	0x00	Programmable biquad 9, N2 coefficient byte[15:8]
0x3B	BQ9_N2_BYT4[7:0]	0x00	Programmable biquad 9, N2 coefficient byte[7:0]
0x3C	BQ9_D1_BYT1[7:0]	0x00	Programmable biquad 9, D1 coefficient byte[31:24]
0x3D	BQ9_D1_BYT2[7:0]	0x00	Programmable biquad 9, D1 coefficient byte[23:16]
0x3E	BQ9_D1_BYT3[7:0]	0x00	Programmable biquad 9, D1 coefficient byte[15:8]
0x3F	BQ9_D1_BYT4[7:0]	0x00	Programmable biquad 9, D1 coefficient byte[7:0]
0x40	BQ9_D2_BYT1[7:0]	0x00	Programmable biquad 9, D2 coefficient byte[31:24]
0x41	BQ9_D2_BYT2[7:0]	0x00	Programmable biquad 9, D2 coefficient byte[23:16]
0x42	BQ9_D2_BYT3[7:0]	0x00	Programmable biquad 9, D2 coefficient byte[15:8]
0x43	BQ9_D2_BYT4[7:0]	0x00	Programmable biquad 9, D2 coefficient byte[7:0]
0x44	BQ10_N0_BYT1[7:0]	0x7F	Programmable biquad 10, N0 coefficient byte[31:24]
0x45	BQ10_N0_BYT2[7:0]	0xFF	Programmable biquad 10, N0 coefficient byte[23:16]
0x46	BQ10_N0_BYT3[7:0]	0xFF	Programmable biquad 10, N0 coefficient byte[15:8]
0x47	BQ10_N0_BYT4[7:0]	0xFF	Programmable biquad 10, N0 coefficient byte[7:0]
0x48	BQ10_N1_BYT1[7:0]	0x00	Programmable biquad 10, N1 coefficient byte[31:24]
0x49	BQ10_N1_BYT2[7:0]	0x00	Programmable biquad 10, N1 coefficient byte[23:16]
0x4A	BQ10_N1_BYT3[7:0]	0x00	Programmable biquad 10, N1 coefficient byte[15:8]
0x4B	BQ10_N1_BYT4[7:0]	0x00	Programmable biquad 10, N1 coefficient byte[7:0]
0x4C	BQ10_N2_BYT1[7:0]	0x00	Programmable biquad 10, N2 coefficient byte[31:24]
0x4D	BQ10_N2_BYT2[7:0]	0x00	Programmable biquad 10, N2 coefficient byte[23:16]
0x4E	BQ10_N2_BYT3[7:0]	0x00	Programmable biquad 10, N2 coefficient byte[15:8]
0x4F	BQ10_N2_BYT4[7:0]	0x00	Programmable biquad 10, N2 coefficient byte[7:0]
0x50	BQ10_D1_BYT1[7:0]	0x00	Programmable biquad 10, D1 coefficient byte[31:24]
0x51	BQ10_D1_BYT2[7:0]	0x00	Programmable biquad 10, D1 coefficient byte[23:16]
0x52	BQ10_D1_BYT3[7:0]	0x00	Programmable biquad 10, D1 coefficient byte[15:8]
0x53	BQ10_D1_BYT4[7:0]	0x00	Programmable biquad 10, D1 coefficient byte[7:0]
0x54	BQ10_D2_BYT1[7:0]	0x00	Programmable biquad 10, D2 coefficient byte[31:24]
0x55	BQ10_D2_BYT2[7:0]	0x00	Programmable biquad 10, D2 coefficient byte[23:16]
0x56	BQ10_D2_BYT3[7:0]	0x00	Programmable biquad 10, D2 coefficient byte[15:8]
0x57	BQ10_D2_BYT4[7:0]	0x00	Programmable biquad 10, D2 coefficient byte[7:0]
0x58	BQ11_N0_BYT1[7:0]	0x7F	Programmable biquad 11, N0 coefficient byte[31:24]
0x59	BQ11_N0_BYT2[7:0]	0xFF	Programmable biquad 11, N0 coefficient byte[23:16]
0x5A	BQ11_N0_BYT3[7:0]	0xFF	Programmable biquad 11, N0 coefficient byte[15:8]
0x5B	BQ11_N0_BYT4[7:0]	0xFF	Programmable biquad 11, N0 coefficient byte[7:0]
0x5C	BQ11_N1_BYT1[7:0]	0x00	Programmable biquad 11, N1 coefficient byte[31:24]
0x5D	BQ11_N1_BYT2[7:0]	0x00	Programmable biquad 11, N1 coefficient byte[23:16]
0x5E	BQ11_N1_BYT3[7:0]	0x00	Programmable biquad 11, N1 coefficient byte[15:8]
0x5F	BQ11_N1_BYT4[7:0]	0x00	Programmable biquad 11, N1 coefficient byte[7:0]

Table 7-78. Page 0x03 Programmable Coefficient Registers (continued)

0x60	BQ11_N2_BYT1[7:0]	0x00	Programmable biquad 11, N2 coefficient byte[31:24]
0x61	BQ11_N2_BYT2[7:0]	0x00	Programmable biquad 11, N2 coefficient byte[23:16]
0x62	BQ11_N2_BYT3[7:0]	0x00	Programmable biquad 11, N2 coefficient byte[15:8]
0x63	BQ11_N2_BYT4[7:0]	0x00	Programmable biquad 11, N2 coefficient byte[7:0]
0x64	BQ11_D1_BYT1[7:0]	0x00	Programmable biquad 11, D1 coefficient byte[31:24]
0x65	BQ11_D1_BYT2[7:0]	0x00	Programmable biquad 11, D1 coefficient byte[23:16]
0x66	BQ11_D1_BYT3[7:0]	0x00	Programmable biquad 11, D1 coefficient byte[15:8]
0x67	BQ11_D1_BYT4[7:0]	0x00	Programmable biquad 11, D1 coefficient byte[7:0]
0x68	BQ11_D2_BYT1[7:0]	0x00	Programmable biquad 11, D2 coefficient byte[31:24]
0x69	BQ11_D2_BYT2[7:0]	0x00	Programmable biquad 11, D2 coefficient byte[23:16]
0x6A	BQ11_D2_BYT3[7:0]	0x00	Programmable biquad 11, D2 coefficient byte[15:8]
0x6B	BQ11_D2_BYT4[7:0]	0x00	Programmable biquad 11, D2 coefficient byte[7:0]
0x6C	BQ12_N0_BYT1[7:0]	0x7F	Programmable biquad 12, N0 coefficient byte[31:24]
0x6D	BQ12_N0_BYT2[7:0]	0xFF	Programmable biquad 12, N0 coefficient byte[23:16]
0x6E	BQ12_N0_BYT3[7:0]	0xFF	Programmable biquad 12, N0 coefficient byte[15:8]
0x6F	BQ12_N0_BYT4[7:0]	0xFF	Programmable biquad 12, N0 coefficient byte[7:0]
0x70	BQ12_N1_BYT1[7:0]	0x00	Programmable biquad 12, N1 coefficient byte[31:24]
0x71	BQ12_N1_BYT2[7:0]	0x00	Programmable biquad 12, N1 coefficient byte[23:16]
0x72	BQ12_N1_BYT3[7:0]	0x00	Programmable biquad 12, N1 coefficient byte[15:8]
0x73	BQ12_N1_BYT4[7:0]	0x00	Programmable biquad 12, N1 coefficient byte[7:0]
0x74	BQ12_N2_BYT1[7:0]	0x00	Programmable biquad 12, N2 coefficient byte[31:24]
0x75	BQ12_N2_BYT2[7:0]	0x00	Programmable biquad 12, N2 coefficient byte[23:16]
0x76	BQ12_N2_BYT3[7:0]	0x00	Programmable biquad 12, N2 coefficient byte[15:8]
0x77	BQ12_N2_BYT4[7:0]	0x00	Programmable biquad 12, N2 coefficient byte[7:0]
0x78	BQ12_D1_BYT1[7:0]	0x00	Programmable biquad 12, D1 coefficient byte[31:24]
0x79	BQ12_D1_BYT2[7:0]	0x00	Programmable biquad 12, D1 coefficient byte[23:16]
0x7A	BQ12_D1_BYT3[7:0]	0x00	Programmable biquad 12, D1 coefficient byte[15:8]
0x7B	BQ12_D1_BYT4[7:0]	0x00	Programmable biquad 12, D1 coefficient byte[7:0]
0x7C	BQ12_D2_BYT1[7:0]	0x00	Programmable biquad 12, D2 coefficient byte[31:24]
0x7D	BQ12_D2_BYT2[7:0]	0x00	Programmable biquad 12, D2 coefficient byte[23:16]
0x7E	BQ12_D2_BYT3[7:0]	0x00	Programmable biquad 12, D2 coefficient byte[15:8]
0x7F	BQ12_D2_BYT4[7:0]	0x00	Programmable biquad 12, D2 coefficient byte[7:0]

