

LMx93-N、LM2903-N 低功耗低失调电压双路比较器

1 特性

- 宽电源
 - 电压范围: 2.0V 至 36V
 - 单电源或双电源: $\pm 1.0V$ 至 $\pm 18V$
- 极低的电源电流消耗 (0.4mA) - 与电源电压无关
- 低输入偏置电流: 25nA
- 低输入失调电流: $\pm 5nA$
- 最大失调电压: $\pm 3mV$
- 输入共模电压范围包括接地
- 差分输入电压范围等于电源电压
- 低输出饱和电压: 250mV (电流为 4mA)
- 输出电压与 TTL、DTL、ECL、MOS 和 CMOS 逻辑系统兼容
- 采用 8 凸点 (12 密耳) DSBGA 封装
- 请参阅 AN-1112 (SNVA009) 中的 DSBGA 注意事项
- 优势
 - 高精度比较器
 - V_{OS} 温漂更低
 - 消除了对双电源的需求
 - 允许接近接地检测
 - 与所有形式的逻辑兼容
 - 具有适用于电池供电的功耗

2 应用

- 电池供电类 应用
- 工业 应用

3 说明

LM193-N 系列由两个独立的精密电压比较器组成，两个比较器的失调电压规格低至 2.0mV (最大值)，这些比较器专为在宽电压范围内由单个电源供电而设计。也可以由分离式电源供电，并且低电源电流消耗与电源电压的大小无关。这些比较器还具有独特的特性：即使由单电源电压供电，输入共模电压范围也包括接地。

应用领域包括限制比较器、简单模数转换器；脉冲、方波和时间延迟发生器；各种 VCO；MOS 时钟计时器；多谐振荡器和高压数字逻辑门。LM193-N 系列可以直接连接 TTL 和 CMOS。由正负电源供电时，LM19-N 系列将直接连接 MOS 逻辑，其低功耗相对于标准比较器具有明显的优势。

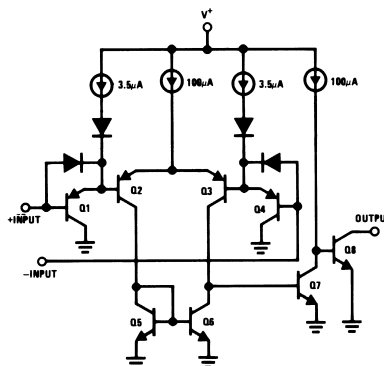
LM393 和 LM2903 器件采用 TI 创新的具有 8 个 (12 密耳) 大凸点的薄 DSBGA 封装。

器件信息⁽¹⁾

器件型号	封装	封装尺寸 (标称值)
LM193-N	TO-99 (8)	9.08mm x 9.08mm
LM293-N		
LM393-N	SOIC (8)	4.90mm x 3.91mm
	DSBGA (8)	1.54mm x 1.54mm
LM2903-N	SOIC (8)	4.90mm x 3.91mm
	DSBGA (8)	1.54mm x 1.54mm

(1) 如需了解所有可用封装，请参阅数据表末尾的可订购产品附录。

简化原理图



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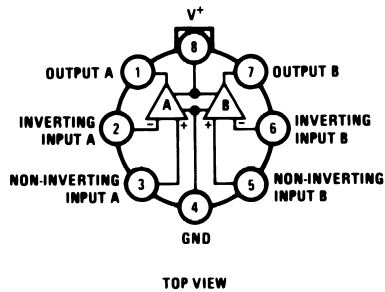
4 修订历史记录

注：之前版本的页码可能与当前版本有所不同。

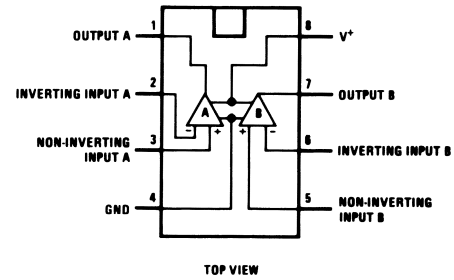
Changes from Revision F (December 2014) to Revision G	Page
• 已添加 添加了在格式转换期间“器件信息”表中无意漏掉的 DSBGA 封装	1
Changes from Revision E (March 2013) to Revision F	Page
• 已添加 引脚配置和功能 部分、ESD 额定值 表、特性 说明 部分、器件功能模式、应用和实施 部分、电源相关建议 部分、布局 部分、器件和文档支持 部分以及机械、封装和可订购信息 部分	1
Changes from Revision D (March 2013) to Revision E	Page
• 已更改 将美国国家半导体数据表的布局更改成了 TI 格式	1

5 Pin Configuration and Functions

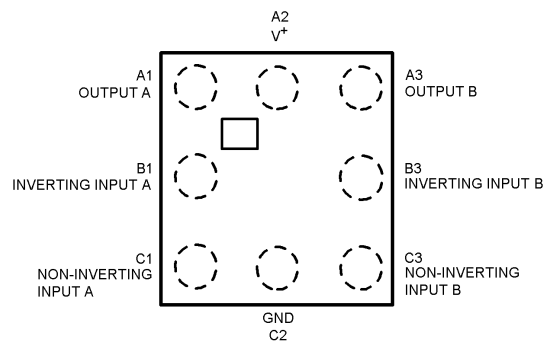
**LMC Package
8-Pin TO-99
Top View**



**P and D Package
8-Pin CDIP, PDIP, SOIC
Top View**



**YZR Package
8-Pin DSBGA
Top View**



Pin Functions

NAME	PIN NO.		I/O	DESCRIPTION
	PDIP/SOIC/ TO-99	DSBGA		
OUTA	1	A1	O	Output, Channel A
-INA	2	B1	I	Inverting Input, Channel A
+INA	3	C1	I	Noninverting Input, Channel A
GND	4	C2	P	Ground
+INB	5	C3	I	Noninverting Input, Channel B
-INB	6	B3	I	Inverting Input, Channel B
OUTB	7	A3	O	Output, Channel B
V+	8	A2	P	Positive power supply

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾⁽²⁾⁽³⁾

			MIN	MAX	UNIT
Differential Input Voltage ⁽⁴⁾				36	V
Input Voltage			-0.3	36	V
Input Current (V _{IN} <-0.3 V) ⁽⁵⁾				50	mA
Power Dissipation ⁽⁶⁾	PDIP			780	mW
	TO-99			660	mW
	SOIC			510	mW
	DSBGA			568	mW
Output Short-Circuit to Ground ⁽⁷⁾				Continu ous	
Lead Temperature (Soldering, 10 seconds)				260	°C
Soldering Information	PDIP Package Soldering (10 seconds)			260	°C
	SOIC Package	Vapor Phase (60 seconds)		215	°C
		Infrared (15 seconds)			220
Storage temperature, T _{stg}			-65	150	°C

- (1) *Absolute Maximum Ratings* indicate limits beyond which damage may occur. *Recommended Operating Conditions* indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and test conditions, see the Electrical Characteristics.
- (2) Refer to RETS193AX for LM193AH military specifications and to RETS193X for LM193H military specifications.
- (3) If Military/Aerospace specified devices are required, please contact the TI Sales Office/Distributors for availability and specifications.
- (4) Positive excursions of input voltage may exceed the power supply level. As long as the other voltage remains within the common-mode range, the comparator will provide a proper output state. The low input voltage state must not be less than -0.3V (or 0.3V below the magnitude of the negative power supply, if used).
- (5) This input current will only exist when the voltage at any of the input leads is driven negative. It is due to the collector-base junction of the input PNP transistors becoming forward biased and thereby acting as input diode clamps. In addition to this diode action, there is also lateral NPN parasitic transistor action on the IC chip. This transistor action can cause the output voltages of the comparators to go to the V^+ voltage level (or to ground for a large overdrive) for the time duration that an input is driven negative. This is not destructive and normal output states will re-establish when the input voltage, which was negative, again returns to a value greater than -0.3V.
- (6) For operating at high temperatures, the LM393 and LM2903 must be derated based on a 125°C maximum junction temperature and a thermal resistance of 170°C/W which applies for the device soldered in a printed circuit board, operating in a still air ambient. The LM193/LM193A/LM293 must be derated based on a 150°C maximum junction temperature. The low bias dissipation and the "ON-OFF" characteristic of the outputs keeps the chip dissipation very small ($P_D \leq 100$ mW), provided the output transistors are allowed to saturate.
- (7) Short circuits from the output to V^+ can cause excessive heating and eventual destruction. When considering short circuits to ground, the maximum output current is approximately 20 mA independent of the magnitude of V^+ .

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±1300	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

	MIN	NOM	MAX	UNIT
Supply Voltage (V^+) - Single Supply	2.0		36	V
Supply Voltage (V^+) - Dual Supply	±1.0		±18	V
Operating Input Voltage on (V_{IN} pin)	0	(V^+) -1.5V		V
Operating junction temperature, T_J : LM193/LM193A	-55		125	°C
Operating junction temperature, T_J : LM2903	-40		85	°C
Operating junction temperature, T_J : LM293	-25		85	°C
Operating junction temperature, T_J : LM393	0		70	°C

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾	UNIT
$R_{\theta JA}$ Junction-to-ambient thermal resistance	170 °C/W

(1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

6.5 Electrical Characteristics: LM193A $V^+ = 5\text{ V}$, $T_A = 25^\circ\text{C}$

Unless otherwise stated.

PARAMETER	TEST CONDITIONS	LM193A			UNIT
		MIN	TYP	MAX	
Input Offset Voltage	See ⁽¹⁾ .		1.0	2.0	mV
Input Bias Current	$I_{IN}(+)$ or $I_{IN}(-)$ with Output in Linear Range, $V_{CM} = 0\text{ V}$ ⁽²⁾		25	100	nA
Input Offset Current	$I_{IN}(+) - I_{IN}(-)$ $V_{CM} = 0\text{ V}$		3.0	25	nA
Input Common Mode Voltage Range	$V^+ = 30\text{ V}$ ⁽³⁾	0		$V^+ - 1.5$	V
Supply Current	$R_L = \infty$ $V^+ = 5\text{ V}$		0.4	1	mA
	$V^+ = 36\text{ V}$		1	2.5	mA
Voltage Gain	$R_L \geq 15\text{ k}\Omega$, $V^+ = 15\text{ V}$ $V_O = 1\text{ V}$ to 11 V	50	200		V/mV
Large Signal Response Time	$V_{IN} = \text{TTL Logic Swing}$, $V_{REF} = 1.4\text{ V}$ $V_{RL} = 5\text{ V}$, $R_L = 5.1\text{ k}\Omega$		300		ns
Response Time	$V_{RL} = 5\text{ V}$, $R_L = 5.1\text{ k}\Omega$ ⁽⁴⁾		1.3		μs
Output Sink Current	$V_{IN}(-) = 1\text{ V}$, $V_{IN}(+) = 0$, $V_O \approx 1.5\text{ V}$	6.0	16		mA
Saturation Voltage	$V_{IN}(-) = 1\text{ V}$, $V_{IN}(+) = 0$, $I_{SINK} \leq 4\text{ mA}$		250	400	mV
Output Leakage Current	$V_{IN}(-) = 0$, $V_{IN}(+) = 1\text{ V}$, $V_O = 5\text{ V}$		0.1		nA

- (1) At output switch point, $V_O \approx 1.4\text{ V}$, $R_S = 0\text{ }\Omega$ with V^+ from 5 V to 30 V ; and over the full input common-mode range (0 V to $V^+ - 1.5\text{ V}$), at 25°C .
- (2) The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.
- (3) The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V . The upper end of the common-mode voltage range is $V^+ - 1.5\text{ V}$ at 25°C , but either or both inputs can go to 36 V without damage, independent of the magnitude of V^+ .
- (4) The response time specified is for a 100 mV input step with 5 mV overdrive. For larger overdrive signals 300 ns can be obtained, see [LMx93 and LM193A Typical Characteristics](#).

6.6 Electrical Characteristics: LM193A ($V^+ = 5\text{ V}$)⁽¹⁾

PARAMETER	TEST CONDITIONS	LM193A			UNIT
		MIN	TYP	MAX	
Input Offset Voltage	See ⁽²⁾			4.0	mV
Input Offset Current	$I_{IN}(+) - I_{IN}(-)$, $V_{CM} = 0\text{ V}$			100	nA
Input Bias Current	$I_{IN}(+)$ or $I_{IN}(-)$ with Output in Linear Range, $V_{CM} = 0\text{ V}$ ⁽³⁾			300	nA
Input Common Mode Voltage Range	$V^+ = 30\text{ V}$ ⁽⁴⁾	0		$V^+ - 2.0$	V
Saturation Voltage	$V_{IN}(-) = 1\text{ V}$, $V_{IN}(+) = 0$, $I_{SINK} \leq 4\text{ mA}$			700	mV
Output Leakage Current	$V_{IN}(-) = 0$, $V_{IN}(+) = 1\text{ V}$, $V_O = 30\text{ V}$			1.0	μA
Differential Input Voltage	Keep All V_{IN} 's $\geq 0\text{ V}$ (or V^- , if Used), ⁽⁵⁾			36	V

- (1) These specifications are limited to $-55^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$, for the LM193/LM193A. With the LM293 all temperature specifications are limited to $-25^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$ and the LM393 temperature specifications are limited to $0^\circ\text{C} \leq T_A \leq 70^\circ\text{C}$. The LM2903 is limited to $-40^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$.
- (2) At output switch point, $V_O \approx 1.4\text{ V}$, $R_S = 0\text{ }\Omega$ with V^+ from 5 V to 30 V ; and over the full input common-mode range (0 V to $V^+ - 1.5\text{ V}$), at 25°C .
- (3) The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.
- (4) The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V . The upper end of the common-mode voltage range is $V^+ - 1.5\text{ V}$ at 25°C , but either or both inputs can go to 36 V without damage, independent of the magnitude of V^+ .
- (5) Positive excursions of input voltage may exceed the power supply level. As long as the other voltage remains within the common-mode range, the comparator will provide a proper output state. The low input voltage state must not be less than -0.3 V (or 0.3 V below the magnitude of the negative power supply, if used).

6.7 Electrical Characteristics: LMx93 and LM2903 $V^+ = 5\text{ V}$, $T_A = 25^\circ\text{C}$

Unless otherwise stated.

PARAMETER	TEST CONDITIONS	LM193-N			LM293-N, LM393-N			LM2903-N			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	See ⁽¹⁾		1.0	5.0		1.0	5.0		2.0	7.0	mV
Input Bias Current	$I_{IN}(+)$ or $I_{IN}(-)$ with Output In Linear Range, $V_{CM} = 0\text{ V}$ ⁽²⁾		25	100		25	250		25	250	nA
Input Offset Current	$I_{IN}(+) - I_{IN}(-)$ $V_{CM} = 0\text{ V}$		3.0	25		5.0	50		5.0	50	nA
Input Common Mode Voltage Range	$V^+ = 30\text{ V}$ ⁽³⁾	0		$V^+ - 1.5$	0		$V^+ - 1.5$	0		$V^+ - 1.5$	V
Supply Current	$R_L = \infty$ $V^+ = 5\text{ V}$		0.4	1		0.4	1		0.4	1.0	mA
	$V^+ = 36\text{ V}$		1	2.5		1	2.5		1	2.5	mA
Voltage Gain	$R_L \geq 15\text{ k}\Omega$, $V^+ = 15\text{ V}$ $V_O = 1\text{ V}$ to 11 V	50	200		50	200		25	100		V/mV
Large Signal Response Time	$V_{IN} = \text{TTL Logic Swing}$, $V_{REF} = 1.4\text{ V}$ $V_{RL} = 5\text{ V}$, $R_L = 5.1\text{ k}\Omega$		300			300			300		ns
Response Time	$V_{RL} = 5\text{ V}$, $R_L = 5.1\text{ k}\Omega$ ⁽⁴⁾		1.3			1.3			1.5		μs
Output Sink Current	$V_{IN}(-) = 1\text{ V}$, $V_{IN}(+) = 0$, $V_O \leq 1.5\text{ V}$	6.0	16		6.0	16		6.0	16		mA
Saturation Voltage	$V_{IN}(-) = 1\text{ V}$, $V_{IN}(+) = 0$, $I_{SINK} \leq 4\text{ mA}$		250	400		250	400		250	400	mV
Output Leakage Current	$V_{IN}(-) = 0$, $V_{IN}(+) = 1\text{ V}$, $V_O = 5\text{ V}$		0.1			0.1			0.1		nA

(1) At output switch point, $V_O = 1.4\text{ V}$, $R_S = 0\text{ }\Omega$ with V^+ from 5 V to 30 V ; and over the full input common-mode range (0 V to $V^+ - 1.5\text{ V}$), at 25°C .

(2) The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.

(3) The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V . The upper end of the common-mode voltage range is $V^+ - 1.5\text{ V}$ at 25°C , but either or both inputs can go to 36 V without damage, independent of the magnitude of V^+ .

(4) The response time specified is for a 100 mV input step with 5 mV overdrive. For larger overdrive signals 300 ns can be obtained, see [LMx93 and LM193A Typical Characteristics](#).

6.8 Electrical Characteristics: LMx93 and LM2903 ($V_+ = 5\text{ V}$)⁽¹⁾

PARAMETER	TEST CONDITIONS	LM193-N			LM293-N, LM393-N			LM290-N			UNIT
		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage	See ⁽²⁾			9			9		9	15	mV
Input Offset Current	$I_{IN(+)} - I_{IN(-)}$, $V_{CM}=0\text{ V}$			100			150		50	200	nA
Input Bias Current	$I_{IN(+)}$ or $I_{IN(-)}$ with Output in Linear Range, $V_{CM}=0\text{ V}$ ⁽³⁾			300			400		200	500	nA
Input Common Mode Voltage Range	$V^+ = 30\text{ V}$ ⁽⁴⁾	0		$V^+ - 2.0$	0		$V^+ - 2.0$	0		$V^+ - 2.0$	V
Saturation Voltage	$V_{IN(-)} = 1\text{ V}$, $V_{IN(+)} = 0$, $I_{SINK} \leq 4\text{ mA}$			700			700		400	700	mV
Output Leakage Current	$V_{IN(-)} = 0$, $V_{IN(+)} = 1\text{ V}$, $V_O = 30\text{ V}$			1.0			1.0			1.0	μA
Differential Input Voltage	Keep All V_{IN} 's $\geq 0\text{ V}$ (or V^- , if Used), ⁽⁵⁾			36			36			36	V

- (1) These specifications are limited to $-55^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$, for the LM193/LM193A. With the LM293 all temperature specifications are limited to $-25^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$ and the LM393 temperature specifications are limited to $0^\circ\text{C} \leq T_A \leq +70^\circ\text{C}$. The LM2903 is limited to $-40^\circ\text{C} \leq T_A \leq +85^\circ\text{C}$.
- (2) At output switch point, $V_O = 1.4\text{ V}$, $R_S = 0\ \Omega$ with V^+ from 5 V to 30 V ; and over the full input common-mode range (0 V to $V^+ - 1.5\text{ V}$), at 25°C .
- (3) The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the reference or input lines.
- (4) The input common-mode voltage or either input signal voltage should not be allowed to go negative by more than 0.3 V . The upper end of the common-mode voltage range is $V^+ - 1.5\text{ V}$ at 25°C , but either or both inputs can go to 36 V without damage, independent of the magnitude of V^+ .
- (5) Positive excursions of input voltage may exceed the power supply level. As long as the other voltage remains within the common-mode range, the comparator will provide a proper output state. The low input voltage state must not be less than -0.3 V (or 0.3 V below the magnitude of the negative power supply, if used).