7.2.3 Programmable Coefficient Registers: Page = 0x04

This register page (shown in [Table 7-79](#)) consists of the programmable coefficients for mixer 1 to mixer 4 and the first-order IIR filter. All mixer coefficients are 32-bit, two's complement numbers using a 1.31 number format. The value of 0x7FFFFFFF is equivalent to +1 (0dB gain), the value 0x00000000 is equivalent to mute (zero data) and all values in between set the mixer attenuation computed using [Equation 4](#). If the MSB is set to '1' then the attenuation remains the same but the signal phase is inverted. All IIR filter programmable coefficients are 32-bit, two's complement numbers. For a successful coefficient register transaction, the host device must write and read all the four bytes starting with the most significant byte (BYT1) for a target coefficient register transaction. While using SPI for a coefficient register read transaction, the device gives out first byte as dummy read byte therefore the host must read five bytes which includes first byte as dummy read byte and the last four bytes corresponds to the coefficient register value starting with the most significant byte (BYT1).

$$\text{hex2dec}(\text{value}) / 2^{31} \quad (4)$$

Table 7-79. Page 0x04 Programmable Coefficient Registers

ADDR	REGISTER	RESET	DESCRIPTION
0x00	PAGE[7:0]	0x00	Device page register
0x08	MIX1_CH1_BYT1[7:0]	0x7F	Digital mixer 1, channel 1 coefficient byte[31:24]
0x09	MIX1_CH1_BYT2[7:0]	0xFF	Digital mixer 1, channel 1 coefficient byte[23:16]
0x0A	MIX1_CH1_BYT3[7:0]	0xFF	Digital mixer 1, channel 1 coefficient byte[15:8]
0x0B	MIX1_CH1_BYT4[7:0]	0xFF	Digital mixer 1, channel 1 coefficient byte[7:0]
0x0C	MIX1_CH2_BYT1[7:0]	0x00	Digital mixer 1, channel 2 coefficient byte[31:24]
0x0D	MIX1_CH2_BYT2[7:0]	0x00	Digital mixer 1, channel 2 coefficient byte[23:16]
0x0E	MIX1_CH2_BYT3[7:0]	0x00	Digital mixer 1, channel 2 coefficient byte[15:8]
0x0F	MIX1_CH2_BYT4[7:0]	0x00	Digital mixer 1, channel 2 coefficient byte[7:0]
0x10	MIX1_CH3_BYT1[7:0]	0x00	Digital mixer 1, channel 3 coefficient byte[31:24]
0x11	MIX1_CH3_BYT2[7:0]	0x00	Digital mixer 1, channel 3 coefficient byte[23:16]
0x12	MIX1_CH3_BYT3[7:0]	0x00	Digital mixer 1, channel 3 coefficient byte[15:8]
0x13	MIX1_CH3_BYT4[7:0]	0x00	Digital mixer 1, channel 3 coefficient byte[7:0]
0x14	MIX1_CH4_BYT1[7:0]	0x00	Digital mixer 1, channel 4 coefficient byte[31:24]
0x15	MIX1_CH4_BYT2[7:0]	0x00	Digital mixer 1, channel 4 coefficient byte[23:16]
0x16	MIX1_CH4_BYT3[7:0]	0x00	Digital mixer 1, channel 4 coefficient byte[15:8]
0x17	MIX1_CH4_BYT4[7:0]	0x00	Digital mixer 1, channel 4 coefficient byte[7:0]
0x18	MIX2_CH1_BYT1[7:0]	0x00	Digital mixer 2, channel 1 coefficient byte[31:24]
0x19	MIX2_CH1_BYT2[7:0]	0x00	Digital mixer 2, channel 1 coefficient byte[23:16]
0x1A	MIX2_CH1_BYT3[7:0]	0x00	Digital mixer 2, channel 1 coefficient byte[15:8]
0x1B	MIX2_CH1_BYT4[7:0]	0x00	Digital mixer 2, channel 1 coefficient byte[7:0]
0x1C	MIX2_CH2_BYT1[7:0]	0x7F	Digital mixer 2, channel 2 coefficient byte[31:24]
0x1D	MIX2_CH2_BYT2[7:0]	0xFF	Digital mixer 2, channel 2 coefficient byte[23:16]
0x1E	MIX2_CH2_BYT3[7:0]	0xFF	Digital mixer 2, channel 2 coefficient byte[15:8]
0x1F	MIX2_CH2_BYT4[7:0]	0xFF	Digital mixer 2, channel 2 coefficient byte[7:0]
0x20	MIX2_CH3_BYT1[7:0]	0x00	Digital mixer 2, channel 3 coefficient byte[31:24]
0x21	MIX2_CH3_BYT2[7:0]	0x00	Digital mixer 2, channel 3 coefficient byte[23:16]
0x22	MIX2_CH3_BYT3[7:0]	0x00	Digital mixer 2, channel 3 coefficient byte[15:8]
0x23	MIX2_CH3_BYT4[7:0]	0x00	Digital mixer 2, channel 3 coefficient byte[7:0]
0x24	MIX2_CH4_BYT1[7:0]	0x00	Digital mixer 2, channel 4 coefficient byte[31:24]
0x25	MIX2_CH4_BYT2[7:0]	0x00	Digital mixer 2, channel 4 coefficient byte[23:16]
0x26	MIX2_CH4_BYT3[7:0]	0x00	Digital mixer 2, channel 4 coefficient byte[15:8]
0x27	MIX2_CH4_BYT4[7:0]	0x00	Digital mixer 2, channel 4 coefficient byte[7:0]
0x28	MIX3_CH1_BYT1[7:0]	0x00	Digital mixer 3, channel 1 coefficient byte[31:24]
0x29	MIX3_CH1_BYT2[7:0]	0x00	Digital mixer 3, channel 1 coefficient byte[23:16]
0x2A	MIX3_CH1_BYT3[7:0]	0x00	Digital mixer 3, channel 1 coefficient byte[15:8]

Table 7-79. Page 0x04 Programmable Coefficient Registers (continued)