6.9 Typical Characteristics: LMx93 and LM193A

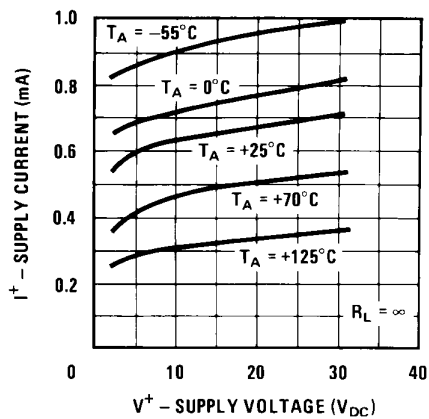


Figure 1. Supply Current

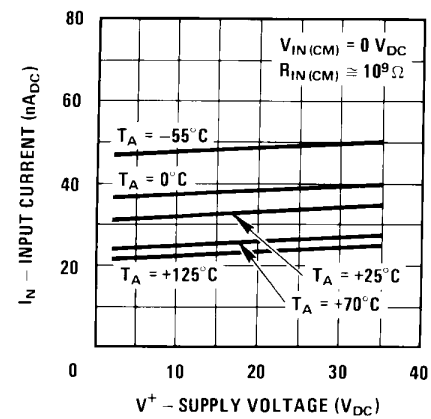


Figure 2. Input Current

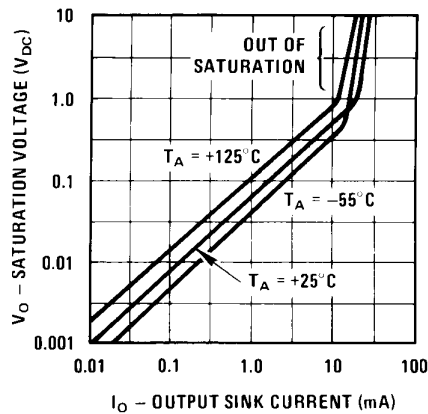


Figure 3. Output Saturation Voltage

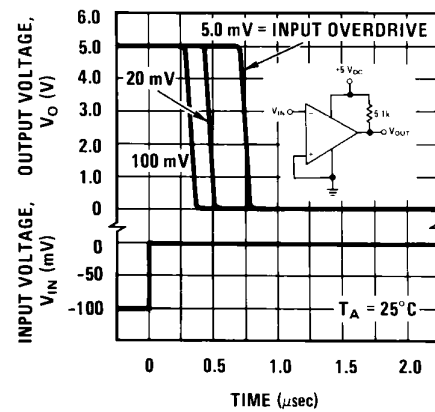


Figure 4. Response Time for Various Input Overdrives—Negative Transition

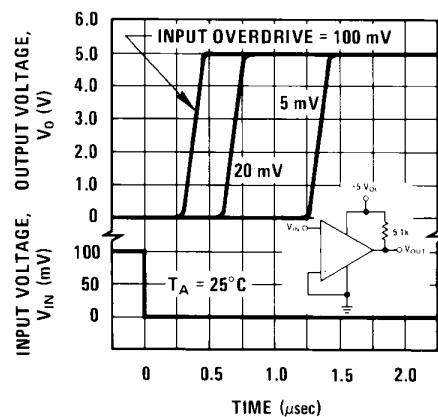


Figure 5. Response Time for Various Input Overdrives—Positive Transition

6.10 Typical Characteristics: LM2903

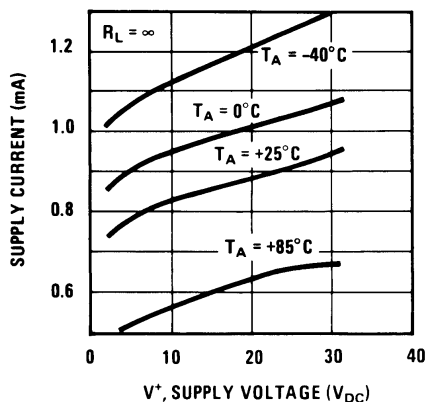


Figure 6. Supply Current

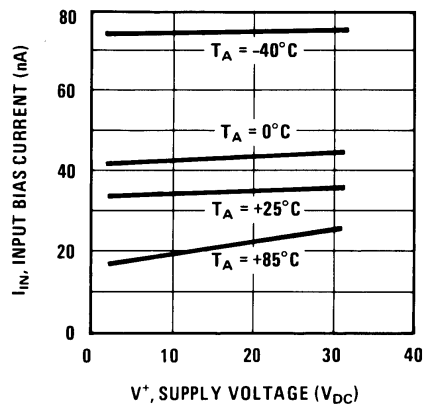


Figure 7. Input Current

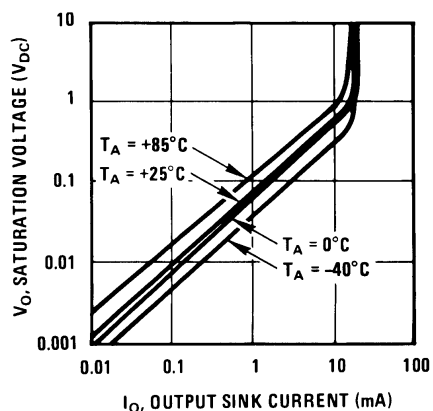


Figure 8. Output Saturation Voltage

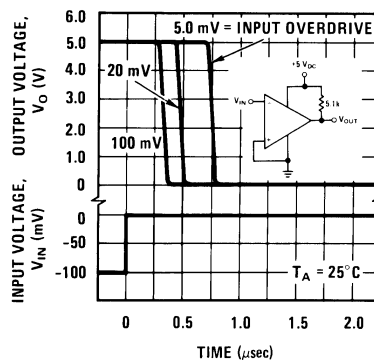


Figure 9. Response Time for Various Input Overdrives—Negative Transition

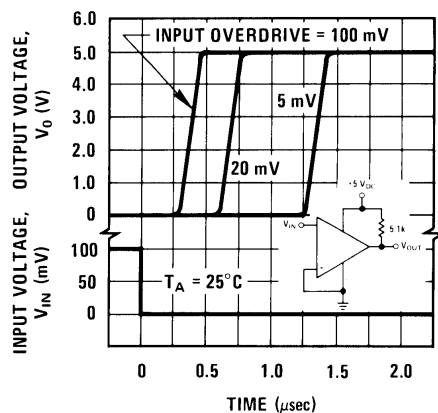


Figure 10. Response Time for Various Input Overdrives—Positive Transition

7 Detailed Description

7.1 Overview

The LM193 provides two independently functioning, high-precision, low V_{OS} drift, low input bias current comparators in a single package. The low power consumption of 0.4 mA at 5 V and the 2.0 V supply operation makes the LM193 suitable for battery powered applications.

7.2 Functional Block Diagram

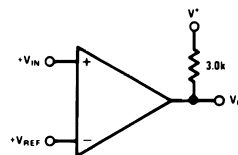


Figure 11. Basic Comparator

7.3 Feature Description

The input bias current of 25 nA enables the LM193 to use even very high impedance nodes as inputs. The differential voltage input range equals the supply voltage range.

The LM193 can be operated with a single supply, where $V+$ can be from 2.0 V to 36 V, or in a dual supply voltage configuration, where GND pin is used as a $V-$ supply. The supply current draws only 0.4 mA for both comparators.

The output of each comparator in the LM193 is the open collector of a grounded-emitter NPN output transistor which can typically draw up to 16 mA.

7.4 Device Functional Modes

A basic comparator circuit is used for converting analog signals to a digital output. The output is HIGH when the voltage on the non-inverting (+IN) input is greater than the inverting (-IN) input. The output is LOW when the voltage on the non-inverting (+IN) input is less than the inverting (-IN) input. The inverting input (-IN) is also commonly referred to as the "reference" or "VREF" input. All pins of any unused comparators should be tied to the negative supply.

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The LM193 series are high gain, wide bandwidth devices which, like most comparators, can easily oscillate if the output lead is inadvertently allowed to capacitively couple to the inputs via stray capacitance. This shows up only during the output voltage transition intervals as the comparator change states. Power supply bypassing is not required to solve this problem. Standard PC board layout is helpful as it reduces stray input-output coupling. Reducing the input resistors to $< 10\text{ k}\Omega$ reduces the feedback signal levels and finally, adding even a small amount (1.0 to 10 mV) of positive feedback (hysteresis) causes such a rapid transition that oscillations due to stray feedback are not possible. Simply socketing the IC and attaching resistors to the pins will cause input-output oscillations during the small transition intervals unless hysteresis is used. If the input signal is a pulse waveform, with relatively fast rise and fall times, hysteresis is not required.

All input pins of any unused comparators should be tied to the negative supply.

The bias network of the LM193 series establishes a drain current which is independent of the magnitude of the power supply voltage over the range of from 2.0 V_{DC} to 30 V_{DC} .

The differential input voltage may be larger than V^+ without damaging the device [Typical Applications](#). Protection should be provided to prevent the input voltages from going negative more than $-0.3\text{ V}_{\text{DC}}$ (at 25°C). An input clamp diode can be used as shown in [Typical Applications](#).

The output of the LM193 series is the uncommitted collector of a grounded-emitter NPN output transistor. Many collectors can be tied together to provide an output OR'ing function. An output pullup resistor can be connected to any available power supply voltage within the permitted supply voltage range and there is no restriction on this voltage due to the magnitude of the voltage which is applied to the V^+ terminal of the LM193 package. The output can also be used as a simple SPST switch to ground (when a pullup resistor is not used). The amount of current which the output device can sink is limited by the drive available (which is independent of V^+) and the β of this device. When the maximum current limit is reached (approximately 16 mA), the output transistor will come out of saturation and the output voltage will rise very rapidly. The output saturation voltage is limited by the approximately $60\text{ }\Omega\text{ }r_{\text{SAT}}$ of the output transistor. The low offset voltage of the output transistor (1.0 mV) allows the output to clamp essentially to ground level for small load currents.

8.2 Typical Applications

8.2.1 Basic Comparator

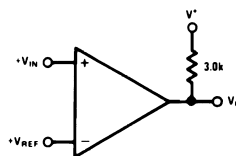


Figure 12. Basic Comparator

8.2.1.1 Design Requirements

The basic usage of a comparator is to indicate when a specific analog signal has exceeded some predefined threshold. In this application, the negative input (IN-) is tied to a reference voltage, and the positive input (IN+) is connected to the input signal. The output is pulled up with a resistor to the logic supply voltage, V^+ with a pullup resistor.

For an example application, the supply voltage is 5V. The input signal varies between 1 V and 3 V, and we want to know when the input exceeds $2.5\text{ V} \pm 1\%$. The supply current draw should not exceed 1 mA.

Typical Applications (continued)

8.2.1.2 Detailed Design Procedure

First, we determine the biasing for the 2.5-V reference. With the 5-V supply voltage, we would use a voltage divider consisting of one resistor from the supply to IN⁻ and an second resistor from IN⁻. The 25 nA of input current bias should be < 1% of the bias current for V_{ref}. With a 100-kΩ resistor from IN⁻ to V⁺ and an additional 100-kΩ resistor from IN⁻ to ground, there would be 25 μA of current through the two resistors. The 3-kΩ pullup shown will need $5\text{ V}/3\text{ k}\Omega \rightarrow 1.67\text{ mA}$, which exceeds our current budget.

With the 400-μA supply current and 25 μA of V_{REF} bias current, there is 575 μA remaining for output pullup resistor; with 5-V supply, we need a pullup larger than 8.7 kΩ. A 10-kΩ pullup is a value that is commonly available and can be used here.

8.2.1.3 Application Curve

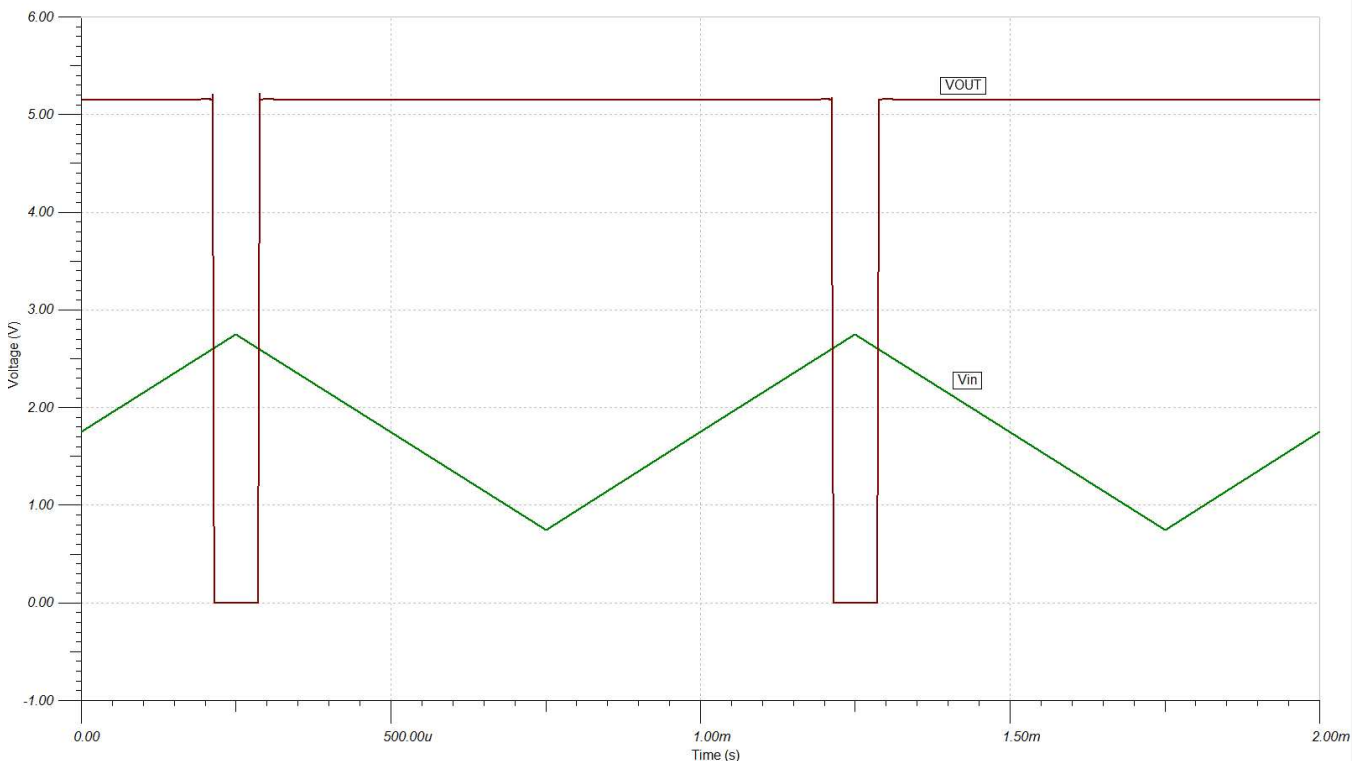
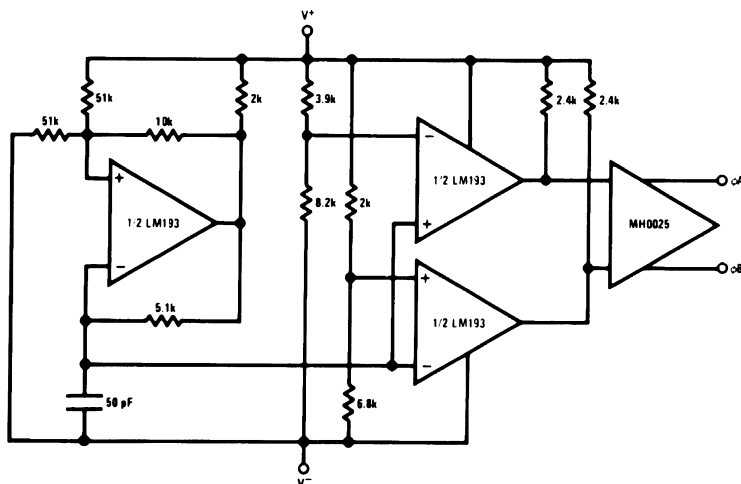


Figure 13. Basic Comparator Response

Typical Applications (continued)

8.2.2 System Examples

8.2.2.1 Split-Supply Application



($V_+ = -15 V_{DC}$ and $V_- = -15 V_{DC}$)

Figure 14. MOS Clock Driver

8.2.2.2 $V_+ = 5.0 V_{DC}$ Application Circuits

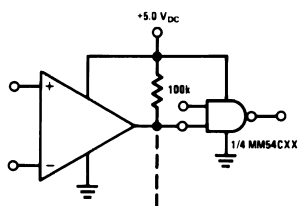


Figure 15. Driving CMOS

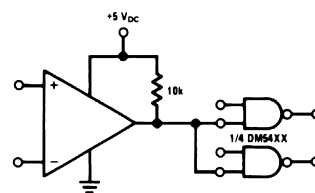


Figure 16. Driving TTL

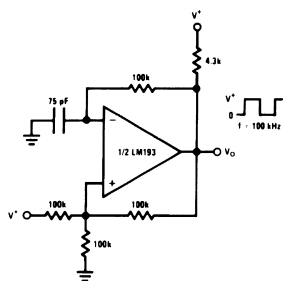
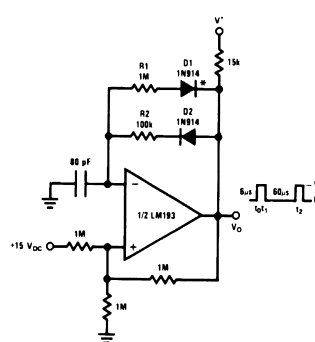


Figure 17. Squarewave Oscillator



* For large ratios of R_1/R_2 ,
D1 can be omitted.