0x2B	MIX3_CH1_BYT4[7:0]	0x00	Digital mixer 3, channel 1 coefficient byte[7:0]
0x2C	MIX3_CH2_BYT1[7:0]	0x00	Digital mixer 3, channel 2 coefficient byte[31:24]
0x2D	MIX3_CH2_BYT2[7:0]	0x00	Digital mixer 3, channel 2 coefficient byte[23:16]
0x2E	MIX3_CH2_BYT3[7:0]	0x00	Digital mixer 3, channel 2 coefficient byte[15:8]
0x2F	MIX3_CH2_BYT4[7:0]	0x00	Digital mixer 3, channel 2 coefficient byte[7:0]
0x30	MIX3_CH3_BYT1[7:0]	0x7F	Digital mixer 3, channel 3 coefficient byte[31:24]
0x31	MIX3_CH3_BYT2[7:0]	0xFF	Digital mixer 3, channel 3 coefficient byte[23:16]
0x32	MIX3_CH3_BYT3[7:0]	0xFF	Digital mixer 3, channel 3 coefficient byte[15:8]
0x33	MIX3_CH3_BYT4[7:0]	0xFF	Digital mixer 3, channel 3 coefficient byte[7:0]
0x34	MIX3_CH4_BYT1[7:0]	0x00	Digital mixer 3, channel 4 coefficient byte[31:24]
0x35	MIX3_CH4_BYT2[7:0]	0x00	Digital mixer 3, channel 4 coefficient byte[23:16]
0x36	MIX3_CH4_BYT3[7:0]	0x00	Digital mixer 3, channel 4 coefficient byte[15:8]
0x37	MIX3_CH4_BYT4[7:0]	0x00	Digital mixer 3, channel 4 coefficient byte[7:0]
0x38	MIX4_CH1_BYT1[7:0]	0x00	Digital mixer 4, channel 1 coefficient byte[31:24]
0x39	MIX4_CH1_BYT2[7:0]	0x00	Digital mixer 4, channel 1 coefficient byte[23:16]
0x3A	MIX4_CH1_BYT3[7:0]	0x00	Digital mixer 4, channel 1 coefficient byte[15:8]
0x3B	MIX4_CH1_BYT4[7:0]	0x00	Digital mixer 4, channel 1 coefficient byte[7:0]
0x3C	MIX4_CH2_BYT1[7:0]	0x00	Digital mixer 4, channel 2 coefficient byte[31:24]
0x3D	MIX4_CH2_BYT2[7:0]	0x00	Digital mixer 4, channel 2 coefficient byte[23:16]
0x3E	MIX4_CH2_BYT3[7:0]	0x00	Digital mixer 4, channel 2 coefficient byte[15:8]
0x3F	MIX4_CH2_BYT4[7:0]	0x00	Digital mixer 4, channel 2 coefficient byte[7:0]
0x40	MIX4_CH3_BYT1[7:0]	0x00	Digital mixer 4, channel 3 coefficient byte[31:24]
0x41	MIX4_CH3_BYT2[7:0]	0x00	Digital mixer 4, channel 3 coefficient byte[23:16]
0x42	MIX4_CH3_BYT3[7:0]	0x00	Digital mixer 4, channel 3 coefficient byte[15:8]
0x43	MIX4_CH3_BYT4[7:0]	0x00	Digital mixer 4, channel 3 coefficient byte[7:0]
0x44	MIX4_CH4_BYT1[7:0]	0x7F	Digital mixer 4, channel 4 coefficient byte[31:24]
0x45	MIX4_CH4_BYT2[7:0]	0xFF	Digital mixer 4, channel 4 coefficient byte[23:16]
0x46	MIX4_CH4_BYT3[7:0]	0xFF	Digital mixer 4, channel 4 coefficient byte[15:8]
0x47	MIX4_CH4_BYT4[7:0]	0xFF	Digital mixer 4, channel 4 coefficient byte[7:0]
0x48	IIR_N0_BYT1[7:0]	0x7F	Programmable first-order IIR, N0 coefficient byte[31:24]
0x49	IIR_N0_BYT2[7:0]	0xFF	Programmable first-order IIR, N0 coefficient byte[23:16]
0x4A	IIR_N0_BYT3[7:0]	0xFF	Programmable first-order IIR, N0 coefficient byte[15:8]
0x4B	IIR_N0_BYT4[7:0]	0xFF	Programmable first-order IIR, N0 coefficient byte[7:0]
0x4C	IIR_N1_BYT1[7:0]	0x00	Programmable first-order IIR, N1 coefficient byte[31:24]
0x4D	IIR_N1_BYT2[7:0]	0x00	Programmable first-order IIR, N1 coefficient byte[23:16]
0x4E	IIR_N1_BYT3[7:0]	0x00	Programmable first-order IIR, N1 coefficient byte[15:8]
0x4F	IIR_N1_BYT4[7:0]	0x00	Programmable first-order IIR, N1 coefficient byte[7:0]
0x50	IIR_D1_BYT1[7:0]	0x00	Programmable first-order IIR, D1 coefficient byte[31:24]
0x51	IIR_D1_BYT2[7:0]	0x00	Programmable first-order IIR, D1 coefficient byte[23:16]
0x52	IIR_D1_BYT3[7:0]	0x00	Programmable first-order IIR, D1 coefficient byte[15:8]
0x53	IIR_D1_BYT4[7:0]	0x00	Programmable first-order IIR, D1 coefficient byte[7:0]

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

8.1 Application Information

The PCMD3180 is a multichannel, pulse-density-modulation (PDM) input to time-division multiplexing (TDM) or I²S audio output converter that supports output sample rates of up to 768kHz. The device supports up to eight digital pulse density modulation (PDM) microphones for simultaneous recording applications.

Communication to the PCMD3180 for configuration of the control registers is supported using an I²C or SPI. The device supports a highly flexible, audio serial interface (TDM, I²S, and LJ) to transmit audio data seamlessly in the system across devices.

8.2 Typical Applications

8.2.1 Eight-Channel Digital PDM Microphone Recording

Figure 8-1 shows a typical configuration of the PCMD3180 for an application using eight digital PDM MEMS microphones with simultaneous recording operation using an I²C control interface and the TDM audio data slave interface. If the MICBIAS output is not used in the system then the 1 μ F capacitor for the MICBIAS pin is not must.