Figure 18. Pulse Generator

Typical Applications (continued)

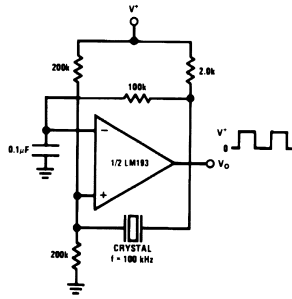
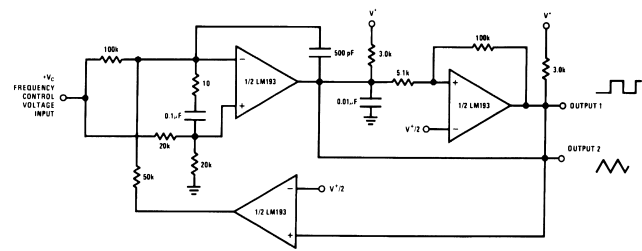


Figure 19. Crystal Controlled Oscillator



$$V^* = +30 V_{DC}$$

$$+250 \text{ mV}_{DC} \leq V_C \leq +50 V_{DC}$$

$$700\text{Hz} \leq f_o \leq 100\text{kHz}$$

Figure 20. Two-Decade High Frequency VCO

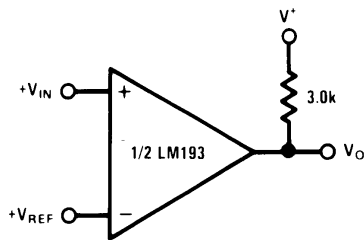


Figure 21. Basic Comparator

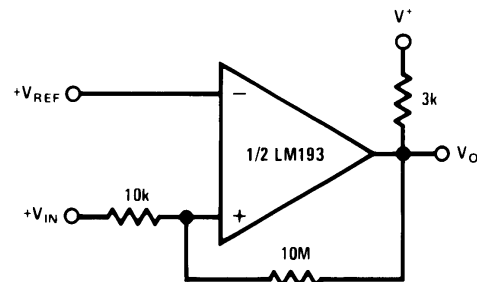


Figure 22. Non-Inverting Comparator With Hysteresis

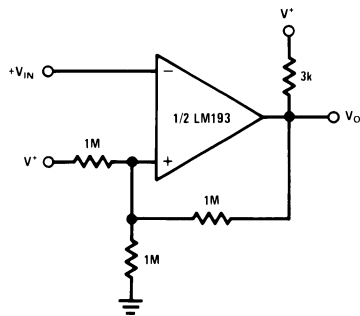
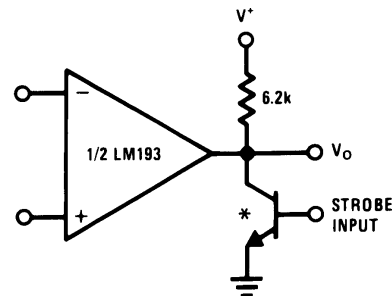


Figure 23. Inverting Comparator With Hysteresis



* OR LOGIC GATE
WITHOUT PULL-UP RESISTOR

Figure 24. Output Strobing

Typical Applications (continued)

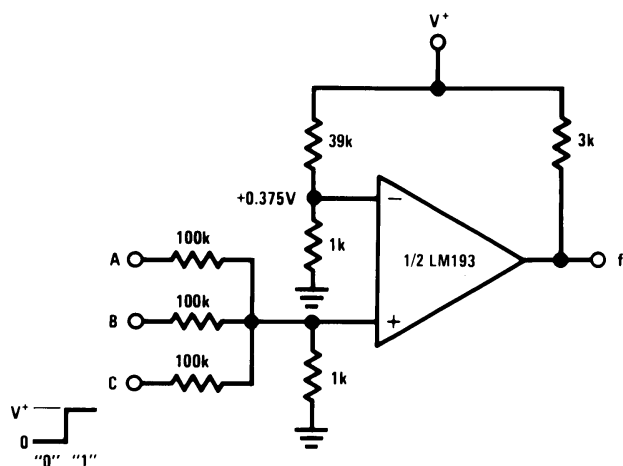


Figure 25. And Gate

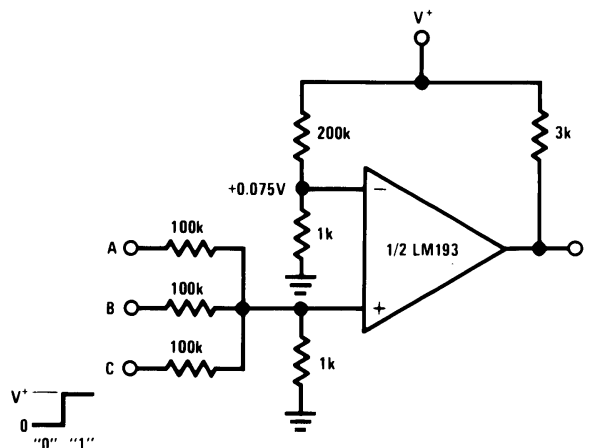


Figure 26. Or Gate

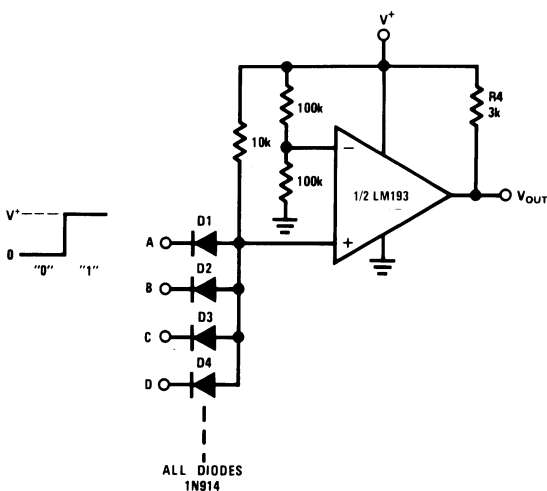


Figure 27. Large Fan-In and Gate

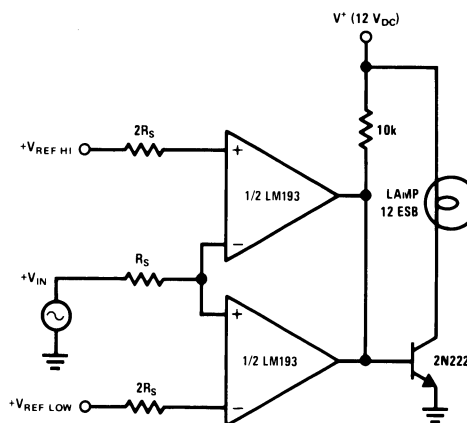


Figure 28. Limit Comparator

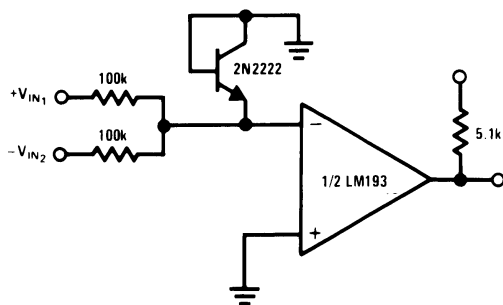


Figure 29. Comparing Input Voltages of Opposite Polarity

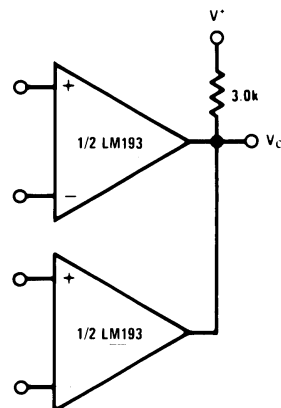
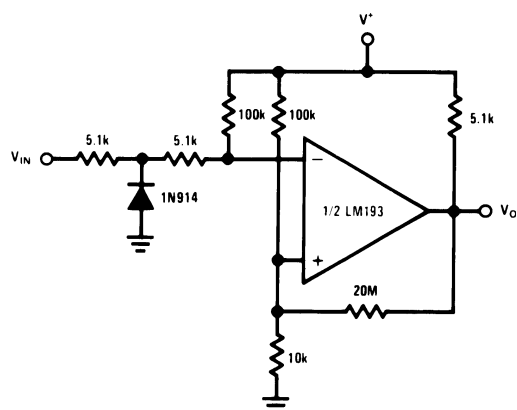
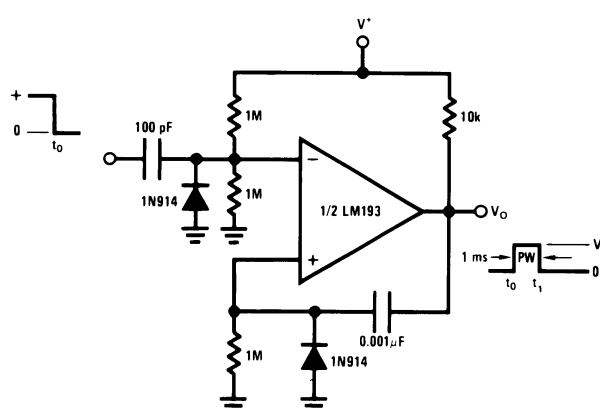
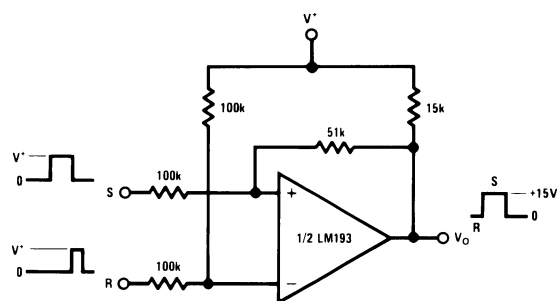
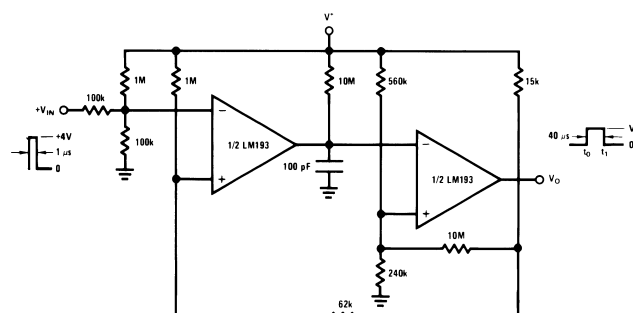
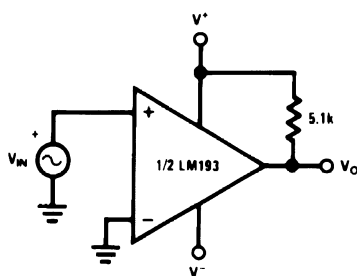
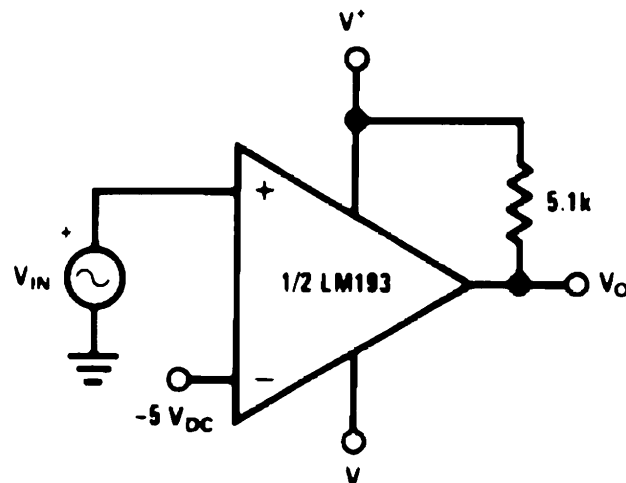


Figure 30. Oring the Outputs

Typical Applications (continued)

Figure 31. Zero Crossing Detector (Single Power Supply)

Figure 32. One-Shot Multivibrator

Figure 33. Bi-Stable Multivibrator

Figure 34. One-Shot Multivibrator With Input Lock Out

Figure 35. Zero Crossing Detector

Figure 36. Comparator With a Negative Reference

Typical Applications (continued)

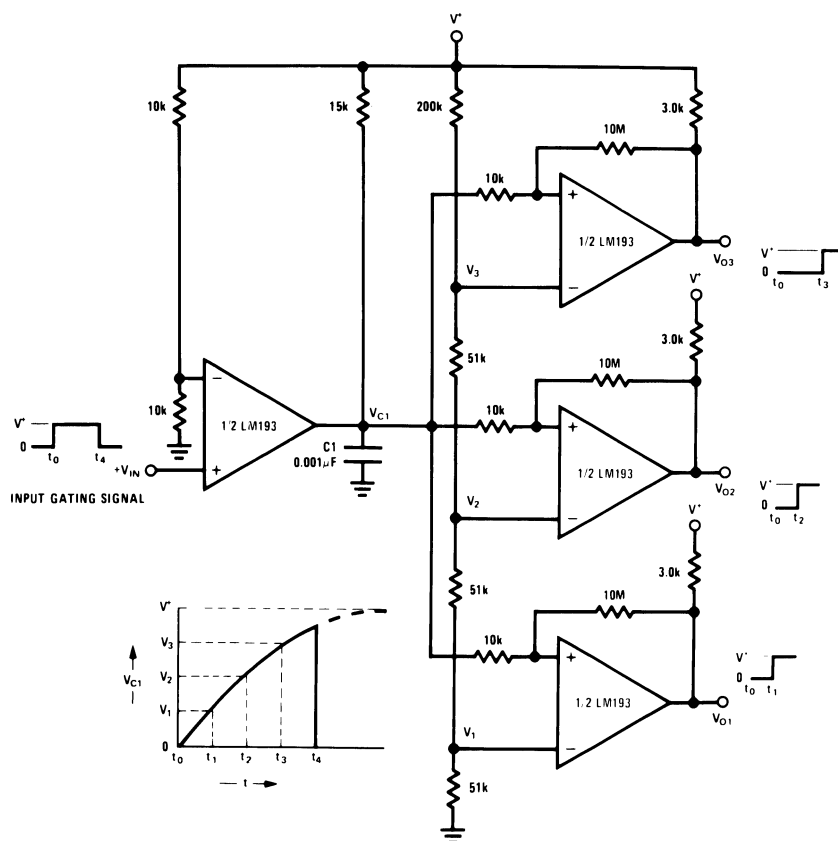


Figure 37. Time Delay Generator

9 Power Supply Recommendations

Even in low frequency applications, the LM139-N can have internal transients which are extremely quick. For this reason, bypassing the power supply with 1.0 μF to ground will provide improved performance; the supply bypass capacitor should be placed as close as possible to the supply pin and have a solid connection to ground. The bypass capacitor should have a low ESR and also a SRF greater than 50MHz.

10 Layout

10.1 Layout Guidelines

Try to minimize parasitic impedances on the inputs to avoid oscillation. Any positive feedback used as hysteresis should place the feedback components as close as possible to the input pins. Care should be taken to ensure that the output pins do not couple to the inputs. This can occur through capacitive coupling if the traces are too close and lead to oscillations on the output. The optimum placement for the bypass capacitor is closest to the V+ and ground pins. Take care to minimize the loop area formed by the bypass capacitor connection between V+ and ground. The ground pin should be connected to the PCB ground plane at the pin of the device. The feedback components should be placed as close to the device as possible minimizing strays.

10.2 Layout Example

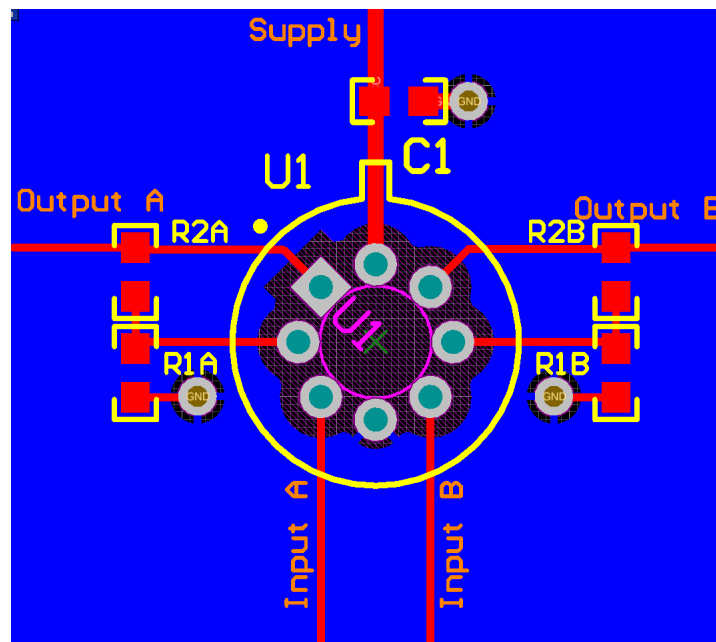


Figure 38. Layout Example

11 器件和文档支持

11.1 相关链接

下表列出了快速访问链接。类别包括技术文档、支持和社区资源、工具和软件，以及立即订购快速访问。

表 1. 相关链接

器件	产品文件夹	立即订购	技术文档	工具与软件	支持和社区
LM193-N	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处
LM2903-N	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处
LM293-N	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处
LM393-N	请单击此处	请单击此处	请单击此处	请单击此处	请单击此处

11.2 接收文档更新通知

要接收文档更新通知，请导航至 TI.com.cn 上的器件产品文件夹。单击右上角的 [通知我](#) 进行注册，即可每周接收产品信息更改摘要。有关更改的详细信息，请查看任何已修订文档中包含的修订历史记录。

11.3 商标

All trademarks are the property of their respective owners.