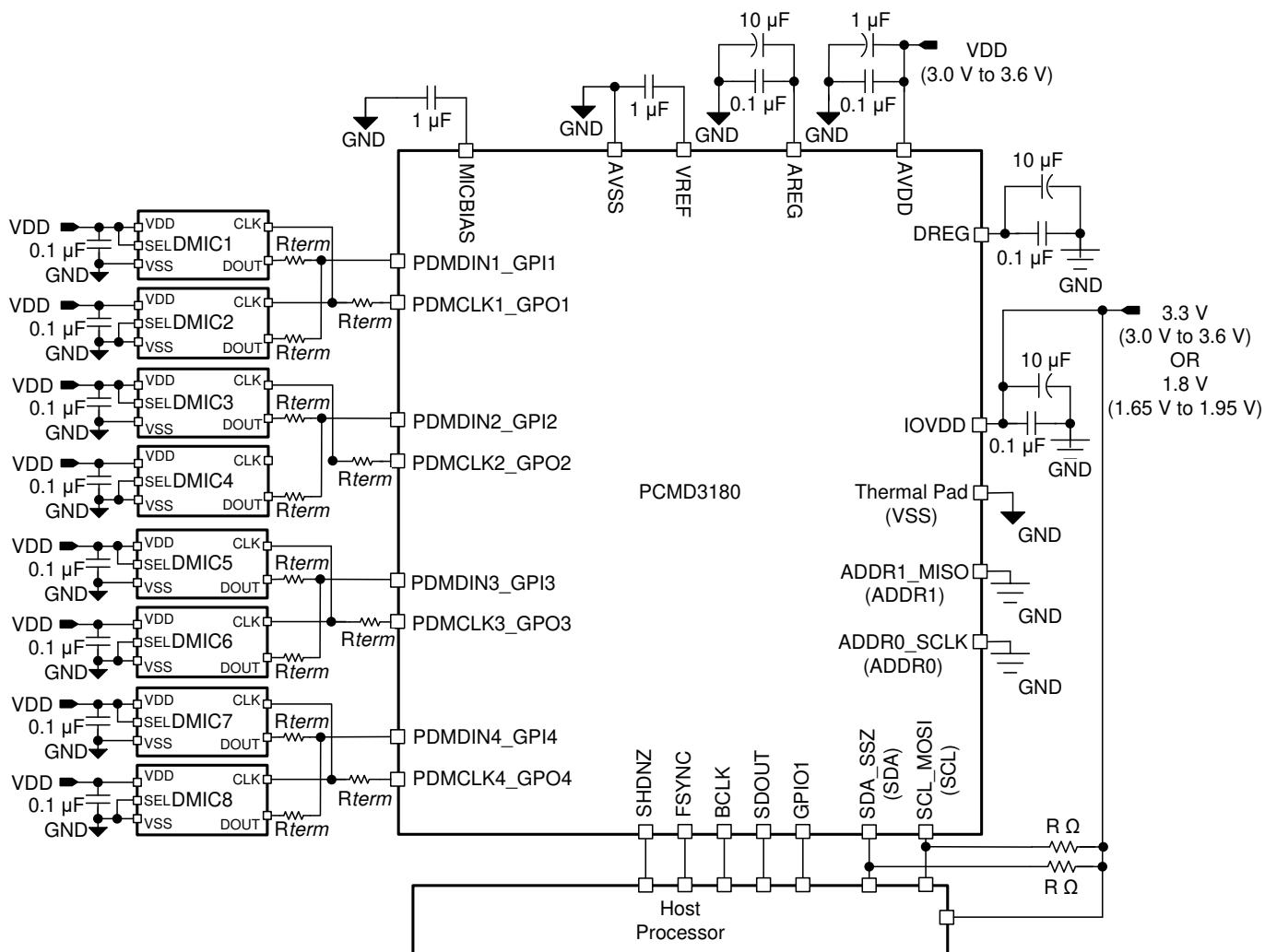


Figure 8-1. Eight-Channel Digital PDM Microphone Recording Diagram

8.2.1.1 Design Requirements

The supply decoupling capacitors must be used ceramic type with low ESR. Table 8-1 lists the design parameters for this application.

Table 8-1. Design Parameters

KEY PARAMETER	SPECIFICATION
AVDD	3.3V
AVDD supply current consumption	11.7mA (PLL on, eight-channel recording, $f_S = 48\text{kHz}$, $\text{PDMCLK}_x = 64 \times f_S$)
IOVDD	1.8V or 3.3V

8.2.1.2 Detailed Design Procedure

This section describes the necessary steps to configure the PCMD3180 for this specific application. The following steps provide a sequence of items that must be executed in the time between powering the device up and reading data from the device or transitioning from one mode to another mode of operation.

1. Apply power to the device:
 - a. Power up the IOVDD and AVDD power supplies, keeping the SHDNZ pin voltage low
 - b. The device now goes into hardware shutdown mode (ultra-low-power mode < μ A)
2. Transition from hardware shutdown mode to sleep mode (or software shutdown mode):
 - a. Release SHDNZ only when the IOVDD and AVDD power supplies settle to the steady-state operating voltage
 - b. Wait for at least 1ms to allow the device to initialize the internal registers initialization
 - c. The device now goes into sleep mode (low-power mode < 10 μ A)
3. Transition from sleep mode to active mode whenever required for the recording operation:
 - a. Wake up the device by writing to P0_R2 to disable sleep mode
 - b. Wait for at least 1ms to allow the device to complete the internal wake-up sequence
 - c. Override the default configuration registers or programmable coefficients value as required (this step is optional)
 - d. Configure channel 1 to channel 4 (CHx_INSRC) for the digital microphone as the input source for recording
 - e. Configure GPO1 to GPO4 (GPOx_CFG) as the PDMCLK output
 - f. Configure GPI1 to GPI4 (GPI1x_CFG) as PDMDIN1 to PDMDIN4, respectively
 - g. Enable all desired input channels by writing to P0_R115
 - h. Enable all desired audio serial interface output channels by writing to P0_R116
 - i. Power-up the PDM converter and PLL by writing to P0_R117
 - j. Apply FSYNC and BCLK with the desired output sample rates and the BCLK to FSYNC ratio

This specific step can be done at any point in the sequence after step a.

See the [Phase-Locked Loop \(PLL\) and Clock Generation](#) section for supported sample rates and the BCLK to FSYNC ratio.

- k. The device recording data is now sent to the host processor using the TDM audio serial data bus
4. Transition from active mode to sleep mode (again) as required in the system for low-power operation:
 - a. Enter sleep mode by writing to P0_R2 to enable sleep mode
 - b. Wait at least 6ms (when FSYNC = 48kHz) for the volume to ramp down and for all blocks to power down
 - c. Read P0_R119 to check the device shutdown and sleep mode status
 - d. If the device P0_R119_D7 status bit is 1'b1 then stop FSYNC and BCLK in the system
 - e. The device now goes into sleep mode (low-power mode < 10 μ A) and retains all register values
5. Transition from sleep mode to active mode (again) as required for the recording operation:
 - a. Wake up the device by writing to P0_R2 to disable sleep mode
 - b. Wait at least 1ms to allow the device to complete the internal wake-up sequence
 - c. Apply FSYNC and BCLK with the desired output sample rates and the BCLK to FSYNC ratio
 - d. The device recording data are now sent to the host processor using the TDM audio serial data bus
6. Repeat step 4 and step 5 as required for mode transitions
7. Assert the SHDNZ pin low to enter hardware shutdown mode (again) at any time
8. Follow step 2 onwards to exit hardware shutdown mode (again)

8.2.1.2.1 Example Device Register Configuration Script for EVM Setup

This section provides a typical EVM I²C register control script that shows how to set up the PCMD3180 in an eight-channel digital PDM microphone recording mode.