11.4 静电放电警告



这些装置包含有限的内置 ESD 保护。存储或装卸时，应将导线一起截短或将装置放置于导电泡棉中，以防止 MOS 门极遭受静电损伤。

11.5 术语表

[SLYZ022](#) — TI 术语表。

这份术语表列出并解释术语、缩写和定义。

12 机械、封装和可订购信息

以下页面包含机械、封装和可订购信息。这些信息是指定器件的最新可用数据。数据如有变更，恕不另行通知，且不会对此文档进行修订。如需获取此数据表的浏览器版本，请查阅左侧的导航栏。

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM193AH	ACTIVE	TO-99	LMC	8	500	Non-RoHS & Green	Call TI	Level-1-NA-UNLIM	-55 to 125	(LM193AH, LM193AH)	Samples
LM193AH/NOPB	ACTIVE	TO-99	LMC	8	500	RoHS & Green	Call TI	Level-1-NA-UNLIM	-55 to 125	(LM193AH, LM193AH)	Samples
LM193H	ACTIVE	TO-99	LMC	8	500	Non-RoHS & Green	Call TI	Level-1-NA-UNLIM	-55 to 125	(LM193H, LM193H)	Samples
LM193H/NOPB	ACTIVE	TO-99	LMC	8	500	RoHS & Green	Call TI	Level-1-NA-UNLIM	-55 to 125	(LM193H, LM193H)	Samples
LM2903ITL/NOPB	ACTIVE	DSBGA	YZR	8	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	C 03	Samples
LM2903ITLX/NOPB	ACTIVE	DSBGA	YZR	8	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85	C 03	Samples
LM2903M	LIFEBUY	SOIC	D	8	95	Non-RoHS & Green	Call TI	Level-1-235C-UNLIM	-40 to 85	LM 2903M	
LM2903M/NOPB	ACTIVE	SOIC	D	8	95	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	LM 2903M	Samples
LM2903MX/NOPB	ACTIVE	SOIC	D	8	2500	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	LM 2903M	Samples
LM2903N/NOPB	ACTIVE	PDIP	P	8	40	RoHS & Green	NIPDAU	Level-1-NA-UNLIM	-40 to 85	LM 2903N	Samples
LM293H	ACTIVE	TO-99	LMC	8	500	Non-RoHS & Green	Call TI	Level-1-NA-UNLIM	-25 to 85	(LM293H, LM293H)	Samples
LM293H/NOPB	ACTIVE	TO-99	LMC	8	500	RoHS & Green	Call TI	Level-1-NA-UNLIM	-25 to 85	(LM293H, LM293H)	Samples
LM393M	LIFEBUY	SOIC	D	8	95	Non-RoHS & Green	Call TI	Level-1-235C-UNLIM	0 to 70	LM 393M	
LM393M/NOPB	ACTIVE	SOIC	D	8	95	RoHS & Green	SN	Level-1-260C-UNLIM	0 to 70	LM 393M	Samples
LM393MX	LIFEBUY	SOIC	D	8	2500	Non-RoHS & Green	Call TI	Level-1-235C-UNLIM	0 to 70	LM 393M	
LM393MX/NOPB	ACTIVE	SOIC	D	8	2500	RoHS & Green	SN	Level-1-260C-UNLIM	0 to 70	LM 393M	Samples
LM393N/NOPB	ACTIVE	PDIP	P	8	40	RoHS & Green	NIPDAU	Level-1-NA-UNLIM	0 to 70	LM 393N	Samples

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM393TL/NOPB	ACTIVE	DSBGA	YZR	8	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	0 to 70	C 02	Samples
LM393TLX/NOPB	ACTIVE	DSBGA	YZR	8	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	0 to 70	C 02	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

⁽³⁾ MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

⁽⁵⁾ Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

⁽⁶⁾ Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF LM2903-N, LM293-N :

- Automotive : [LM2903-Q1](#)
- Enhanced Product : [LM293-EP](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects
- Enhanced Product - Supports Defense, Aerospace and Medical Applications

TAPE AND REEL INFORMATION



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2903ITL/NOPB	DSBGA	YZR	8	250	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1
LM2903ITLX/NOPB	DSBGA	YZR	8	3000	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1
LM2903MX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LM393MX	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LM393MX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1
LM393TL/NOPB	DSBGA	YZR	8	250	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1
LM393TLX/NOPB	DSBGA	YZR	8	3000	178.0	8.4	1.7	1.7	0.76	4.0	8.0	Q1

TAPE AND REEL BOX DIMENSIONS



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM2903ITL/NOPB	DSBGA	YZR	8	250	208.0	191.0	35.0
LM2903ITLX/NOPB	DSBGA	YZR	8	3000	208.0	191.0	35.0
LM2903MX/NOPB	SOIC	D	8	2500	367.0	367.0	35.0
LM393MX	SOIC	D	8	2500	367.0	367.0	35.0
LM393MX/NOPB	SOIC	D	8	2500	367.0	367.0	35.0
LM393TL/NOPB	DSBGA	YZR	8	250	208.0	191.0	35.0
LM393TLX/NOPB	DSBGA	YZR	8	3000	208.0	191.0	35.0

TUBE



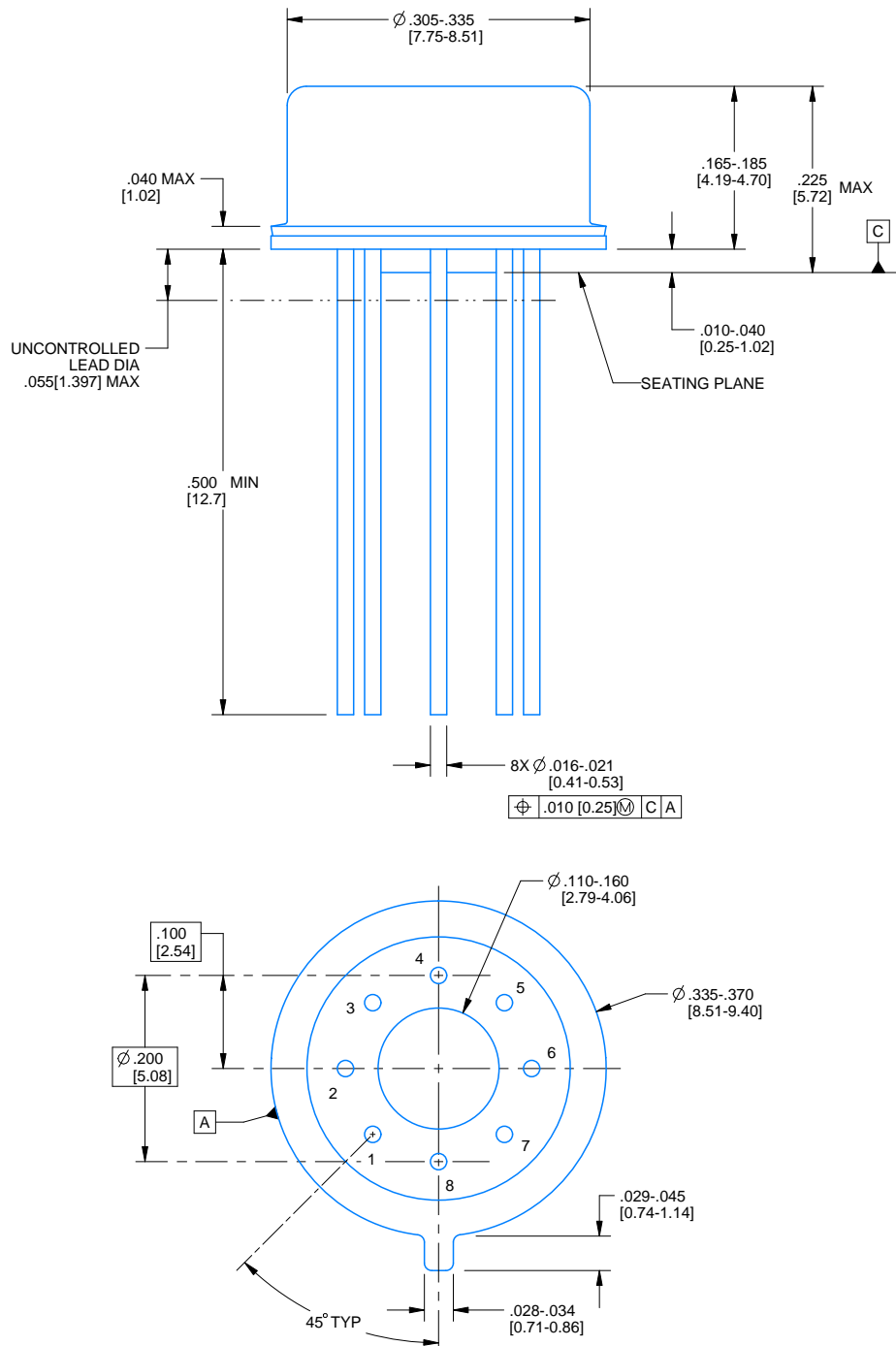
*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
LM2903M	D	SOIC	8	95	495	8	4064	3.05
LM2903M	D	SOIC	8	95	495	8	4064	3.05
LM2903M/NOPB	D	SOIC	8	95	495	8	4064	3.05
LM2903N/NOPB	P	PDIP	8	40	502	14	11938	4.32
LM393M	D	SOIC	8	95	495	8	4064	3.05
LM393M	D	SOIC	8	95	495	8	4064	3.05
LM393M/NOPB	D	SOIC	8	95	495	8	4064	3.05
LM393N/NOPB	P	PDIP	8	40	502	14	11938	4.32