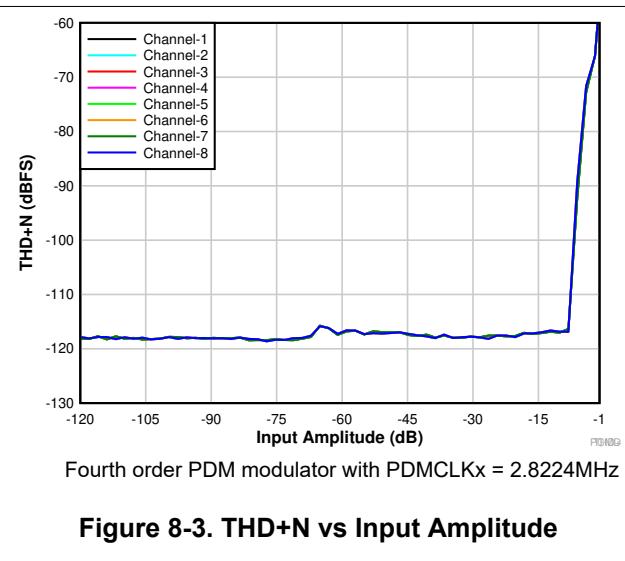
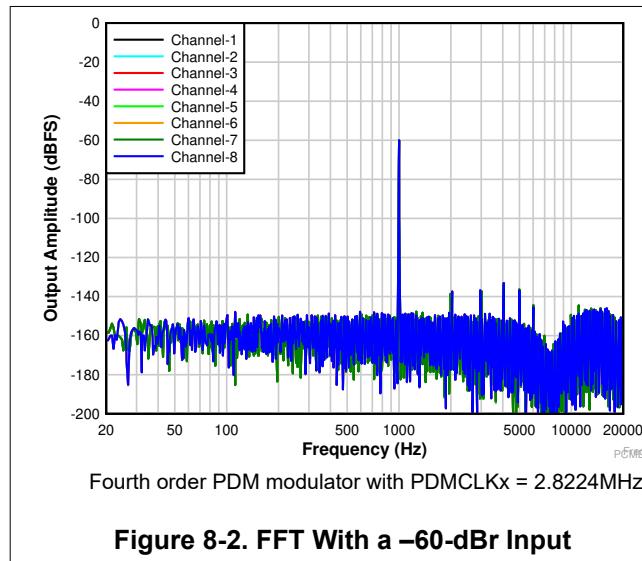
```

# Key: w 98 XX YY ==> write to I2C address 0x98, to register 0XXX, data 0XYY
#           # ==> comment delimiter
#
# The following list gives an example sequence of items that must be executed in the time
# between powering the device up and reading data from the device. Note that there are
# other valid sequences depending on which features are used.
#
# See the PCMD3180EVM user guide for jumper settings and audio connections.
#
# PDM 8-channel : PDMDIN1 - Ch1 and Ch2, PDMDIN2 - Ch3 and Ch4,
#                 PDMDIN3 - Ch5 and Ch6, PDMDIN4 - Ch7 and Ch8
# PDMCLKx = 2.8224MHz (PDMCLKx/FSYNC = 64)
# FSYNC = 44.1kHz (Output Data Sample Rate), BCLK = 11.2896MHz (BCLK/FSYNC = 256)
#####
#
#
# Power up IOVDD and AVDD power supplies keeping SHDNZ pin voltage LOW
# Wait for IOVDD and AVDD power supplies to settle to steady state operating voltage range.
# Release SHDNZ to HIGH.
# Wait for 1ms.
#
# Wake-up device by I2C write into P0_R2 using internal AREG
w 98 02 81
#
# Configure CH1_INSRC as Digital PDM Input by I2C write into P0_R60
w 98 3C 40
#
# Configure CH2_INSRC as Digital PDM Input by I2C write into P0_R65
w 98 41 40
#
# Configure CH3_INSRC as Digital PDM Input by I2C write into P0_R70
w 98 46 40
#
# Configure CH4_INSRC as Digital PDM Input by I2C write into P0_R75
w 98 4B 40
#
# Configure PDMCLK1_GPO1 as PDMCLK by I2C write into P0_R34
w 98 22 41
#
# Configure PDMCLK1_GPO2 as PDMCLK by I2C write into P0_R35
w 98 23 41
#
# Configure PDMCLK1_GPO3 as PDMCLK by I2C write into P0_R36
w 98 24 41
#
# Configure PDMCLK1_GPO4 as PDMCLK by I2C write into P0_R37
w 98 25 41
#
# Configure PDMDIN1_GPI1 and PDMDIN2_GPI2 as PDMDIN1 and PDMDIN2 by I2C write into P0_R43
w 98 2B 45
#
# Configure PDMDIN3_GPI3 and PDMDIN4_GPI4 as PDMDIN3 and PDMDIN4 by I2C write into P0_R44
w 98 2C 67
#
# Enable Input Ch-1 to Ch-8 by I2C write into P0_R115
w 98 73 FF
#
# Enable ASI output Ch-1 to Ch-8 slots by I2C write into P0_R116
w 98 74 FF
#
# Power-up PDM converter and PLL by I2C write into P0_R117
w 98 75 60
#
# Apply FSYNC = 44.1kHz and BCLK = 11.2896MHz and
# Start recording data by host on ASI bus with TDM protocol 32-bits channel wordlength

```

8.2.1.3 Application Curves

Measurements are done on the EVM by feeding the device PDM digital input signal using audio precision. In the system application, the device performance is expected to be limited by single-bit PDM modulator digital microphone output performance.



8.3 What to Do and What Not to Do

In master mode operation with I²S or LJ format, the device generates FSYNC half a cycle earlier than the normal protocol timing behavior expected. This timing behavior can still function for most of the system, however for further details and a suggested workaround for this weakness, see the [Configuring and Operating the TLV320ADCx140 as Audio Bus Master application note](#).

8.4 Power Supply Recommendations

The power-supply sequence between the IOVDD and AVDD rails can be applied in any order. However, keep the SHDNZ pin low until the IOVDD supply voltage settles to a stable and supported operating voltage range. After all supplies are stable, set the SHDNZ pin high to initialize the device.

For the supply power-up requirement, t_1 and t_2 must be at least 100 μ s. For the supply power-down requirement, t_3 and t_4 must be at least 10ms. This timing (as shown in Figure 8-4) allows the device to ramp down the volume on the record data, power down the analog and digital blocks, and put the device into hardware shutdown mode. The device can also be immediately put into hardware shutdown mode from active mode if SHDNZ_CFG[1:0] is set to 2'b00 using the P0_R5_D[3:2] bits. In that case, t_3 and t_4 are required to be at least 100 μ s.

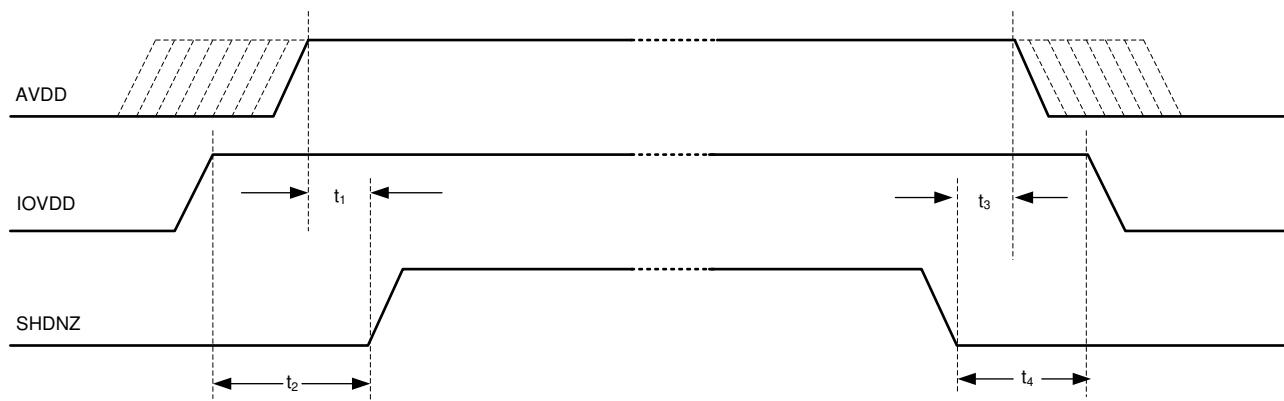


Figure 8-4. Power-Supply Sequencing Requirement Timing Diagram

Make sure that the supply ramp rate is slower than 1V/ μ s and that the wait time between a power-down and a power-up event is at least 100ms. For supply ramp rate slower than 0.1V/ms, host device must apply a software reset as first transaction before doing any device configuration.

After releasing SHDNZ, or after a software reset, delay any additional I²C or SPI transactions to the device for at least 2ms to allow the device to initialize the internal registers. See the [Device Functional Modes](#) section for details on how the device operates in various modes after the device power supplies are settled to the recommended operating voltage levels.

The PCMD3180 supports a single AVDD supply operation by integrating an on-chip digital regulator, DREG, and an analog regulator, AREG. However, if the AVDD voltage is less than 1.98V in the system, then short the AREG and AVDD pins onboard and do not enable the internal AREG by keeping the AREG_SELECT bit to 1b0 (default value) of P0_R2. If the AVDD supply used in the system is higher than 2.7V, then the host device can set AREG_SELECT to 1'b1 while exiting sleep mode to allow the device internal regulator to generate the AREG supply.

8.5 Layout

8.5.1 Layout Guidelines

Each system design and printed circuit board (PCB) layout is unique. The layout must be carefully reviewed in the context of a specific PCB design. However, the following guidelines can optimize the device performance:

- Connect the thermal pad to ground. Use a via pattern to connect the device thermal pad, which is the area directly under the device, to the ground planes. This connection helps dissipate heat from the device.
- The decoupling capacitors for the power supplies must be placed close to the device pins.

- The supply decoupling capacitors must be used ceramic type with low ESR.
- Route the analog differential audio signals differentially on the PCB for better noise immunity. Avoid crossing digital and analog signals to prevent undesirable crosstalk.
- The device internal voltage references must be filtered using external capacitors. Place the filter capacitors near the VREF pin for best performance.
- Directly tap the MICBIAS pin to avoid common impedance when routing the biasing or supply traces for multiple microphones to avoid coupling across microphones.
- Directly short the VREF and MICBIAS external capacitors ground terminal to the AVSS pin without using any vias for this connection trace.
- Place the MICBIAS capacitor (with low equivalent series resistance) close to the device with minimal trace impedance.
- Use ground planes to provide the lowest impedance for power and signal current between the device and the decoupling capacitors. Treat the area directly under the device as a central ground area for the device, and all device grounds must be connected directly to that area.

8.5.2 Layout Example

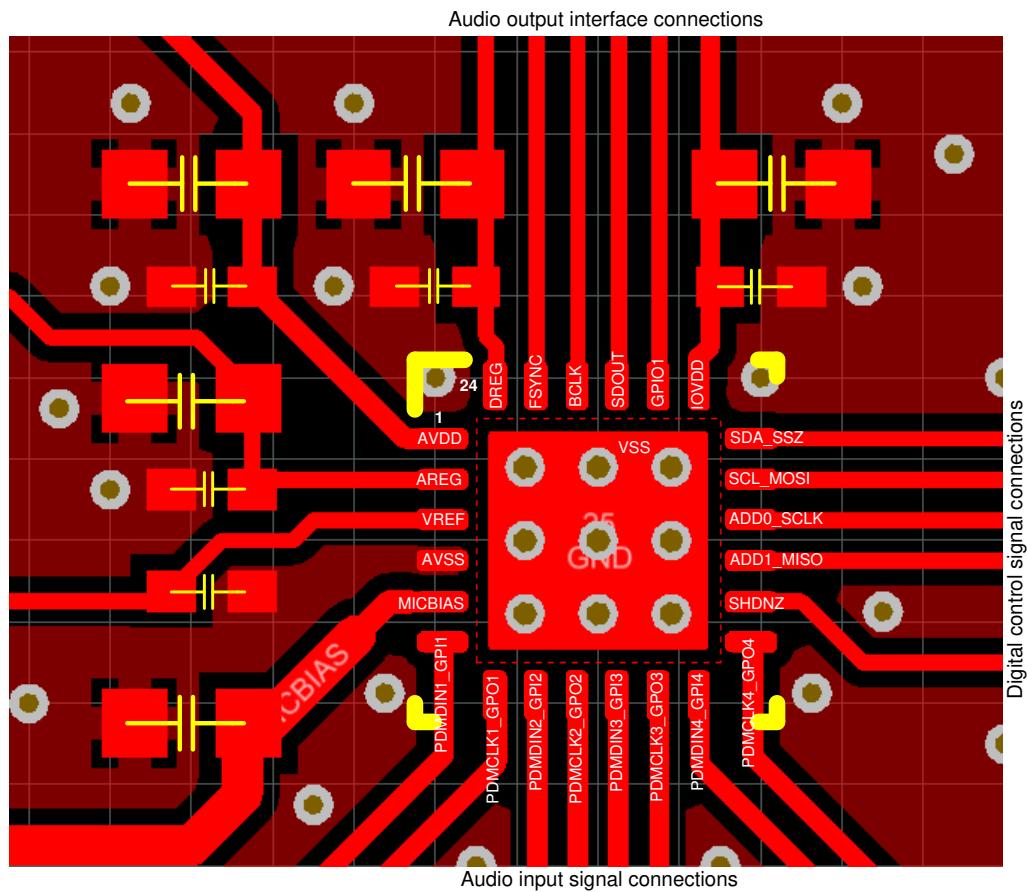


Figure 8-5. Layout Example

9 Device and Documentation Support

9.1 Documentation Support

9.1.1 Related Documentation

For related documentation see the following:

- Texas Instruments, [Multiple TLV320ADCx140 Devices With Shared TDM and I²C Bus](#) application note
- Texas Instruments, [Configuring and Operating the TLV320ADCx140 as Audio Bus Master](#) application note
- Texas Instruments, [TLV320ADCx140 Sampling Rates and Programmable Processing Blocks Supported](#) application note
- Texas Instruments, [TLV320ADCx140 Programmable Bi-Quad Filter Configuration and Application](#) application note
- Texas Instruments, [Analog Microphone and ADC System in Far-field Application](#) application note
- Texas Instruments, [TLV320ADCx140 Power Consumption Matrix Across Various Usage Scenarios](#) application note
- Texas Instruments, [ADCx140EVM-PDK](#) user's guide
- Texas Instruments, [PurePath™ Console Graphical Development Suite for Audio System Design and Development](#)

9.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on *Notifications* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

9.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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9.4 Trademarks

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9.5 Electrostatic Discharge Caution

 This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

9.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

10 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision * (May 2020) to Revision A (January 2026)	Page
• Updated the numbering format for tables, figures, and cross-references throughout the document.....	1

11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable part number	Status (1)	Material type (2)	Package Pins	Package qty Carrier	RoHS (3)	Lead finish/ Ball material (4)	MSL rating/ Peak reflow (5)	Op temp (°C)	Part marking (6)
PCMD3180IRTWR	Active	Production	WQFN (RTW) 24	3000 LARGE T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PDM3180
PCMD3180IRTWR.A	Active	Production	null (null)	3000 LARGE T&R	-	NIPDAU	Level-1-260C-UNLIM	See PCMD3180IRTWR	PDM3180
PCMD3180IRTWT	Active	Production	WQFN (RTW) 24	250 SMALL T&R	Yes	NIPDAU	Level-1-260C-UNLIM	-40 to 125	PDM3180
PCMD3180IRTWT.A	Active	Production	null (null)	250 SMALL T&R	-	NIPDAU	Level-1-260C-UNLIM	See PCMD3180IRTWT	PDM3180

⁽¹⁾ **Status:** For more details on status, see our [product life cycle](#).

⁽²⁾ **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

⁽³⁾ **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

⁽⁴⁾ **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

⁽⁵⁾ **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

⁽⁶⁾ **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

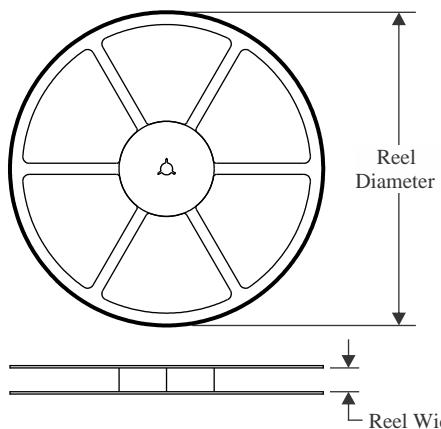
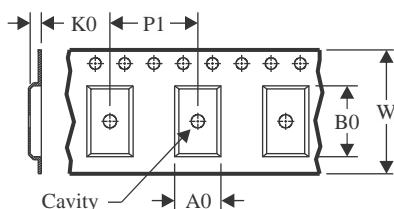
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF PCMD3180 :

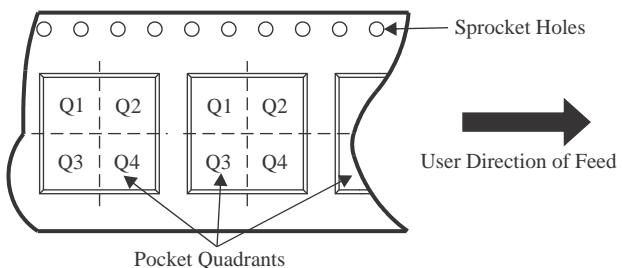
- Automotive : [PCMD3180-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