LMC0008A

TO-CAN - 5.72 mm max height

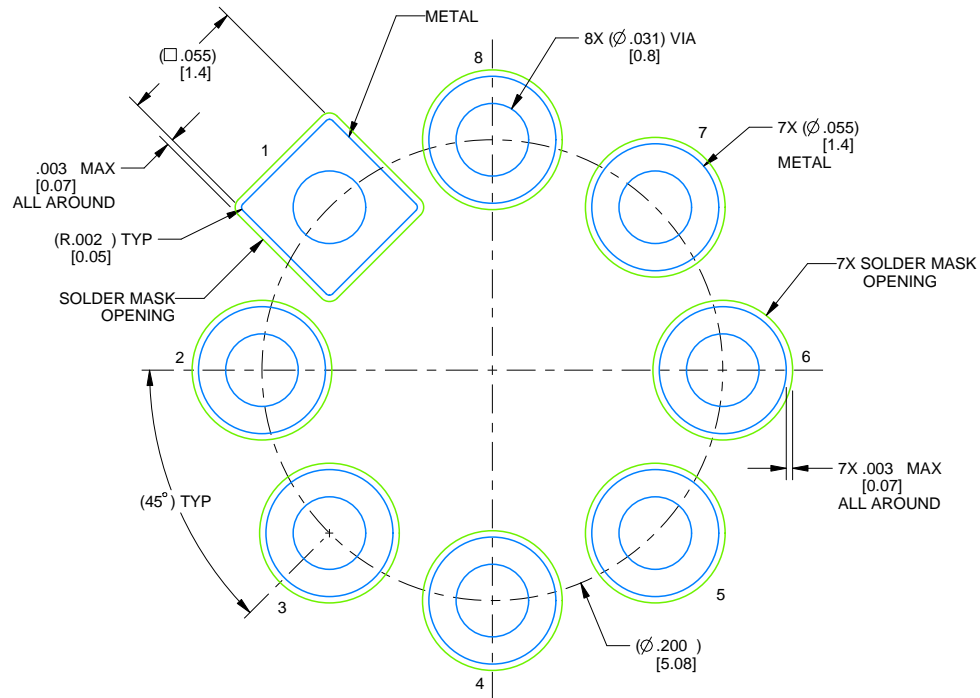
TRANSISTOR OUTLINE



4220610/B 09/2024

NOTES:

1. All linear dimensions are in inches [millimeters]. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Pin numbers shown for reference only. Numbers may not be marked on package.
4. Reference JEDEC registration MO-002/TO-99.



LAND PATTERN EXAMPLE
NON-SOLDER MASK DEFINED
SCALE: 12X

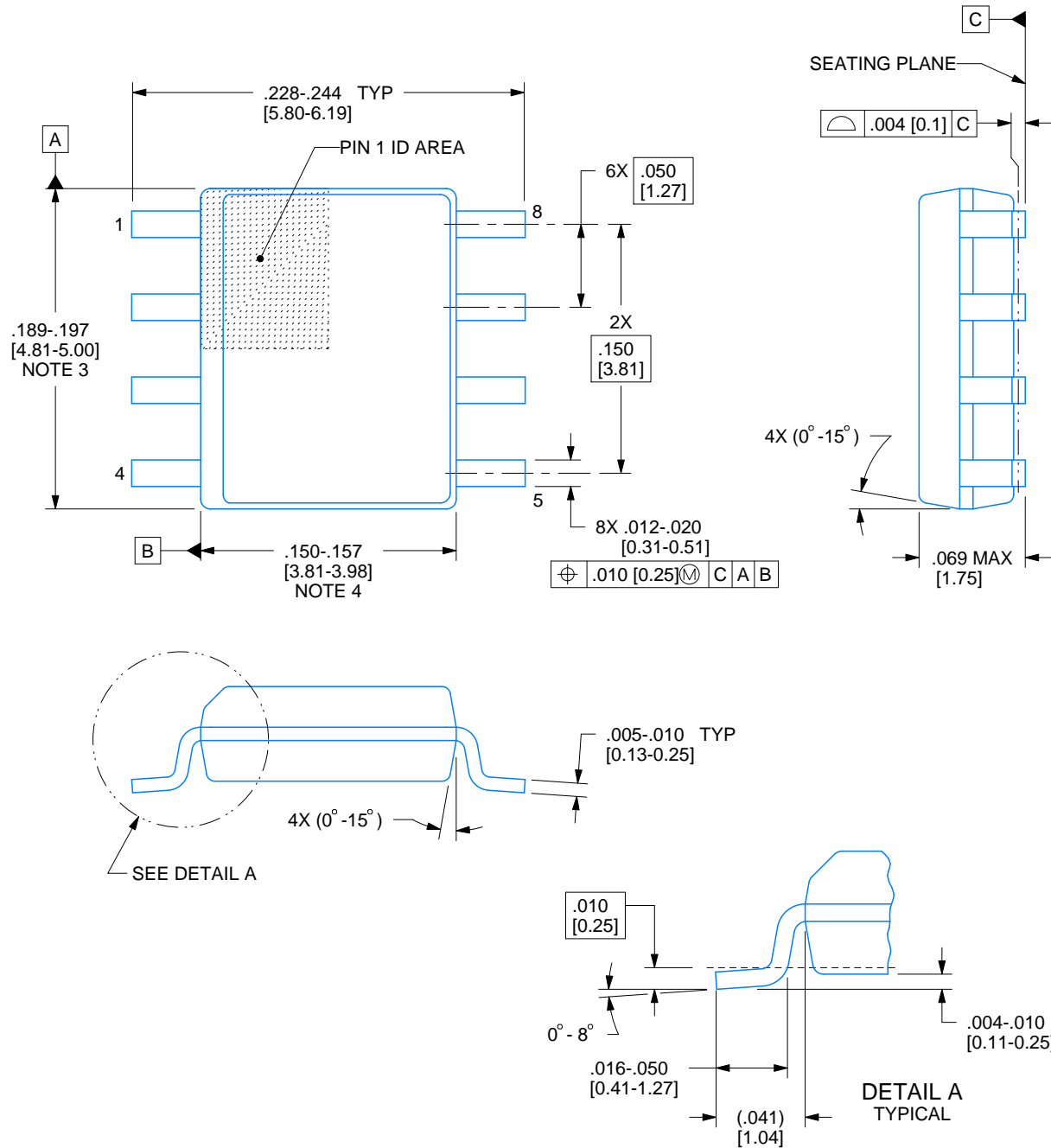
4220610/B 09/2024

D0008A

PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

1. Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
4. This dimension does not include interlead flash.
5. Reference JEDEC registration MS-012, variation AA.

EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE
BASED ON .005 INCH [0.125 MM] THICK STENCIL
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

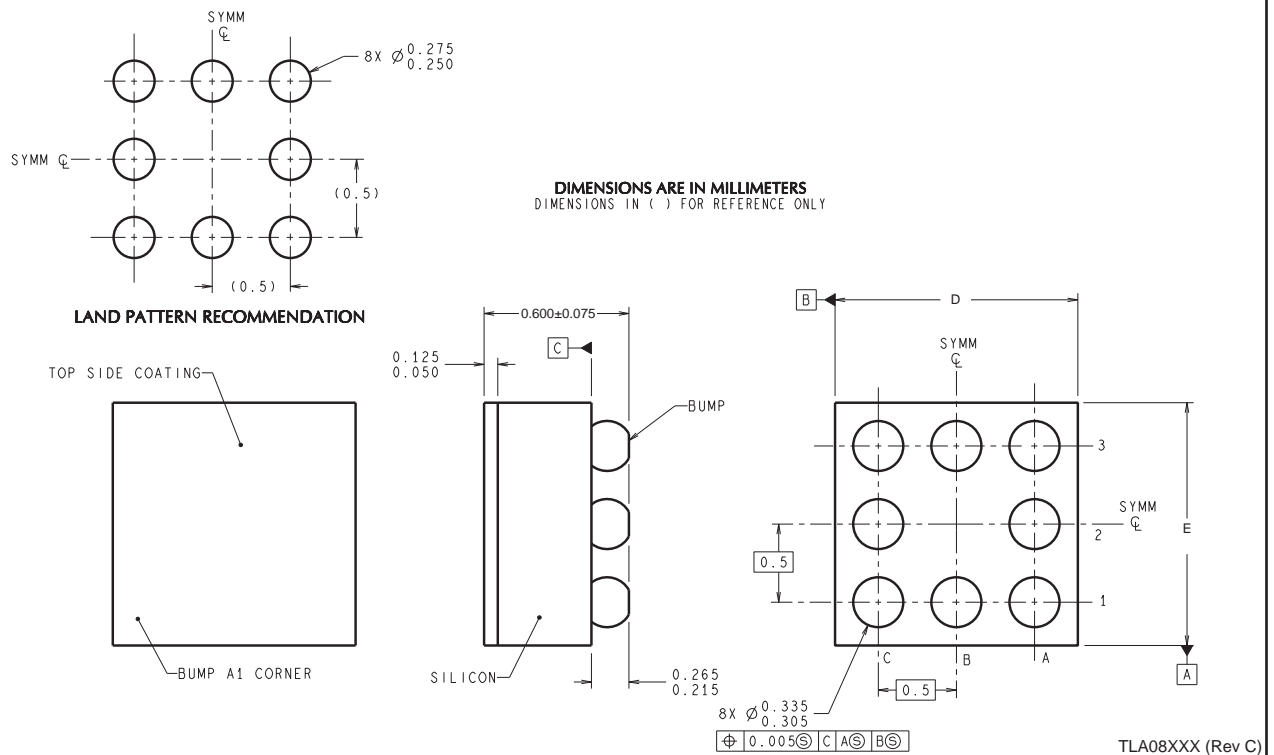
P (R-PDIP-T8)

PLASTIC DUAL-IN-LINE PACKAGE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - C. Falls within JEDEC MS-001 variation BA.

YZR0008



D: Max = 1.54 mm, Min = 1.479 mm

E: Max = 1.54 mm, Min = 1.479 mm

4215045/A 12/12

NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
B. This drawing is subject to change without notice.

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