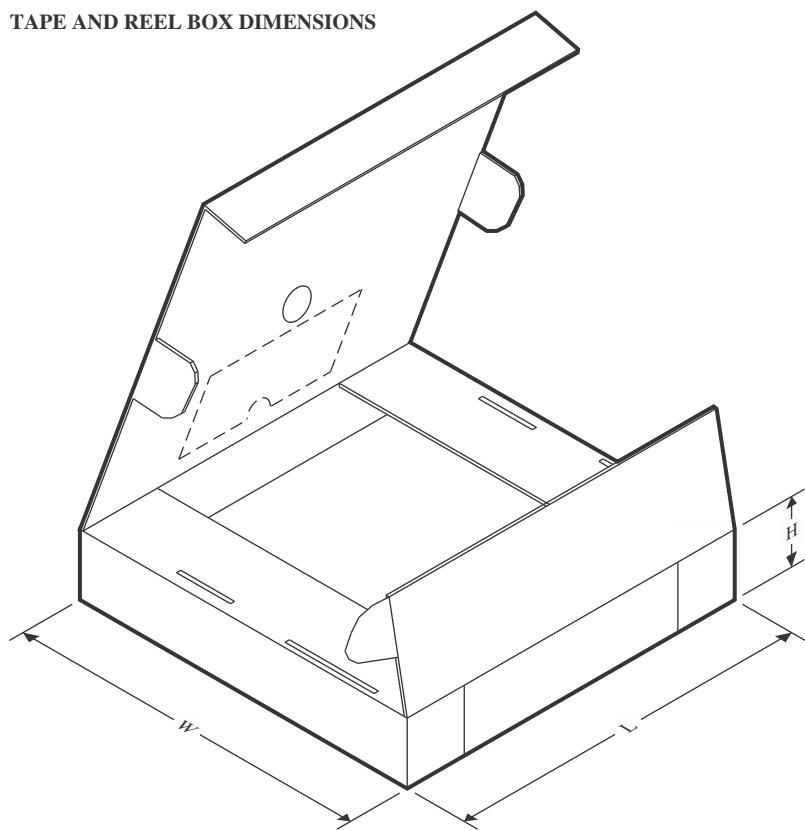
TAPE AND REEL INFORMATION
REEL DIMENSIONS

TAPE DIMENSIONS


A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
PCMD3180IRTWR	WQFN	RTW	24	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
PCMD3180IRTWT	WQFN	RTW	24	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
PCMD3180IRTWR	WQFN	RTW	24	3000	367.0	367.0	35.0
PCMD3180IRTWT	WQFN	RTW	24	250	210.0	185.0	35.0

GENERIC PACKAGE VIEW

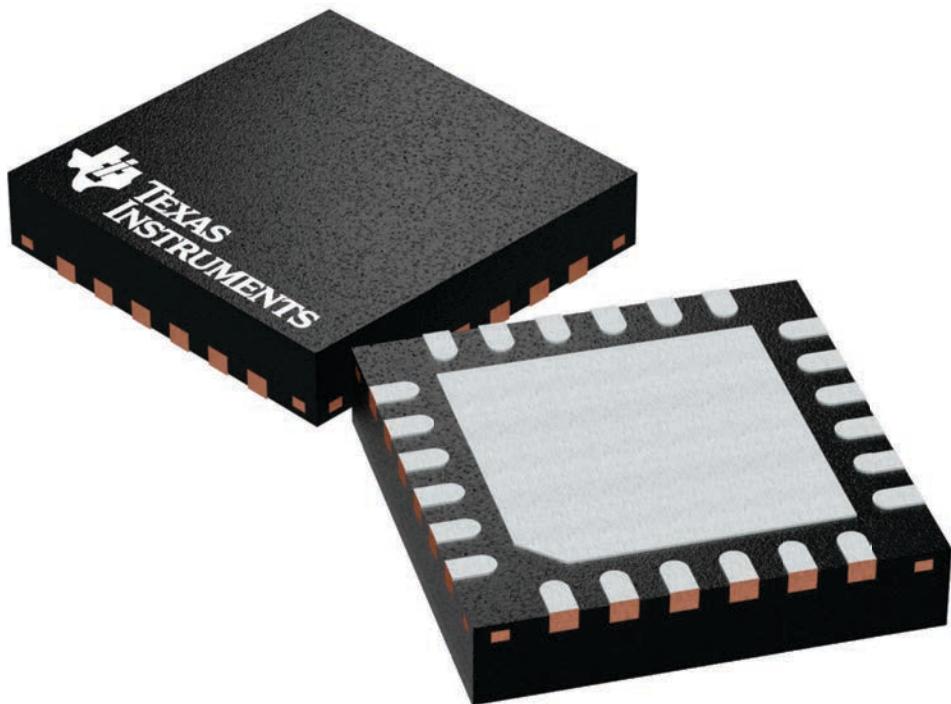
RTW 24

WQFN - 0.8 mm max height

4 x 4, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD

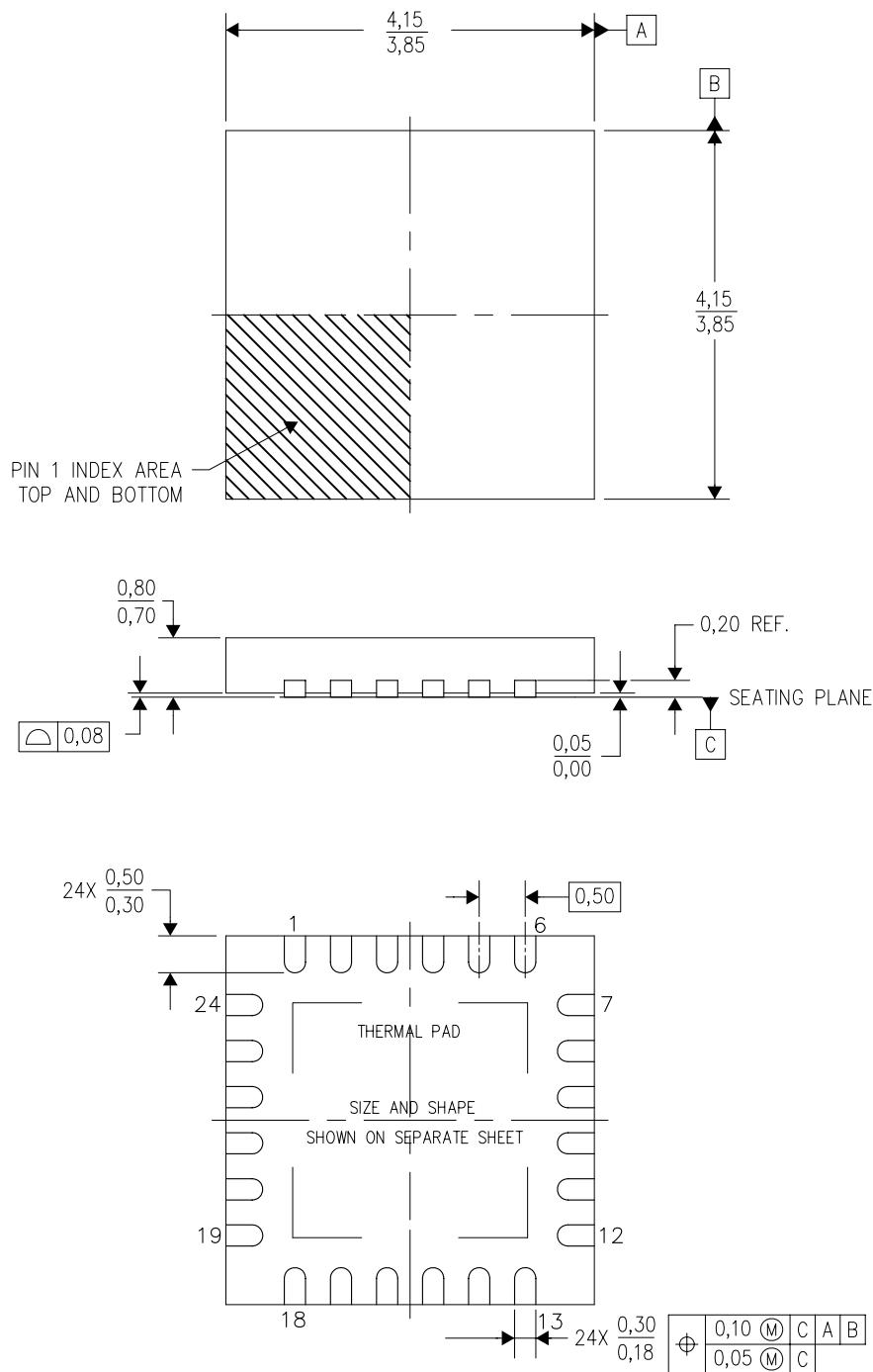
This image is a representation of the package family, actual package may vary.
Refer to the product data sheet for package details.



4224801/A

RTW (S-PWQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



4206244/C 07/11

NOTES:

- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
- This drawing is subject to change without notice.
- Quad Flatpack, No-Leads (QFN) package configuration.
- The package thermal pad must be soldered to the board for thermal and mechanical performance.
- See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- Falls within JEDEC MO-220.

RTW (S-PWQFN-N24)

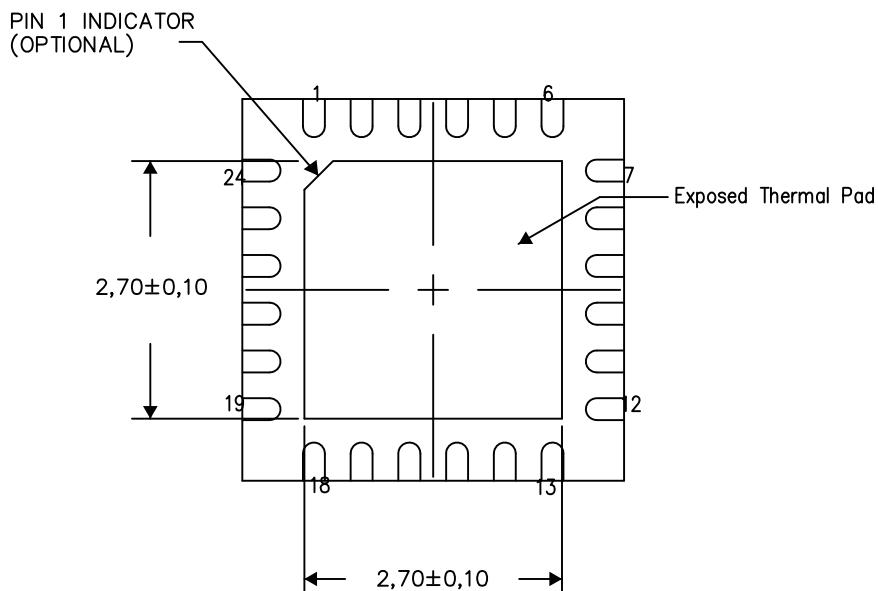
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View
Exposed Thermal Pad Dimensions

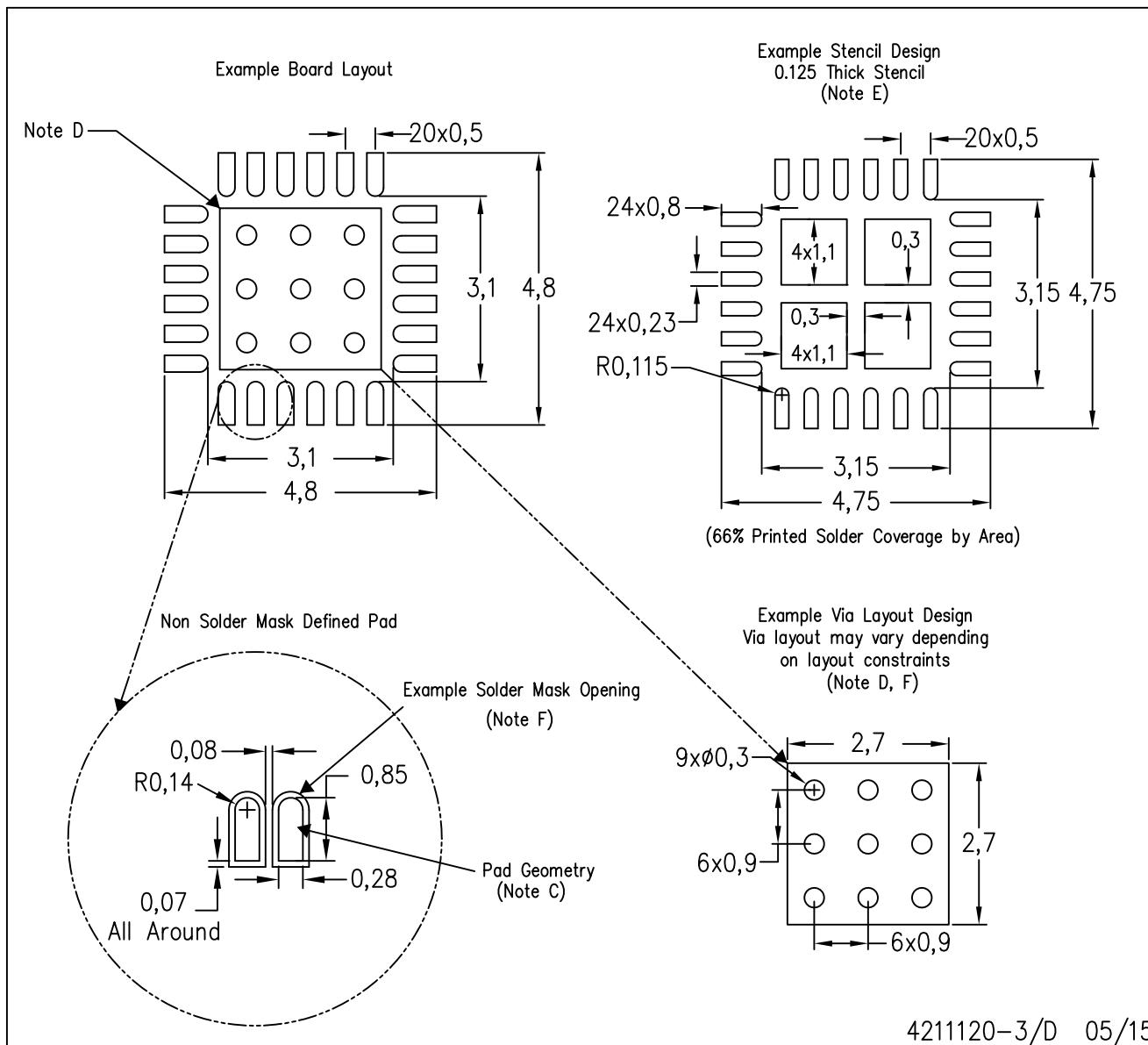
4206249-5/P 05/15

NOTES: A. All linear dimensions are in millimeters

LAND PATTERN DATA

RTW (S-PWQFN-N24)

PLASTIC QUAD FLATPACK NO-LEAD



4211120-3/D 05/15

NOTES:

- All linear dimensions are in millimeters.
- This drawing is subject to change without notice.
- Publication IPC-7351 is recommended for alternate designs.
- This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
- Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.

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